

#### Description

The Atmel® SAM4C16C and SAM4C8C microcontrollers are system-on-chip solutions for smart energy applications, built around two high-performance 32-bit ARM® Cortex®-M4 RISC processors.

These devices operate at a maximum speed of 120 MHz and feature up to 1 Mbyte of embedded Flash, 152 Kbytes of SRAM and on-chip cache for each core.

The dual ARM Cortex-M4 architecture allows for integration of application layer, communications layers and security functions in a single device, with the ability to extend program and data memory via a 16-bit external bus interface.

The peripheral set includes advanced cryptographic engine, anti-tamper, floating point unit (FPU), five USARTs, two UARTs, two TWIs, up to seven SPIs, as well as a PWM timer, two 3-channel general-purpose 16-bit timers, temperature compensable low-power RTC running on backup area down to 0.5  $\mu$ A, and a 50 x 6 segmented LCD controller.

The SAM4C series is a scalable platform providing, alongside Atmel's industry leading SAM4 standard microcontrollers, unprecedented cost structure, performance and flexibility to smart meter designers worldwide.

## Features

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- Application/ Master Core
  - ARM Cortex-M4 running at up to 120 MHz<sup>(1)</sup>
  - Memory Protection Unit (MPU)
  - DSP Instruction
  - Thumb®-2 instruction set
  - Instruction and Data Cache Controller with 2 Kbytes Cache Memory
  - Memories
    - Up to 1 Mbyte of Embedded Flash for Program Code (I-Code bus) and Program Data (D-Code bus) with Built-in ECC (2-bit error detection and 1-bit correction per 128 bits)
    - 128 Kbytes of Embedded SRAM (SRAM0) for Program Data (System bus)
    - 8 Kbytes ROM with embedded boot loader routines (UART) and In-Application Programming (IAP) routines
- Coprocessor (provides ability to separate application, communication or metrology functions)
  - ARM Cortex-M4F running at up to 120 MHz<sup>(1)</sup>
  - IEEE® 754 Compliant, Single precision Floating-Point Unit (FPU)
  - DSP Instruction
  - Thumb-2 instruction set
  - Instruction and Data Cache Controller with 2 Kbytes Cache Memory
  - Memories
    - 16 Kbytes of Embedded SRAM (SRAM1) for Program Code (I-Code bus) and Program Data (D-Code bus and System bus)
    - 8 Kbytes of Embedded SRAM (SRAM2) for Program Data (System bus)
- Symmetrical/Asynchronous Dual Core Architecture
  - Interrupt-based Interprocessor Communication
  - Asynchronous Clocking
  - One Interrupt Controller (NVIC) for each core
  - Each Peripheral IRQ routed to each NVIC Input
- Cryptography
  - High-performance AES 128 to 256 with various modes (GCM, CBC, ECB, CFB, CBC-MAC, CTR)
  - TRNG (up to 38 Mbit/s stream, with tested diehard and fips)
  - Public Key Crypto accelerator and associated ROM library for RSA, ECC, DSA, ECDSA
  - Integrity Check Module (ICM) based on Secure Hash Algorithm (SHA1, SHA224, SHA256), DMA assisted
- Safety
  - 4 Physical Anti-tamper Detection I/O with Time Stamping and General Backup Registers Immediate Clear
  - Security bit for Device Protection from JTAG accesses
- Shared System Controller
  - Power Supply
    - Embedded Core and LCD Voltage Regulator for single supply operation
    - Power-on-Reset (POR), Brownout Detector (BOD) and Dual Watchdog for safe operation
    - Ultra-low-power Backup mode (< 0.5 µA Typical @ 25°C)
  - Clock
    - Optional 3 to 20 MHz Quartz or ceramic resonator oscillators with Clock Failure Detection
    - Ultra-low-power 32.768 kHz crystal oscillator for RTC with Frequency Monitoring
    - High precision 4/8/12 MHz factory trimmed internal RC oscillator with on-the-fly trimming capability

- One High Frequency PLL up to 240 MHz, One 8 MHz PLL with internal 32 kHz input, as source for High Frequency PLL
  - Low power Slow Clock Internal RC oscillator as permanent clock
- Ultra low-power RTC with Gregorian and Persian Calendar, waveform generation in low-power modes and clock calibration circuitry for 32.768 kHz crystal frequency compensation circuitry
- Up to 23 peripheral DMA (PDC) channels
- Shared Peripherals
  - One Segmented LCD Controller
    - Display Capacity of Fifty Segments and Six Common Terminals
    - Software Selectable LCD Output Voltage (Contrast)
    - Low Current Consumption in Steady State Mode
    - Can be used in Backup Mode
  - Up to five USARTs with ISO7816, IrDA®, RS-485, SPI and Manchester Mode
  - Two 2-wire UARTs with one UART (UART1) supporting optical transceiver allowing to establish an electrically isolated serial communication with hand-held equipment, such as calibrators, compliant with ANSI-C12.18 or IEC62056-21 norms.
  - Two 400 kHz Master/Slave and Multi-Master Two-wire Interface (I2C compatible)
  - Up to seven Serial Peripheral Interface (SPI)
  - Two 3-Channel 16-bit Timer/Counter with capture, waveform, compare and PWM mode. Quadrature Decoder Logic and 2-bit Gray Up/Down Counter for Stepper Motor
  - 4-channel 16-bit Pulse Width Modulator
  - 32-bit Real-time Timer
- Analog Conversion Block
  - 8-channel, 500 kS/s, Low Power, 10-Bit SAR ADC with Digital averager providing 12-bit resolution @ 30 kS/s
  - Software Controlled On-Chip Reference ranging from 1.6V to 3.4V
  - Temperature Sensor and Backup Battery Voltage Measurement Channel
- Debug
  - Star Topology AHB-AP Debug Access Port Implementation with common SW-DP / SWJ-DP providing higher performance than daisy-chain topology.
  - Debug Synchronization between both Cores (cross triggering to/from each core for Halt and Run Mode)
- I/O
  - 74 I/O lines with external interrupt capability (edge or level sensitivity), schmitt trigger, internal pull-up/pull-down, debouncing, glitch filtering and on-die Series Resistor Termination
- Packages
  - 100-lead LQFP, 14 x 14 mm, pitch 0.5 mm

Note: 1. 120 MHz: -40/+85°C, VDDCORE = 1.2V

## 1. Configuration Summary

The SAM4C series devices differ in memory size, package and features. [Table 1-1](#) summarizes configuration of the device family.

**Table 1-1. Configuration Summary**

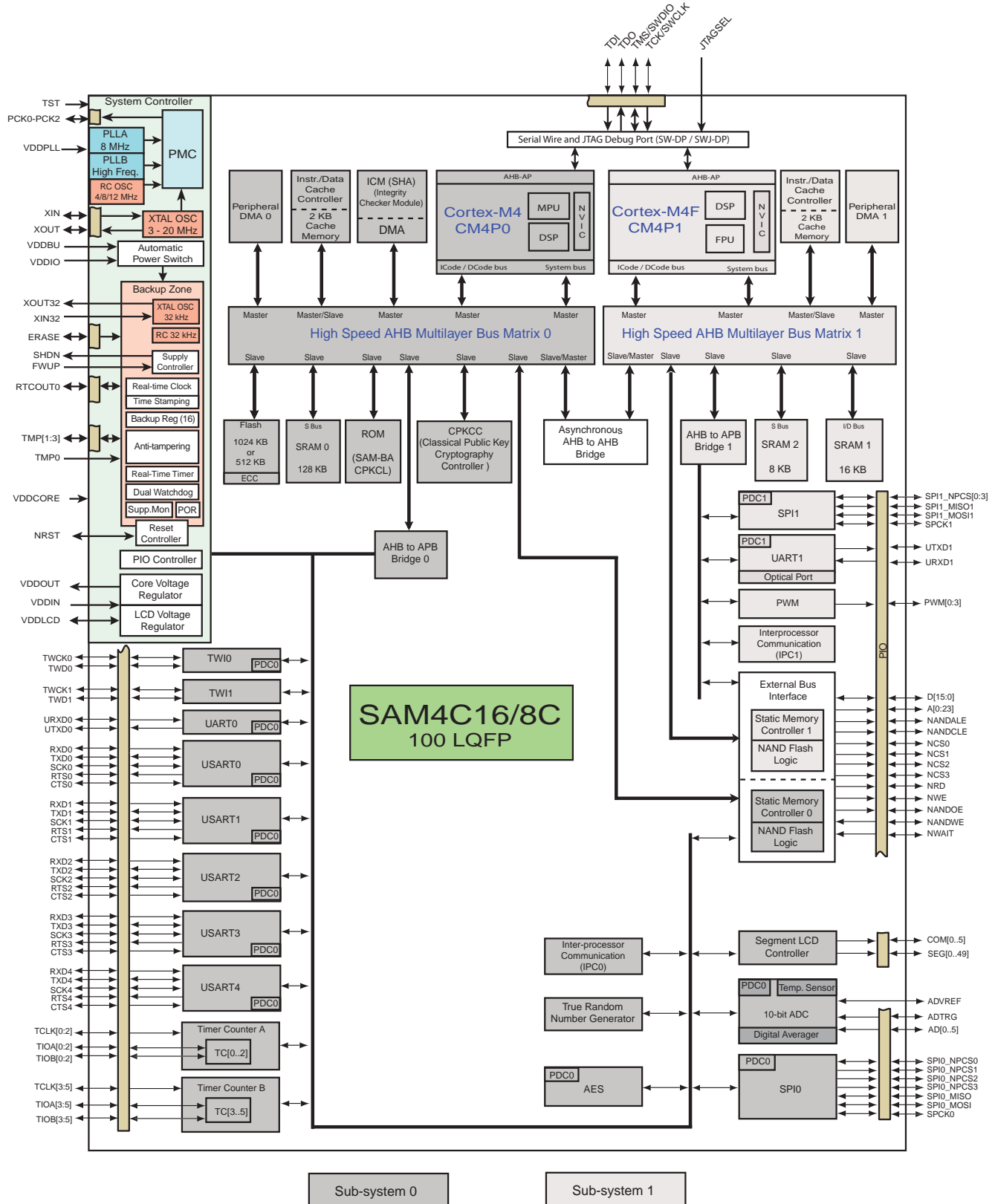
Feature	SAM4C16C	SAM4C8C
Flash	1024 Kbytes	512 Kbytes
SRAM	128 + 16 + 8 Kbytes	128 + 16 + 8 Kbytes
Package	LQFP 100	LQFP 100
Number of PIOs		
External Bus Interface	16-bit data	
16-bit Timer	6 ch.	
16-bit PWM	4 ch.	
UART/USART	2/5	
SPI <sup>(1)</sup>	7 (2/5 + 5)	
TWI	2	
10-bit ADC Channels <sup>(2)</sup>	8	
Cryptography	AES, CPKCC, ICM (SHA), TRNG	
Segmented LCD	50x6	
Anti-Tampering Input	4	
Flash Page Size	512	
Flash Pages	2048	1024
Flash Lock Region Size	8192	8192
Flash Lock Bits	128	64

Notes: 1.  $2 / 5 + 5$  = Number of SPI Controllers / Number of Chip Selects + Number of USART with SPI Mode.

2. One channel is reserved for internal temperature sensor and one channel for VDDBU measurement.

## 2. Block Diagram

Figure 2-1. SAM4C16/8 100-pin Block Diagram



### 3. Signal Description

Table 3-1 provides details on signal names classified by peripheral.

Table 3-1. Signal Description List

Signal Name	Function	Type	Active Level	Voltage Reference	Comments
Power Supplies					
VDDIO	See <a href="#">Table 5-1 on page 12</a>	Power			
VDDBU		Power			
VDDIN		Power			
VDDLCD		Power			
VDDOUT		Power			
VDDPLL		Power			
VDDCORE		Power			
GND		Ground			
Clocks, Oscillators and PLLs					
XIN	Main Crystal Oscillator Input	Analog Digital		VDDIO	
XOUT	Main Crystal Oscillator Output				
XIN32	Slow Clock Crystal Oscillator Input	Analog Digital		VDDBU	
XOUT32	Slow Clock Crystal Oscillator Output				
PCK0 - PCK2	Programmable Clock Output	Output		VDDIO	
Real-time Clock					
RTCOUT0	Programmable RTC waveform output	Digital Output		VDDIO	
FWUP	Force Wake-up input	Digital Input	Low	VDDBU	External Pull-up needed
TMP0	Anti-tampering Input 0	Digital Input		VDDBU	
TMP[1-3]	Anti-tampering Input 1 to 3	Digital Input		VDDIO	
SHDN	Active Low Shut-down Control	Digital Output		VDDBU	0: The device is in Backup mode 1: The device is running (not in Backup mode)

**Table 3-1. Signal Description List (Continued)**

Signal Name	Function	Type	Active Level	Voltage Reference	Comments
Serial Wire/JTAG Debug Port - SWJ-DP					
TCK/SWCLK	Test Clock/Serial Wire Clock	Digital Input		VDDIO	
TDI	Test Data In				
TDO/TRACESWO	Test Data Out / Trace Asynchronous Data Out	Digital Output			
TMS/SWDIO	Test Mode Select input / Serial Wire Input/Output	Digital I/O			
JTAGSEL	JTAG Selection	Digital Input	High	VDDBU	Permanent Internal pull-down
Flash Memory					
ERASE	Flash and NVM Configuration Bits Erase Command	Digital Input	High	VDDIO	
Reset/Test					
NRST	Synchronous Microcontroller Reset	Digital I/O	Low	VDDIO	Permanent Internal pull-up
TST	Test Select	Digital Input		VDDBU	Permanent Internal pull-down
Universal Asynchronous Receiver Transceiver - UARTx					
URXDx	UART Receive Data	Digital/ Analog Input		VDDIO	Analog Mode for Optical Receiver
UTXDx	UART Transmit Data	Digital Output			
PIO Controller - PIOA - PIOB - PIOC					
PA0 - PA31	Parallel IO Controller A	Digital I/O		VDDIO	
PB0 - PB31	Parallel IO Controller B				
PC0 - PC9	Parallel IO Controller C				
External Bus Interface					
D0-D15	Data Bus	Digital I/O		VDDIO	
A0-A23	Address Bus	Digital Output			
NWAIT	External Wait signal	Digital Input	Low		
Static Memory Controller - SMC					
NCS0 - NCS3	Chip Select Lines	Digital Output	Low	VDDIO	
NRD	Read Signal				
NWE	Write Enable				
NBS0- NBS1	Byte Mask Signal				
NWR0-NWR1	Write Signal				

**Table 3-1. Signal Description List (Continued)**

Signal Name	Function	Type	Active Level	Voltage Reference	Comments
NAND Flash Logic					
NANDOE	NAND Flash Output Enable	Digital Output	Low	VDDIO	
NANDWE	NAND Flash Write Enable				
NANDCS	NAND Chip Select				SMC chip select used for Nand Flash Logic
Universal Synchronous Asynchronous Receiver Transmitter - USARTx					
SCKx	USARTx Serial Clock	Digital I/O		VDDIO	
TXDx	USARTx Transmit Data	Digital Output			
RXDx	USARTx Receive Data	Digital Input		VDDIO	
RTSx	USARTx Request To Send	Digital Output			
CTSx	USARTx Clear To Send	Digital Input			
Timer/Counter - TC					
TCLKx	TC Channel x External Clock Input	Digital Input		VDDIO	
TIOAx	TC Channel x I/O Line A	Digital I/O			
TIOBx	TC Channel x I/O Line B				
Pulse Width Modulation Controller - PWMx					
PWMx	PWM Waveform Output for channel x	Digital Output		VDDIO	
Serial Peripheral Interface - SPI					
SPIx_MISOx	Master In Slave Out	Digital Input		VDDIO	
SPI_MOSIx	Master Out Slave In	Digital Output			
SPCKx	SPI Serial Clock				
SPIx_NPCS0	SPI Peripheral Chip Select 0		Low		NPCS0 is also NSS for slave mode
SPIx_NPCS1 - SPIx_NPCS3	SPI Peripheral Chip Select	Output	Low		
Segmented LCD Controller - SLCDC					
COM[5:0]	Common Terminals	Output		VDDIO	
SEG[49:0]	Segment Terminals	Output			
Two-wire Interface - TWI					
TWDx	TWlx Two-wire Serial Data	Digital I/O		VDDIO	
TWCKx	TWlx Two-wire Serial Clock	Digital Output			



**Table 3-1. Signal Description List (Continued)**

Signal Name	Function	Type	Active Level	Voltage Reference	Comments
Analog					
ADVREF	External Voltage Reference for ADC	Analog Input		VDDIN	
10-bit Analog-to-Digital Converter - ADC					
AD0-AD5	Analog Inputs	Analog, Digital		VDDIO	ADC input range limited to [0 - ADVREF]
ADTRG	ADC Trigger	Input			
Fast Flash Programming Interface - FFPI					
PGMEN0-PGMEN1	Programming Enabling	Digital Input		VDDIO	
PGMM0-PGMM3	Programming Mode				
PGMD0-PGMD15	Programming Data	Digital I/O			
PGMRDY	Programming Ready	Digital Output	High		
PGMNVALID	Data Direction		Low		
PGMNOE	Programming Read	Digital Input	Low		
PGMCK	Programming Clock				
PGMNCMD	Programming Command		Low		

## 4. Package and Pinout

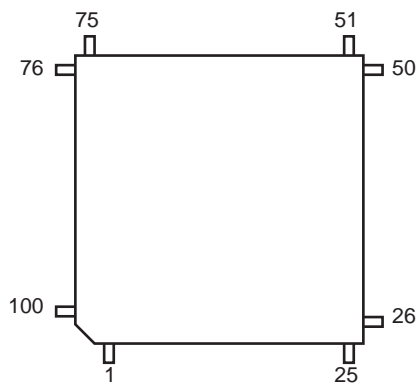
### 4.1 SAM4C 100-lead LQFP Package and Pinout

#### 4.1.1 100-lead LQFP Package Outline

The SAM4C 100-lead LQFP package has a 0.5 mm ball pitch and respects Green Standards.

Figure 4-1 shows the orientation of the 100-lead LQFP package. Refer to the “Mechanical Characteristics” section of the datasheet for the SAM4C 100-lead LQFP package mechanical drawing.

**Figure 4-1. Orientation of the 100-lead LQFP Package**



## 4.1.2 100-lead LQFP Pinout

Table 4-1. SAM4C16/C8 100-lead LQFP Pinout

1	PB6	26	TDI/PB0	51	PA31	76	VDDIO
2	PB7	27	TCK/SWCLK/PB3	52	GND	77	ADVREF
3	PB18	28	TMS/SWDIO/PB2	53	VDDPLL	78	GND
4	GND	29	ERASE/PC9	54	PC8	79	PB31/AD5
5	PB19	30	TDO/TRACESWO/ PB1/RTCOUT0	55	PC5	80	PB23/AD4
6	PB8	31	PC1	56	PC4	81	PB13/AD3
7	PB22	32	PC6	57	PC3	82	PA5/AD2/PGMRDY
8	PB30	33	VDDIO	58	PC2	83	PA4/AD1/PGMNCMD
9	PB25	34	VDDBU	59	PA29	84	PA12/AD0/PGMD0
10	PB24	35	FWUP	60	PA28	85	VDDIN
11	VDDCORE	36	JTAGSEL	61	PA27/PGMD15	86	VDDOUT
12	PB29	37	SHDN	62	PA6/PGMNOE	87	PB21
13	PB9	38	TST	63	VDDCORE	88	PB20
14	PB10	39	TMP0	64	PA3	89	VDDCORE
15	PB11	40	XIN32	65	PA21/PGMD9	90	PA0/PGMEN0
16	PB12	41	XOUT32	66	PA22/PGMD10	91	PB27/TMP2
17	PB14	42	GND	67	PA23/PGMD11	92	VDDLCD
18	PB15	43	PB4	68	PA9/PGMM1	93	PB26
19	PA26/PGMD14	44	VDDCORE	69	PA10/PGMM2	94	PB28/TMP3
20	PA25/PGMD13	45	PB5	70	PA11/PGMM3	95	PB16/TMP1
21	PA24/PGMD12	46	PC7	71	PA13/PGMD1	96	PA1/PGMEN1
22	PA20/PGMD8	47	PC0	72	PA14/PGMD2	97	PB17
23	PA19/PGMD7	48	NRST	73	PA15/PGMD3	98	PA7/PGMNVALID
24	PA18/PGMD6	49	VDDIO	74	PA16/PGMD4	99	VDDIO
25	PA8/PGMM0	50	PA30	75	PA17/PGMD5	100	PA2

## 5. Power Supply and Power Control

### 5.1 Power Supplies

The SAM4C has several types of power supply pins. In most cases, a single supply scheme for all power supplies (except VDDBU) is possible. Figure 5-1 below shows power domains according to the different power supply pins.

Figure 5-1. Power Domains

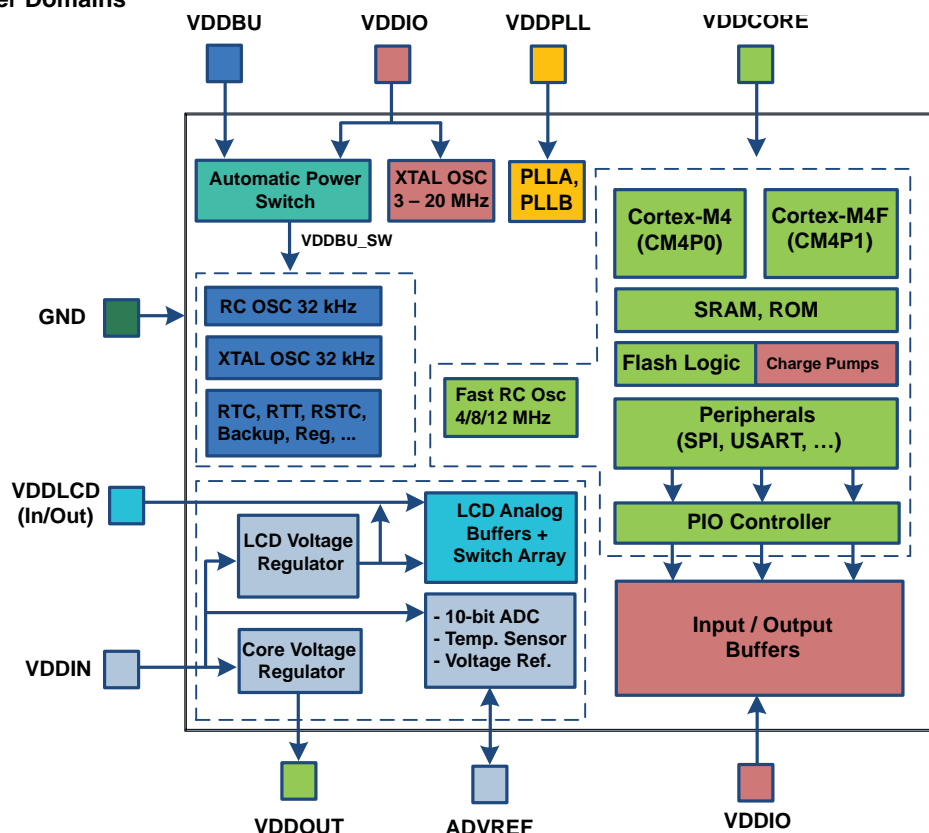


Table 5-1. Power Supply Voltage Ranges

Power Supplies	Ranges	Comments
VDDIO	1.6V to 3.6V	<ul style="list-style-type: none"> <li>- Flash Memory Charge Pumps Supply for Erase and Program Operations, and Read operation.</li> <li>- Input Output buffers Supply</li> </ul>
VDDBU	1.6V to 3.6V	Backup Area power supply. VDDBU is automatically disconnected when VDDIO is present (> 1.9V)
VDDIN	1.6V to 3.6V	1.6V min. if LCD and ADC not used, else 2.5V
VDDLCD	2.5V to 3.6V	<ul style="list-style-type: none"> <li>- LCD Voltage Regulator Output</li> <li>- External LCD power supply input (LCD regulator not used)</li> <li>- VDDIO/VDDIN need to be supplied when the LCD Controller is used</li> </ul>
VDDOUT	1.2V Output	120 mA Output Current
VDDPLL	1.08V to 1.32V	
VDDCORE	1.08V to 1.32V	

### 5.1.1 Core Voltage Regulator

The SAM4C embeds a core voltage regulator that is managed by the Supply Controller.

It features two operating modes:

- In Normal mode, the quiescent current of the voltage regulator is less than 500  $\mu$ A when sourcing maximum load current, i.e. 120 mA. Internal adaptive biasing adjusts the regulator quiescent current depending on the required load current. In Wait Mode quiescent current is only 5  $\mu$ A.
- In Backup mode, the voltage regulator consumes less than 100 nA while its output (VDDOUT) is driven internally to GND.

The default output voltage is 1.20V and the start-up time to reach Normal mode is less than 500  $\mu$ s.

For adequate input and output power supply decoupling/bypassing, refer to the “Voltage Regulator” section of the product electrical characteristics.

### 5.1.2 LCD Voltage Regulator

The SAM4C embeds an adjustable LCD voltage regulator that is managed by the Supply Controller.

The LCD voltage regulator output voltage is software selectable from 2.4V to 3.4V with 16 levels. Its input (VDDIN) must be supplied in the range of 2.5 to 3.6V. The maximum drop-out is 150 mV with a maximum load of 100  $\mu$ A (corresponding to the LCD I/Os toggling).

This internal regulator is designed to supply the Segment LCD outputs when they are used so that it can be used to adjust the contrast.

The operational current of the LCD voltage regulator is 3  $\mu$ A (typical case).

If not used, its output (VDDLCD) can be bypassed (Hi-z mode) and an external power supply can be provided onto the VDDLCD pin. In this case, VDDIO still needs to be supplied.

The LCD voltage regulator can be used in every power modes (Backup, Wait, Sleep and Active).

For adequate input and output power supply decoupling/bypassing, refer to the “Voltage Regulator” section of the product electrical characteristics.

### 5.1.3 Automatic Power Switch

The SAM4C features an automatic power switch between VDDBU and VDDIO. When VDDIO is present ( $>1.9$ V min), the backup zone power supply is powered by VDDIO and current consumption on VDDBU is about zero (around 100 nA, typ.). Switching between VDDIO and VDDBU is transparent to the user.

### 5.1.4 Typical Powering Schematics

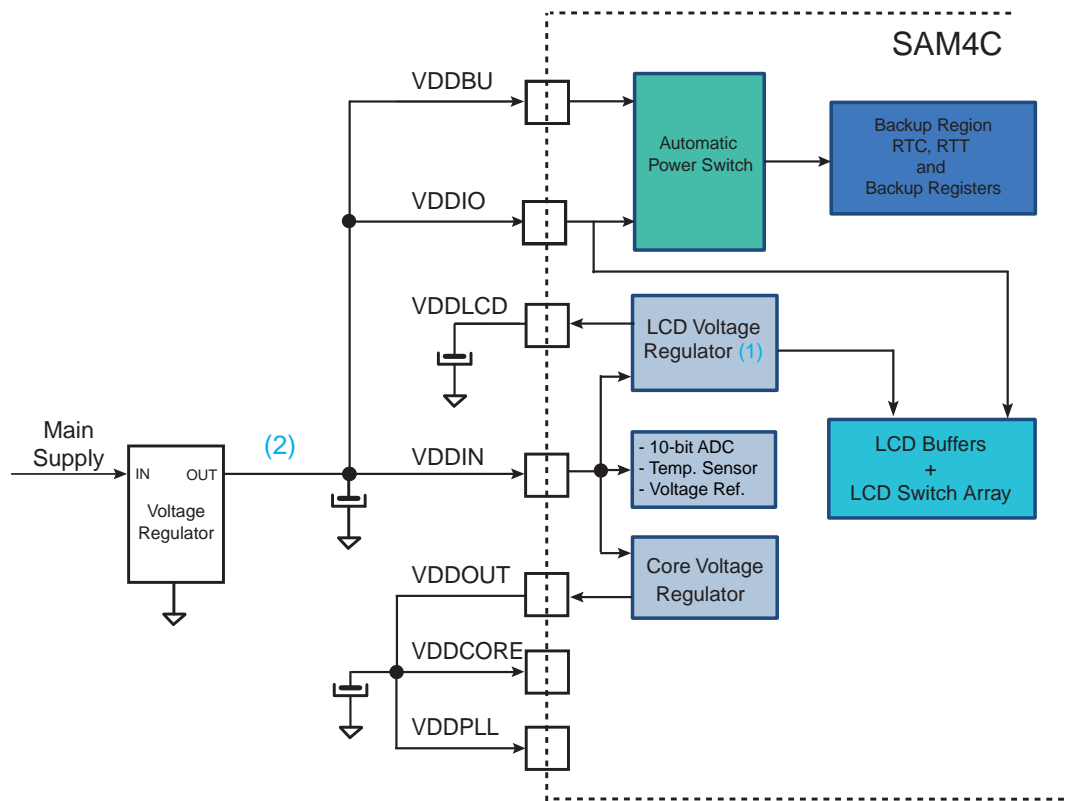
The SAM4C supports a 1.6V to 3.6V single supply mode. This range is reduced to 2.5V to 3.6V when either the LCD Controller or the ADC Controller is used.

Note: The schematics shown in [Figure 5-3](#), [Figure 5-4](#) and [Figure 5-2](#) are principal schematics.

### 5.1.4.1 Single Supply Operation

Figure 5-2 below shows a typical power supply scheme with a single power source. VDDIO, VDDIN, and VDDDBU are derived from the main power source (typically a 3.3V regulator output) while VDDCORE, VDDPLL, VDDLCD are fed by the embedded regulators outputs.

**Figure 5-2. Single Power Supply**

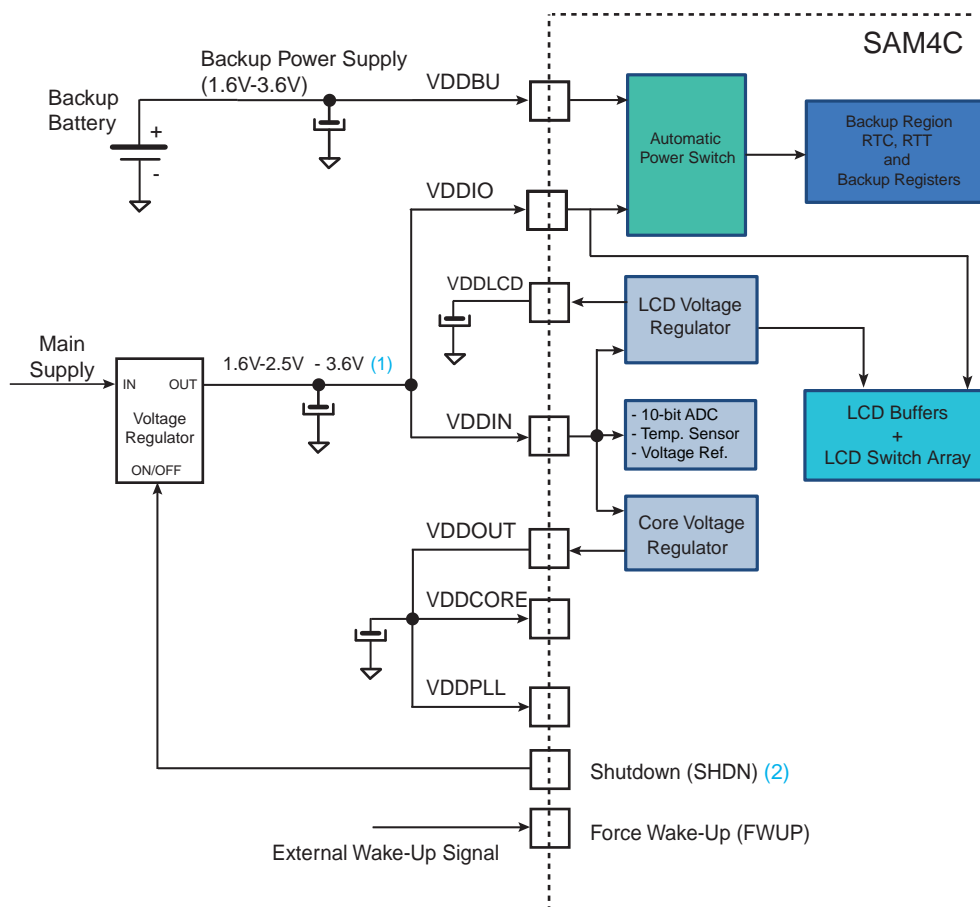


- Notes:
1. Internal LCD Voltage Regulator can be disabled to save its operating current. VDDLCD must then be provided externally.
  2. If ADC and LCD Controllers are used: 2.5V to 3.6V, otherwise 1.6V to 3.6V.

### 5.1.4.2 Single Supply with Backup Battery

Figure 5-3 shows a typical single power supply scheme for VDDIO, VDDIN, VDDCORE and VDDPLL. VDDBU is supplied with a separated backup battery. In this supply scheme, the internal LCD voltage regulator can be used. Note that if anti-tamper pins (TMP1 to TMP3) and the RTCOUT0 output have to be used in backup mode, VDDIO must be kept. The reference voltage of the TMP1 to TMP3 and RTCOUT0 pins is VDDIO.

**Figure 5-3. Single Supply with Backup Battery**



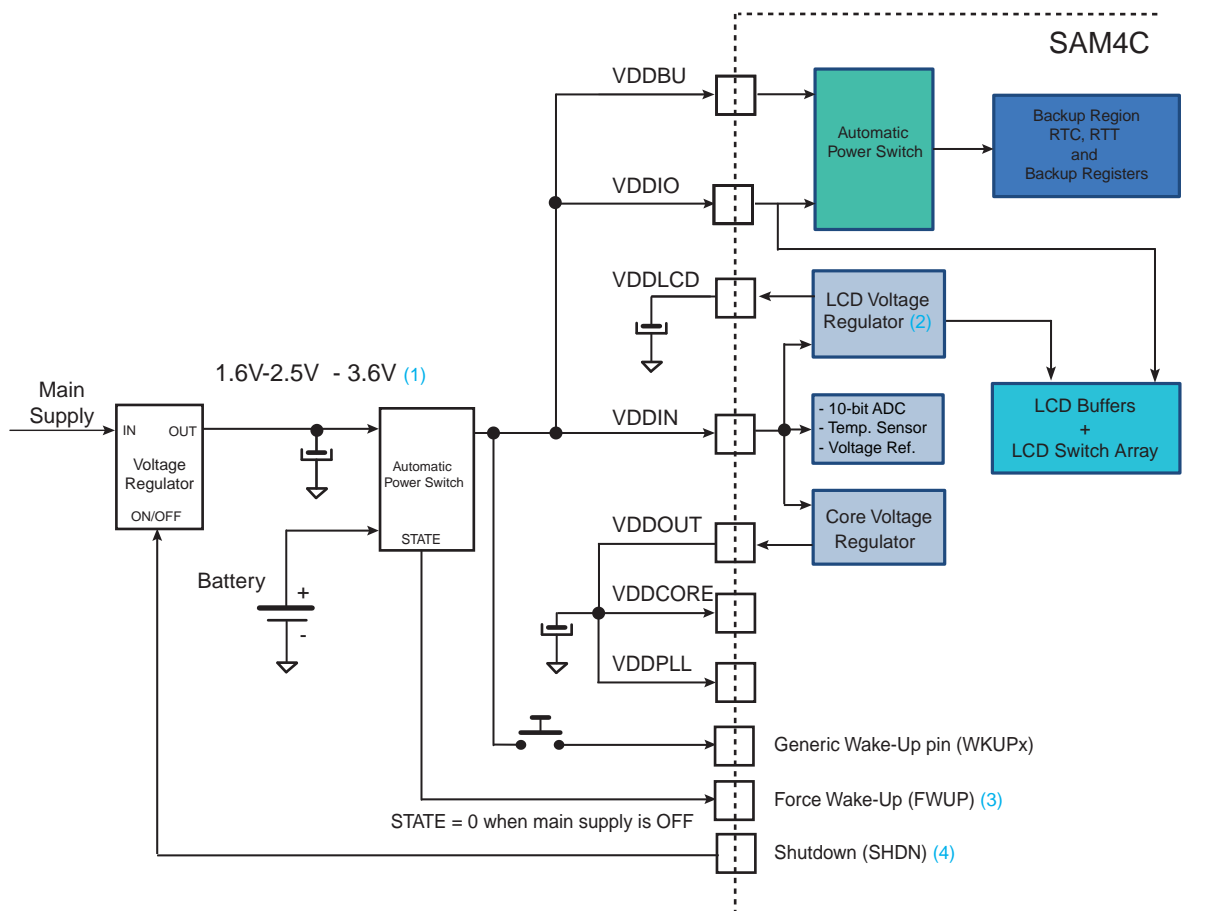
- Notes:
1. If ADC and LCD Controllers are used: 2.5V to 3.6V, otherwise 1.6V to 3.6V.
  2. Example with the SHDN pin used to control the main regulator enable pin. SHDN defaults to VDDBU at startup and when MCU wakes up from a wake-up event (external pin, RTC alarm, etc.). When MCU is in backup mode, SHDN defaults to 0.

### 5.1.4.3 Single Power Supply using One Main Battery and LCD Controller in Backup Mode

Figure 5-4 below shows a typical power supply scheme when the system needs to continue working and/or has to maintain display in backup mode when the main voltage is not present.

In this power supply scheme, the SAM4C can wake up both from an internal wake-up source, such as RTT, RTC and Supply Monitor, and from an external source, such as generic wake-up pins (WKUPx), anti-tamper inputs (TMPx) or force wake-up (FWUP).

**Figure 5-4. Single Power Supply using Battery and LCD Controller in Backup Mode**



- Notes:
1. If ADC and LCD Controllers are used: 2.5V to 3.6V, otherwise 1.6V to 3.6V.
  2. Internal LCD Voltage Regulator can be disabled to save its operating current. VDDLCD must then be provided externally.
  3. The STATE output of the power switch indicates to the MCU that the main supply is back and forces the system to wake up.
  4. Example with the SHDN pin used to control the main regulator enable pin. SHDN defaults to VDDBU at startup and when MCU wakes up from a wake-up event (external pin, RTC alarm, etc.). When the MCU is in backup mode, SHDN defaults to 0.



#### 5.1.4.4 Wake-up, Anti-tamper and RTCOUT0 Pins

In all power supply figures shown above, if generic wake-up pins other than WKUP0/TMP0 are used either as a wake-up or a fast startup input, or as anti-tamper inputs, VDDIO must be present. This also applies to the RTCOUT0 pin.

#### 5.1.4.5 General Purpose IOs (GPIO) State in Low-power Modes

In dual power supply schemes shown in [Figure 5-3](#) and [Figure 5-4](#), where backup or wait mode has to be used, configuration of the GPIO lines is kept in the same state as before entering backup or wait mode. Thus, to avoid extra current consumption on the VDDIO power rail, the user must configure the GPIOs either as an input with pull-up or pull-down enabled, or as output low or high levels corresponding to the external on-board devices.

#### 5.1.4.6 Default General Purpose IOs (GPIO) State after Reset

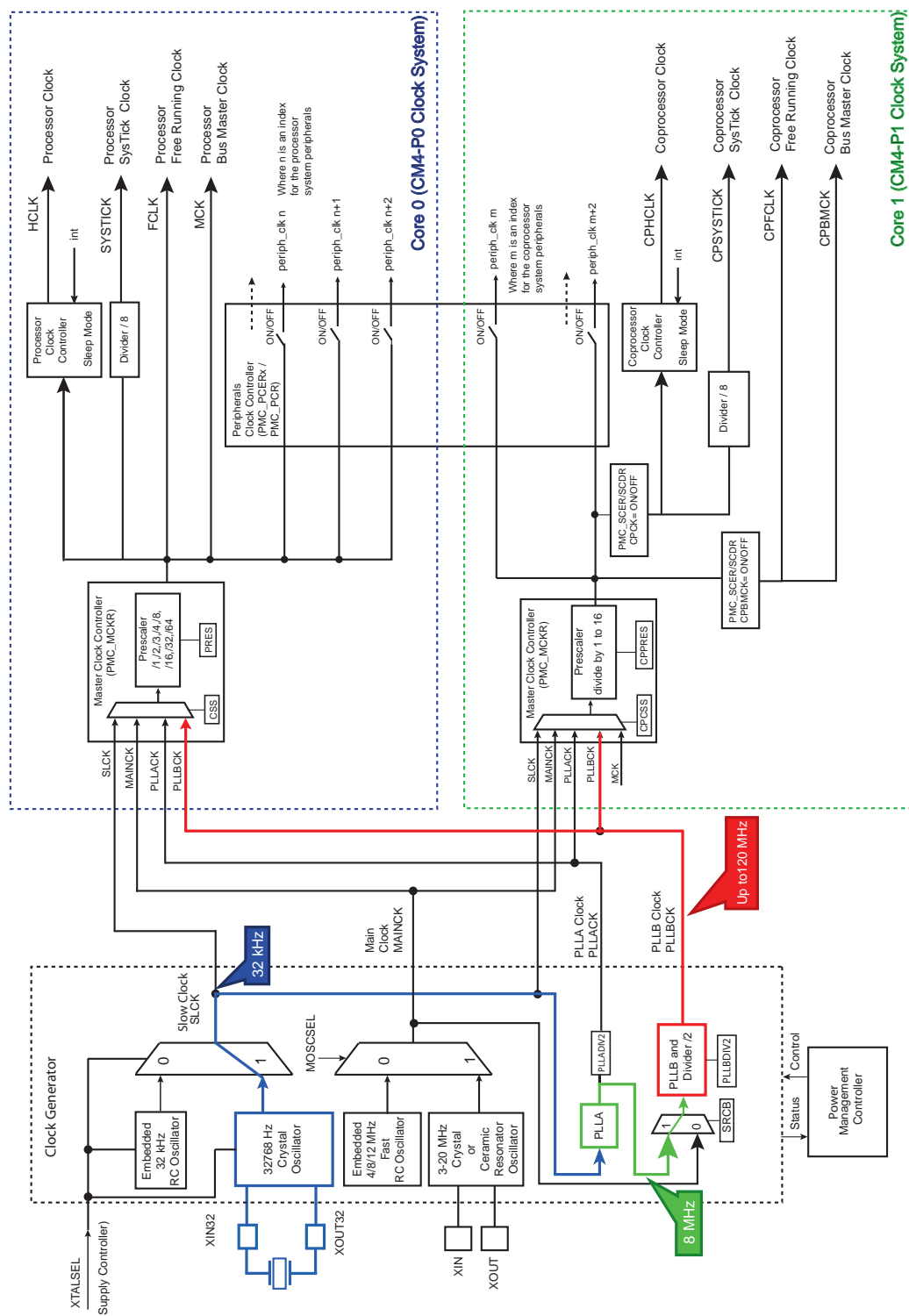
The reset state of the GPIO lines after reset is given in [Table 11-4](#), “[Multiplexing on PIO Controller A \(PIOA\)](#)”, [Section 11-5](#) “[Multiplexing on PIO Controller B \(PIOB\)](#)” and [Table 11-6](#), “[Multiplexing on PIO Controller C \(PIOC\)](#)”. For further details about the General Purpose IO and System lines, wake-up sources and wake-up time, and typical power consumption in different low-power modes, refer to [Table 5-2](#), “[Low-power Mode Configuration Summary](#)”.

## 5.2 Clock System

As shown in Figure 5-5 below, the SAM4C clock system allows single crystal operation:

- the 32 kHz oscillator can be the source clock of the 8 MHz Digital PLL (PLLA)
- the 8 MHz clock can feed the high frequency PLL (PLLB) input
- the output of the PLLB can be used as a main clock for both cores and peripherals

Figure 5-5. SAM4C16/8 Global Clock System



## 5.3 System State at Power-up

### 5.3.1 Device Configuration after the First Power-up

After the first power-up, the SAM4C is booting from the ROM. The device configuration is defined by SAM-BA<sup>®</sup> boot program.

### 5.3.2 Device Configuration after a Power Cycle when Booting from Flash Memory

After a power cycle of all the power supply rails, the system peripherals, such as the Flash Controller, the Clock Generator, the Power Management Controller and the Supply Controller, are in the following state:

- Slow Clock (SLCK) source is the internal 32 kHz RC Oscillator
- Main Clock (MAINCK) source is set to the 4 MHz internal RC Oscillator
- Crystal oscillators and PLLs are disabled
- Core Brownout Detector and Core Reset are enabled
- Backup Power-on-reset is enabled
- VDDIO Supply Monitor is disabled
- Flash Wait State (FWS) bit in the EEFC Flash Mode Register is set to 0
- Core 0 Cache Controller (CMCC0) is enabled (only used if the application link address for the Core 0 is 0x11000000)
- Sub-system 1 is in the reset state and not clocked

### 5.3.3 Device Configuration after a Reset

The system state after a reset or a wake-up from backup mode is the same as after a power cycle, except that the configuration of the peripherals in the backup area remains the same as before a reset:

- Slow Clock (SLCK) source: as after a power cycle (32 kHz RC or Crystal oscillator)
- Main Clock (MAINCK) source is set to the 4 MHz internal RC Oscillator
- Crystal oscillators and PLLs are disabled
- Core Brownout Detector: as after a power cycle
- Backup Power-on-reset: as after a power cycle
- VDDIO Supply Monitor: as after a power cycle

## 5.4 Active Mode

Active mode is the normal running mode with single or dual core executing code. System clock can be the fast RC Oscillator, the Main Crystal Oscillator or the PLLs. The Power Management Controller can be used to adapt the frequency and to disable the peripheral clocks when unused.

## 5.5 Low-power Modes

The various low-power modes (backup, wait and sleep modes) of the SAM4C are described below. Note that the Segmented LCD Controller can be used in all low-power modes.

**Note:** The Wait For Event instruction (WFE) of the Cortex-M4 core can be used to enter any of the low-power modes, however this may add complexity in the design of application state machines. This is due to the fact that the WFE instruction goes along with an event flag of the Cortex core (cannot be managed by the software application). The event flag can be set by interrupts, a debug event or an event signal from another processor. Since an interrupt can occur just before the execution of WFE, WFE takes into account events that happened in the past. As a result, WFE prevents the device from entering low-power mode if an interrupt event has occurred. Atmel has made provision to avoid using the WFE instruction. The workarounds to ease application design are given in the following description of the low-power mode sequences. Using the WFE instruction is given as well.

### 5.5.1 Backup Mode

The purpose of backup mode is to achieve the lowest possible power consumption in a system that executes periodic wake-ups to perform tasks but which does not require fast start-up time. The total current consumption is 0.5  $\mu$ A typical on VDDBU.

The Supply Controller, power-on reset, RTT, RTC, backup registers and the 32 kHz oscillator (RC or crystal oscillator selected by software in the Supply Controller) are running. The regulator and the core supplies are off. The power-on-reset on VDDBU can be deactivated by software.

The SAM4C can be awakened from backup mode through the Force Wake-up (FWUP) pin, WKUP0, WKUP1 to WKUP15 pins, the VDDIO Supply Monitor (SM) if VDDIO is supplied, or through an RTT or RTC wake-up event. Wake-up pins multiplexed with anti-tampering functions are possible sources of a wake-up as well in case if an anti-tampering event is detected. The TMP0 pad is supplied by the backup power supply (VDDBU). Other anti-tamper input pads are supplied by VDDIO.

The LCD Controller can be used in this mode. The purpose is to maintain the displayed message on the LCD display after entering backup mode. The current consumption on VDDIN to maintain the LCD is 10  $\mu$ A typical.

In case if the VDDIO power supply is kept on with VDDBU when entering backup mode, it is up to the application to configure all PIO lines in a stable and known state to avoid extra power consumption or possible current path with the input/output lines of the external on-board devices.

#### 5.5.1.1 Backup Mode Sequence

- Application dependant: set the PIO lines in the correct mode and configuration (input pull-up or pull-down, output low or high levels)
- Disable the Main Crystal Oscillator (enabled by SAM-BA boot if device is booting from ROM)
- Configure PA30/PA31 (XIN/XOUT) into PIO mode according to their use
- Disable JTAG lines via the SFR1 register in Matrix 0 (by default, internal pull-up or pull-down is disabled on JTAG lines)
- Enable RTT in 1 Hz mode
- Disable Normal Mode of RTT (RTT will run in 1 Hz mode)
- Disable POR backup

To enter backup mode using the VROFF bit:

- Write a 1 to the VROFF bit of SUPC\_CR

To enter backup mode using the WFE instruction:

- Write a 1 to the SLEEPDEEP bit of the Cortex-M4 processor
- Execute the WFE instruction of the processor

After this step, the Core voltage regulator is shut down and the SHDN pin goes low. All the digital internal logics (cores, peripherals and memories) are not powered. The LCD controller can be enabled if needed before entering backup mode.

In both cases, either the VROFF bit or the WFE instruction was used to enter backup mode, the system exits backup mode if one of the following enabled wake-up events occurs:

- WKUP[0-15] pins
- Force Wake-up pin
- VDDIO Supply Monitor (if VDDIO is present)
- Anti-tamper event detection
- RTC alarm
- RTT alarm

After exiting backup mode, the device is in the reset state. Only the configuration of the backup area peripherals remains unchanged.

Note that the device does not automatically enter backup mode if VDDIN is disconnected or fell down under minimum voltage. The Shutdown pin (SHDN) remains high in this case.

For current consumption in backup mode, refer to the electrical characteristics of this datasheet.

## 5.5.2 Wait Mode

The purpose of wait mode is to achieve very low power consumption while maintaining the whole device in a powered state for a start-up time of less than 10  $\mu$ s. For current consumption in wait mode, refer to the product electrical characteristics.

In this mode, the bus and peripheral clocks of Sub-system 0 and Sub-system 1 (MCK/CPBMCK), the clocks of Core 0 and Core 1 (HCLK/CPHCLK) are stopped when the Entering Wait Mode sequence is performed (see [Section 5.5.2.1](#)). However, the power supply of core, peripherals and memories are maintained using standby mode of the core voltage regulator.

The SAM4C is able to handle external and internal events in order to perform a wake-up. This is done by configuring the external WKUPx lines as fast startup wake-up pins (refer to [Section 5.7 “Fast Start-up”](#)). RTC alarm, RTT alarm and anti-tamper events can wake the device up as well.

Wait mode can be used together with Flash in Read-Idle mode, Standby mode or Deep Power mode to further reduce the current consumption. Flash in Read-Idle mode provides a faster start-up and Standby mode offers a lower power consumption. For further details, see the “Low-power Wake-up Time” section of the product electrical characteristics.

### 5.5.2.1 Wait Mode Sequence

- Stop Sub-system 1
- Select the 4/8/12 MHz fast RC Oscillator as Main Clock<sup>(1)</sup>
- Application dependant: set the PIO lines in the correct mode and configuration (input pull-up or pull-down, output low or high level)
- Disable the Main Crystal Oscillator (enabled by SAM-BA boot if device is booting from ROM)
- Configure PA30/PA31 (XIN/XOUT) into PIO mode according to their use
- Disable JTAG lines via the SFR1 register in Matrix 0 (by default, internal pull-up or pull-down is disabled on JTAG lines)
- Set the FLPM field in the PMC Fast Startup Mode Register (PMC\_FSMR)<sup>(2)</sup>
- Set the Flash Wait State (FWS) bit in the EEFC Flash Mode Register to 0

To enter wait mode using the WAITMODE bit:

- Set the WAITMODE bit to 1 in the PMC Main Oscillator Register (CKGR\_MOR)
- Wait for Master Clock Ready MCKRDY = 1 in the PMC Status Register (PMC\_SR)

To enter wait mode using the WFE instruction:

- Select the 4/8/12 MHz fast RC Oscillator as Main Clock
- Set the FLPM field in the PMC Fast Startup Mode Register (PMC\_FSMR)
- Set Flash Wait State at 0
- Set the LPM bit in the PMC Fast Startup Mode Register (PMC\_FSMR)
- Write a 0 to the SLEEPDEEP bit of the Cortex-M4 processor
- Execute the Wait-For-Event (WFE) instruction of the processor

Notes:

1. Any frequency can be chosen. The 12 MHz frequency will provide a faster start-up compared to the 4 MHz, but with the increased current consumption (in the  $\mu$ A range). See electrical characteristics of the product.
2. Depending on the Flash Low-power Mode (FLPM) value, the flash enters three different modes:
  - If FLPM = 0, the flash enters Stand-by mode (Low consumption)
  - If FLPM = 1, the flash enters Deep Power-down mode (Extra low consumption)
  - If FLPM = 2, the flash enters Idle mode. Memory is ready for Read access

In both cases, either the WAITMODE bit or the WFE instruction was used to enter wait mode, the system exits wait mode if one of the following enabled wake-up events occurs:

- WKUP[0-15] pins in Fast wake-up mode
- Anti-tamper event detection
- RTC alarm
- RTT alarm

After exiting wait mode, the PIO controller has the same configuration state as before entering wait mode. The SAM4C is clocked back to the RC oscillator frequency which was used before entering wait mode. The core will start fetching from

flash at this frequency. Depending on configuration of the Flash Low-power Mode (FLPM) bits used to enter wait mode, the application has to reconfigure it back to read-idle mode.

### 5.5.3 Sleep Mode

The purpose of sleep mode is to optimize power consumption of the device versus response time. In this mode, only the core clocks of CM4P0 and/or CM4P1 are stopped. Some of the peripheral clocks can be enabled depending on the application needs. The current consumption in this mode is application dependent. This mode is entered via Wait for Interrupt (WFI) or Wait for Event (WFE) instructions of the Cortex-M4.

The processor can be awakened from an interrupt if the WFI instruction of the Cortex-M4 is used to enter sleep mode, or from a wake-up event if the WFE instruction is used. The WFI instruction can also be used to enter sleep mode with the SLEEPONEXIT bit set to 1 in the System Control Register (SCB\_SCR) of the Cortex-M. If the SLEEPONEXIT bit of the SCB\_SCR is set to 1, when the processor completes the execution of an exception handler it returns to thread mode and enters immediately sleep mode. This mechanism can be used in applications that require the processor to run only when an exception occurs. Setting the SLEEPONEXIT bit to 1 enables an interrupt-driven application in order to avoid returning to an empty main application.

## 5.5.4 Low-power Mode Summary Table

The modes detailed above are the main low-power modes. Table 5-2 below provides a configuration summary of the low-power modes.

**Table 5-2. Low-power Mode Configuration Summary**

Mode	SUPC, 32 kHz Oscillator, RTC, RTT Backup Registers, POR (Backup Region)	Core Regulator / LCD Regulator	Core 0/1 Memory Peripherals	Mode Entry <sup>(12)</sup>	Potential Wake-up Sources	Core at Wake-up	PIO State in Low-power Mode	PIO State at Wake-up	Consumption <sup>(2)</sup>	Wake-up Time <sup>(1)</sup>
Backup Mode	ON	OFF/OFF	OFF / OFF (Not powered)	VROFF bit = 1 or SLEEPDEEP = 1 + WFE	- FWUP pin. - WKUP0-15 pins <sup>(10)</sup> . - Supply Monitor - Anti-tamper inputs <sup>(10)</sup> - RTC or RTT alarm	Reset	Previous state saved	Reset state <sup>(13)</sup>	<1 µA typ <sup>(3)</sup>	< 1,5 ms
Backup Mode with LCD	ON	OFF/ON	OFF / OFF (Not powered)	VROFF bit = 1 or SLEEPDEEP = 1 + WFE	- FWUP pin. - WKUP0-15 pins <sup>(10)</sup> . - Supply Monitor - Anti-Tamper inputs <sup>(10)</sup> - RTC or RTT alarm	Reset	Previous state saved	Unchanged (LCD Pins)/ Inputs with pull ups	<10 µA typ <sup>(9)</sup>	< 1,5 ms
Wait Mode Flash in Standby Mode <sup>(11)</sup>	ON	ON/ <sup>(8)</sup>	Core 0 and 1, memories and peripherals: Powered, but Not clocked	WAITMODE = 1 + FLPM = 0 or SLEEPDEEP = 0 + LPM = 1 + FLPM = 0 + WFE	Any Event from: - Fast start-up through WKUP0-15 pins. - Anti-Tamper inputs <sup>(10)</sup> - RTC or RTT alarm	Clocked back	Previous state saved	Unchanged	45 µA/ 66 µA <sup>(4)</sup>	< 10 µs
Wait Mode Flash in Deep Power-down Mode <sup>(11)</sup>	ON	ON/ <sup>(8)</sup>	Core 0 and 1, memories and peripherals: Powered, but Not clocked	WAITMODE = 1 + FLPM = 1 or SLEEPDEEP = 0 + LPM = 1 + FLPM = 1 + WFE	Any Event from: - Fast start-up through WKUP0-15 pins. - Anti-Tamper inputs <sup>(10)</sup> - RTC or RTT alarm	Clocked back	Previous state saved	Unchanged	45 µA/ 62 <sup>(5)</sup>	< 75 µs
Sleep Mode	ON	ON/ <sup>(8)</sup>	Core 0 and/or Core 1: Powered (Not clocked) <sup>(7)</sup>	SLEEPDEEP = 0 + LPM = 0 + WFE or WFI	Entry mode = WFI Any Enabled Interrupts;  Entry mode = WFE Any Enabled Events: - Fast start-up through WKUP0-15 pins. - Anti-Tamper inputs <sup>(10)</sup> - RTC or RTT alarm	Clocked back	Previous state saved	Unchanged	<sup>(6)</sup>	<sup>(6)</sup>

- Notes:
1. When considering wake-up time, the time required to start the PLL is not taken into account. Once started, the device works either from the 4, 8 or 12 MHz fast RC oscillator. The user has to add the PLL start-up time if it is needed in the system. The wake-up time is defined as the time taken for wake-up until the first instruction is fetched.
  2. Current consumption of the Supply Monitor on VDDIO is not included.
  3. On VDDBU (in backup mode, if VDDIO is powered, consumption is on VDDIO due to the automatic power switch). When VDDIO is kept in backup mode to use generic wake-up, anti-tamper or RTCOUT0, total power consumption on VDDIO is less than 3 µA. See electrical characteristics of the product.
  4. 45 µA on VDDCORE, 66 µA for total current consumption (using internal voltage regulator), TBD µA for total current consumption (without using internal voltage regulator).
  5. 45 µA on VDDCORE, 62 µA for total current consumption (using internal voltage regulator), TBD µA for total current consumption (without using internal voltage regulator).
  6. Depends on MCK frequency.
  7. In this mode, the core is supplied and not clocked but some peripherals can be clocked.

8. LCD voltage regulator can be OFF if VDDLCD is supplied externally thus saving current consumption of the LCD voltage regulator.
9. On VDDIN, VDDIO, VDDLCD
10. Refer to [Table 3-1, “Signal Description List”](#). Some anti-tamper pin pads are VDDIO powered.
11. Fast RC Osc. set to 4 MHz Frequency.
12. Refer to the note in [Section 5.5 “Low-power Modes”](#).
13. See PIO Controller Multiplexing tables in [Section 11.4 “Peripheral Signal Multiplexing on I/O Lines”](#).

## 5.6 Wake-up Sources

Wake-up events allow the device to exit backup mode. When a wake-up event is detected, the Supply Controller performs a sequence which automatically reenables the device.

## 5.7 Fast Start-up

The SAM4C allows the processor to restart in a few microseconds while the processor is in wait mode or in sleep mode. A fast start-up occurs upon detection of one of the wake-up inputs.

The fast restart circuitry is fully asynchronous and provides a fast start-up signal to the Power Management Controller. As soon as the fast start-up signal is asserted, the PMC automatically restarts the embedded 4/8/12 MHz Fast RC oscillator, switches the master clock on this 4 MHz clock and re-enables the processor clock.



## 6. Input/Output Lines

The SAM4C has two types of input/output (I/O) lines: general purpose I/Os (GPIO) and system I/Os. GPIOs have alternate functionality due to multiplexing capabilities of the PIO controllers. The same PIO line can be used whether in I/O mode or by the multiplexed peripheral. System I/Os include pins such as test pins, oscillators, erase or analog inputs.

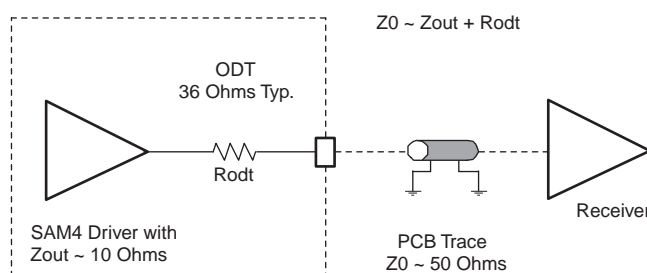
### 6.1 General Purpose I/O Lines

GPIO lines are managed by PIO Controllers. All I/Os have several input or output modes such as pull-up or pull-down, input Schmitt triggers, multi-drive (open-drain), glitch filters, debouncing or input change interrupt. Programming of these modes is performed independently for each I/O line through the PIO controller user interface. For more details, refer to the “Parallel Input/Output (PIO) Controller” section of this datasheet.

The input/output buffers of the PIO lines are supplied through VDDIO power supply rail when used as general purpose I/Os (GPIOs). When used as extra functions like LCD or analog modes, GPIO lines have either VDDLCD or VDDIN voltage range.

Each I/O line embeds an ODT (On-die Termination) shown in [Figure 6-1](#) below. ODT consists of an internal series resistor termination scheme for impedance matching between the driver output (SAM4C) and the PCB trace impedance preventing signal reflection. The series resistor helps to reduce I/Os switching current (di/dt) thereby reducing EMI. It also decreases overshoot and undershoot (ringing) due to inductance of interconnect between devices or between boards. Finally, ODT helps diminish signal integrity issues.

Figure 6-1. On-die Termination



### 6.2 System I/O Lines

System I/O lines are pins used by oscillators, test mode, reset and JTAG, to name but a few. Described below in [Table 6-1](#) are the SAM4C system I/O lines shared with PIO lines. These pins are software configurable as general purpose I/O or system pins. At start-up, the default function of these pins is always used.

Table 6-1. System I/O Configuration Pin List

SYSTEM_IO Bit Number	Default Function after Reset	Other Function	Constraints for Normal Start	Configuration
0	TDI	PB0	-	In Matrix User Interface Registers (Refer to the System I/O Configuration Register in the “Bus Matrix” section of this datasheet)
1	TDO/TRACESWO	PB1	-	
2	TMS/SWDIO	PB2	-	
3	TCK/SWCLK	PB3	-	
4	ERASE	PC9	Low level at Start-up <sup>(1)</sup>	See footnote <sup>(2)</sup> below
-	PA31	XIN	-	
-	PA30	XOUT	-	

Notes: 1. If PC9 is used as PIO input in user applications, a low level must be ensured at start-up to prevent Flash erase before the user application sets PC9 into PIO mode.

2. In the product datasheet, refer to: “3 to 20 MHz Crystal Oscillator” subsection in the “Power Management Controller PMC” section.

### 6.2.1 Serial Wire JTAG Debug Port (SWJ-DP) and Serial Wire Debug Port (SW-DP) Pins

The SWJ-DP pins are TCK/SWCLK, TMS/SWDIO, TDO/SWO, TDI and commonly provided on a standard 20-pin JTAG connector defined by ARM. For more details about voltage reference and reset state, refer to [Table 11-5, “Multiplexing on PIO Controller B \(PIOB\)”](#).

At start-up, SWJ-DP pins are configured in SWJ-DP mode to allow connection with debugging probe. Refer to the “Debug and Test” section of this datasheet.

SWJ-DP pins can be used as standard I/Os to provide users with more general input/output pins when the debug port is not needed in the end application. Mode selection between SWJ-DP mode (System IO mode) and general IO mode is performed through the AHB Matrix Special Function Registers (MATRIX\_SFR). Configuration of the pad for pull-up, triggers, debouncing and glitch filters is possible regardless of the mode.

The JTAGSEL pin is used to select the JTAG boundary scan when asserted at a high level. It integrates a permanent pull-down resistor of about 15 k $\Omega$  to GND, so that it can be left unconnected for normal operations.

By default, the JTAG Debug Port is active. If the debugger host wants to switch to the Serial Wire Debug Port, it must provide a dedicated JTAG sequence on TMS/SWDIO and TCK/SWCLK which disables the JTAG-DP and enables the SW-DP. When the Serial Wire Debug Port is active, TDO/TRACESWO can be used for trace.

The asynchronous TRACE output (TRACESWO) is multiplexed with TDO. So the asynchronous trace can only be used with SW-DP, not JTAG-DP. For more information about SW-DP and JTAG-DP switching, refer to the “Debug and Test” section of this datasheet. The SW-DP/SWJ-DP pins are used for debug access to both cores.

## 6.3 Test Pin

The TST pin is used for JTAG Boundary Scan Manufacturing Test or Fast Flash programming mode of the SAM4C series. For details on entering fast programming mode, see the “Fast Flash Programming Interface (FFPI)” section of this datasheet. For more information on the manufacturing and test modes, refer to the “Debug and Test” section of this datasheet.

## 6.4 NRST Pin

The NRST pin is bidirectional. It is handled by the on-chip reset controller and can be driven low to provide a reset signal to the external components or asserted low externally to reset the microcontroller. It resets the core and the peripherals except the Backup region (RTC, RTT and Supply Controller). There is no constraint on the length of the reset pulse and the Reset controller can guarantee a minimum pulse length. The NRST pin integrates a permanent pull-up resistor to VDDIO of about 100 k $\Omega$ . By default, the NRST pin is configured as an input.

## 6.5 TMPx Pins: Anti-tamper Pins

Anti-tamper pins detect intrusion, for example, into a smart meter case. Upon detection through a tamper switch, automatic, asynchronous and immediate clear of registers in the backup area, and time stamping in the RTC will be performed. Anti-tamper pins can be used in all modes. Date and number of tampering events are stored automatically. Anti-tampering events can be programmed so that half of the General Purpose Backup Registers (GPBR) are erased automatically. TMP1 to TMP3 signals are shared with a PIO pin meaning that VDDIO must be supplied, whereas TMP0 is in the VDDBU domain.

## 6.6 RTCOUT0 Pin

The RTCOUT0 pin shared in the PIO (supplied by VDDIO) can be used to generate waveforms from the RTC in order to take advantage of the RTC inherent prescalers while the RTC is the only powered circuitry (low-power mode of operation, backup mode) or in any active mode. Entering backup or low-power operating modes does not affect the

waveform generation outputs (VDDIO needs still to be supplied). Anti-tampering pin detection can be synchronised with this signal.

**Note:** To use the RTCCOUT0 signal during application development via JTAG-ICE interface, the programmer must use Serial Wire Debug (SWD) mode. In this case, the TDO pin is not used as a JTAG signal by the ICE interface.

## 6.7 Shutdown (SHDN) Pin

The SHDN pin reflects the MCU backup mode of operation: when the MCU is in backup mode, SHDN = 0, otherwise SHDN = 1 (VDDDBU). This pin is designed to control the enable pin of the main external voltage regulator. When the MCU enters backup mode, the SHDN pin disables the external voltage regulator and, upon the wake-up event, it re-enables the voltage regulator. The SHDN pin is used to control an external main voltage regulator and/or power switch when entering backup mode.

The SHDN pin is asserted low when the VROFF bit in the Supply Controller Control Register (SUPC\_CR) is set to 1.

## 6.8 Force Wake-up (FWUP) Pin

The FWUP pin can be used as a wake-up source in all low-power modes as it is supplied by VDDDBU.

## 6.9 ERASE Pin

The ERASE pin is used to reinitialize the Flash content (and some of its NVM bits) to an erased state (all bits read as logic level 1). The ERASE pin integrates a pull-down resistor of about 100 kΩ into GND, so that it can be left unconnected for normal operations.

This pin is debounced by SLCK to improve the glitch tolerance. When the ERASE pin is tied high during less than 100 ms, it is not taken into account. The pin must be tied high during more than 220 ms to perform a Flash erase operation.

The ERASE pin is a system I/O pin and can be used as a standard I/O. At start-up, the ERASE pin is not configured as a PIO pin. If the ERASE pin is used as a standard I/O, the start-up level of this pin must be low to prevent unwanted erasing. Refer to [Section 11.3 “APB/AHB Bridge”](#). If the ERASE pin is used as a standard I/O output, asserting the pin to low does not erase the Flash.

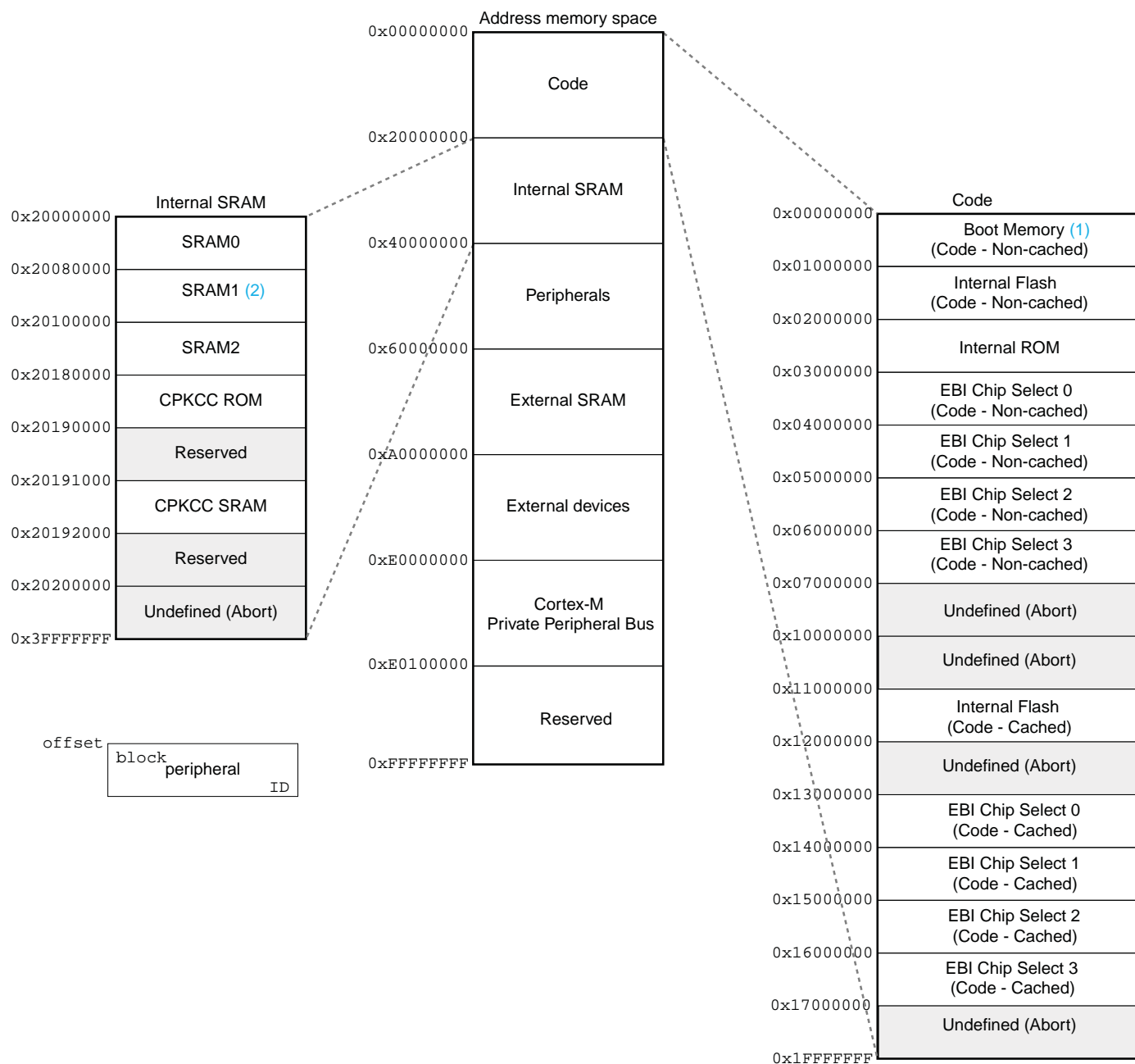
## 7. Product Mapping and Peripheral Access

Figure 7-1 shows the default memory mapping of the ARM Cortex-M Core.

**Figure 7-1. Cortex-M Memory Mapping**

0xFFFFFFFF	System level	Private peripherals including build-in interrupt controller (NVIC), MPU control registers, and debug components
0xE0000000 0xDFFFFFFF		
0xA0000000 0x9FFFFFFF	External device	Mainly used as external peripherals
0x60000000 0x5FFFFFFF 0x40000000	External RAM	Mainly used as external memory
0x3FFFFFFF 0x20000000	Peripherals	Mainly used as peripherals
0x1FFFFFFF 0x00000000	SRAM	Mainly used as static RAM
	CODE	Mainly used for program code. Also provides exception vector table after power up

**Figure 7-2. SAM4C16/8 Memory Mapping of CODE and SRAM Area**



- Notes:
1. Boot Memory for Core 0.
  2. Boot Memory for Core 1 @ 0x00000000.

In [Figure 7-2](#) above, 'Code' means 'Program Code over I-Code bus' and 'Program Data over D-Code bus'.

SRAM1 shown in the mapping above, can be seen at the address 0x20080000 (through S-bus) and the address 0x00000000 (through I/D Bus) for Core1. Instruction fetch from Core 1 to the SRAM address range is possible but leads to reduced performance due to the fact that instructions and read/write data go through the System Bus (S-Bus). Maximum performance for Core 1 (Metrology Core) is obtained by mapping the instruction code to the address 0x00000000 (SRAM1 through I/D-Code) and read/write data from the address 0x20100000 (SRAM2 through S-Bus).

For Core 0 (Application Core), maximum performance is achieved when the instruction code is mapped to the flash address and read/write data is mapped into SRAM0.

Each cores can access the following memories and peripherals:

- Core 0 (Application Core):
  - All internal memories
  - External memories or memory devices mapped on SMC 0 or SMC 1
  - All internal peripherals
- Core 1 (Metrology/Coprocessor Core):
  - All internal memories
  - External memories or memory devices mapped on SMC 0 or SMC 1
  - All internal peripherals

Note that Peripheral DMA 0 on Matrix 0 cannot access SRAM1 or SRAM2, Peripheral DMA 1 on Matrix 1 cannot access SRAM0, SRAM2 or SRAM0 can be the Data RAM for Inter-core Communication.

If the Metrology Core is not to be used (Clock Stopped and Reset active), all the peripherals, SRAM1 and SRAM2 of the Sub-system 1 can be used by the Application Core (Core 0) as long as the peripheral bus clock and reset are configured.

Detailed Memory Mapping and Memory Access versus Matrix Masters/Slaves is given in the "Bus Matrix (MATRIX)" section of this datasheet.

**Figure 7-3. SAM4C16/8 Memory Mapping of the Peripherals Area**

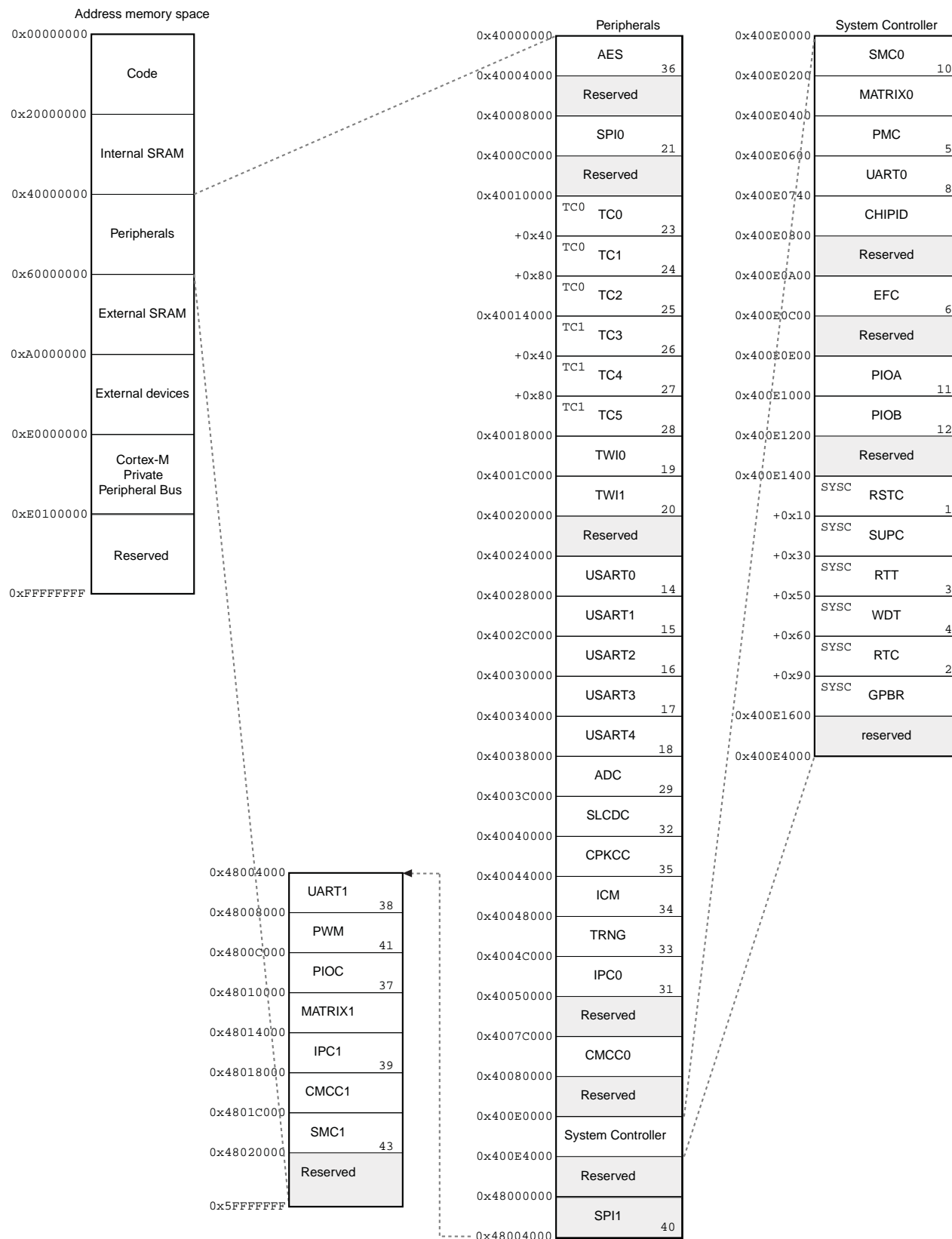
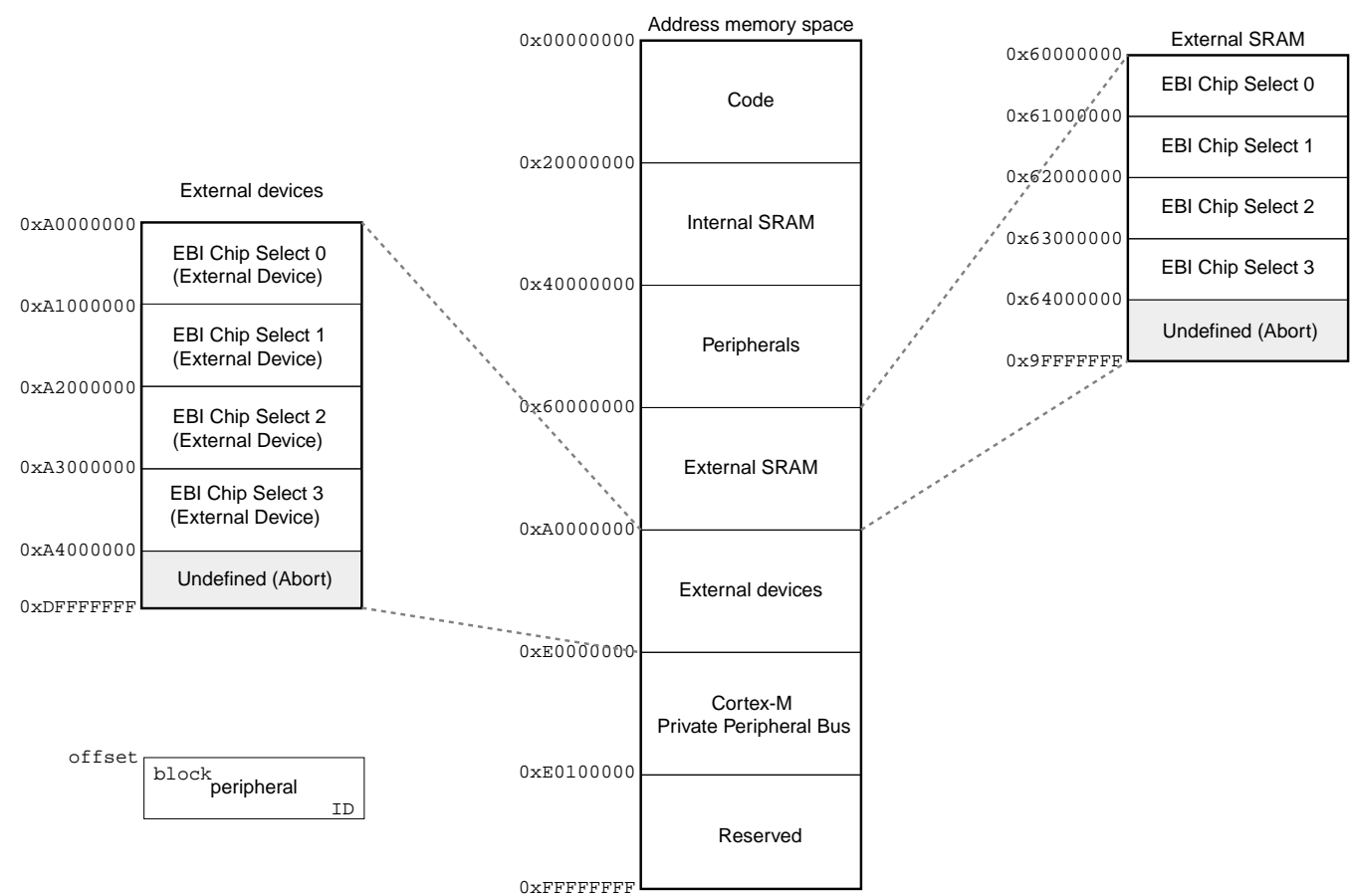


Figure 7-4. SAM4C16/8 Memory Mapping of External SRAM and External Devices Area





## 8. Memories

The memory map shown in [Figure 7-2, “SAM4C16/8 Memory Mapping of CODE and SRAM Area”](#) is global to both Cortex-M4 processors except the “Boot Memory” block. For more information on Boot Memory refer to [Section 8.1.5 “Boot Strategy” on page 37](#).

Each processor uses its own ARM private bus memory map for the NVIC and other system functions.

### 8.1 Embedded Memories

#### 8.1.1 Internal SRAM

The SAM4C embeds a total of 152 Kbytes high-speed SRAM with zero wait state access time.

SRAM0 on Matrix0 is 128 Kbytes. It is dedicated to the application processor (CM4P0) or other peripherals on Matrix0 but can be identified and used by masters on Matrix1. Refer to “Bus Matrix (MATRIX)” section of this datasheet for more details.

SRAM1 on Matrix1 is 16 Kbytes. It is mainly dedicated to be the code region of the CM4P1 processor but can be identified and used by on Matrix0. Refer to “Bus Matrix (MATRIX)” section of this datasheet for more details.

SRAM2 on Matrix1 is 8 Kbytes. It is mainly dedicated to be the data region of the CM4P1 processor or other peripherals on Matrix1 but can be identified and used by masters on Matrix0. Refer to “Bus Matrix (MATRIX)” section of this datasheet for more details.

If the CM4P1 processor is in the reset state and not used, the application core can use it.

The SRAM is located in the bit band region. The bit band alias region is from 0x2200\_0000 to 0x23FF\_FFFF.

#### 8.1.2 System ROM

The SAM4C embeds an Internal ROM for the master processor (CM4P0), which contains the SAM Boot Assistant (SAM-BA®), In Application Programming routines (IAP), and Fast Flash Programming Interface (FFPI).

The ROM is always mapped at the address 0x02000000.

#### 8.1.3 CPKCC ROM

The ROM contains a cryptographic library using the CPKCC cryptographic accelerator peripheral (CPKCC) to provide support for Rivest Shamir Adleman (RSA), Elliptic Curve Cryptography (ECC), Digital Signature Algorithm (DSA) and Elliptic Curve Digital Signature Algorithm (ECDSA).

#### 8.1.4 Embedded Flash

##### 8.1.4.1 Flash Overview

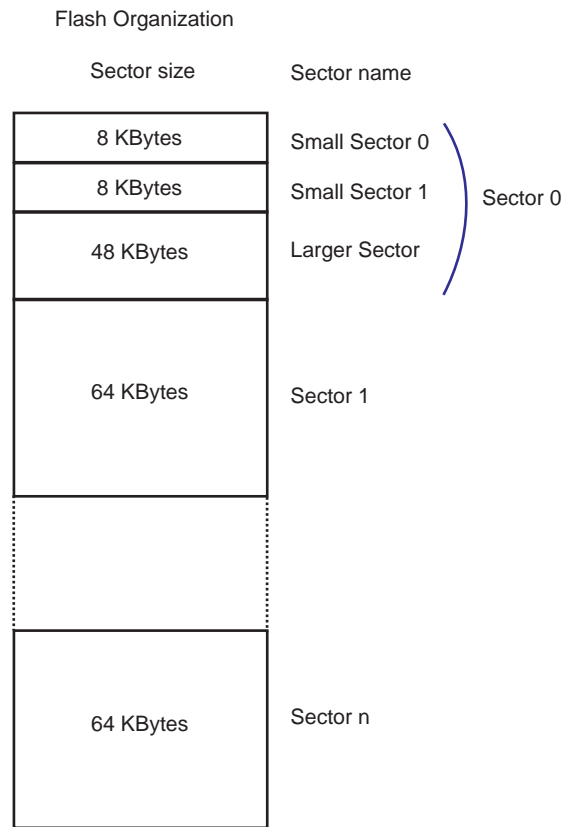
The embedded Flash is the boot memory for the Cortex-M4 Core 0 (CM4P0). The Flash memory can be accessed through the Cache Memory Controller (CMCC0) of the CM4P0 and can also be identified by the Cortex-M4F Core 1 (CM4P1) through its Cache Memory Controller (CMCC1).

The memory plane is organized in sectors. Each sector has a size of 64 Kbytes. The first sector of 64 Kbytes is divided into 3 smaller sectors.

The three smaller sectors are organized into 2 sectors of 8 Kbytes and 1 sector of 48 Kbytes. Refer to [Figure 8-1](#) below.

The Flash memory has a built-in error code correction providing 2-bit error detection and 1-bit correction per 128 bits.

Figure 8-1. Memory Plane Organization



Each sector is organized in pages of 512 Bytes.

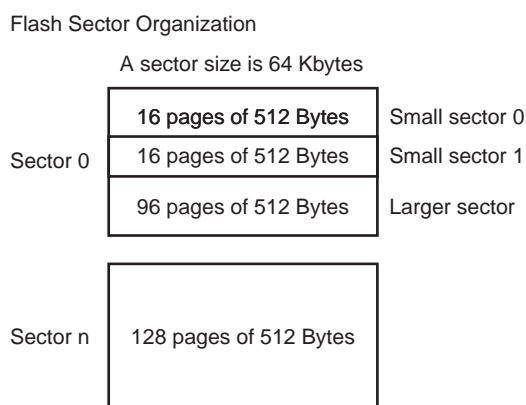
For sector 0:

- The small sector 0 has 16 pages of 512 Bytes, 8 Kbytes in total
- The small sector 1 has 16 pages of 512 Bytes, 8 Kbytes in total
- The larger sector has 96 pages of 512 Bytes, 48 Kbytes in total

From sector 1 to n:

The rest of the array is composed of 64-Kbyte sector where each sector comprises 128 pages of 512 bytes. Refer to [Figure 8-2, "Flash Sector Organization"](#) below.

**Figure 8-2. Flash Sector Organization**

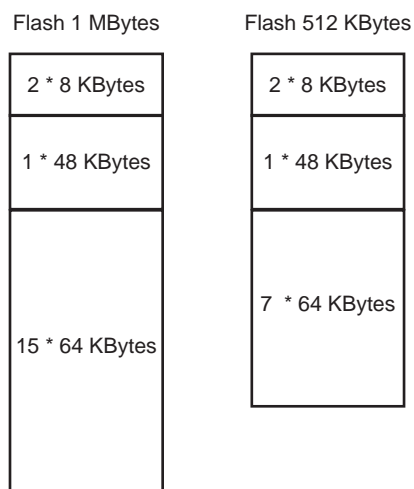


The Flash size varies by product:

- SAM4C8: Flash size is 512 Kbytes
- SAM4C16: Flash size is 1024 Kbytes

Refer to [Figure 8-3](#) below for the organization of the Flash depending on its size.

**Figure 8-3. Flash Size**



Erasing the memory can be performed as follows:

- Per 512-byte page inside a sector of 8 Kbytes.  
EWP and EWPL commands can be only used in 8 Kbytes sectors
- Per 4-Kbyte block inside a sector of 8 Kbytes/48 Kbytes/64 Kbytes.  
Erase Page commands can be only used with FARG[1:0] = 1

- Per sector of 8 Kbytes/48 Kbytes/64 Kbytes.  
Erase Page commands can be only used with FARG[1:0] = 2
- Per Full-Memory

#### 8.1.4.2 Enhanced Embedded Flash Controller

The Enhanced Embedded Flash Controller manages accesses performed by masters of the system. It enables reading the Flash and writing the write buffer. It also contains a User Interface, mapped on the APB.

The Enhanced Embedded Flash Controller ensures the interface of the Flash block. It manages the programming, erasing, locking and unlocking sequences of the Flash using the full set of commands.

One of the commands returns the embedded Flash descriptor definition that informs the system about the Flash organization, thus making the software generic.

#### 8.1.4.3 Flash Speed

The user needs to set the number of wait states depending on the frequency used on the SAM4C.

For more details, refer to the “AC Characteristics” section of the product electrical characteristics.

#### 8.1.4.4 Lock Regions

Several lock bits are used to protect write and erase operations on lock regions. A lock region is composed of several consecutive pages, and each lock region has its associated lock bit.

**Table 8-1. Lock bit number**

Product	Number of Lock Bits	Lock Region Size
SAM4C16	128	8 Kbytes
SAM4C8	64	8 Kbytes

The lock bits are software programmable through the EEFC User Interface. The command “Set Lock Bit” enables the protection. The command “Clear Lock Bit” unlocks the lock region.

Asserting the ERASE pin clears the lock bits, thus unlocking the entire Flash.

#### 8.1.4.5 Security Bit Feature

The SAM4C features a security bit based on a specific General Purpose NVM bit (GPNVM bit 0). When the security is enabled, any access to the Flash, SRAM, core registers and internal peripherals, either through the SW-DP/JTAG-DP interface or through the Fast Flash Programming Interface, is forbidden. This ensures the confidentiality of the code programmed in the Flash.

This security bit can only be enabled through the command “Set General Purpose NVM Bit 0” of the EEFC User Interface. Disabling the security bit can only be achieved by asserting the ERASE pin at 1, and after a full Flash erase is performed. When the security bit is deactivated, all accesses to the Flash, SRAM, Core registers, Internal Peripherals are permitted.

It is important to note that the assertion of the ERASE pin should always be longer than 220 ms.

As the ERASE pin integrates a permanent pull-down, it can be left unconnected during normal operation. However, it is safer to connect it directly to GND for the final application.

#### 8.1.4.6 Unique Identifier

Each device integrates its own 128-bit unique identifier. These bits are factory configured and cannot be changed by the user. The ERASE pin has no effect on the unique identifier.

#### 8.1.4.7 User Signature

The memory has one additional reprogrammable page that can be used as page signature by the user. It is accessible through specific modes, for erase, write and read operations. Erase pin assertion will not erase the User Signature page.

#### 8.1.4.8 Fast Flash Programming Interface

The Fast Flash Programming Interface allows programming the device through either a serial JTAG interface or through a multiplexed fully-handshaked parallel port. It allows gang programming with market-standard industrial programmers.

The FFPI supports read, page program, page erase, full erase, lock, unlock and protect commands.

#### 8.1.4.9 SAM-BA Boot

The SAM-BA Boot is a default Boot Program for the master processor (CM4P0) which provides an easy way to program in-situ the on-chip Flash memory.

The SAM-BA Boot Assistant supports serial communication via the UART0.

The SAM-BA Boot provides an interface with SAM-BA Graphic User Interface (GUI).

The SAM-BA Boot is in ROM and is mapped in Flash at address 0x0 when GPNVM bit 1 is set to 0.

#### 8.1.4.10 GPNVM Bits

The SAM4C features two GPNVM bits. These bits can be cleared or set respectively through the commands “Clear GPNVM Bit” and “Set GPNVM Bit” of the EEFC User Interface (see the “EEFC Flash Command Register” section of this datasheet).

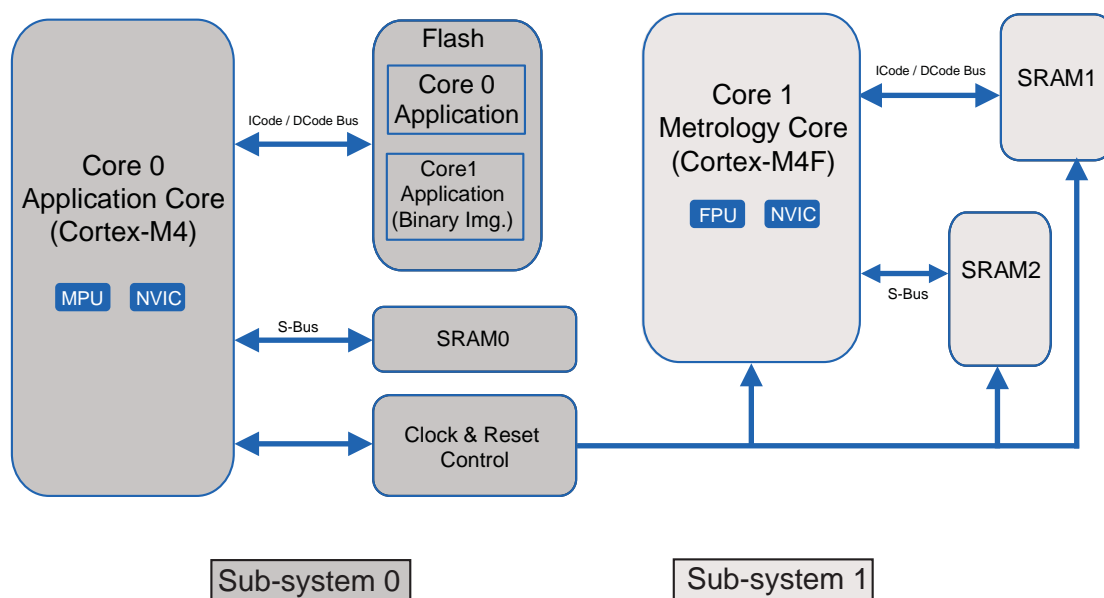
**Figure 8-4. General-purpose Nonvolatile Memory Bits**

GPNVMBit[#]	Function
0	Security bit
1	Boot mode selection

#### 8.1.5 Boot Strategy

Figure 8-5 below shows a load view of the memory at boot time.

**Figure 8-5. Simplified Load View at Boot Time**



Note: Matrices, AHB and APB Bridges are not represented.

### 8.1.5.1 Application Core (Core 0) Boot Process

The application processor (CM4P0) always boots at the address 0x0. To ensure maximum boot possibilities, the memory layout can be changed via GPNVM. A General Purpose NVM (GPNVM) bit is used to boot either on the ROM (default) or from the Flash. The GPNVM bit can be cleared or set through the commands “Clear General-purpose NVM Bit” and “Set General-purpose NVM Bit” of the EEFC User Interface respectively. Setting GPNVM Bit 1 selects the boot from the Flash whereas clearing this bit selects the boot from the ROM. Asserting ERASE clears the GPNVM Bit 1 and thus selects the boot from the ROM by default.

### 8.1.5.2 Metrology/Coprocessor Core (Core 1) Boot Process

After reset, the Sub-system 1 is hold in reset and with no clock. It is up to the Master Application (Core 0 Application) running on the Core 0 to enable the Sub-system 1. Then the application code can be downloaded into the CM4P1 Boot memory (SRAM1), and CM4P0 can afterwards de-assert the CM4P1 reset line. The secondary processor (CM4P1) always identifies SRAM1 as “Boot memory”.

### 8.1.5.3 Sub-system 1 Startup Sequence

After the Core 0 is booted from Flash, the Core 0 Application must perform the following steps:

1. Enable Core 1 System Clock (Bus and peripherals)
2. Enable Core 1 Clock
3. Release Core 1 System Reset (Bus and peripherals)
4. Enable SRAM1 and SRAM2 Clock
5. Copy Core 1 Application from Flash into SRAM1
6. Release Core 1 Reset

After Step 6, the Core 1 boots from SRAM1 Memory

Pseudo-code

```
1- // Enable Coprocessor Bus Master Clock in PMC System Clock Enable Register
   (CPBMCK bit)

2- // Enables Coprocessor Clocks
   • PMC System Clock Enable Register (CPCK bit)
     // Set Coprocessor Clock Prescaler and Source
   • In PMC MCKR: Coprocessor Programmable Clock Prescaler (CPPRES bit fields)
     // Choose coprocessor main clock source
   • In PMC MCKR: Coprocessor Master Clock Selection (CPCSS bit fields)

3- // Release coprocessor peripheral reset
   • In Reset Controller Coprocessor Mode Register (CPEREN bit)

4- // Enable Core 1 SRAM1 and SRAM2 Memories
   • In PMC PCER: Peripheral ID 42 (SRAM)

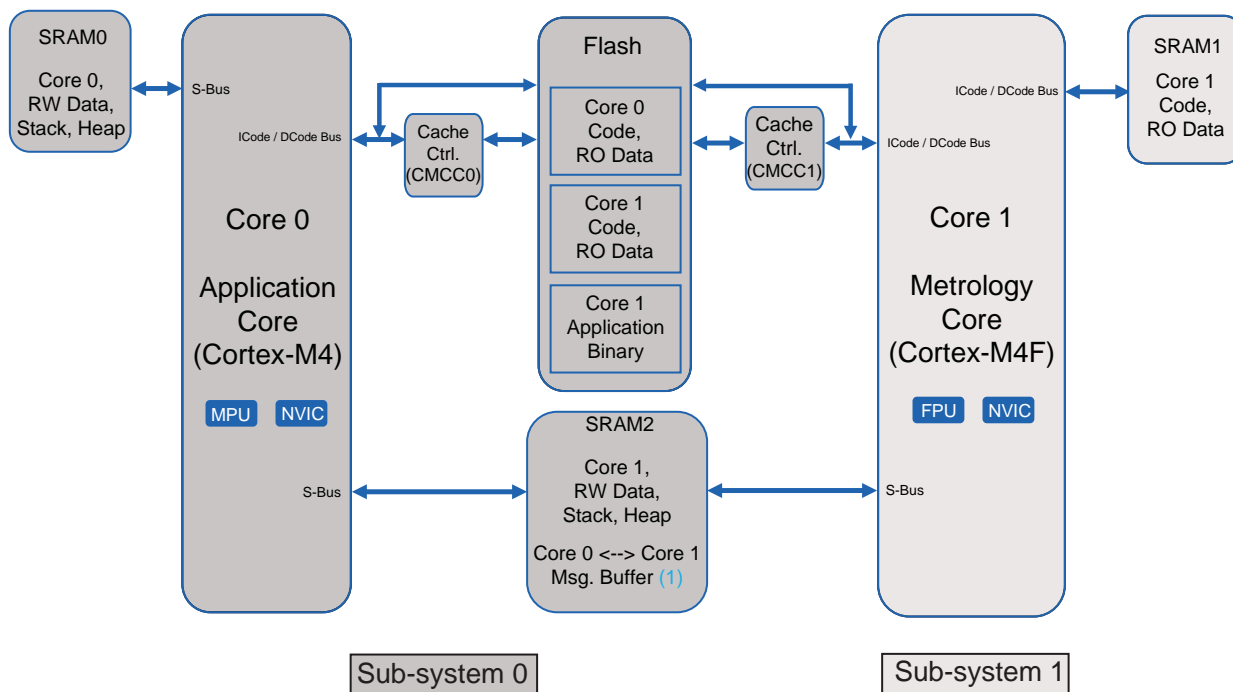
5- // AT THIS POINT Core 1 application code must be loaded from Flash into
   SRAM1.

6- // Release coprocessor reset
   • In Reset Controller Coprocessor Mode Register (CPROCEN bit)
```

#### 8.1.5.4 Typical Execution View

Figure 8-6 below provides the code execution view for both Cortex-M4 cores. AHB to APB, AHB to AHB and Matrices are not represented in this view.

**Figure 8-6. Execution View**



Note: 1. SRAM0 can also be used as Message Buffer Exchange.

Note: Matrices, AHB and APB Bridges are not represented.

## 8.2 External Memories

The SAM4C features the External Bus Interface to provide the interface to a wide range of external memories and to any parallel peripheral. Instruction fetch from external memories connected to the EBI/SMC is either possible through the Cache Controller or not. See [Figure 7-2, "SAM4C16/8 Memory Mapping of CODE and SRAM Area"](#)

### 8.2.1 Static Memory Controller

- Support for Code Fetch through the Cache controller
- 16-bit Data Bus
- Up to 24-bit Address Bus (up to 16 Mbytes linear per chip select)
- Up to 4 chip selects, Configurable Assignment
- Chip Select, Write enable or Read enable Control Mode
- Control signals programmable setup, pulse and hold time for each Memory Bank
- Multiple Wait State Management
  - Programmable Wait State Generation
  - External Wait Request
  - Programmable Data Float Time
- Slow Clock mode supported
- Additional Logic for NAND Flash

The Static Memory Controllers (SMC0/1) / External Bus Interface (EBI) can be used by either the CM4P0 or CM4P1 but only one path is optimized, CM4P0-->SMC0 or CM4P1-->SMC1.

The SMC0 and SMC1 uses the same pin on the EBI. Only one can be used at the same time.

The selection is done in the Matrix User Interface Registers (in the System I/O Configuration Register).

The SMC0 is used by default.



## 9. Real-time Event Management

The events generated by peripherals are designed to be directly routed to peripherals managing/using these events without processor intervention. Peripherals receiving events contain logic to select the required event.

### 9.1 Embedded Characteristics

- Timers generate event triggers which are directly routed to event managers, such as ADC, to start measurement/conversion without processor intervention
- UART, USART, SPI, TWI, and PIO generate event triggers directly connected to Peripheral DMA controller (PDC) for data transfer without processor intervention
- PMC Security Event (Clock Failure Detection) can be programmed to switch the MCK on reliable main RC internal clock

### 9.2 Real-time Event Mapping List

Table 9-1. Real-time Event Mapping List

Event Generator	Event Manager	Function
Anti-tamper Inputs (TMPx)	General Purpose Backup Register (GPBR)	Security / Immediate GPBR clear (asynchronous) on Anti-tamper detection through pins
Power Management Controller (PMC)	PMC	Safety / Automatic Switch to Reliable Main RC oscillator in case of Main Crystal Clock Failure
IO (ADTRG)	ADC	Trigger for measurement. Selection in ADC module
TC Output 0	ADC	Trigger for measurement. Selection in ADC module
TC Output 1	ADC	Trigger for measurement. Selection in ADC module
TC Output 2	ADC	Trigger for measurement. Selection in ADC module
TC Output 3	ADC	Trigger for measurement. Selection in ADC module
TC Output 4	ADC	Trigger for measurement. Selection in ADC module
TC Output 5	ADC	Trigger for measurement. Selection in ADC module

## 10. System Controller

The System Controller comprises a set of peripherals. It handles key elements of the system, such as power, resets, clocks, time, interrupts, watchdog, reinforced safety watchdog, etc.

### 10.1 System Controller and Peripheral Mapping

Refer to [Section 7-2 “SAM4C16/8 Memory Mapping of CODE and SRAM Area” on page 29](#).

All the peripherals are in the bit band region and are mapped in the bit band alias region.

### 10.2 Power Supply Monitoring

The SAM4C embeds Supply Monitor, Power-on-Reset and Brownout detectors for power supplies monitoring allowing to warn and/or reset the chip.

#### 10.2.1 Power-on-Reset on VDDCORE

The Power-on-Reset monitors VDDCORE. It is always activated and monitors voltage at start-up but also during power-down. If VDDCORE goes below the threshold voltage, the entire chip (except VDDDBU domain) is reset. For more information, refer to the “Electrical Characteristics” section of the product datasheet.

#### 10.2.2 Brownout Detector on VDDCORE

The Brownout Detector monitors VDDCORE. It is active by default. It can be deactivated by software through the Supply Controller (SUPC\_MR).

If VDDCORE goes below the threshold voltage, the reset of the core is asserted.

#### 10.2.3 Power-on-Reset on VDDIO

The Power-on-Reset monitors VDDIO. It is always activated and monitors voltage at start-up but also during power-down. If VDDIO goes below the threshold voltage, the IOs are reset but the core continues to run. Voltage detection is fixed.

#### 10.2.4 Supply Monitor on VDDIO

The supply monitor on VDDIO is fully programmable with 16 steps for the threshold (between 1.6V to 3.4V). It provides the user the flexibility to set a voltage level detection higher than the power-on-reset on VDDIO. Either a reset or an interrupt can be generated upon detection. It can be activated by software and it is controlled by the Supply Controller (SUPC). A sample mode is possible. It allows to divide the supply monitor power consumption by a factor of up to 2048.

#### 10.2.5 Power-on-Reset and Brownout Detector on VDDDBU

The Power-on-Reset monitors VDDDBU. It is active by default and monitors voltage at start-up but also during power-down. It can be deactivated by software through the Supply Controller (SUPC\_MR). If VDDDBU goes below the threshold voltage, the entire chip is reset.

## 10.3 Reset Controller

The Reset Controller uses the Power-on-Reset cells and Brownout Monitor.

The Reset Controller returns to software either the source of the last reset, or of a general reset, a wake-up reset, a software reset, a user reset, a watchdog or reinforced watchdog reset.

The Reset Controller controls the internal resets of the system (or independent reset of CM4P1 processor) and the NRST pin input/output. It shapes a reset signal for the external devices, simplifying to a minimum connection of a push-button on the NRST pin to implement a manual reset.

The configuration of the Reset Controller is saved as its is supplied by VDDBU.

## 10.4 Supply Controller (SUPC)

The Supply Controller controls the power supplies of each section of the processor.

The Supply Controller starts up the device by sequentially enabling the internal power switches and the Voltage Regulator, then it generates the proper reset signals to the core power supply.

It also sets the system in different low-power modes, wakes it up from a wide range of events.

## 11. Peripherals

### 11.1 Peripheral Identifiers

Table 11-1 defines the Peripheral Identifiers of the SAM4C. A peripheral identifier is required for the control of the peripheral interrupt with the Nested Vectored Interrupt Controller, and for the control of the peripheral clock with the Power Management Controller.

The two ARM Cortex-M4 processors share the same interrupt mapping, and thus, they share all the interrupts of the peripherals.

**Note:** Some peripherals are on the Bus Matrix 0/AHB to ABP Bridge 0 and other peripherals are on the Bus Matrix 1/AHB to ABP Bridge 1. If Core 0 needs to access a peripheral on the Bus Matrix 1/AHB to ABP Bridge 1, the Core 0 application must enable the Core 1 System Clock (Bus and peripherals) and release Core 1 System Reset (Bus and peripherals). Peripherals on Sub-system 0 or Sub-system 1 are mentioned in the Instance description table that follows.

Table 11-1. Peripheral Identifiers

Instance ID	Instance Name	NVIC Interrupt	PMC Clock Control	Instance Description
0	SUPC	X		Supply Controller
1	RSTC	X		Reset Controller
2	RTC	X		Real-time Clock
3	RTT	X		Real-time Timer
4	WDT	X		Watchdog Timer/Reinforced Watchdog Timer
5	PMC	X		Power Management Controller
6	EFC	X		Enhanced Embedded Flash Controller 0
7	–	–	–	Reserved
8	UART0	X	X	UART 0 (Sub-system 0 Clock)
9	–	–	–	Reserved
10	SMC0	–	X	Static Memory Controller 0 (Sub-system 0 Clock)
11	PIOA	X	X	Parallel I/O Controller A (Sub-system 0 Clock)
12	PIOB	X	X	Parallel I/O Controller B (Sub-system 0 Clock)
13	–	–	–	Reserved
14	USART0	X	X	USART 0 (Sub-system 0 Clock)
15	USART1	X	X	USART 1 (Sub-system 0 Clock)
16	USART2	X	X	USART 2 (Sub-system 0 Clock)
17	USART3	X	X	USART 3 (Sub-system 0 Clock)
18	USART4	X	X	USART 4 (Sub-system 0 Clock)
19	TWI0	X	X	Two Wire Interface 0 (Sub-system 0 Clock)
20	TWI1	X	X	Two Wire Interface 1 (Sub-system 0 Clock)
21	SPI0	X	X	Serial Peripheral Interface 0 (Sub-system 0 Clock)
22	–	–	–	Reserved
23	TC0	X	X	Timer/Counter 0 (Sub-system 0 Clock)
24	TC1	X	X	Timer/Counter 1 (Sub-system 0 Clock)
25	TC2	X	X	Timer/Counter 2 (Sub-system 0 Clock)

Table 11-1. Peripheral Identifiers (Continued)

Instance ID	Instance Name	NVIC Interrupt	PMC Clock Control	Instance Description
26	TC3	X	X	Timer/Counter 3 (Sub-system 0 Clock)
27	TC4	X	X	Timer/Counter 4 (Sub-system 0 Clock)
28	TC5	X	X	Timer/Counter 5 (Sub-system 0 Clock)
29	ADC	X	X	Analog To Digital Converter (Sub-system 0 Clock)
30	ARM	X	–	FPU signals (only on CM4P1 core): FPIX, FPOFC, FPUFC, FPIOC, FPDZC, FPIDC, FPIX
31	IPC0	X	X	Interprocessor communication 0 (Sub-system 0 Clock)
32	SLCDC	X	X	Segment LCD Controller (Sub-system 0 Clock)
33	TRNG	X	X	True Random Generator (Sub-system 0 Clock)
34	ICM	X	X	Integrity Check Module (Sub-system 0 Clock)
35	CPKCC	X	X	Classical Public Key Cryptography Controller (Sub-system 0 Clock)
36	AES	X	X	Advanced Enhanced Standard (Sub-system 0 Clock)
37	PIOC	X	X	Parallel I/O Controller C (Sub-system 1 Clock)
38	UART1	X	X	UART 1 (Sub-system 1 Clock)
39	IPC1	X	X	Interprocessor communication 1 (Sub-system 1 Clock)
40	SPI1	X	X	Serial Peripheral Interface 1 (Sub-system 1 Clock)
41	PWM	X	X	Pulse Width Modulation (Sub-system 1 Clock)
42	SRAM	–	X	SRAM1 (I/D Code bus of CM4P1), SRAM2 (System bus of CM4P1) (Sub-system 1 Clock)
43	SMC1	–	X	Static Memory Controller 1 (Sub-system 1 Clock)

## 11.2 Peripheral DMA Controller

Two PDC are available:

- PDC0: dedicated to peripherals on APB0
- PDC1: dedicated to peripherals on APB1

Features:

- Handles data transfer between peripherals and memories
- Low bus arbitration overhead
  - One Master Clock cycle needed for a transfer from memory to peripheral
  - Two Master Clock cycles needed for a transfer from peripheral to memory
- Next Pointer management for reducing interrupt latency requirement

Note that Peripheral DMA 0 on Matrix 0 cannot access SRAM1 or SRAM2. Peripheral DMA 1 on Matrix 1 cannot access SRAM0.

The Peripheral DMA Controller handles transfer requests from the channel according to the following priorities (Low to High priorities):

**Table 11-2. Peripheral DMA Controller (PDC0)**

Instance name	Channel T/R
AES	Transmit
TWI0	Transmit
UART0	Transmit
USART1	Transmit
USART0	Transmit
USART2	Transmit
USART3	Transmit
USART4	Transmit
SPI0	Transmit
AES	Receive
TWI0	Receive
UART0	Receive
USART4	Receive
USART3	Receive
USART2	Receive
USART1	Receive
USART0	Receive
ADC	Receive
SPI0	Receive

**Table 11-3. Peripheral DMA Controller (PDC1)**

Instance name	Channel T/R
UART1	Transmit
SPI1	Transmit
UART1	Receive
SPI1	Receive

## 11.3 APB/AHB Bridge

The SAM4C embeds two peripheral bridges: one on each Matrix (Matrix 0 for CM4P0 and Matrix 1 for CM4P1).

The peripherals of the bridge corresponding to CM4P0 (APB0) are clocked by MCK, and the peripherals of the bridge corresponding to CM4P1 (APB1) are clocked by CPBMCK.

## 11.4 Peripheral Signal Multiplexing on I/O Lines

The SAM4C can multiplex the I/O lines of the peripheral set.

The SAM4C PIO Controllers control up to 32 lines. Each line can be assigned to one of two peripheral functions: A or B. The multiplexing tables in the paragraphs that follow define how the I/O lines of the peripherals A and B are multiplexed on the PIO Controllers. The column “Comments” has been inserted in this table for the user’s own comments; it may be used to track how pins are defined in an application.

Note that some peripheral functions which are output only, might be duplicated within the tables.

### 11.4.1 Pad Features

In the tables that follow, the column “Feature” indicates if the corresponding I/O Line has programmable Pull-up, Pull-Down and Schmitt Trigger with mnemonics.

- “PUP”: Programmable (P) / Not Programmable (NP) Pull-up.
- “PDN”: Programmable (P) / Not Programmable (NP) Pull-down.
- “ST”: Programmable (P) / Not Programmable (NP) Schmitt trigger.
- “LDRV/MDRV/HDRV”: Programmable (P) / Not Programmable (NP) Drive (Low/Medium/High).

### 11.4.2 Reset State

In the tables that follow, the column “Reset State” indicates the reset state of the line with mnemonics.

- “PIO” or signal name: Indicates whether the PIO Line resets in I/O mode or in peripheral mode.  
If “PIO” is mentioned, the PIO Line is in general purpose I/O (GPIO). If a signal name is mentioned in the “Reset State” column, the PIO Line is assigned to this function.
- “I”/“O”: Indicates whether the signal is input or output state.
- “PU”/“PD”: Indicates whether Pull-up, Pull-down or nothing is enabled.
- “ST”: Indicates if Schmitt Trigger is enabled.

### 11.4.3 PIO Controller A Multiplexing

Table 11-4. Multiplexing on PIO Controller A (PIOA)

I/O Line	Peripheral A	Peripheral B	Peripheral C	Extra Function	System Function	Feature	Reset State	Comments
PA0	RTS3	PCK2	A10	COM0	WKUP5	- PUP(P) / PDN(P) - ST(P) - HDRV(NP)	PIO, I, PU	
PA1	CTS3	NCS1	A9	COM1		- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)		
PA2	SCK3	NCS2	A8	COM2				
PA3	RXD3	NCS3	A7	COM3	WKUP6			
PA4	TXD3		A6	COM4/AD1				
PA5	SPI0_NPCS0		A5	COM5/AD2				
PA6	SPI0_MISO		A4	SEG0				
PA7	SPI0_MOSI		A3	SEG1				
PA8	SPI0_SPCK		A2	SEG2		- PUP(P) / PDN(P) - ST(P) - HDRV(NP)		
PA9	RXD2		A1	SEG3	WKUP2	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)		
PA10	TXD2		A0/NBS0	SEG4				
PA11	RXD1		A23	SEG5	WKUP9			
PA12	TXD1		A22/ NANDCLE	SEG6/AD0				
PA13	SCK2	TIOA0	A21/ NANDALE	SEG7				
PA14	RTS2	TIOB0	A20	SEG8	WKUP3			
PA15	CTS2	TIOA4	A19	SEG9				
PA16	SCK1	TIOB4	A18	SEG10				
PA17	RTS1	TCLK4	A17	SEG11	WKUP7			
PA18	CTS1	TIOA5	A16	SEG12				
PA19	RTS0	TCLK5	A15	SEG13	WKUP4			
PA20	CTS0	TIOB5	A14	SEG14				
PA21	SPI0_NPCS1		A13	SEG15				
PA22	SPI0_NPCS2		A12	SEG16				
PA23	SPI0_NPCS3		A11	SEG17				
PA24	TWD0		A10	SEG18	WKUP1			
PA25	TWCK0		A9	SEG19				
PA26	CTS4		A8	SEG20				
PA27			NCS0	SEG21				
PA28			NRD	SEG22				
PA29	PCK1		NWAIT	SEG23		- PUP(P) / PDN(P) - ST(P) - HDRV(NP)		
PA30	PCK1		A15		XOUT	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)	XOUT	
PA31	PCK0		A14		XIN	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)	XIN	



## 11.4.4 PIO Controller B Multiplexing

Table 11-5. Multiplexing on PIO Controller B (PIOB)

I/O Line	Peripheral A	Peripheral B	Peripheral C	Extra Function	System Function	Feature	Reset State	Comments
PB0	TWD1				TDI	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)	JTAG, I	
PB1	TWCK1			RTCOUT0	TDO/ TRACESWO	- PUP(P)/PDN(P) - LDRV(NP)	JTAG, O	
PB2					TMS/SWDIO	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)	JTAG, I	
PB3					TCK/SWCLK			
PB4	URXD0	TCLK0	A17		WKUP8		PIO, I, PU	
PB5	UTXD0		A16					
PB6			D0	SEG24				
PB7	TIOA1		D1	SEG25				
PB8	TIOB1		D2	SEG26				
PB9	TCLK1		D3	SEG27				
PB10	TIOA2		D4	SEG28				
PB11	TIOB2		D5	SEG29				
PB12	TCLK2		D6	SEG30				
PB13	PCK0		D7	SEG31/AD3		- PUP(P) / PDN(P) - ST(P) - HDRV(NP)		
PB14			NWR0/ NWE	SEG32		- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)		
PB15			NWR1/ NBS1	SEG33				
PB16	RXD0		D8	SEG34	WKUP10/ TMP1		PIO, I, PD	
PB17	TXD0		D9	SEG35				
PB18	SCK0	PCK2	D10	SEG36				
PB19	RXD4		D11	SEG37				
PB20	TXD4		D12	SEG38				
PB21	SCK4	NANDOE	D13	SEG39	WKUP11		PIO, I, PU	
PB22	RTS4	NANDWE	D14	SEG40				
PB23	ADTRG		D15	SEG41/AD4				
PB24	TIOA3		A7	SEG42				
PB25	TIOB3		A6	SEG43				
PB26	TCLK3		A5	SEG44	WKUP13			
PB27			A4	SEG45	WKUP14/ TMP2			
PB28			A3	SEG46	WKUP15/ TMP3			
PB29			A2	SEG47				
PB30			A1	SEG48				
PB31			A0/ NBS0	SEG49/AD5				

## 11.4.5 PIO Controller C Multiplexing

Table 11-6. Multiplexing on PIO Controller C (PIOC)

I/O Line	Peripheral A	Peripheral B	Peripheral C	Extra Function	System Function	Feature	Reset State	Comments
PC0	UTXD1	PWM0				- PUP(P) - HDRV(NP)	PIO, I, PU	
PC1	URXD1	PWM1			WKUP12	- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)		
PC2	SPI1_NPCS0	PWM2						
PC3	SPI1_MISO	PWM3						
PC4	SPI1_MOSI							
PC5	SPI1_SPCK					- PUP(P) - HDRV(NP)		
PC6	PWM0	SPI1_NPCS1				- PUP(P) / PDN(P) - ST(P) - LDRV(P) / MDRV(P)		
PC7	PWM1	SPI1_NPCS2						
PC8	PWM2	SPI1_NPCS3						
PC9	PWM3				ERASE		ERASE, PD	

## 12. ARM Cortex-M4

### 12.1 Description

The Cortex-M4 processor is a high performance 32-bit processor designed for the microcontroller market. It offers significant benefits to developers, including outstanding processing performance combined with fast interrupt handling, enhanced system debug with extensive breakpoint and trace capabilities, efficient processor core, system and memories, ultra-low power consumption with integrated sleep modes, and platform security robustness, with integrated memory protection unit (MPU).

The Cortex-M4 processor is built on a high-performance processor core, with a 3-stage pipeline Harvard architecture, making it ideal for demanding embedded applications. The processor delivers exceptional power efficiency through an efficient instruction set and extensively optimized design, providing high-end processing hardware including IEEE754-compliant single-precision floating-point computation, a range of single-cycle and SIMD multiplication and multiply-with-accumulate capabilities, saturating arithmetic and dedicated hardware division.

To facilitate the design of cost-sensitive devices, the Cortex-M4 processor implements tightly-coupled system components that reduce processor area while significantly improving interrupt handling and system debug capabilities. The Cortex-M4 processor implements a version of the Thumb instruction set based on Thumb-2 technology, ensuring high code density and reduced program memory requirements. The Cortex-M4 instruction set provides the exceptional performance expected of a modern 32-bit architecture, with the high code density of 8-bit and 16-bit microcontrollers.

The Cortex-M4 processor closely integrates a configurable NVIC, to deliver industry-leading interrupt performance. The NVIC includes a non-maskable interrupt (NMI), and provides up to 256 interrupt priority levels. The tight integration of the processor core and NVIC provides fast execution of interrupt service routines (ISRs), dramatically reducing the interrupt latency. This is achieved through the hardware stacking of registers, and the ability to suspend load-multiple and store-multiple operations. Interrupt handlers do not require wrapping in assembler code, removing any code overhead from the ISRs. A tail-chain optimization also significantly reduces the overhead when switching from one ISR to another.

To optimize low-power designs, the NVIC integrates with the sleep modes, that include a deep sleep function that enables the entire device to be rapidly powered down while still retaining program state.

#### 12.1.1 System Level Interface

The Cortex-M4 processor provides multiple interfaces using AMBA® technology to provide high speed, low latency memory accesses. It supports unaligned data accesses and implements atomic bit manipulation that enables faster peripheral controls, system spinlocks and thread-safe Boolean data handling.

The Cortex-M4 processor has a Memory Protection Unit (MPU) that provides fine grain memory control, enabling applications to utilize multiple privilege levels, separating and protecting code, data and stack on a task-by-task basis. Such requirements are becoming critical in many embedded applications such as automotive.

#### 12.1.2 Integrated Configurable Debug

The Cortex-M4 processor implements a complete hardware debug solution. This provides high system visibility of the processor and memory through either a traditional JTAG port or a 2-pin Serial Wire Debug (SWD) port that is ideal for microcontrollers and other small package devices.

For system trace the processor integrates an Instrumentation Trace Macrocell (ITM) alongside data watchpoints and a profiling unit. To enable simple and cost-effective profiling of the system events these generate, a Serial Wire Viewer (SWV) can export a stream of software-generated messages, data trace, and profiling information through a single pin.

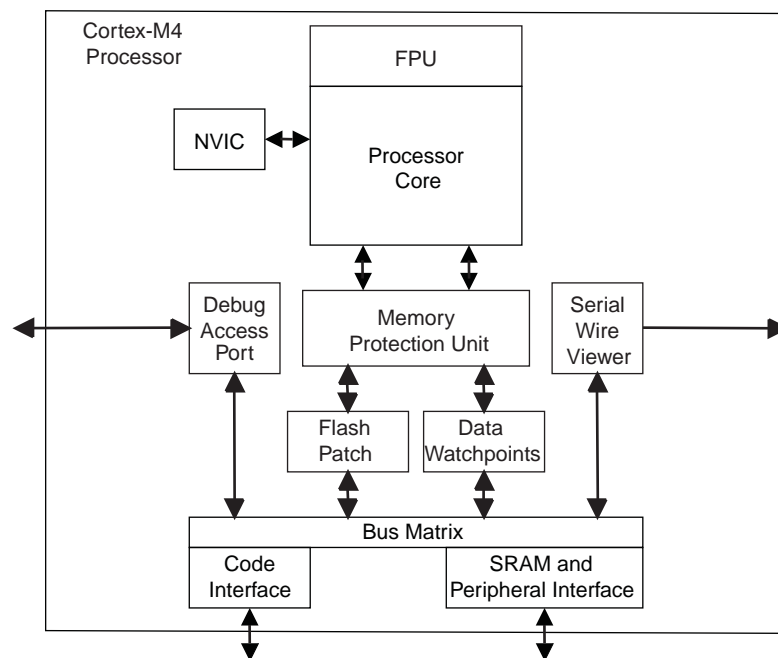
The Flash Patch and Breakpoint Unit (FPB) provides up to 8 hardware breakpoint comparators that debuggers can use. The comparators in the FPB also provide remap functions of up to 8 words in the program code in the CODE memory region. This enables applications stored on a non-erasable, ROM-based microcontroller to be patched if a small programmable memory, for example flash, is available in the device. During initialization, the application in ROM detects, from the programmable memory, whether a patch is required. If a patch is required, the application programs the FPB to remap a number of addresses. When those addresses are accessed, the accesses are redirected to a remap table specified in the FPB configuration, which means the program in the non-modifiable ROM can be patched.

## 12.2 Embedded Characteristics

- Tight integration of system peripherals reduces area and development costs
- Thumb instruction set combines high code density with 32-bit performance
- IEEE754-compliant single-precision FPU
- Code-patch ability for ROM system updates
- Power control optimization of system components
- Integrated sleep modes for low power consumption
- Fast code execution permits slower processor clock or increases sleep mode time
- Hardware division and fast digital-signal-processing oriented multiply accumulate
- Saturating arithmetic for signal processing
- Deterministic, high-performance interrupt handling for time-critical applications
- Memory Protection Unit (MPU) for safety-critical applications
- Extensive debug and trace capabilities:
  - Serial Wire Debug and Serial Wire Trace reduce the number of pins required for debugging, tracing, and code profiling.

## 12.3 Block Diagram

Figure 12-1. Typical Cortex-M4 Implementation



## 12.4 Cortex-M4 Models

### 12.4.1 Programmers Model

This section describes the Cortex-M4 programmers model. In addition to the individual core register descriptions, it contains information about the processor modes and privilege levels for software execution and stacks.

#### 12.4.1.1 Processor Modes and Privilege Levels for Software Execution

The processor *modes* are:

- Thread mode  
Used to execute application software. The processor enters the Thread mode when it comes out of reset.
- Handler mode  
Used to handle exceptions. The processor returns to the Thread mode when it has finished exception processing.

The *privilege levels* for software execution are:

- Unprivileged  
The software:
  - Has limited access to the MSR and MRS instructions, and cannot use the CPS instruction
  - Cannot access the System Timer, NVIC, or System Control Block
  - Might have a restricted access to memory or peripherals.

*Unprivileged software* executes at the unprivileged level.

- Privileged  
The software can use all the instructions and has access to all resources. *Privileged software* executes at the privileged level.

In Thread mode, the CONTROL register controls whether the software execution is privileged or unprivileged, see [Section 12.4.1.16 "CONTROL Register"](#). In Handler mode, software execution is always privileged.

Only privileged software can write to the CONTROL register to change the privilege level for software execution in Thread mode. Unprivileged software can use the SVC instruction to make a *supervisor call* to transfer control to privileged software.

#### 12.4.1.2 Stacks

The processor uses a full descending stack. This means the stack pointer holds the address of the last stacked item in memory. When the processor pushes a new item onto the stack, it decrements the stack pointer and then writes the item to the new memory location. The processor implements two stacks, the *main stack* and the *process stack*, with a pointer for each held in independent registers, see [Section 12.4.1.5 "Stack Pointer"](#).

In Thread mode, the CONTROL register controls whether the processor uses the main stack or the process stack, see [Section 12.4.1.16 "CONTROL Register"](#).

In Handler mode, the processor always uses the main stack.

The options for processor operations are:

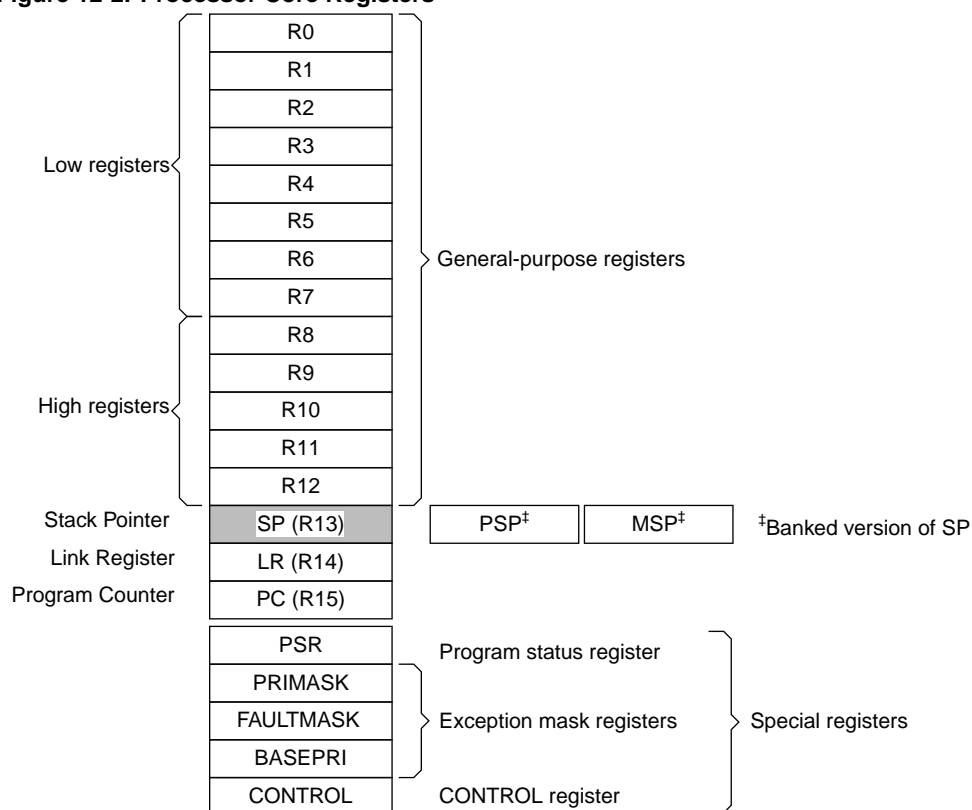
**Table 12-1. Summary of processor mode, execution privilege level, and stack use options**

Processor Mode	Used to Execute	Privilege Level for Software Execution	Stack Used
Thread	Applications	Privileged or unprivileged <sup>(1)</sup>	Main stack or process stack <sup>(1)</sup>
Handler	Exception handlers	Always privileged	Main stack

Note: 1. See [Section 12.4.1.16 "CONTROL Register"](#).

### 12.4.1.3 Core Registers

**Figure 12-2. Processor Core Registers**



**Table 12-2. Core Processor Registers**

Register	Name	Access <sup>(1)</sup>	Required Privilege <sup>(2)</sup>	Reset
General-purpose registers	R0-R12	Read-write	Either	Unknown
Stack Pointer	MSP	Read-write	Privileged	See description
Stack Pointer	PSP	Read-write	Either	Unknown
Link Register	LR	Read-write	Either	0xFFFFFFFF
Program Counter	PC	Read-write	Either	See description
Program Status Register	PSR	Read-write	Privileged	0x01000000
Application Program Status Register	APSR	Read-write	Either	0x00000000
Interrupt Program Status Register	IPSR	Read-only	Privileged	0x00000000
Execution Program Status Register	EPSR	Read-only	Privileged	0x01000000
Priority Mask Register	PRIMASK	Read-write	Privileged	0x00000000
Fault Mask Register	FAULTMASK	Read-write	Privileged	0x00000000
Base Priority Mask Register	BASEPRI	Read-write	Privileged	0x00000000
CONTROL register	CONTROL	Read-write	Privileged	0x00000000

Notes: 1. Describes access type during program execution in thread mode and Handler mode. Debug access can differ.

2. An entry of Either means privileged and unprivileged software can access the register.

#### 12.4.1.4 General-purpose Registers

R0-R12 are 32-bit general-purpose registers for data operations.

#### 12.4.1.5 Stack Pointer

The *Stack Pointer* (SP) is register R13. In Thread mode, bit[1] of the CONTROL register indicates the stack pointer to use:

- 0 = *Main Stack Pointer* (MSP). This is the reset value.
- 1 = *Process Stack Pointer* (PSP).

On reset, the processor loads the MSP with the value from address 0x00000000.

#### 12.4.1.6 Link Register

The *Link Register* (LR) is register R14. It stores the return information for subroutines, function calls, and exceptions. On reset, the processor loads the LR value 0xFFFFFFFF.

#### 12.4.1.7 Program Counter

The *Program Counter* (PC) is register R15. It contains the current program address. On reset, the processor loads the PC with the value of the reset vector, which is at address 0x00000004. Bit[0] of the value is loaded into the EPSR T-bit at reset and must be 1.

### 12.4.1.8 Program Status Register

**Name:** PSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
N	Z	C	V	Q	ICI/IT		T
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
ICI/IT						—	ISR_NUMBER
7	6	5	4	3	2	1	0
ISR_NUMBER							

The *Program Status Register* (PSR) combines:

- *Application Program Status Register* (APSR)
- *Interrupt Program Status Register* (IPSR)
- *Execution Program Status Register* (EPSR).

These registers are mutually exclusive bitfields in the 32-bit PSR.

The PSR register accesses these registers individually or as a combination of any two or all three registers, using the register name as an argument to the MSR or MRS instructions. For example:

- Read of all the registers using PSR with the MRS instruction
- Write to the APSR N, Z, C, V and Q bits using APSR\_nzcvq with the MSR instruction.

The PSR combinations and attributes are:

Name	Access	Combination
PSR	Read-write <sup>(1)(2)</sup>	APSR, EPSR, and IPSR
IEPSR	Read-only	EPSR and IPSR
IAPSR	Read-write <sup>(1)</sup>	APSR and IPSR
EAPSR	Read-write <sup>(2)</sup>	APSR and EPSR

- Notes:
1. The processor ignores writes to the IPSR bits.
  2. Reads of the EPSR bits return zero, and the processor ignores writes to these bits

See the instruction descriptions [Section 12.6.12.6 "MRS"](#) and [Section 12.6.12.7 "MSR"](#) for more information about how to access the program status registers.



#### 12.4.1.9 Application Program Status Register

**Name:** APSR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
N	Z	C	V	Q	–		
23	22	21	20	19	18	17	16
–				GE[3:0]			
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
–							

The APSR contains the current state of the condition flags from previous instruction executions.

- **N: Negative Flag**

0: Operation result was positive, zero, greater than, or equal

1: Operation result was negative or less than.

- **Z: Zero Flag**

0: Operation result was not zero

1: Operation result was zero.

- **C: Carry or Borrow Flag**

Carry or borrow flag:

0: Add operation did not result in a carry bit or subtract operation resulted in a borrow bit

1: Add operation resulted in a carry bit or subtract operation did not result in a borrow bit.

- **V: Overflow Flag**

0: Operation did not result in an overflow

1: Operation resulted in an overflow.

- **Q: DSP Overflow and Saturation Flag**

Sticky saturation flag:

0: Indicates that saturation has not occurred since reset or since the bit was last cleared to zero

1: Indicates when an SSAT or USAT instruction results in saturation.

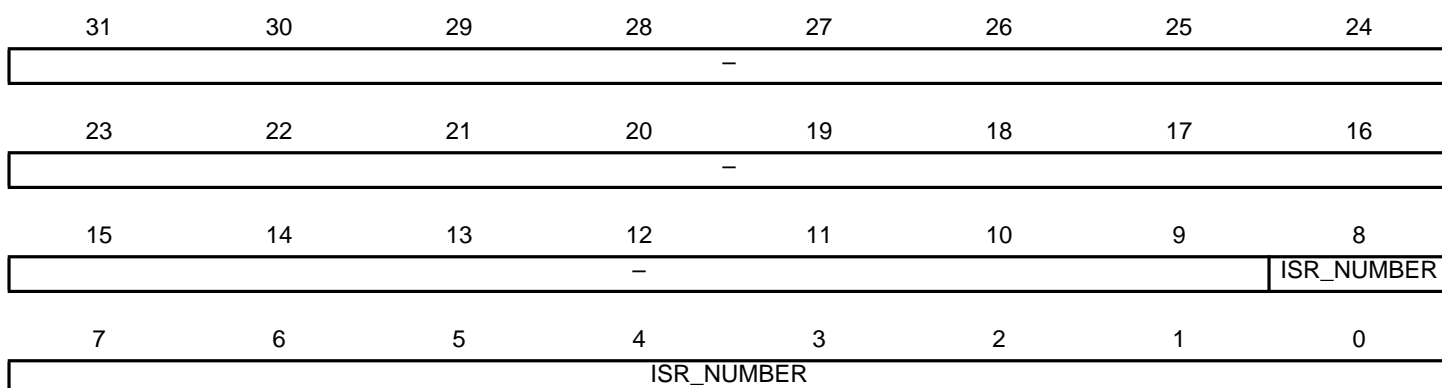
This bit is cleared to zero by software using an MRS instruction.

- **GE[19:16]: Greater Than or Equal Flags**

See [Section 12.6.5.21 "SEL"](#) for more information.

#### 12.4.1.10 Interrupt Program Status Register

**Name:** IPSR  
**Access:** Read-write  
**Reset:** 0x00000000



The IPSR contains the exception type number of the current *Interrupt Service Routine* (ISR).

- **ISR\_NUMBER: Number of the Current Exception**

- 0 = Thread mode
- 1 = Reserved
- 2 = NMI
- 3 = Hard fault
- 4 = Memory management fault
- 5 = Bus fault
- 6 = Usage fault
- 7-10 = Reserved
- 11 = SVCall
- 12 = Reserved for Debug
- 13 = Reserved
- 14 = PendSV
- 15 = SysTick
- 16 = IRQ0
- 53 = IRQ37

See [Section 12.4.3.2 "Exception Types"](#) for more information.

#### 12.4.1.11 Execution Program Status Register

**Name:** EPSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–					ICI/IT		T
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
ICI/IT						–	
7	6	5	4	3	2	1	0
–							

The EPSR contains the Thumb state bit, and the execution state bits for either the *If-Then* (IT) instruction, or the *Interruptible-Continuable Instruction* (ICI) field for an interrupted load multiple or store multiple instruction.

Attempts to read the EPSR directly through application software using the MSR instruction always return zero. Attempts to write the EPSR using the MSR instruction in the application software are ignored. Fault handlers can examine the EPSR value in the stacked PSR to indicate the operation that is at fault. See [Section 12.4.3.7 "Exception Entry and Return"](#)

##### • ICI: Interruptible-continuable Instruction

When an interrupt occurs during the execution of an LDM, STM, PUSH, POP, VLDM, VSTM, VPUSH, or VPOP instruction, the processor:

- Stops the load multiple or store multiple instruction operation temporarily
- Stores the next register operand in the multiple operation to EPSR bits[15:12].

After servicing the interrupt, the processor:

- Returns to the register pointed to by bits[15:12]
- Resumes the execution of the multiple load or store instruction.

When the EPSR holds the ICI execution state, bits[26:25,11:10] are zero.

##### • IT: If-Then Instruction

Indicates the execution state bits of the IT instruction.

The If-Then block contains up to four instructions following an IT instruction. Each instruction in the block is conditional. The conditions for the instructions are either all the same, or some can be the inverse of others. See [Section 12.6.10.3 "IT"](#) for more information.

##### • T: Thumb State

The Cortex-M4 processor only supports the execution of instructions in Thumb state. The following can clear the T bit to 0:

- Instructions BLX, BX and POP{PC}
- Restoration from the stacked xPSR value on an exception return
- Bit[0] of the vector value on an exception entry or reset.

Attempting to execute instructions when the T bit is 0 results in a fault or lockup. See subsection ["Lockup"](#) for more information.

### 12.4.1.12 Exception Mask Registers

The exception mask registers disable the handling of exceptions by the processor. Disable exceptions where they might impact on timing critical tasks.

To access the exception mask registers use the MSR and MRS instructions, or the CPS instruction to change the value of PRIMASK or FAULTMASK. See [Section 12.6.12.6 "MRS"](#), [Section 12.6.12.7 "MSR"](#), and [Section 12.6.12.2 "CPS"](#) for more information.

### 12.4.1.13 Priority Mask Register

**Name:** PRIMASK

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
—							PRIMASK

The PRIMASK register prevents the activation of all exceptions with a configurable priority.

- **PRIMASK**

0: No effect

1: Prevents the activation of all exceptions with a configurable priority.

#### 12.4.1.14 Fault Mask Register

**Name:** FAULTMASK

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
—							FAULTMASK

The FAULTMASK register prevents the activation of all exceptions except for Non-Maskable Interrupt (NMI).

- **FAULTMASK**

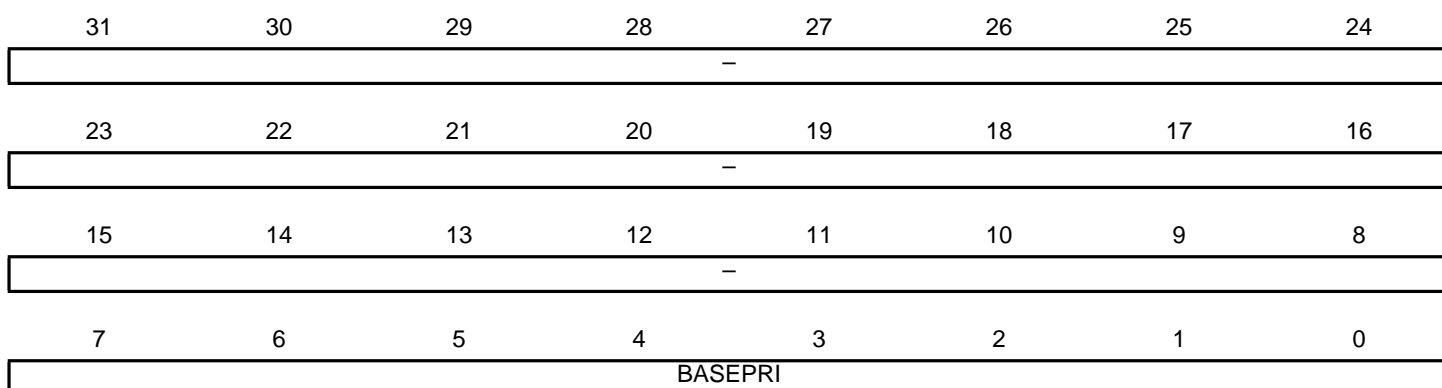
0: No effect.

1: Prevents the activation of all exceptions except for NMI.

The processor clears the FAULTMASK bit to 0 on exit from any exception handler except the NMI handler.

#### 12.4.1.15 Base Priority Mask Register

**Name:** BASEPRI  
**Access:** Read-write  
**Reset:** 0x00000000



The BASEPRI register defines the minimum priority for exception processing. When BASEPRI is set to a nonzero value, it prevents the activation of all exceptions with same or lower priority level as the BASEPRI value.

##### • BASEPRI

Priority mask bits:

- 0x0000 = No effect.
- Nonzero = Defines the base priority for exception processing.

The processor does not process any exception with a priority value greater than or equal to BASEPRI.

This field is similar to the priority fields in the interrupt priority registers. The processor implements only bits[7:4] of this field, bits[3:0] read as zero and ignore writes. See [Section 12.8.3.6 "Interrupt Priority Registers"](#) for more information. Remember that higher priority field values correspond to lower exception priorities.

#### 12.4.1.16 CONTROL Register

**Name:** CONTROL  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
—					FPCA	SPSEL	nPRIV

The CONTROL register controls the stack used and the privilege level for software execution when the processor is in Thread mode and indicates whether the FPU state is active.

- **FPCA: Floating-point Context Active**

Indicates whether the floating-point context is currently active:

- 0: No floating-point context active.
- 1: Floating-point context active.

The Cortex-M4 uses this bit to determine whether to preserve the floating-point state when processing an exception.

- **SPSEL: Active Stack Pointer**

Defines the current stack:

- 0: MSP is the current stack pointer.
- 1: PSP is the current stack pointer.

In Handler mode, this bit reads as zero and ignores writes. The Cortex-M4 updates this bit automatically on exception return.

- **nPRIV: Thread Mode Privilege Level**

Defines the Thread mode privilege level:

- 0: Privileged.
- 1: Unprivileged.

Handler mode always uses the MSP, so the processor ignores explicit writes to the active stack pointer bit of the CONTROL register when in Handler mode. The exception entry and return mechanisms update the CONTROL register based on the EXC\_RETURN value.

In an OS environment, ARM recommends that threads running in Thread mode use the process stack, and the kernel and exception handlers use the main stack.

By default, the Thread mode uses the MSP. To switch the stack pointer used in Thread mode to the PSP, either:

- Use the MSR instruction to set the Active stack pointer bit to 1, see [Section 12.6.12.7 "MSR"](#), or
- Perform an exception return to Thread mode with the appropriate EXC\_RETURN value, see [Table 12-10](#).

**Note:** When changing the stack pointer, the software must use an ISB instruction immediately after the MSR instruction. This ensures that instructions after the ISB execute using the new stack pointer. See [Section 12.6.12.5 "ISB"](#).

#### 12.4.1.17 Exceptions and Interrupts

The Cortex-M4 processor supports interrupts and system exceptions. The processor and the *Nested Vectored Interrupt Controller* (NVIC) prioritize and handle all exceptions. An exception changes the normal flow of software control. The processor uses the Handler mode to handle all exceptions except for reset. See [Section "Exception Entry"](#) and [Section "Exception Return"](#) for more information.

The NVIC registers control interrupt handling. See [Section 12.8 "Nested Vectored Interrupt Controller \(NVIC\)"](#) for more information.

#### 12.4.1.18 Data Types

The processor supports the following data types:

- 32-bit words
- 16-bit halfwords
- 8-bit bytes
- The processor manages all data memory accesses as little-endian. Instruction memory and *Private Peripheral Bus* (PPB) accesses are always little-endian. See [Section 12.4.2.1 "Memory Regions, Types and Attributes"](#) for more information.

#### 12.4.1.19 Cortex Microcontroller Software Interface Standard (CMSIS)

For a Cortex-M4 microcontroller system, the *Cortex Microcontroller Software Interface Standard* (CMSIS) defines:

- A common way to:
  - Access peripheral registers
  - Define exception vectors
- The names of:
  - The registers of the core peripherals
  - The core exception vectors
- A device-independent interface for RTOS kernels, including a debug channel.

The CMSIS includes address definitions and data structures for the core peripherals in the Cortex-M4 processor.

The CMSIS simplifies the software development by enabling the reuse of template code and the combination of CMSIS-compliant software components from various middleware vendors. Software vendors can expand the CMSIS to include their peripheral definitions and access functions for those peripherals.

This document includes the register names defined by the CMSIS, and gives short descriptions of the CMSIS functions that address the processor core and the core peripherals.

**Note:** This document uses the register short names defined by the CMSIS. In a few cases, these differ from the architectural short names that might be used in other documents.

The following sections give more information about the CMSIS:

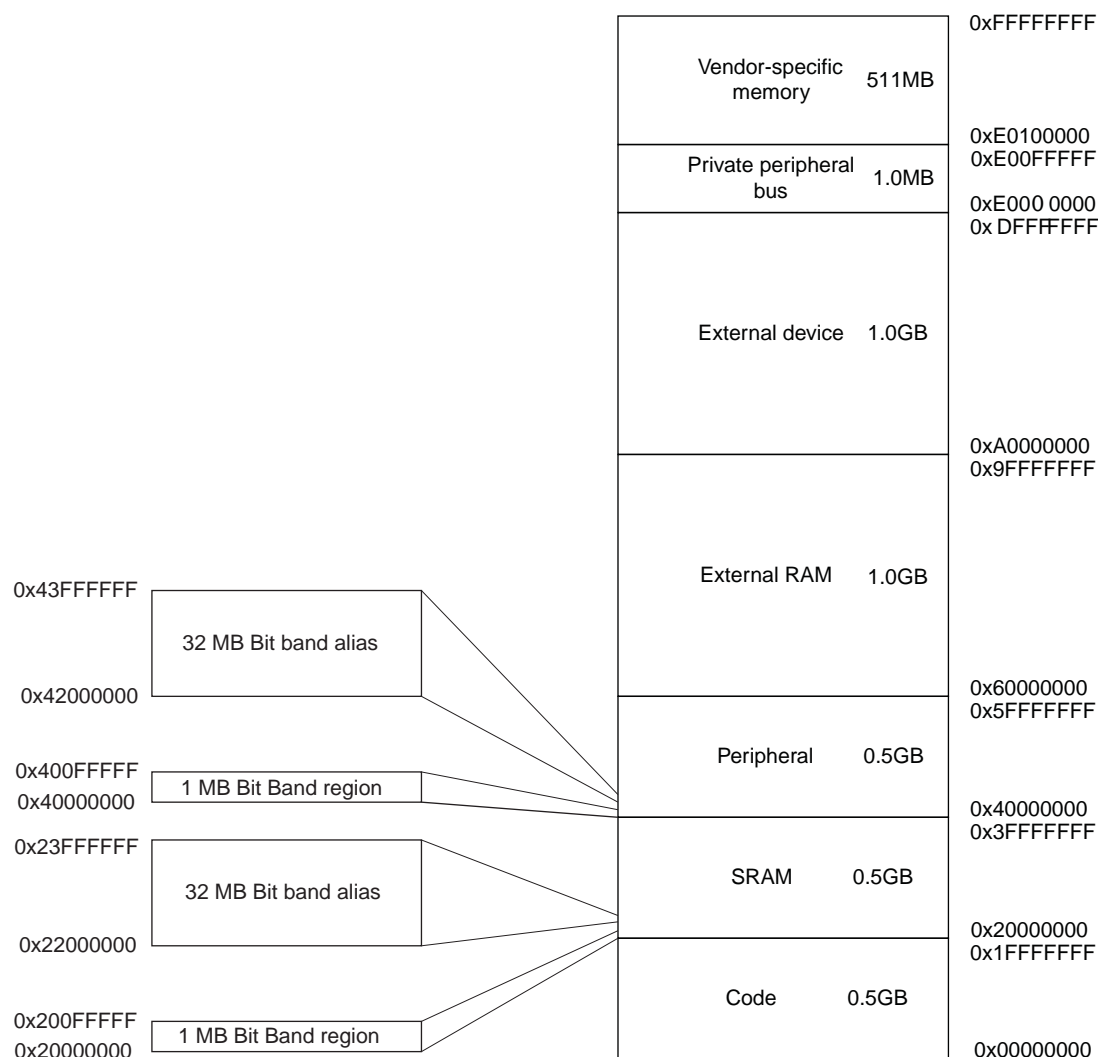
- [Section 12.5.3 "Power Management Programming Hints"](#)
- [Section 12.6.2 "CMSIS Functions"](#)
- [Section 12.8.2.1 "NVIC Programming Hints"](#).



## 12.4.2 Memory Model

This section describes the processor memory map, the behavior of memory accesses, and the bit-banding features. The processor has a fixed memory map that provides up to 4GB of addressable memory.

**Figure 12-3. Memory Map**



The regions for SRAM and peripherals include bit-band regions. Bit-banding provides atomic operations to bit data, see [Section 12.4.2.5 "Bit-banding"](#).

The processor reserves regions of the *Private peripheral bus* (PPB) address range for core peripheral registers.

This memory mapping is generic to ARM Cortex-M4 products. To get the specific memory mapping of this product, refer to the Memories section of the datasheet.

### 12.4.2.1 Memory Regions, Types and Attributes

The memory map and the programming of the MPU split the memory map into regions. Each region has a defined memory type, and some regions have additional memory attributes. The memory type and attributes determine the behavior of accesses to the region.

#### Memory Types

- **Normal**  
The processor can re-order transactions for efficiency, or perform speculative reads.
- **Device**  
The processor preserves transaction order relative to other transactions to Device or Strongly-ordered memory.
- **Strongly-ordered**  
The processor preserves transaction order relative to all other transactions.

The different ordering requirements for Device and Strongly-ordered memory mean that the memory system can buffer a write to Device memory, but must not buffer a write to Strongly-ordered memory.

#### Additional Memory Attributes

- **Shareable**  
For a shareable memory region, the memory system provides data synchronization between bus masters in a system with multiple bus masters, for example, a processor with a DMA controller.  
Strongly-ordered memory is always shareable.  
If multiple bus masters can access a non-shareable memory region, the software must ensure data coherency between the bus masters.
- **Execute Never (XN)**  
Means the processor prevents instruction accesses. A fault exception is generated only on execution of an instruction executed from an XN region.

### 12.4.2.2 Memory System Ordering of Memory Accesses

For most memory accesses caused by explicit memory access instructions, the memory system does not guarantee that the order in which the accesses complete matches the program order of the instructions, providing this does not affect the behavior of the instruction sequence. Normally, if correct program execution depends on two memory accesses completing in program order, the software must insert a memory barrier instruction between the memory access instructions, see [Section 12.4.2.4 "Software Ordering of Memory Accesses"](#).

However, the memory system does guarantee some ordering of accesses to Device and Strongly-ordered memory. For two memory access instructions A1 and A2, if A1 occurs before A2 in program order, the ordering of the memory accesses is described below.

**Table 12-3. Ordering of the Memory Accesses Caused by Two Instructions**

A2	Normal Access	Device Access		Strongly-ordered Access
		Non-shareable	Shareable	
A1				
Normal Access	–	–	–	–
Device access, non-shareable	–	<	–	<
Device access, shareable	–	–	<	<
Strongly-ordered access	–	<	<	<

Where:

- Means that the memory system does not guarantee the ordering of the accesses.
- < Means that accesses are observed in program order, that is, A1 is always observed before A2.

### 12.4.2.3 Behavior of Memory Accesses

The behavior of accesses to each region in the memory map is:

**Table 12-4. Memory Access Behavior**

Address Range	Memory Region	Memory Type	XN	Description
0x00000000 - 0x1FFFFFFF	Code	Normal <sup>(1)</sup>	-	Executable region for program code. Data can also be put here.
0x20000000 - 0x3FFFFFFF	SRAM	Normal <sup>(1)</sup>	-	Executable region for data. Code can also be put here. This region includes bit band and bit band alias areas, see <a href="#">Table 12-6</a> .
0x40000000 - 0x5FFFFFFF	Peripheral	Device <sup>(1)</sup>	XN	This region includes bit band and bit band alias areas, see <a href="#">Table 12-6</a> .
0x60000000 - 0x9FFFFFFF	External RAM	Normal <sup>(1)</sup>	-	Executable region for data.
0xA0000000 - 0xDFFFFFFF	External device	Device <sup>(1)</sup>	XN	External Device memory
0xE0000000 - 0xE00FFFFF	Private Peripheral Bus	Strongly-ordered <sup>(1)</sup>	XN	This region includes the NVIC, System timer, and system control block.
0xE0100000 - 0xFFFFFFFF	Reserved	Device <sup>(1)</sup>	XN	Reserved

Note: 1. See [Section 12.4.2.1 "Memory Regions, Types and Attributes"](#) for more information.

The Code, SRAM, and external RAM regions can hold programs. However, ARM recommends that programs always use the Code region. This is because the processor has separate buses that enable instruction fetches and data accesses to occur simultaneously.

The MPU can override the default memory access behavior described in this section. For more information, see [Section 12.11 "Memory Protection Unit \(MPU\)"](#).

#### *Additional Memory Access Constraints For Shared Memory*

When a system includes shared memory, some memory regions have additional access constraints, and some regions are subdivided, as [Table 12-5](#) shows:

**Table 12-5. Memory Region Shareability Policies**

Address Range	Memory Region	Memory Type	Shareability	
0x00000000-0x1FFFFFFF	Code	Normal <sup>(1)</sup>	-	<a href="#">(2)</a>
0x20000000-0x3FFFFFFF	SRAM	Normal <sup>(1)</sup>	-	<a href="#">(2)</a>
0x40000000-0x5FFFFFFF	Peripheral	Device <sup>(1)</sup>	-	
0x60000000-0x7FFFFFFF	External RAM	Normal <sup>(1)</sup>	-	WBWA <a href="#">(2)</a>
0x80000000-0x9FFFFFFF				WT <a href="#">(2)</a>
0xA0000000-0xBFFFFFFF	External device	Device <sup>(1)</sup>	Shareable <sup>(1)</sup>	-
0xC0000000-0xDFFFFFFF			Non-shareable <sup>(1)</sup>	

**Table 12-5. Memory Region Shareability Policies (Continued)**

Address Range	Memory Region	Memory Type	Shareability	
0xE0000000-0xE00FFFFF	Private Peripheral Bus	Strongly- ordered <sup>(1)</sup>	Shareable <sup>(1)</sup>	-
0xE0100000-0xFFFFFFF	Vendor-specific device	Device <sup>(1)</sup>	-	-

Notes: 1. See [Section 12.4.2.1 "Memory Regions, Types and Attributes"](#) for more information.

2. WT = Write through, no write allocate. WBWA = Write back, write allocate. See the [Section 12.13 "Glossary"](#) for more information.

### *Instruction Prefetch and Branch Prediction*

The Cortex-M4 processor:

- Prefetches instructions ahead of execution
- Speculatively prefetches from branch target addresses.

#### **12.4.2.4 Software Ordering of Memory Accesses**

The order of instructions in the program flow does not always guarantee the order of the corresponding memory transactions. This is because:

- The processor can reorder some memory accesses to improve efficiency, providing this does not affect the behavior of the instruction sequence.
- The processor has multiple bus interfaces
- Memory or devices in the memory map have different wait states
- Some memory accesses are buffered or speculative.

[Section 12.4.2.2 "Memory System Ordering of Memory Accesses"](#) describes the cases where the memory system guarantees the order of memory accesses. Otherwise, if the order of memory accesses is critical, the software must include memory barrier instructions to force that ordering. The processor provides the following memory barrier instructions:

#### *DMB*

The *Data Memory Barrier* (DMB) instruction ensures that outstanding memory transactions complete before subsequent memory transactions. See [Section 12.6.12.3 "DMB"](#).

#### *DSB*

The *Data Synchronization Barrier* (DSB) instruction ensures that outstanding memory transactions complete before subsequent instructions execute. See [Section 12.6.12.4 "DSB"](#).

#### *ISB*

The *Instruction Synchronization Barrier* (ISB) ensures that the effect of all completed memory transactions is recognizable by subsequent instructions. See [Section 12.6.12.5 "ISB"](#).

### *MPU Programming*

Use a DSB followed by an ISB instruction or exception return to ensure that the new MPU configuration is used by subsequent instructions.

#### **12.4.2.5 Bit-banding**

A bit-band region maps each word in a *bit-band alias* region to a single bit in the *bit-band region*. The bit-band regions occupy the lowest 1 MB of the SRAM and peripheral memory regions.

The memory map has two 32 MB alias regions that map to two 1 MB bit-band regions:

- Accesses to the 32 MB SRAM alias region map to the 1 MB SRAM bit-band region, as shown in [Table 12-6](#).

- Accesses to the 32 MB peripheral alias region map to the 1 MB peripheral bit-band region, as shown in [Table 12-7](#).

**Table 12-6. SRAM Memory Bit-banding Regions**

Address Range	Memory Region	Instruction and Data Accesses
0x20000000-0x200FFFFFFF	SRAM bit-band region	Direct accesses to this memory range behave as SRAM memory accesses, but this region is also bit-addressable through bit-band alias.
0x22000000-0x23FFFFFFF	SRAM bit-band alias	Data accesses to this region are remapped to bit-band region. A write operation is performed as read-modify-write. Instruction accesses are not remapped.

**Table 12-7. Peripheral Memory Bit-banding Regions**

Address Range	Memory Region	Instruction and Data Accesses
0x40000000-0x400FFFFFFF	Peripheral bit-band alias	Direct accesses to this memory range behave as peripheral memory accesses, but this region is also bit-addressable through bit-band alias.
0x42000000-0x43FFFFFFF	Peripheral bit-band region	Data accesses to this region are remapped to bit-band region. A write operation is performed as read-modify-write. Instruction accesses are not permitted.

- Notes:
1. A word access to the SRAM or peripheral bit-band alias regions map to a single bit in the SRAM or peripheral bit-band region.
  2. Bit-band accesses can use byte, halfword, or word transfers. The bit-band transfer size matches the transfer size of the instruction making the bit-band access.

The following formula shows how the alias region maps onto the bit-band region:

$$\begin{aligned}\text{bit\_word\_offset} &= (\text{byte\_offset} \times 32) + (\text{bit\_number} \times 4) \\ \text{bit\_word\_addr} &= \text{bit\_band\_base} + \text{bit\_word\_offset}\end{aligned}$$

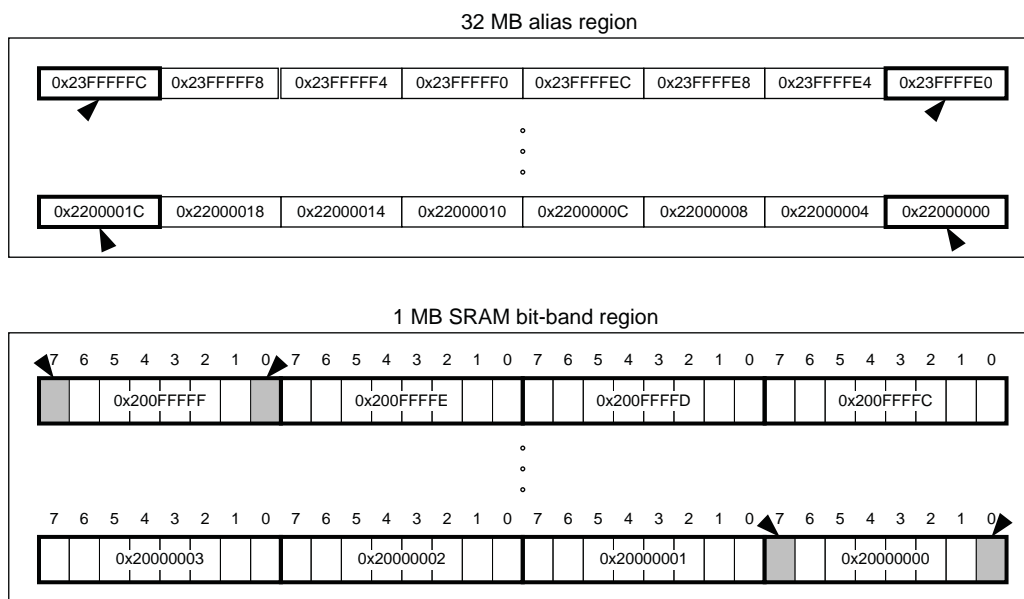
where:

- `Bit_word_offset` is the position of the target bit in the bit-band memory region.
- `Bit_word_addr` is the address of the word in the alias memory region that maps to the targeted bit.
- `Bit_band_base` is the starting address of the alias region.
- `Byte_offset` is the number of the byte in the bit-band region that contains the targeted bit.
- `Bit_number` is the bit position, 0-7, of the targeted bit.

[Figure 12-4](#) shows examples of bit-band mapping between the SRAM bit-band alias region and the SRAM bit-band region:

- The alias word at 0x23FFFE0 maps to bit[0] of the bit-band byte at 0x200FFFFF:  $0x23FFFE0 = 0x22000000 + (0xFFFF \times 32) + (0 \times 4)$ .
- The alias word at 0x23FFFFC maps to bit[7] of the bit-band byte at 0x200FFFFF:  $0x23FFFFC = 0x22000000 + (0xFFFF \times 32) + (7 \times 4)$ .
- The alias word at 0x2200000 maps to bit[0] of the bit-band byte at 0x20000000:  $0x2200000 = 0x22000000 + (0 \times 32) + (0 \times 4)$ .
- The alias word at 0x220001C maps to bit[7] of the bit-band byte at 0x20000000:  $0x220001C = 0x22000000 + (0 \times 32) + (7 \times 4)$ .

**Figure 12-4. Bit-band Mapping**



### *Directly Accessing an Alias Region*

Writing to a word in the alias region updates a single bit in the bit-band region.

Bit[0] of the value written to a word in the alias region determines the value written to the targeted bit in the bit-band region. Writing a value with bit[0] set to 1 writes a 1 to the bit-band bit, and writing a value with bit[0] set to 0 writes a 0 to the bit-band bit.

Bits[31:1] of the alias word have no effect on the bit-band bit. Writing 0x01 has the same effect as writing 0xFF. Writing 0x00 has the same effect as writing 0x0E.

Reading a word in the alias region:

- 0x00000000 indicates that the targeted bit in the bit-band region is set to 0
- 0x00000001 indicates that the targeted bit in the bit-band region is set to 1

### *Directly Accessing a Bit-band Region*

[Section 12.4.2.3 "Behavior of Memory Accesses"](#) describes the behavior of direct byte, halfword, or word accesses to the bit-band regions.

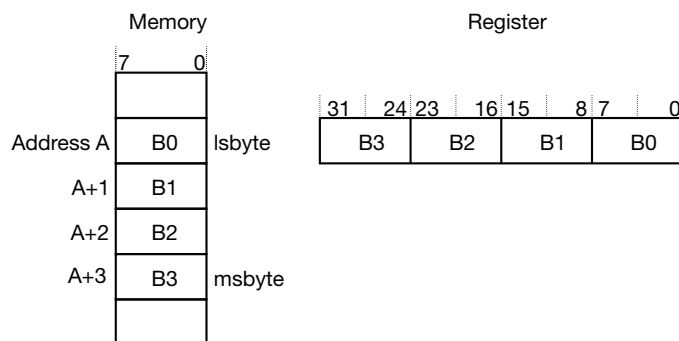
### 12.4.2.6 Memory Endianness

The processor views memory as a linear collection of bytes numbered in ascending order from zero. For example, bytes 0-3 hold the first stored word, and bytes 4-7 hold the second stored word. The subsection “[Little-endian Format](#)” describes how words of data are stored in memory.

#### *Little-endian Format*

In little-endian format, the processor stores the least significant byte of a word at the lowest-numbered byte, and the most significant byte at the highest-numbered byte. An example is given in [Figure 12-5 “Little-endian Format”](#) below.

**Figure 12-5. Little-endian Format**



### 12.4.2.7 Synchronization Primitives

The Cortex-M4 instruction set includes pairs of *synchronization primitives*. These provide a non-blocking mechanism that a thread or process can use to obtain exclusive access to a memory location. The software can use them to perform a guaranteed read-modify-write memory update sequence, or for a semaphore mechanism.

A pair of synchronization primitives comprises:

**A Load-exclusive Instruction**, used to read the value of a memory location, requesting exclusive access to that location.

**A Store-Exclusive instruction**, used to attempt to write to the same memory location, returning a status bit to a register. If this bit is:

- 0: It indicates that the thread or process gained exclusive access to the memory, and the write succeeds,
- 1: It indicates that the thread or process did not gain exclusive access to the memory, and no write is performed.

The pairs of Load-Exclusive and Store-Exclusive instructions are:

- The word instructions LDREX and STREX
- The halfword instructions LDREXH and STREXH
- The byte instructions LDREXB and STREXB.

The software must use a Load-Exclusive instruction with the corresponding Store-Exclusive instruction.

To perform an exclusive read-modify-write of a memory location, the software must:

1. Use a Load-Exclusive instruction to read the value of the location.
2. Update the value, as required.
3. Use a Store-Exclusive instruction to attempt to write the new value back to the memory location
4. Test the returned status bit. If this bit is:

0: The read-modify-write completed successfully.

1: No write was performed. This indicates that the value returned at step 1 might be out of date. The software must retry the read-modify-write sequence.

The software can use the synchronization primitives to implement a semaphore as follows:

1. Use a Load-Exclusive instruction to read from the semaphore address to check whether the semaphore is free.
2. If the semaphore is free, use a Store-Exclusive instruction to write the claim value to the semaphore address.
3. If the returned status bit from step 2 indicates that the Store-Exclusive instruction succeeded then the software has claimed the semaphore. However, if the Store-Exclusive instruction failed, another process might have claimed the semaphore after the software performed the first step.

The Cortex-M4 includes an exclusive access monitor, that tags the fact that the processor has executed a Load-Exclusive instruction. If the processor is part of a multiprocessor system, the system also globally tags the memory locations addressed by exclusive accesses by each processor.

The processor removes its exclusive access tag if:

- It executes a CLREX instruction
- It executes a Store-Exclusive instruction, regardless of whether the write succeeds.
- An exception occurs. This means that the processor can resolve semaphore conflicts between different threads.

In a multiprocessor implementation:

- Executing a CLREX instruction removes only the local exclusive access tag for the processor
- Executing a Store-Exclusive instruction, or an exception, removes the local exclusive access tags, and all global exclusive access tags for the processor.

For more information about the synchronization primitive instructions, see [Section 12.6.4.8 "LDREX and STREX"](#) and [Section 12.6.4.9 "CLREX"](#).

#### 12.4.2.8 Programming Hints for the Synchronization Primitives

ISO/IEC C cannot directly generate the exclusive access instructions. CMSIS provides intrinsic functions for generation of these instructions:

**Table 12-8. CMSIS Functions for Exclusive Access Instructions**

Instruction	CMSIS Function
LDREX	uint32_t __LDREXW (uint32_t *addr)
LDREXH	uint16_t __LDREXH (uint16_t *addr)
LDREXB	uint8_t __LDREXB (uint8_t *addr)
STREX	uint32_t __STREXW (uint32_t value, uint32_t *addr)
STREXH	uint32_t __STREXH (uint16_t value, uint16_t *addr)
STREXB	uint32_t __STREXB (uint8_t value, uint8_t *addr)
CLREX	void __CLREX (void)

The actual exclusive access instruction generated depends on the data type of the pointer passed to the intrinsic function. For example, the following C code generates the required LDREXB operation:

```
__ldrex((volatile char *) 0xFF);
```



### 12.4.3 Exception Model

This section describes the exception model.

#### 12.4.3.1 Exception States

Each exception is in one of the following states:

##### *Inactive*

The exception is not active and not pending.

##### *Pending*

The exception is waiting to be serviced by the processor.

An interrupt request from a peripheral or from software can change the state of the corresponding interrupt to pending.

##### *Active*

An exception is being serviced by the processor but has not completed.

An exception handler can interrupt the execution of another exception handler. In this case, both exceptions are in the active state.

##### *Active and Pending*

The exception is being serviced by the processor and there is a pending exception from the same source.

#### 12.4.3.2 Exception Types

The exception types are:

##### *Reset*

Reset is invoked on power up or a warm reset. The exception model treats reset as a special form of exception. When reset is asserted, the operation of the processor stops, potentially at any point in an instruction. When reset is deasserted, execution restarts from the address provided by the reset entry in the vector table. Execution restarts as privileged execution in Thread mode.

##### *Non Maskable Interrupt (NMI)*

A non maskable interrupt (NMI) can be signalled by a peripheral or triggered by software. This is the highest priority exception other than reset. It is permanently enabled and has a fixed priority of -2.

NMIs cannot be:

- Masked or prevented from activation by any other exception.
- Preempted by any exception other than Reset.

##### *Hard Fault*

A hard fault is an exception that occurs because of an error during exception processing, or because an exception cannot be managed by any other exception mechanism. Hard Faults have a fixed priority of -1, meaning they have higher priority than any exception with configurable priority.

##### *Memory Management Fault (MemManage)*

A Memory Management Fault is an exception that occurs because of a memory protection related fault. The MPU or the fixed memory protection constraints determines this fault, for both instruction and data memory transactions. This fault is used to abort instruction accesses to *Execute Never* (XN) memory regions, even if the MPU is disabled.

##### *Bus Fault*

A Bus Fault is an exception that occurs because of a memory related fault for an instruction or data memory transaction. This might be from an error detected on a bus in the memory system.

##### *Usage Fault*

A Usage Fault is an exception that occurs because of a fault related to an instruction execution. This includes:

- An undefined instruction

- An illegal unaligned access
- An invalid state on instruction execution
- An error on exception return.

The following can cause a Usage Fault when the core is configured to report them:

- An unaligned address on word and halfword memory access
- A division by zero.

### SVCall

A *supervisor call* (SVC) is an exception that is triggered by the SVC instruction. In an OS environment, applications can use SVC instructions to access OS kernel functions and device drivers.

### PendSV

PendSV is an interrupt-driven request for system-level service. In an OS environment, use PendSV for context switching when no other exception is active.

### SysTick

A SysTick exception is an exception the system timer generates when it reaches zero. Software can also generate a SysTick exception. In an OS environment, the processor can use this exception as system tick.

### Interrupt (IRQ)

An interrupt, or IRQ, is an exception signalled by a peripheral, or generated by a software request. All interrupts are asynchronous to instruction execution. In the system, peripherals use interrupts to communicate with the processor.

**Table 12-9. Properties of the Different Exception Types**

Exception Number <sup>(1)</sup>	Irq Number <sup>(1)</sup>	Exception Type	Priority	Vector Address or Offset <sup>(2)</sup>	Activation
1	–	Reset	-3, the highest	0x00000004	Asynchronous
2	-14	NMI	-2	0x00000008	Asynchronous
3	-13	Hard fault	-1	0x0000000C	–
4	-12	Memory management fault	Configurable <sup>(3)</sup>	0x00000010	Synchronous
5	-11	Bus fault	Configurable <sup>(3)</sup>	0x00000014	Synchronous when precise, asynchronous when imprecise
6	-10	Usage fault	Configurable <sup>(3)</sup>	0x00000018	Synchronous
7-10	-	–	–	Reserved	–
11	-5	SVCall	Configurable <sup>(3)</sup>	0x0000002C	Synchronous
12-13	-	–	–	Reserved	–
14	-2	PendSV	Configurable <sup>(3)</sup>	0x00000038	Asynchronous
15	-1	SysTick	Configurable <sup>(3)</sup>	0x0000003C	Asynchronous
16 and above	0 and above	Interrupt (IRQ)	Configurable <sup>(4)</sup>	0x00000040 and above <sup>(5)</sup>	Asynchronous

Notes: 1. To simplify the software layer, the CMSIS only uses IRQ numbers and therefore uses negative values for exceptions other than interrupts. The IPSR returns the Exception number, see [Section 12.4.1.10 "Interrupt Program Status Register"](#).

2. See [Section 12.4.3.4 "Vector Table"](#) for more information

3. See [Section 12.9.1.8 "System Handler Priority Registers"](#)

4. See [Section 12.8.3.6 "Interrupt Priority Registers"](#)

5. Increasing in steps of 4.

For an asynchronous exception, other than reset, the processor can execute another instruction between when the exception is triggered and when the processor enters the exception handler.

Privileged software can disable the exceptions that [Table 12-9](#) shows as having configurable priority, see:

- [Section 12.9.1.12 "System Handler Control and State Register"](#)
- [Section 12.8.3.2 on page 206](#).

For more information about hard faults, memory management faults, bus faults, and usage faults, see [Section 12.4.3.8 "Fault Handling"](#).

### 12.4.3.3 Exception Handlers

The processor handles exceptions using:

- **Interrupt Service Routines (ISRs)**  
Interrupts IRQ0 to IRQ37 are the exceptions handled by ISRs.
- **Fault Handlers**  
Hard fault, memory management fault, usage fault, bus fault are fault exceptions handled by the fault handlers.
- **System Handlers**  
NMI, PendSV, SVCall SysTick, and the fault exceptions are all system exceptions that are handled by system handlers.

### 12.4.3.4 Vector Table

The vector table contains the reset value of the stack pointer, and the start addresses, also called exception vectors, for all exception handlers. [Figure 12-6](#) shows the order of the exception vectors in the vector table. The least-significant bit of each vector must be 1, indicating that the exception handler is Thumb code.

**Figure 12-6. Vector Table**

Exception number	IRQ number	Offset	Vector
255	239	0x03FC	IRQ239
.		.	.
.		.	.
.		.	.
18	2	0x004C	IRQ2
17	1	0x0048	IRQ1
16	0	0x0044	IRQ0
15	-1	0x0040	SysTick
14	-2	0x003C	PendSV
13		0x0038	Reserved
12			Reserved for Debug
11	-5	0x002C	SVCall
10			Reserved
9			
8			
7			
6	-10	0x0018	Usage fault
5	-11	0x0014	Bus fault
4	-12	0x0010	Memory management fault
3	-13	0x000C	Hard fault
2	-14	0x0008	NMI
1		0x0004	Reset
		0x0000	Initial SP value

On system reset, the vector table is fixed at address 0x00000000. Privileged software can write to the SCB\_VTOR register to relocate the vector table start address to a different memory location, in the range 0x00000080 to 0x3FFFFFF80, see [Section 12.9.1.4 "Vector Table Offset Register"](#).

#### 12.4.3.5 Exception Priorities

As [Table 12-9](#) shows, all exceptions have an associated priority, with:

- A lower priority value indicating a higher priority
- Configurable priorities for all exceptions except Reset, Hard fault and NMI.

If the software does not configure any priorities, then all exceptions with a configurable priority have a priority of 0. For information about configuring exception priorities see [Section 12.9.1.8 "System Handler Priority Registers"](#), and [Section 12.8.3.6 "Interrupt Priority Registers"](#).

**Note:** Configurable priority values are in the range 0-15. This means that the Reset, Hard fault, and NMI exceptions, with fixed negative priority values, always have higher priority than any other exception.

For example, assigning a higher priority value to IRQ[0] and a lower priority value to IRQ[1] means that IRQ[1] has higher priority than IRQ[0]. If both IRQ[1] and IRQ[0] are asserted, IRQ[1] is processed before IRQ[0].

If multiple pending exceptions have the same priority, the pending exception with the lowest exception number takes precedence. For example, if both IRQ[0] and IRQ[1] are pending and have the same priority, then IRQ[0] is processed before IRQ[1].

When the processor is executing an exception handler, the exception handler is preempted if a higher priority exception occurs. If an exception occurs with the same priority as the exception being handled, the handler is not preempted, irrespective of the exception number. However, the status of the new interrupt changes to pending.

#### 12.4.3.6 Interrupt Priority Grouping

To increase priority control in systems with interrupts, the NVIC supports priority grouping. This divides each interrupt priority register entry into two fields:

- An upper field that defines the *group priority*
- A lower field that defines a *subpriority* within the group.

Only the group priority determines preemption of interrupt exceptions. When the processor is executing an interrupt exception handler, another interrupt with the same group priority as the interrupt being handled does not preempt the handler.

If multiple pending interrupts have the same group priority, the subpriority field determines the order in which they are processed. If multiple pending interrupts have the same group priority and subpriority, the interrupt with the lowest IRQ number is processed first.

For information about splitting the interrupt priority fields into group priority and subpriority, see [Section 12.9.1.5 "Application Interrupt and Reset Control Register"](#).

#### 12.4.3.7 Exception Entry and Return

Descriptions of exception handling use the following terms:

##### *Preemption*

When the processor is executing an exception handler, an exception can preempt the exception handler if its priority is higher than the priority of the exception being handled. See ["Interrupt Priority Grouping"](#) for more information about preemption by an interrupt.

When one exception preempts another, the exceptions are called nested exceptions. See [Section "Exception Entry"](#) for more information.

##### *Return*

This occurs when the exception handler is completed, and:

- There is no pending exception with sufficient priority to be serviced

- The completed exception handler was not handling a late-arriving exception.

The processor pops the stack and restores the processor state to the state it had before the interrupt occurred. See subsection “[Exception Return](#)” for more information.

#### *Tail-chaining*

This mechanism speeds up exception servicing. On completion of an exception handler, if there is a pending exception that meets the requirements for exception entry, the stack pop is skipped and control transfers to the new exception handler.

#### *Late-arriving*

This mechanism speeds up preemption. If a higher priority exception occurs during state saving for a previous exception, the processor switches to handle the higher priority exception and initiates the vector fetch for that exception. State saving is not affected by late arrival because the state saved is the same for both exceptions. Therefore the state saving continues uninterrupted. The processor can accept a late arriving exception until the first instruction of the exception handler of the original exception enters the execute stage of the processor. On return from the exception handler of the late-arriving exception, the normal tail-chaining rules apply.

#### *Exception Entry*

An Exception entry occurs when there is a pending exception with sufficient priority and either the processor is in Thread mode, or the new exception is of a higher priority than the exception being handled, in which case the new exception preempts the original exception.

When one exception preempts another, the exceptions are nested.

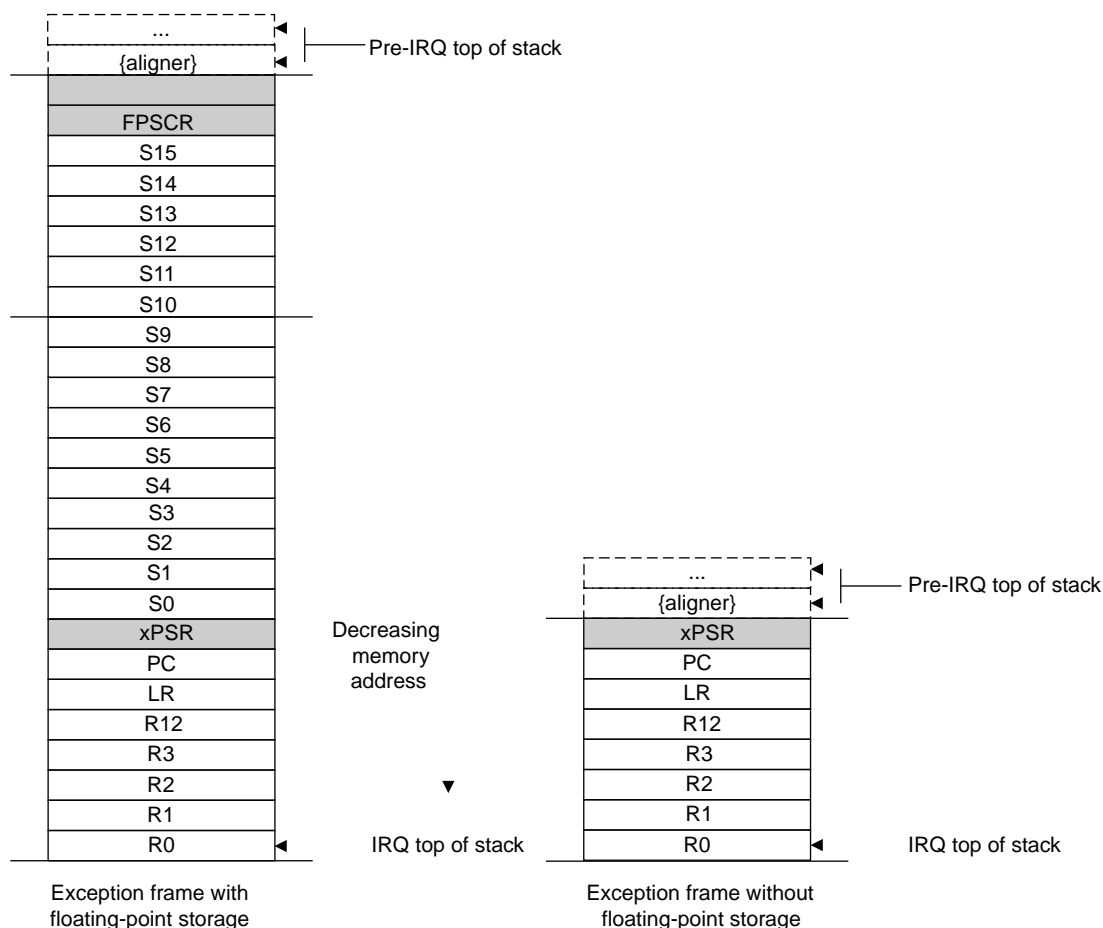
Sufficient priority means that the exception has more priority than any limits set by the mask registers, see [Section 12.4.1.12 “Exception Mask Registers”](#). An exception with less priority than this is pending but is not handled by the processor.

When the processor takes an exception, unless the exception is a tail-chained or a late-arriving exception, the processor pushes information onto the current stack. This operation is referred as *stacking* and the structure of eight data words is referred to as *stack frame*.

When using floating-point routines, the Cortex-M4 processor automatically stacks the architected floating-point state on exception entry. Figure 2-3 on page 2-27 shows the Cortex-M4 stack frame layout when floating-point state is preserved on the stack as the result of an interrupt or an exception.

**Note:** Where stack space for floating-point state is not allocated, the stack frame is the same as that of ARMv7-M implementations without an FPU. Figure 2-3 on page 2-27 shows this stack frame also.

**Figure 12-7. Exception Stack Frame**



Immediately after stacking, the stack pointer indicates the lowest address in the stack frame. The alignment of the stack frame is controlled via the STKALIGN bit of the Configuration Control Register (CCR).

The stack frame includes the return address. This is the address of the next instruction in the interrupted program. This value is restored to the PC at exception return so that the interrupted program resumes.

In parallel to the stacking operation, the processor performs a vector fetch that reads the exception handler start address from the vector table. When stacking is complete, the processor starts executing the exception handler. At the same time, the processor writes an EXC\_RETURN value to the LR. This indicates which stack pointer corresponds to the stack frame and what operation mode the processor was in before the entry occurred.

If no higher priority exception occurs during the exception entry, the processor starts executing the exception handler and automatically changes the status of the corresponding pending interrupt to active.

If another higher priority exception occurs during the exception entry, the processor starts executing the exception handler for this exception and does not change the pending status of the earlier exception. This is the late arrival case.

### Exception Return

An Exception return occurs when the processor is in Handler mode and executes one of the following instructions to load the EXC\_RETURN value into the PC:

- An LDM or POP instruction that loads the PC
- An LDR instruction with the PC as the destination.
- A BX instruction using any register.

EXC\_RETURN is the value loaded into the LR on exception entry. The exception mechanism relies on this value to detect when the processor has completed an exception handler. The lowest five bits of this value provide information on the return stack and processor mode. [Table 12-10](#) shows the EXC\_RETURN values with a description of the exception return behavior.

All EXC\_RETURN values have bits[31:5] set to one. When this value is loaded into the PC, it indicates to the processor that the exception is complete, and the processor initiates the appropriate exception return sequence.

**Table 12-10. Exception Return Behavior**

EXC_RETURN[31:0]	Description
0xFFFFFFFF1	Return to Handler mode, exception return uses non-floating-point state from the MSP and execution uses MSP after return.
0xFFFFFFFF9	Return to Thread mode, exception return uses state from MSP and execution uses MSP after return.
0xFFFFFFFFD	Return to Thread mode, exception return uses state from the PSP and execution uses PSP after return.
0xFFFFFEE1	Return to Handler mode, exception return uses floating-point-state from MSP and execution uses MSP after return.
0xFFFFFEE9	Return to Thread mode, exception return uses floating-point state from MSP and execution uses MSP after return.
0xFFFFFED	Return to Thread mode, exception return uses floating-point state from PSP and execution uses PSP after return.

#### 12.4.3.8 Fault Handling

Faults are a subset of the exceptions, see [Section 12.4.3 "Exception Model"](#). The following generate a fault:

- A bus error on:
  - An instruction fetch or vector table load
  - A data access
- An internally-detected error such as an undefined instruction
- An attempt to execute an instruction from a memory region marked as *Non-Executable* (XN).
- A privilege violation or an attempt to access an unmanaged region causing an MPU fault.

#### Fault Types

[Table 12-11](#) shows the types of fault, the handler used for the fault, the corresponding fault status register, and the register bit that indicates that the fault has occurred. See [Section 12.9.1.13 "Configurable Fault Status Register"](#) for more information about the fault status registers.

**Table 12-11. Faults**

Fault	Handler	Bit Name	Fault Status Register
Bus error on a vector read	Hard fault	VECTTBL	<a href="#">Section 12.9.1.15 "Hard Fault Status Register"</a>
Fault escalated to a hard fault		FORCED	
MPU or default memory map mismatch:		-	
on instruction access	Memory management fault	IACCVIOL	<a href="#">"MMFSR: Memory Management Fault Status Subregister"</a>
on data access		DACCVIOL <sup>(2)</sup>	
during exception stacking		MSTKERR	
during exception unstacking		MUNSKERR	
during lazy floating-point state preservation		MLSPERR	

Table 12-11. Faults (Continued)

Fault	Handler	Bit Name	Fault Status Register
Bus error:	Bus fault	-	-
during exception stacking		STKERR	"BFSR: Bus Fault Status Subregister"
during exception unstacking		UNSTKERR	
during instruction prefetch		IBUSERR	
during lazy floating-point state preservation		LSPERR	
Precise data bus error		PRECISERR	
Imprecise data bus error	Usage fault	IMPRECISERR	"UFSR: Usage Fault Status Subregister"
Attempt to access a coprocessor		NOCOP	
Undefined instruction		UNDEFINSTR	
Attempt to enter an invalid instruction set state <sup>(1)</sup>		INVSTATE	
Invalid EXC_RETURN value		INVPC	
Illegal unaligned load or store		UNALIGNED	
Divide By 0		DIVBYZERO	

Notes: 1. Occurs on an access to an XN region even if the processor does not include an MPU or the MPU is disabled.  
2. Attempt to use an instruction set other than the Thumb instruction set, or return to a non load/store-multiple instruction with ICI continuation.

#### Fault Escalation and Hard Faults

All faults exceptions except for hard fault have configurable exception priority, see [Section 12.9.1.8 "System Handler Priority Registers"](#). The software can disable the execution of the handlers for these faults, see [Section 12.9.1.12 "System Handler Control and State Register"](#).

Usually, the exception priority, together with the values of the exception mask registers, determines whether the processor enters the fault handler, and whether a fault handler can preempt another fault handler, as described in [Section 12.4.3 "Exception Model"](#).

In some situations, a fault with configurable priority is treated as a hard fault. This is called *priority escalation*, and the fault is described as *escalated to hard fault*. Escalation to hard fault occurs when:

- A fault handler causes the same kind of fault as the one it is servicing. This escalation to hard fault occurs because a fault handler cannot preempt itself; it must have the same priority as the current priority level.
- A fault handler causes a fault with the same or lower priority as the fault it is servicing. This is because the handler for the new fault cannot preempt the currently executing fault handler.
- An exception handler causes a fault for which the priority is the same as or lower than the currently executing exception.
- A fault occurs and the handler for that fault is not enabled.

If a bus fault occurs during a stack push when entering a bus fault handler, the bus fault does not escalate to a hard fault. This means that if a corrupted stack causes a fault, the fault handler executes even though the stack push for the handler failed. The fault handler operates but the stack contents are corrupted.

Note: Only Reset and NMI can preempt the fixed priority hard fault. A hard fault can preempt any exception other than Reset, NMI, or another hard fault.



## Fault Status Registers and Fault Address Registers

The fault status registers indicate the cause of a fault. For bus faults and memory management faults, the fault address register indicates the address accessed by the operation that caused the fault, as shown in [Table 12-12](#).

**Table 12-12. Fault Status and Fault Address Registers**

Handler	Status Register Name	Address Register Name	Register Description
Hard fault	SCB_HFSR	-	"Hard Fault Status Register"
Memory management fault	MMFSR	SCB_MMFAR	"MMFSR: Memory Management Fault Status Subregister" "MemManage Fault Address Register"
Bus fault	BFSR	SCB_BFAR	"BFSR: Bus Fault Status Subregister" "Bus Fault Address Register"
Usage fault	UFSR	-	"UFSR: Usage Fault Status Subregister"

### Lockup

The processor enters a lockup state if a hard fault occurs when executing the NMI or hard fault handlers. When the processor is in lockup state, it does not execute any instructions. The processor remains in lockup state until either:

- It is reset
- An NMI occurs
- It is halted by a debugger.

Note: If the lockup state occurs from the NMI handler, a subsequent NMI does not cause the processor to leave the lockup state.

## 12.5 Power Management

The Cortex-M4 processor sleep modes reduce the power consumption:

- Sleep mode stops the processor clock
- Deep sleep mode stops the system clock and switches off the PLL and flash memory.

The SLEEPDEEP bit of the SCR selects which sleep mode is used; see [Section 12.9.1.6 "System Control Register"](#).

This section describes the mechanisms for entering sleep mode, and the conditions for waking up from sleep mode.

### 12.5.1 Entering Sleep Mode

This section describes the mechanisms software can use to put the processor into sleep mode.

The system can generate spurious wakeup events, for example a debug operation wakes up the processor. Therefore, the software must be able to put the processor back into sleep mode after such an event. A program might have an idle loop to put the processor back to sleep mode.

#### 12.5.1.1 Wait for Interrupt

The *wait for interrupt* instruction, WFI, causes immediate entry to sleep mode. When the processor executes a WFI instruction it stops executing instructions and enters sleep mode. See [Section 12.6.12.12 "WFI"](#) for more information.

#### 12.5.1.2 Wait for Event

The *wait for event* instruction, WFE, causes entry to sleep mode conditional on the value of an one-bit event register. When the processor executes a WFE instruction, it checks this register:

- If the register is 0, the processor stops executing instructions and enters sleep mode
- If the register is 1, the processor clears the register to 0 and continues executing instructions without entering sleep mode.

See [Section 12.6.12.11 "WFE"](#) for more information.

### 12.5.1.3 Sleep-on-exit

If the SLEEPONEXIT bit of the SCR is set to 1 when the processor completes the execution of an exception handler, it returns to Thread mode and immediately enters sleep mode. Use this mechanism in applications that only require the processor to run when an exception occurs.

## 12.5.2 Wakeup from Sleep Mode

The conditions for the processor to wake up depend on the mechanism that cause it to enter sleep mode.

### 12.5.2.1 Wakeup from WFI or Sleep-on-exit

Normally, the processor wakes up only when it detects an exception with sufficient priority to cause exception entry.

Some embedded systems might have to execute system restore tasks after the processor wakes up, and before it executes an interrupt handler. To achieve this, set the PRIMASK bit to 1 and the FAULTMASK bit to 0. If an interrupt arrives that is enabled and has a higher priority than the current exception priority, the processor wakes up but does not execute the interrupt handler until the processor sets PRIMASK to zero. For more information about PRIMASK and FAULTMASK, see [Section 12.4.1.12 "Exception Mask Registers"](#).

### 12.5.2.2 Wakeup from WFE

The processor wakes up if:

- It detects an exception with sufficient priority to cause an exception entry
- It detects an external event signal. See [Section 12.5.2.3 "External Event Input"](#)
- In a multiprocessor system, another processor in the system executes an SEV instruction.

In addition, if the SEVONPEND bit in the SCR is set to 1, any new pending interrupt triggers an event and wakes up the processor, even if the interrupt is disabled or has insufficient priority to cause an exception entry. For more information about the SCR, see [Section 12.9.1.6 "System Control Register"](#).

### 12.5.2.3 External Event Input

The processor provides an external event input signal. Peripherals can drive this signal, either to wake the processor from WFE, or to set the internal WFE event register to 1 to indicate that the processor must not enter sleep mode on a later WFE instruction. See [Section 12.5.1.2 "Wait for Event"](#) for more information.

## 12.5.3 Power Management Programming Hints

ISO/IEC C cannot directly generate the WFI and WFE instructions. The CMSIS provides the following functions for these instructions:

```
void __WFE(void) // Wait for Event
void __WFI(void) // Wait for Interrupt
```

## 12.6 Cortex-M4 Instruction Set

### 12.6.1 Instruction Set Summary

The processor implements a version of the Thumb instruction set. [Table 12-13](#) lists the supported instructions.

- Angle brackets, <>, enclose alternative forms of the operand
- Braces, {}, enclose optional operands
- The Operands column is not exhaustive
- Op2 is a flexible second operand that can be either a register or a constant
- Most instructions can use an optional condition code suffix.

For more information on the instructions and operands, see the instruction descriptions.

**Table 12-13. Cortex-M4 Instructions**

Mnemonic	Operands	Description	Flags
ADC, ADCS	{Rd,} Rn, Op2	Add with Carry	N,Z,C,V
ADD, ADDS	{Rd,} Rn, Op2	Add	N,Z,C,V
ADD, ADDW	{Rd,} Rn, #imm12	Add	N,Z,C,V
ADR	Rd, label	Load PC-relative address	—
AND, ANDS	{Rd,} Rn, Op2	Logical AND	N,Z,C
ASR, ASRS	Rd, Rm, <Rs #n>	Arithmetic Shift Right	N,Z,C
B	label	Branch	—
BFC	Rd, #lsb, #width	Bit Field Clear	—
BFI	Rd, Rn, #lsb, #width	Bit Field Insert	—
BIC, BICS	{Rd,} Rn, Op2	Bit Clear	N,Z,C
BKPT	#imm	Breakpoint	—
BL	label	Branch with Link	—
BLX	Rm	Branch indirect with Link	—
BX	Rm	Branch indirect	—
CBNZ	Rn, label	Compare and Branch if Non Zero	—
CBZ	Rn, label	Compare and Branch if Zero	—
CLREX	-	Clear Exclusive	—
CLZ	Rd, Rm	Count leading zeros	—
CMN	Rn, Op2	Compare Negative	N,Z,C,V
CMP	Rn, Op2	Compare	N,Z,C,V
CPSID	i	Change Processor State, Disable Interrupts	—
CPSIE	i	Change Processor State, Enable Interrupts	—
DMB	-	Data Memory Barrier	—
DSB	-	Data Synchronization Barrier	—
EOR, EORS	{Rd,} Rn, Op2	Exclusive OR	N,Z,C
ISB	-	Instruction Synchronization Barrier	—
IT	-	If-Then condition block	—
LDM	Rn{!}, reglist	Load Multiple registers, increment after	—

Table 12-13. Cortex-M4 Instructions (Continued)

Mnemonic	Operands	Description	Flags
LDMDB, LDMEA	Rn{!}, reglist	Load Multiple registers, decrement before	–
LDMFD, LDMIA	Rn{!}, reglist	Load Multiple registers, increment after	–
LDR	Rt, [Rn, #offset]	Load Register with word	–
LDRB, LDRBT	Rt, [Rn, #offset]	Load Register with byte	–
LDRD	Rt, Rt2, [Rn, #offset]	Load Register with two bytes	–
LDREX	Rt, [Rn, #offset]	Load Register Exclusive	–
LDREXB	Rt, [Rn]	Load Register Exclusive with byte	–
LDREXH	Rt, [Rn]	Load Register Exclusive with halfword	–
LDRH, LDRHT	Rt, [Rn, #offset]	Load Register with halfword	–
LDRSB, DRSBT	Rt, [Rn, #offset]	Load Register with signed byte	–
LDRSH, LDRSHT	Rt, [Rn, #offset]	Load Register with signed halfword	–
LDRT	Rt, [Rn, #offset]	Load Register with word	–
LSL, LSLS	Rd, Rm, <Rs n>	Logical Shift Left	N,Z,C
LSR, LSRS	Rd, Rm, <Rs n>	Logical Shift Right	N,Z,C
MLA	Rd, Rn, Rm, Ra	Multiply with Accumulate, 32-bit result	–
MLS	Rd, Rn, Rm, Ra	Multiply and Subtract, 32-bit result	–
MOV, MOVs	Rd, Op2	Move	N,Z,C
MOVT	Rd, #imm16	Move Top	–
MOVW, MOV	Rd, #imm16	Move 16-bit constant	N,Z,C
MRS	Rd, spec_reg	Move from special register to general register	–
MSR	spec_reg, Rm	Move from general register to special register	N,Z,C,V
MUL, MULS	{Rd,} Rn, Rm	Multiply, 32-bit result	N,Z
MVN, MVNS	Rd, Op2	Move NOT	N,Z,C
NOP	-	No Operation	–
ORN, ORNS	{Rd,} Rn, Op2	Logical OR NOT	N,Z,C
ORR, ORRS	{Rd,} Rn, Op2	Logical OR	N,Z,C
PKHTB, PKHBT	{Rd,} Rn, Rm, Op2	Pack Halfword	–
POP	reglist	Pop registers from stack	–
PUSH	reglist	Push registers onto stack	–
QADD	{Rd,} Rn, Rm	Saturating double and Add	Q
QADD16	{Rd,} Rn, Rm	Saturating Add 16	–
QADD8	{Rd,} Rn, Rm	Saturating Add 8	–
QASX	{Rd,} Rn, Rm	Saturating Add and Subtract with Exchange	–
QDADD	{Rd,} Rn, Rm	Saturating Add	Q
QDSUB	{Rd,} Rn, Rm	Saturating double and Subtract	Q
QSAX	{Rd,} Rn, Rm	Saturating Subtract and Add with Exchange	–
QSUB	{Rd,} Rn, Rm	Saturating Subtract	Q

Table 12-13. Cortex-M4 Instructions (Continued)

Mnemonic	Operands	Description	Flags
QSUB16	{Rd,} Rn, Rm	Saturating Subtract 16	–
QSUB8	{Rd,} Rn, Rm	Saturating Subtract 8	–
RBIT	Rd, Rn	Reverse Bits	–
REV	Rd, Rn	Reverse byte order in a word	–
REV16	Rd, Rn	Reverse byte order in each halfword	–
REVSH	Rd, Rn	Reverse byte order in bottom halfword and sign extend	–
ROR, RORS	Rd, Rm, <Rs >#n	Rotate Right	N,Z,C
RRX, RRXS	Rd, Rm	Rotate Right with Extend	N,Z,C
RSB, RSBS	{Rd,} Rn, Op2	Reverse Subtract	N,Z,C,V
SADD16	{Rd,} Rn, Rm	Signed Add 16	GE
SADD8	{Rd,} Rn, Rm	Signed Add 8 and Subtract with Exchange	GE
SASX	{Rd,} Rn, Rm	Signed Add	GE
SBC, SBSCS	{Rd,} Rn, Op2	Subtract with Carry	N,Z,C,V
SBFX	Rd, Rn, #lsb, #width	Signed Bit Field Extract	–
SDIV	{Rd,} Rn, Rm	Signed Divide	–
SEL	{Rd,} Rn, Rm	Select bytes	–
SEV	-	Send Event	–
SHADD16	{Rd,} Rn, Rm	Signed Halving Add 16	–
SHADD8	{Rd,} Rn, Rm	Signed Halving Add 8	–
SHASX	{Rd,} Rn, Rm	Signed Halving Add and Subtract with Exchange	–
SHSAX	{Rd,} Rn, Rm	Signed Halving Subtract and Add with Exchange	–
SHSUB16	{Rd,} Rn, Rm	Signed Halving Subtract 16	–
SHSUB8	{Rd,} Rn, Rm	Signed Halving Subtract 8	–
SMLABB, SMLABT, SMLATB, SMLATT	Rd, Rn, Rm, Ra	Signed Multiply Accumulate Long (halfwords)	Q
SMLAD, SMLADX	Rd, Rn, Rm, Ra	Signed Multiply Accumulate Dual	Q
SMLAL	RdLo, RdHi, Rn, Rm	Signed Multiply with Accumulate (32 x 32 + 64), 64-bit result	–
SMLALBB, SMLALBT, SMLALTB, SMLALTT	RdLo, RdHi, Rn, Rm	Signed Multiply Accumulate Long, halfwords	–
SMLALD, SMLALDX	RdLo, RdHi, Rn, Rm	Signed Multiply Accumulate Long Dual	–
SMLAWB, SMLAWT	Rd, Rn, Rm, Ra	Signed Multiply Accumulate, word by halfword	Q
SMLSD	Rd, Rn, Rm, Ra	Signed Multiply Subtract Dual	Q
SMLSLD	RdLo, RdHi, Rn, Rm	Signed Multiply Subtract Long Dual	–
SMMLA	Rd, Rn, Rm, Ra	Signed Most significant word Multiply Accumulate	–
SMMLS, SMMLR	Rd, Rn, Rm, Ra	Signed Most significant word Multiply Subtract	–
SMMUL, SMMULR	{Rd,} Rn, Rm	Signed Most significant word Multiply	–
SMUAD	{Rd,} Rn, Rm	Signed dual Multiply Add	Q

Table 12-13. Cortex-M4 Instructions (Continued)

Mnemonic	Operands	Description	Flags
SMULBB, SMULBT SMULTB, SMULTT	{Rd,} Rn, Rm	Signed Multiply (halfwords)	–
SMULL	RdLo, RdHi, Rn, Rm	Signed Multiply (32 x 32), 64-bit result	–
SMULWB, SMULWT	{Rd,} Rn, Rm	Signed Multiply word by halfword	–
SMUSD, SMUSDX	{Rd,} Rn, Rm	Signed dual Multiply Subtract	–
SSAT	Rd, #n, Rm {,shift #s}	Signed Saturate	Q
SSAT16	Rd, #n, Rm	Signed Saturate 16	Q
SSAX	{Rd,} Rn, Rm	Signed Subtract and Add with Exchange	GE
SSUB16	{Rd,} Rn, Rm	Signed Subtract 16	–
SSUB8	{Rd,} Rn, Rm	Signed Subtract 8	–
STM	Rn{!}, regist	Store Multiple registers, increment after	–
STMDB, STMEA	Rn{!}, regist	Store Multiple registers, decrement before	–
STMTD, STMIA	Rn{!}, regist	Store Multiple registers, increment after	–
STR	Rt, [Rn, #offset]	Store Register word	–
STRB, STRBT	Rt, [Rn, #offset]	Store Register byte	–
STRD	Rt, Rt2, [Rn, #offset]	Store Register two words	–
STREX	Rd, Rt, [Rn, #offset]	Store Register Exclusive	–
STREXB	Rd, Rt, [Rn]	Store Register Exclusive byte	–
STREXH	Rd, Rt, [Rn]	Store Register Exclusive halfword	–
STRH, STRHT	Rt, [Rn, #offset]	Store Register halfword	–
STRT	Rt, [Rn, #offset]	Store Register word	–
SUB, SUBS	{Rd,} Rn, Op2	Subtract	N,Z,C,V
SUB, SUBW	{Rd,} Rn, #imm12	Subtract	N,Z,C,V
SVC	#imm	Supervisor Call	–
SXTAB	{Rd,} Rn, Rm,{,ROR #}	Extend 8 bits to 32 and add	–
SXTAB16	{Rd,} Rn, Rm,{,ROR #}	Dual extend 8 bits to 16 and add	–
SXTAH	{Rd,} Rn, Rm,{,ROR #}	Extend 16 bits to 32 and add	–
SXTB16	{Rd,} Rm {,ROR #n}	Signed Extend Byte 16	–
SXTB	{Rd,} Rm {,ROR #n}	Sign extend a byte	–
SXTH	{Rd,} Rm {,ROR #n}	Sign extend a halfword	–
TBB	[Rn, Rm]	Table Branch Byte	–
TBH	[Rn, Rm, LSL #1]	Table Branch Halfword	–
TEQ	Rn, Op2	Test Equivalence	N,Z,C
TST	Rn, Op2	Test	N,Z,C
UADD16	{Rd,} Rn, Rm	Unsigned Add 16	GE
UADD8	{Rd,} Rn, Rm	Unsigned Add 8	GE
USAX	{Rd,} Rn, Rm	Unsigned Subtract and Add with Exchange	GE

Table 12-13. Cortex-M4 Instructions (Continued)

Mnemonic	Operands	Description	Flags
UHADD16	{Rd}, Rn, Rm	Unsigned Halving Add 16	–
UHADD8	{Rd}, Rn, Rm	Unsigned Halving Add 8	–
UHASX	{Rd}, Rn, Rm	Unsigned Halving Add and Subtract with Exchange	–
UHSAX	{Rd}, Rn, Rm	Unsigned Halving Subtract and Add with Exchange	–
UHSUB16	{Rd}, Rn, Rm	Unsigned Halving Subtract 16	–
UHSUB8	{Rd}, Rn, Rm	Unsigned Halving Subtract 8	–
UBFX	Rd, Rn, #lsb, #width	Unsigned Bit Field Extract	–
UDIV	{Rd}, Rn, Rm	Unsigned Divide	–
UMAAL	RdLo, RdHi, Rn, Rm	Unsigned Multiply Accumulate Accumulate Long (32 x 32 + 32 +32), 64-bit result	–
UMLAL	RdLo, RdHi, Rn, Rm	Unsigned Multiply with Accumulate (32 x 32 + 64), 64-bit result	–
UMULL	RdLo, RdHi, Rn, Rm	Unsigned Multiply (32 x 32), 64-bit result	–
UQADD16	{Rd}, Rn, Rm	Unsigned Saturating Add 16	–
UQADD8	{Rd}, Rn, Rm	Unsigned Saturating Add 8	–
UQASX	{Rd}, Rn, Rm	Unsigned Saturating Add and Subtract with Exchange	–
UQSAX	{Rd}, Rn, Rm	Unsigned Saturating Subtract and Add with Exchange	–
UQSUB16	{Rd}, Rn, Rm	Unsigned Saturating Subtract 16	–
UQSUB8	{Rd}, Rn, Rm	Unsigned Saturating Subtract 8	–
USAD8	{Rd}, Rn, Rm	Unsigned Sum of Absolute Differences	–
USADA8	{Rd}, Rn, Rm, Ra	Unsigned Sum of Absolute Differences and Accumulate	–
USAT	Rd, #n, Rm {,shift #s}	Unsigned Saturate	Q
USAT16	Rd, #n, Rm	Unsigned Saturate 16	Q
UASX	{Rd}, Rn, Rm	Unsigned Add and Subtract with Exchange	GE
USUB16	{Rd}, Rn, Rm	Unsigned Subtract 16	GE
USUB8	{Rd}, Rn, Rm	Unsigned Subtract 8	GE
UXTAB	{Rd}, Rn, Rm,{,ROR #}	Rotate, extend 8 bits to 32 and Add	–
UXTAB16	{Rd}, Rn, Rm,{,ROR #}	Rotate, dual extend 8 bits to 16 and Add	–
UXTAH	{Rd}, Rn, Rm,{,ROR #}	Rotate, unsigned extend and Add Halfword	–
UXTB	{Rd}, Rm {,ROR #n}	Zero extend a byte	–
UXTB16	{Rd}, Rm {,ROR #n}	Unsigned Extend Byte 16	–
UXTH	{Rd}, Rm {,ROR #n}	Zero extend a halfword	–
VABS.F32	Sd, Sm	Floating-point Absolute	–
VADD.F32	{Sd}, Sn, Sm	Floating-point Add	–
VCMP.F32	Sd, <Sm   #0.0>	Compare two floating-point registers, or one floating-point register and zero	FPSCR
VCMPE.F32	Sd, <Sm   #0.0>	Compare two floating-point registers, or one floating-point register and zero with Invalid Operation check	FPSCR

Table 12-13. Cortex-M4 Instructions (Continued)

Mnemonic	Operands	Description	Flags
VCVT.S32.F32	Sd, Sm	Convert between floating-point and integer	—
VCVT.S16.F32	Sd, Sd, #fbits	Convert between floating-point and fixed point	—
VCVTR.S32.F32	Sd, Sm	Convert between floating-point and integer with rounding	—
VCVT<B H>.F32.F16	Sd, Sm	Converts half-precision value to single-precision	—
VCVTT<B T>.F32.F16	Sd, Sm	Converts single-precision register to half-precision	—
VDIV.F32	{Sd,} Sn, Sm	Floating-point Divide	—
VFMA.F32	{Sd,} Sn, Sm	Floating-point Fused Multiply Accumulate	—
VFNMA.F32	{Sd,} Sn, Sm	Floating-point Fused Negate Multiply Accumulate	—
VFMS.F32	{Sd,} Sn, Sm	Floating-point Fused Multiply Subtract	—
VFNMS.F32	{Sd,} Sn, Sm	Floating-point Fused Negate Multiply Subtract	—
VLDM.F<32 64>	Rn{!}, list	Load Multiple extension registers	—
VLDR.F<32 64>	<Dd Sd>, [Rn]	Load an extension register from memory	—
VLMA.F32	{Sd,} Sn, Sm	Floating-point Multiply Accumulate	—
VLMS.F32	{Sd,} Sn, Sm	Floating-point Multiply Subtract	—
VMOV.F32	Sd, #imm	Floating-point Move immediate	—
VMOV	Sd, Sm	Floating-point Move register	—
VMOV	Sn, Rt	Copy ARM core register to single precision	—
VMOV	Sm, Sm1, Rt, Rt2	Copy 2 ARM core registers to 2 single precision	—
VMOV	Dd[x], Rt	Copy ARM core register to scalar	—
VMOV	Rt, Dn[x]	Copy scalar to ARM core register	—
VMRS	Rt, FPSCR	Move FPSCR to ARM core register or APSR	N,Z,C,V
VMSR	FPSCR, Rt	Move to FPSCR from ARM Core register	FPSCR
VMUL.F32	{Sd,} Sn, Sm	Floating-point Multiply	—
VNEG.F32	Sd, Sm	Floating-point Negate	—
VNMLA.F32	Sd, Sn, Sm	Floating-point Multiply and Add	—
VNMLS.F32	Sd, Sn, Sm	Floating-point Multiply and Subtract	—
VNMUL	{Sd,} Sn, Sm	Floating-point Multiply	—
VPOP	list	Pop extension registers	—
VPUSH	list	Push extension registers	—
VSQRT.F32	Sd, Sm	Calculates floating-point Square Root	—
VSTM	Rn{!}, list	Floating-point register Store Multiple	—
VSTR.F<32 64>	Sd, [Rn]	Stores an extension register to memory	—
VSUB.F<32 64>	{Sd,} Sn, Sm	Floating-point Subtract	—
WFE	—	Wait For Event	—
WFI	—	Wait For Interrupt	—



## 12.6.2 CMSIS Functions

ISO/IEC cannot directly access some Cortex-M4 instructions. This section describes intrinsic functions that can generate these instructions, provided by the CMSIS and that might be provided by a C compiler. If a C compiler does not support an appropriate intrinsic function, the user might have to use inline assembler to access some instructions.

The CMSIS provides the following intrinsic functions to generate instructions that ISO/IEC C code cannot directly access:

**Table 12-14. CMSIS Functions to Generate some Cortex-M4 Instructions**

Instruction	CMSIS Function
CPSIE I	void __enable_irq(void)
CPSID I	void __disable_irq(void)
CPSIE F	void __enable_fault_irq(void)
CPSID F	void __disable_fault_irq(void)
ISB	void __ISB(void)
DSB	void __DSB(void)
DMB	void __DMB(void)
REV	uint32_t __REV(uint32_t int value)
REV16	uint32_t __REV16(uint32_t int value)
REVSH	uint32_t __REVSH(uint32_t int value)
RBIT	uint32_t __RBIT(uint32_t int value)
SEV	void __SEV(void)
WFE	void __WFE(void)
WFI	void __WFI(void)

The CMSIS also provides a number of functions for accessing the special registers using MRS and MSR instructions:

**Table 12-15. CMSIS Intrinsic Functions to Access the Special Registers**

Special Register	Access	CMSIS Function
PRIMASK	Read	uint32_t __get_PRIMASK (void)
	Write	void __set_PRIMASK (uint32_t value)
FAULTMASK	Read	uint32_t __get_FAULTMASK (void)
	Write	void __set_FAULTMASK (uint32_t value)
BASEPRI	Read	uint32_t __get_BASEPRI (void)
	Write	void __set_BASEPRI (uint32_t value)
CONTROL	Read	uint32_t __get_CONTROL (void)
	Write	void __set_CONTROL (uint32_t value)
MSP	Read	uint32_t __get_MSP (void)
	Write	void __set_MSP (uint32_t TopOfMainStack)
PSP	Read	uint32_t __get_PSP (void)
	Write	void __set_PSP (uint32_t TopOfProcStack)

## 12.6.3 Instruction Descriptions

### 12.6.3.1 Operands

An instruction operand can be an ARM register, a constant, or another instruction-specific parameter. Instructions act on the operands and often store the result in a destination register. When there is a destination register in the instruction, it is usually specified before the operands.

Operands in some instructions are flexible, can either be a register or a constant. See [Section 12.6.3.3 "Flexible Second Operand"](#).

### 12.6.3.2 Restrictions when Using PC or SP

Many instructions have restrictions on whether the *Program Counter* (PC) or *Stack Pointer* (SP) for the operands or destination register can be used. See instruction descriptions for more information.

Note: Bit[0] of any address written to the PC with a BX, BLX, LDM, LDR, or POP instruction must be 1 for correct execution, because this bit indicates the required instruction set, and the Cortex-M4 processor only supports Thumb instructions.

### 12.6.3.3 Flexible Second Operand

Many general data processing instructions have a flexible second operand. This is shown as *Operand2* in the descriptions of the syntax of each instruction.

*Operand2* can be a:

- "Constant"
- "Register with Optional Shift"

#### Constant

Specify an *Operand2* constant in the form:

`#constant`

where *constant* can be:

- Any constant that can be produced by shifting an 8-bit value left by any number of bits within a 32-bit word
- Any constant of the form 0x00XY00XY
- Any constant of the form 0xXY00XY00
- Any constant of the form 0xXYXYXYXY.

Note: In the constants shown above, X and Y are hexadecimal digits.

In addition, in a small number of instructions, *constant* can take a wider range of values. These are described in the individual instruction descriptions.

When an *Operand2* constant is used with the instructions MOVs, MVNS, ANDS, ORRS, ORNS, EORS, BICS, TEQ or TST, the carry flag is updated to bit[31] of the constant, if the constant is greater than 255 and can be produced by shifting an 8-bit value. These instructions do not affect the carry flag if *Operand2* is any other constant.

#### Instruction Substitution

The assembler might be able to produce an equivalent instruction in cases where the user specifies a constant that is not permitted. For example, an assembler might assemble the instruction `CMP Rd, #0xFFFFFFFFE` as the equivalent instruction `CMN Rd, #0x2`.

#### Register with Optional Shift

Specify an *Operand2* register in the form:

`Rm {, shift}`

where:

Rm is the register holding the data for the second operand.

shift is an optional shift to be applied to *Rm*. It can be one of:

ASR # <i>n</i>	arithmetic shift right <i>n</i> bits, $1 \leq n \leq 32$ .
LSL # <i>n</i>	logical shift left <i>n</i> bits, $1 \leq n \leq 31$ .
LSR # <i>n</i>	logical shift right <i>n</i> bits, $1 \leq n \leq 32$ .
ROR # <i>n</i>	rotate right <i>n</i> bits, $1 \leq n \leq 31$ .
RRX	rotate right one bit, with extend.
-	if omitted, no shift occurs, equivalent to LSL #0.

If the user omits the shift, or specifies LSL #0, the instruction uses the value in *Rm*.

If the user specifies a shift, the shift is applied to the value in *Rm*, and the resulting 32-bit value is used by the instruction. However, the contents in the register *Rm* remains unchanged. Specifying a register with shift also updates the carry flag when used with certain instructions. For information on the shift operations and how they affect the carry flag, see [Section 12.6.3.3 "Flexible Second Operand"](#)

#### 12.6.3.4 Shift Operations

Register shift operations move the bits in a register left or right by a specified number of bits, the *shift length*. Register shift can be performed:

- Directly by the instructions ASR, LSR, LSL, ROR, and RRX, and the result is written to a destination register
- During the calculation of *Operand2* by the instructions that specify the second operand as a register with shift. See [Section 12.6.3.3 "Flexible Second Operand"](#). The result is used by the instruction.

The permitted shift lengths depend on the shift type and the instruction. If the shift length is 0, no shift occurs. Register shift operations update the carry flag except when the specified shift length is 0. The following sub-sections describe the various shift operations and how they affect the carry flag. In these descriptions, *Rm* is the register containing the value to be shifted, and *n* is the shift length.

##### ASR

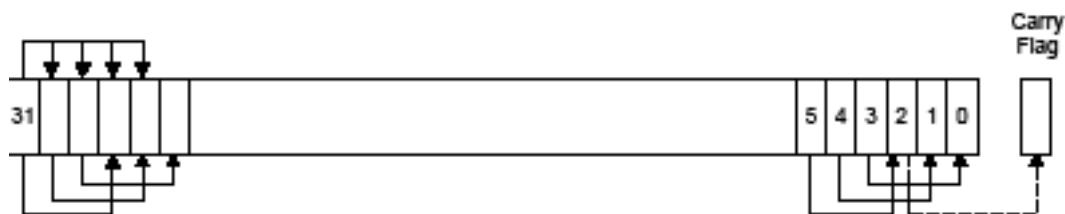
Arithmetic shift right by *n* bits moves the left-hand 32-*n* bits of the register, *Rm*, to the right by *n* places, into the right-hand 32-*n* bits of the result. And it copies the original bit[31] of the register into the left-hand *n* bits of the result. See [Figure 12-8](#).

The ASR #*n* operation can be used to divide the value in the register *Rm* by  $2^n$ , with the result being rounded towards negative-infinity.

When the instruction is ASRS or when ASR #*n* is used in *Operand2* with the instructions MOVS, MVNS, ANDS, ORRS, ORNS, EORS, BICS, TEQ or TST, the carry flag is updated to the last bit shifted out, bit[*n*-1], of the register *Rm*.

- If *n* is 32 or more, then all the bits in the result are set to the value of bit[31] of *Rm*.
- If *n* is 32 or more and the carry flag is updated, it is updated to the value of bit[31] of *Rm*.

Figure 12-8. ASR #3



##### LSR

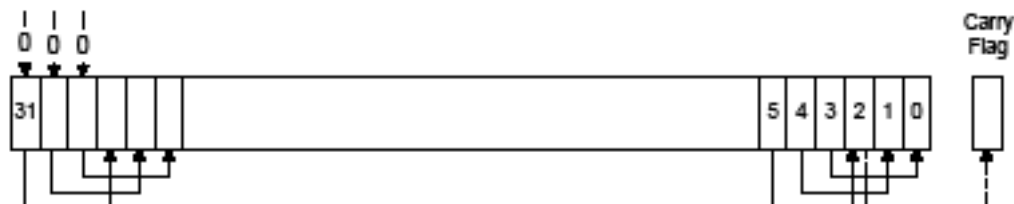
Logical shift right by *n* bits moves the left-hand 32-*n* bits of the register *Rm*, to the right by *n* places, into the right-hand 32-*n* bits of the result. And it sets the left-hand *n* bits of the result to 0. See [Figure 12-9](#).

The LSR #*n* operation can be used to divide the value in the register *Rm* by  $2^n$ , if the value is regarded as an unsigned integer.

When the instruction is LSRS or when LSR #*n* is used in *Operand2* with the instructions MOV<sub>S</sub>, MVNS, AND<sub>S</sub>, ORR<sub>S</sub>, ORN<sub>S</sub>, EOR<sub>S</sub>, BIC<sub>S</sub>, TEQ or TST, the carry flag is updated to the last bit shifted out, bit[*n*-1], of the register *Rm*.

- If *n* is 32 or more, then all the bits in the result are cleared to 0.
- If *n* is 33 or more and the carry flag is updated, it is updated to 0.

**Figure 12-9. LSR #3**



## LSL

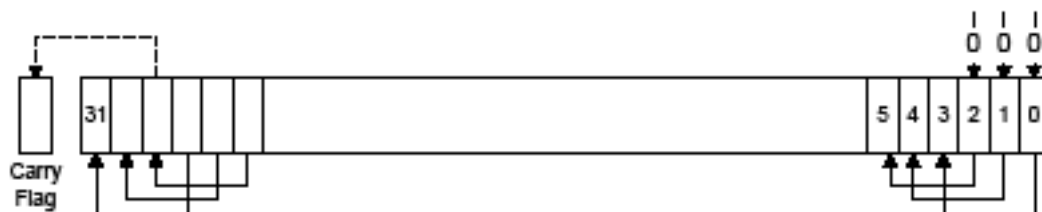
Logical shift left by *n* bits moves the right-hand 32-*n* bits of the register *Rm*, to the left by *n* places, into the left-hand 32-*n* bits of the result; and it sets the right-hand *n* bits of the result to 0. See [Figure 12-10](#).

The LSL #*n* operation can be used to multiply the value in the register *Rm* by  $2^n$ , if the value is regarded as an unsigned integer or a two's complement signed integer. Overflow can occur without warning.

When the instruction is LSLS or when LSL #*n*, with non-zero *n*, is used in *Operand2* with the instructions MOV<sub>S</sub>, MVNS, AND<sub>S</sub>, ORR<sub>S</sub>, ORN<sub>S</sub>, EOR<sub>S</sub>, BIC<sub>S</sub>, TEQ or TST, the carry flag is updated to the last bit shifted out, bit[32-*n*], of the register *Rm*. These instructions do not affect the carry flag when used with LSL #0.

- If *n* is 32 or more, then all the bits in the result are cleared to 0.
- If *n* is 33 or more and the carry flag is updated, it is updated to 0.

**Figure 12-10. LSL #3**



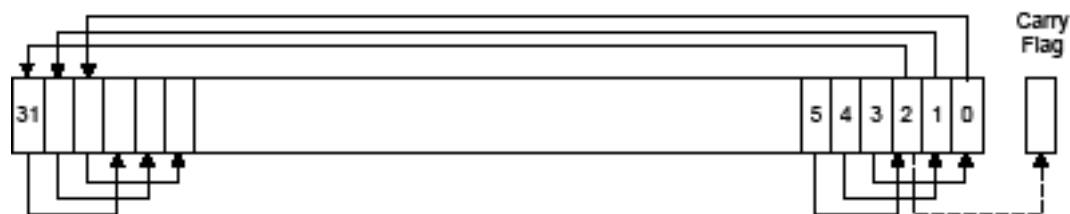
## ROR

Rotate right by *n* bits moves the left-hand 32-*n* bits of the register *Rm*, to the right by *n* places, into the right-hand 32-*n* bits of the result; and it moves the right-hand *n* bits of the register into the left-hand *n* bits of the result. See [Figure 12-11](#).

When the instruction is RORS or when ROR #*n* is used in *Operand2* with the instructions MOV<sub>S</sub>, MVNS, AND<sub>S</sub>, ORR<sub>S</sub>, ORN<sub>S</sub>, EOR<sub>S</sub>, BIC<sub>S</sub>, TEQ or TST, the carry flag is updated to the last bit rotation, bit[*n*-1], of the register *Rm*.

- If *n* is 32, then the value of the result is same as the value in *Rm*, and if the carry flag is updated, it is updated to bit[31] of *Rm*.
- ROR with shift length, *n*, more than 32 is the same as ROR with shift length *n*-32.

Figure 12-11. ROR #3

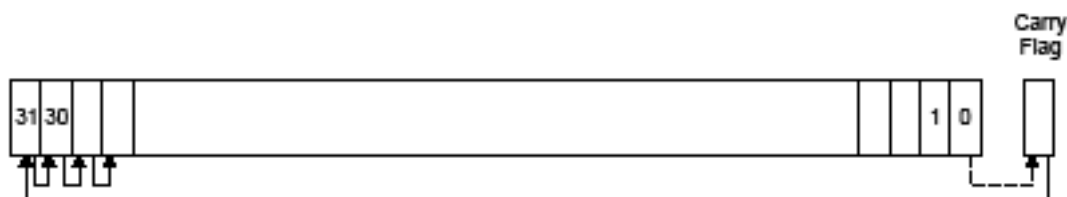


## RRX

Rotate right with extend moves the bits of the register *Rm* to the right by one bit; and it copies the carry flag into bit[31] of the result. See [Figure 12-12](#).

When the instruction is RRXS or when RRX is used in *Operand2* with the instructions MOVS, MVNS, ANDS, ORRS, ORNS, EORS, BICS, TEQ or TST, the carry flag is updated to bit[0] of the register *Rm*.

Figure 12-12. RRX



### 12.6.3.5 Address Alignment

An aligned access is an operation where a word-aligned address is used for a word, dual word, or multiple word access, or where a halfword-aligned address is used for a halfword access. Byte accesses are always aligned.

The Cortex-M4 processor supports unaligned access only for the following instructions:

- LDR, LDRT
- LDRH, LDRHT
- LDRSH, LDRSHT
- STR, STRT
- STRH, STRHT

All other load and store instructions generate a usage fault exception if they perform an unaligned access, and therefore their accesses must be address-aligned. For more information about usage faults, see [Section 12.4.3.8 "Fault Handling"](#).

Unaligned accesses are usually slower than aligned accesses. In addition, some memory regions might not support unaligned accesses. Therefore, ARM recommends that programmers ensure that accesses are aligned. To avoid accidental generation of unaligned accesses, use the UNALIGN\_TRP bit in the Configuration and Control Register to trap all unaligned accesses, see [Section 12.9.1.7 on page 222](#).

### 12.6.3.6 PC-relative Expressions

A PC-relative expression or *label* is a symbol that represents the address of an instruction or literal data. It is represented in the instruction as the PC value plus or minus a numeric offset. The assembler calculates the required offset from the label and the address of the current instruction. If the offset is too big, the assembler produces an error.

- For B, BL, CBNZ, and CBZ instructions, the value of the PC is the address of the current instruction plus 4 bytes.
- For all other instructions that use labels, the value of the PC is the address of the current instruction plus 4 bytes, with bit[1] of the result cleared to 0 to make it word-aligned.
- Your assembler might permit other syntaxes for PC-relative expressions, such as a label plus or minus a number, or an expression of the form [PC, #number].

### 12.6.3.7 Conditional Execution

Most data processing instructions can optionally update the condition flags in the *Application Program Status Register* (APSR) according to the result of the operation, see [Section 12.4.1.9 "Application Program Status Register"](#). Some instructions update all flags, and some only update a subset. If a flag is not updated, the original value is preserved. See the instruction descriptions for the flags they affect.

An instruction can be executed conditionally, based on the condition flags set in another instruction, either:

- Immediately after the instruction that updated the flags
- After any number of intervening instructions that have not updated the flags.

Conditional execution is available by using conditional branches or by adding condition code suffixes to instructions. See [Table 12-16](#) for a list of the suffixes to add to instructions to make them conditional instructions. The condition code suffix enables the processor to test a condition based on the flags. If the condition test of a conditional instruction fails, the instruction:

- Does not execute
- Does not write any value to its destination register
- Does not affect any of the flags
- Does not generate any exception.

Conditional instructions, except for conditional branches, must be inside an If-Then instruction block. See ["IT"](#) for more information and restrictions when using the IT instruction. Depending on the vendor, the assembler might automatically insert an IT instruction if there are conditional instructions outside the IT block.

The CBZ and CBNZ instructions are used to compare the value of a register against zero and branch on the result.

This section describes:

- ["Condition Flags"](#)
- ["Condition Code Suffixes"](#) .

#### Condition Flags

The APSR contains the following condition flags:

N	Set to 1 when the result of the operation was negative, cleared to 0 otherwise.
Z	Set to 1 when the result of the operation was zero, cleared to 0 otherwise.
C	Set to 1 when the operation resulted in a carry, cleared to 0 otherwise.
V	Set to 1 when the operation caused overflow, cleared to 0 otherwise.

For more information about the APSR, see [Section 12.4.1.8 "Program Status Register"](#).

A carry occurs:

- If the result of an addition is greater than or equal to  $2^{32}$
- If the result of a subtraction is positive or zero
- As the result of an inline barrel shifter operation in a move or logical instruction.

An overflow occurs when the sign of the result, in bit[31], does not match the sign of the result, had the operation been performed at infinite precision, for example:

- If adding two negative values results in a positive value
- If adding two positive values results in a negative value
- If subtracting a positive value from a negative value generates a positive value
- If subtracting a negative value from a positive value generates a negative value.

The Compare operations are identical to subtracting, for CMP, or adding, for CMN, except that the result is discarded. See the instruction descriptions for more information.

**Note:** Most instructions update the status flags only if the S suffix is specified. See the instruction descriptions for more information.

## Condition Code Suffixes

The instructions that can be conditional have an optional condition code, shown in syntax descriptions as {cond}. Conditional execution requires a preceding IT instruction. An instruction with a condition code is only executed if the condition code flags in the APSR meet the specified condition. Table 12-16 shows the condition codes to use.

A conditional execution can be used with the IT instruction to reduce the number of branch instructions in code.

Table 12-16 also shows the relationship between condition code suffixes and the N, Z, C, and V flags.

**Table 12-16. Condition Code Suffixes**

Suffix	Flags	Meaning
EQ	Z = 1	Equal
NE	Z = 0	Not equal
CS or HS	C = 1	Higher or same, unsigned $\geq$
CC or LO	C = 0	Lower, unsigned $<$
MI	N = 1	Negative
PL	N = 0	Positive or zero
VS	V = 1	Overflow
VC	V = 0	No overflow
HI	C = 1 and Z = 0	Higher, unsigned $>$
LS	C = 0 or Z = 1	Lower or same, unsigned $\leq$
GE	N = V	Greater than or equal, signed $\geq$
LT	N $\neq$ V	Less than, signed $<$
GT	Z = 0 and N = V	Greater than, signed $>$
LE	Z = 1 and N $\neq$ V	Less than or equal, signed $\leq$
AL	Can have any value	Always. This is the default when no suffix is specified.

### Absolute Value

The example below shows the use of a conditional instruction to find the absolute value of a number. R0 = ABS(R1).

```
MOVS    R0, R1        ; R0 = R1, setting flags
IT      MI            ; IT instruction for the negative condition
RSBMI   R0, R1, #0     ; If negative, R0 = -R1
```

### Compare and Update Value

The example below shows the use of conditional instructions to update the value of R4 if the signed values R0 is greater than R1 and R2 is greater than R3.

```
CMP      R0, R1        ; Compare R0 and R1, setting flags
ITT      GT            ; IT instruction for the two GT conditions
CMPGT    R2, R3        ; If 'greater than', compare R2 and R3, setting flags
MOVGT    R4, R5        ; If still 'greater than', do R4 = R5
```

#### 12.6.3.8 Instruction Width Selection

There are many instructions that can generate either a 16-bit encoding or a 32-bit encoding depending on the operands and destination register specified. For some of these instructions, the user can force a specific instruction size by using an instruction width suffix. The .W suffix forces a 32-bit instruction encoding. The .N suffix forces a 16-bit instruction encoding.

If the user specifies an instruction width suffix and the assembler cannot generate an instruction encoding of the requested width, it generates an error.

**Note:** In some cases, it might be necessary to specify the .W suffix, for example if the operand is the label of an instruction or literal data, as in the case of branch instructions. This is because the assembler might not automatically generate the right size encoding.

To use an instruction width suffix, place it immediately after the instruction mnemonic and condition code, if any. The example below shows instructions with the instruction width suffix.

```
BCS.W label      ; creates a 32-bit instruction even for a short
                  ; branch
ADDS.W R0, R0, R1 ; creates a 32-bit instruction even though the same
                  ; operation can be done by a 16-bit instruction
```

#### 12.6.4 Memory Access Instructions

The table below shows the memory access instructions:

**Table 12-17. Memory Access Instructions**

Mnemonic	Description
ADR	Load PC-relative address
CLREX	Clear Exclusive
LDM{mode}	Load Multiple registers
LDR{type}	Load Register using immediate offset
LDR{type}	Load Register using register offset
LDR{type}T	Load Register with unprivileged access
LDR	Load Register using PC-relative address
LDRD	Load Register Dual
LDREX{type}	Load Register Exclusive
POP	Pop registers from stack
PUSH	Push registers onto stack
STM{mode}	Store Multiple registers
STR{type}	Store Register using immediate offset
STR{type}	Store Register using register offset
STR{type}T	Store Register with unprivileged access
STREX{type}	Store Register Exclusive



### 12.6.4.1 ADR

Load PC-relative address.

Syntax

`ADR{cond} Rd, label`

where:

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*label* is a PC-relative expression. See [“PC-relative Expressions”](#).

Operation

ADR determines the address by adding an immediate value to the PC, and writes the result to the destination register.

ADR produces position-independent code, because the address is PC-relative.

If ADR is used to generate a target address for a BX or BLX instruction, ensure that bit[0] of the address generated is set to 1 for correct execution.

Values of *label* must be within the range of –4095 to +4095 from the address in the PC.

Note: The user might have to use the .W suffix to get the maximum offset range or to generate addresses that are not word-aligned. See [“Instruction Width Selection”](#).

Restrictions

*Rd* must not be SP and must not be PC.

Condition Flags

This instruction does not change the flags.

Examples

```
ADR    R1, TextMessage    ; Write address value of a location labelled as
                           ; TextMessage to R1
```

#### 12.6.4.2 LDR and STR, Immediate Offset

Load and Store with immediate offset, pre-indexed immediate offset, or post-indexed immediate offset.

Syntax

<code>op{type}{cond} Rt, [Rn {, #offset}]</code>	<code>; immediate offset</code>
<code>op{type}{cond} Rt, [Rn, #offset]!</code>	<code>; pre-indexed</code>
<code>op{type}{cond} Rt, [Rn], #offset</code>	<code>; post-indexed</code>
<code>opD{cond} Rt, Rt2, [Rn {, #offset}]</code>	<code>; immediate offset, two words</code>
<code>opD{cond} Rt, Rt2, [Rn, #offset]!</code>	<code>; pre-indexed, two words</code>
<code>opD{cond} Rt, Rt2, [Rn], #offset</code>	<code>; post-indexed, two words</code>

where:

op is one of:

LDR Load Register.

STR Store Register.

type is one of:

B unsigned byte, zero extend to 32 bits on loads.

SB signed byte, sign extend to 32 bits (LDR only).

H unsigned halfword, zero extend to 32 bits on loads.

SH signed halfword, sign extend to 32 bits (LDR only).

- omit, for word.

cond is an optional condition code, see [“Conditional Execution”](#).

Rt is the register to load or store.

Rn is the register on which the memory address is based.

offset is an offset from *Rn*. If *offset* is omitted, the address is the contents of *Rn*.

Rt2 is the additional register to load or store for two-word operations.

Operation

LDR instructions load one or two registers with a value from memory.

STR instructions store one or two register values to memory.

Load and store instructions with immediate offset can use the following addressing modes:

Offset Addressing

The offset value is added to or subtracted from the address obtained from the register *Rn*. The result is used as the address for the memory access. The register *Rn* is unaltered. The assembly language syntax for this mode is:

`[Rn, #offset]`

**Pre-indexed Addressing**

The offset value is added to or subtracted from the address obtained from the register *Rn*. The result is used as the address for the memory access and written back into the register *Rn*. The assembly language syntax for this mode is:

`[Rn, #offset]!`

**Post-indexed Addressing**

The address obtained from the register *Rn* is used as the address for the memory access. The offset value is added to or subtracted from the address, and written back into the register *Rn*. The assembly language syntax for this mode is:

`[Rn], #offset`

The value to load or store can be a byte, halfword, word, or two words. Bytes and halfwords can either be signed or unsigned. See [“Address Alignment”](#).

Table 12-18 below shows the ranges of offset for immediate, pre-indexed and post-indexed forms.

**Table 12-18. Offset Ranges**

Instruction Type	Immediate Offset	Pre-indexed	Post-indexed
Word, halfword, signed halfword, byte, or signed byte	-255 to 4095	-255 to 255	-255 to 255
Two words	multiple of 4 in the range -1020 to 1020	multiple of 4 in the range -1020 to 1020	multiple of 4 in the range -1020 to 1020

#### Restrictions

For load instructions:

- *Rt* can be SP or PC for word loads only
- *Rt* must be different from *Rt2* for two-word loads
- *Rn* must be different from *Rt* and *Rt2* in the pre-indexed or post-indexed forms.

When *Rt* is PC in a word load instruction:

- Bit[0] of the loaded value must be 1 for correct execution
- A branch occurs to the address created by changing bit[0] of the loaded value to 0
- If the instruction is conditional, it must be the last instruction in the IT block.

For store instructions:

- *Rt* can be SP for word stores only
- *Rt* must not be PC
- *Rn* must not be PC
- *Rn* must be different from *Rt* and *Rt2* in the pre-indexed or post-indexed forms.

#### Condition Flags

These instructions do not change the flags.

## Examples

```
LDR      R8, [R10]           ; Loads R8 from the address in R10.
LDRNE    R2, [R5, #960]!    ; Loads (conditionally) R2 from a word
                             ; 960 bytes above the address in R5, and
                             ; increments R5 by 960.

STR      R2, [R9, #const-struct] ; const-struct is an expression evaluating
                             ; to a constant in the range 0-4095.
STRH     R3, [R4], #4        ; Store R3 as halfword data into address in
                             ; R4, then increment R4 by 4
LDRD     R8, R9, [R3, #0x20] ; Load R8 from a word 32 bytes above the
                             ; address in R3, and load R9 from a word 36
                             ; bytes above the address in R3
STRD     R0, R1, [R8], #-16   ; Store R0 to address in R8, and store R1 to
                             ; a word 4 bytes above the address in R8,
                             ; and then decrement R8 by 16.
```

### 12.6.4.3 LDR and STR, Register Offset

Load and Store with register offset.

Syntax

*op*{*type*}{*cond*} *Rt*, [*Rn*, *Rm* {, LSL #*n*}]

where:

*op* is one of:

LDR Load Register.

STR Store Register.

*type* is one of:

B unsigned byte, zero extend to 32 bits on loads.

SB signed byte, sign extend to 32 bits (LDR only).

H unsigned halfword, zero extend to 32 bits on loads.

SH signed halfword, sign extend to 32 bits (LDR only).

- omit, for word.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rt* is the register to load or store.

*Rn* is the register on which the memory address is based.

*Rm* is a register containing a value to be used as the offset.

LSL #*n* is an optional shift, with *n* in the range 0 to 3.

Operation

LDR instructions load a register with a value from memory.

STR instructions store a register value into memory.

The memory address to load from or store to is at an offset from the register *Rn*. The offset is specified by the register *Rm* and can be shifted left by up to 3 bits using LSL.

The value to load or store can be a byte, halfword, or word. For load instructions, bytes and halfwords can either be signed or unsigned. See [“Address Alignment”](#).

Restrictions

In these instructions:

- *Rn* must not be PC
- *Rm* must not be SP and must not be PC

- *Rt* can be SP only for word loads and word stores
- *Rt* can be PC only for word loads.

When *Rt* is PC in a word load instruction:

- Bit[0] of the loaded value must be 1 for correct execution, and a branch occurs to this halfword-aligned address
- If the instruction is conditional, it must be the last instruction in the IT block.

#### Condition Flags

These instructions do not change the flags.

#### Examples

```
STR    R0, [R5, R1]           ; Store value of R0 into an address equal to
                                ; sum of R5 and R1
LDRSB  R0, [R5, R1, LSL #1]    ; Read byte value from an address equal to
                                ; sum of R5 and two times R1, sign extended it
                                ; to a word value and put it in R0
STR    R0, [R1, R2, LSL #2]    ; Stores R0 to an address equal to sum of R1
                                ; and four times R2
```

#### 12.6.4.4 LDR and STR, Unprivileged

Load and Store with unprivileged access.

Syntax

```
op{type}T{cond} Rt, [Rn {, #offset}] ; immediate offset
```

where:

op is one of:

LDR Load Register.

STR Store Register.

type is one of:

B unsigned byte, zero extend to 32 bits on loads.

SB signed byte, sign extend to 32 bits (LDR only).

H unsigned halfword, zero extend to 32 bits on loads.

SH signed halfword, sign extend to 32 bits (LDR only).

- omit, for word.

cond is an optional condition code, see [“Conditional Execution”](#).

Rt is the register to load or store.

Rn is the register on which the memory address is based.

offset is an offset from *Rn* and can be 0 to 255.

If *offset* is omitted, the address is the value in *Rn*.

Operation

These load and store instructions perform the same function as the memory access instructions with immediate offset, see [“LDR and STR, Immediate Offset”](#). The difference is that these instructions have only unprivileged access even when used in privileged software.

When used in unprivileged software, these instructions behave in exactly the same way as normal memory access instructions with immediate offset.

Restrictions

In these instructions:

- *Rn* must not be PC
- *Rt* must not be SP and must not be PC.

Condition Flags

These instructions do not change the flags.

Examples

```
STRBTEQ R4, [R7] ; Conditionally store least significant byte in  
; R4 to an address in R7, with unprivileged access  
LDRHT R2, [R2, #8] ; Load halfword value from an address equal to  
; sum of R2 and 8 into R2, with unprivileged access
```

#### 12.6.4.5 LDR, PC-relative

Load register from memory.

Syntax

```
LDR{type}{cond} Rt, label
LDRD{cond} Rt, Rt2, label      ; Load two words
```

where:

type is one of:

- B unsigned byte, zero extend to 32 bits.
- SB signed byte, sign extend to 32 bits.
- H unsigned halfword, zero extend to 32 bits.
- SH signed halfword, sign extend to 32 bits.
- omit, for word.

cond is an optional condition code, see [“Conditional Execution”](#).

Rt is the register to load or store.

Rt2 is the second register to load or store.

label is a PC-relative expression. See [“PC-relative Expressions”](#).

Operation

LDR loads a register with a value from a PC-relative memory address. The memory address is specified by a label or by an offset from the PC.

The value to load or store can be a byte, halfword, or word. For load instructions, bytes and halfwords can either be signed or unsigned. See [“Address Alignment”](#).

*label* must be within a limited range of the current instruction. The table below shows the possible offsets between *label* and the PC.

**Table 12-19. Offset Ranges**

Instruction Type	Offset Range
Word, halfword, signed halfword, byte, signed byte	-4095 to 4095
Two words	-1020 to 1020

The user might have to use the .W suffix to get the maximum offset range. See [“Instruction Width Selection”](#).

Restrictions

In these instructions:

- Rt can be SP or PC only for word loads
- Rt2 must not be SP and must not be PC
- Rt must be different from Rt2.

When *Rt* is PC in a word load instruction:

- Bit[0] of the loaded value must be 1 for correct execution, and a branch occurs to this halfword-aligned address
- If the instruction is conditional, it must be the last instruction in the IT block.

#### Condition Flags

These instructions do not change the flags.

#### Examples

```
LDR      R0, LookUpTable    ; Load R0 with a word of data from an address
                                ; labelled as LookUpTable
LDRSB    R7, localdata      ; Load a byte value from an address labelled
                                ; as localdata, sign extend it to a word
                                ; value, and put it in R7
```

### 12.6.4.6 LDM and STM

Load and Store Multiple registers.

#### Syntax

*op*{*addr\_mode*}{*cond*} *Rn*{!}, *reglist*

where:

*op* is one of:

LDM Load Multiple registers.

STM Store Multiple registers.

*addr\_mode* is any one of the following:

IA Increment address After each access. This is the default.

DB Decrement address Before each access.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rn* is the register on which the memory addresses are based.

! is an optional writeback suffix.

If ! is present, the final address, that is loaded from or stored to, is written back into *Rn*.

*reglist* is a list of one or more registers to be loaded or stored, enclosed in braces. It can contain register ranges. It must be comma separated if it contains more than one register or register range, see [“Examples”](#).

LDM and LDMFD are synonyms for LDMIA. LDMFD refers to its use for popping data from Full Descending stacks.

LDMEA is a synonym for LDMDB, and refers to its use for popping data from Empty Ascending stacks.

STM and STMEA are synonyms for STMIA. STMEA refers to its use for pushing data onto Empty Ascending stacks.

STMFD is a synonym for STMDB, and refers to its use for pushing data onto Full Descending stacks

#### Operation

LDM instructions load the registers in *reglist* with word values from memory addresses based on *Rn*.

STM instructions store the word values in the registers in *reglist* to memory addresses based on *Rn*.

For LDM, LDMIA, LDMFD, STM, STMIA, and STMEA the memory addresses used for the accesses are at 4-byte intervals ranging from *Rn* to *Rn* + 4 \* (*n*-1), where *n* is the number of registers in *reglist*. The accesses happen in order of increasing register numbers, with the lowest numbered register using the lowest memory address and the highest number register using the highest memory address. If the writeback suffix is specified, the value of *Rn* + 4 \* (*n*-1) is written back to *Rn*.

For LDMDB, LDMEA, STMDB, and STMFD the memory addresses used for the accesses are at 4-byte intervals ranging from *Rn* to *Rn* - 4 \* (*n*-1), where *n* is the number of registers in *reglist*. The accesses happen in order of decreasing



register numbers, with the highest numbered register using the highest memory address and the lowest number register using the lowest memory address. If the writeback suffix is specified, the value of  $Rn - 4 * (n-1)$  is written back to  $Rn$ .

The PUSH and POP instructions can be expressed in this form. See “[PUSH and POP](#)” for details.

#### Restrictions

In these instructions:

- $Rn$  must not be PC
- *reglist* must not contain SP
- In any STM instruction, *reglist* must not contain PC
- In any LDM instruction, *reglist* must not contain PC if it contains LR
- *reglist* must not contain  $Rn$  if the writeback suffix is specified.

When PC is in *reglist* in an LDM instruction:

- Bit[0] of the value loaded to the PC must be 1 for correct execution, and a branch occurs to this halfword-aligned address
- If the instruction is conditional, it must be the last instruction in the IT block.

#### Condition Flags

These instructions do not change the flags.

#### Examples

```
LDM      R8, {R0,R2,R9}      ; LDMIA is a synonym for LDM
STMDB    R1!, {R3-R6,R11,R12}
```

#### Incorrect Examples

```
STM      R5!, {R5,R4,R9} ; Value stored for R5 is unpredictable
LDM      R2, {}           ; There must be at least one register in the list
```

#### 12.6.4.7 PUSH and POP

Push registers onto, and pop registers off a full-descending stack.

Syntax

```
PUSH{cond} reglist
POP{cond} reglist
```

where:

cond is an optional condition code, see “[Conditional Execution](#)”.

reglist is a non-empty list of registers, enclosed in braces. It can contain register ranges. It must be comma separated if it contains more than one register or register range.

PUSH and POP are synonyms for STMDB and LDM (or LDMIA) with the memory addresses for the access based on SP, and with the final address for the access written back to the SP. PUSH and POP are the preferred mnemonics in these cases.

Operation

PUSH stores registers on the stack in order of decreasing the register numbers, with the highest numbered register using the highest memory address and the lowest numbered register using the lowest memory address.

POP loads registers from the stack in order of increasing register numbers, with the lowest numbered register using the lowest memory address and the highest numbered register using the highest memory address.

See “[LDM and STM](#)” for more information.

Restrictions

In these instructions:

- *reglist* must not contain SP
- For the PUSH instruction, *reglist* must not contain PC
- For the POP instruction, *reglist* must not contain PC if it contains LR.

When PC is in *reglist* in a POP instruction:

- Bit[0] of the value loaded to the PC must be 1 for correct execution, and a branch occurs to this halfword-aligned address
- If the instruction is conditional, it must be the last instruction in the IT block.

Condition Flags

These instructions do not change the flags.

Examples

```
PUSH    {R0,R4-R7}
PUSH    {R2,LR}
POP     {R0,R10,PC}
```

#### 12.6.4.8 LDREX and STREX

Load and Store Register Exclusive.

Syntax

```
LDREX{cond} Rt, [Rn {, #offset}]
STREX{cond} Rd, Rt, [Rn {, #offset}]
LDREXB{cond} Rt, [Rn]
STREXB{cond} Rd, Rt, [Rn]
LDREXH{cond} Rt, [Rn]
STREXH{cond} Rd, Rt, [Rn]
```

where:

**cond** is an optional condition code, see [“Conditional Execution”](#).

**Rd** is the destination register for the returned status.

**Rt** is the register to load or store.

**Rn** is the register on which the memory address is based.

**offset** is an optional offset applied to the value in *Rn*.  
If *offset* is omitted, the address is the value in *Rn*.

Operation

LDREX, LDREXB, and LDREXH load a word, byte, and halfword respectively from a memory address.

STREX, STREXB, and STREXH attempt to store a word, byte, and halfword respectively to a memory address. The address used in any Store-Exclusive instruction must be the same as the address in the most recently executed Load-exclusive instruction. The value stored by the Store-Exclusive instruction must also have the same data size as the value loaded by the preceding Load-exclusive instruction. This means software must always use a Load-exclusive instruction and a matching Store-Exclusive instruction to perform a synchronization operation, see [“Synchronization Primitives”](#).

If an Store-Exclusive instruction performs the store, it writes 0 to its destination register. If it does not perform the store, it writes 1 to its destination register. If the Store-Exclusive instruction writes 0 to the destination register, it is guaranteed that no other process in the system has accessed the memory location between the Load-exclusive and Store-Exclusive instructions.

For reasons of performance, keep the number of instructions between corresponding Load-Exclusive and Store-Exclusive instruction to a minimum.

The result of executing a Store-Exclusive instruction to an address that is different from that used in the preceding Load-Exclusive instruction is unpredictable.

Restrictions

In these instructions:

- Do not use PC
- Do not use SP for *Rd* and *Rt*
- For STREX, *Rd* must be different from both *Rt* and *Rn*
- The value of *offset* must be a multiple of four in the range 0-1020.

Condition Flags

These instructions do not change the flags.

Examples

```
MOV      R1, #0x1           ; Initialize the 'lock taken' value try
LDREX    R0, [LockAddr]     ; Load the lock value
CMP      R0, #0             ; Is the lock free?
ITT      EQ                 ; IT instruction for STREXEQ and CMPEQ
STREXEQ  R0, R1, [LockAddr] ; Try and claim the lock
CMPEQ    R0, #0             ; Did this succeed?
BNE      try                ; No - try again
....      ; Yes - we have the lock
```

#### 12.6.4.9 CLREX

Clear Exclusive.

Syntax

CLREX{*cond*}

where:

*cond* is an optional condition code, see “Conditional Execution”.

Operation

Use CLREX to make the next STREX, STREXB, or STREXH instruction write 1 to its destination register and fail to perform the store. It is useful in exception handler code to force the failure of the store exclusive if the exception occurs between a load exclusive instruction and the matching store exclusive instruction in a synchronization operation.

See “Synchronization Primitives” for more information.

Condition Flags

These instructions do not change the flags.

Examples

CLREX

#### 12.6.5 General Data Processing Instructions

The table below shows the data processing instructions:

**Table 12-20. Data Processing Instructions**

Mnemonic	Description
ADC	Add with Carry
ADD	Add
ADDW	Add
AND	Logical AND
ASR	Arithmetic Shift Right
BIC	Bit Clear
CLZ	Count leading zeros
CMN	Compare Negative
CMP	Compare
EOR	Exclusive OR
LSL	Logical Shift Left
LSR	Logical Shift Right
MOV	Move
MOVT	Move Top
MOVW	Move 16-bit constant
MVN	Move NOT
ORN	Logical OR NOT
ORR	Logical OR
RBIT	Reverse Bits
REV	Reverse byte order in a word

**Table 12-20. Data Processing Instructions (Continued)**

<b>Mnemonic</b>	<b>Description</b>
REV16	Reverse byte order in each halfword
REVSH	Reverse byte order in bottom halfword and sign extend
ROR	Rotate Right
RRX	Rotate Right with Extend
RSB	Reverse Subtract
SADD16	Signed Add 16
SADD8	Signed Add 8
SASX	Signed Add and Subtract with Exchange
SSAX	Signed Subtract and Add with Exchange
SBC	Subtract with Carry
SHADD16	Signed Halving Add 16
SHADD8	Signed Halving Add 8
SHASX	Signed Halving Add and Subtract with Exchange
SHSAX	Signed Halving Subtract and Add with Exchange
SHSUB16	Signed Halving Subtract 16
SHSUB8	Signed Halving Subtract 8
SSUB16	Signed Subtract 16
SSUB8	Signed Subtract 8
SUB	Subtract
SUBW	Subtract
TEQ	Test Equivalence
TST	Test
UADD16	Unsigned Add 16
UADD8	Unsigned Add 8
UASX	Unsigned Add and Subtract with Exchange
USAX	Unsigned Subtract and Add with Exchange
UHADD16	Unsigned Halving Add 16
UHADD8	Unsigned Halving Add 8
UHASX	Unsigned Halving Add and Subtract with Exchange
UHSAX	Unsigned Halving Subtract and Add with Exchange
UHSUB16	Unsigned Halving Subtract 16
UHSUB8	Unsigned Halving Subtract 8
USAD8	Unsigned Sum of Absolute Differences
USADA8	Unsigned Sum of Absolute Differences and Accumulate
USUB16	Unsigned Subtract 16
USUB8	Unsigned Subtract 8

### 12.6.5.1 ADD, ADC, SUB, SBC, and RSB

Add, Add with carry, Subtract, Subtract with carry, and Reverse Subtract.

Syntax

```
op{S}{cond} {Rd,} Rn, Operand2
op{cond} {Rd,} Rn, #imm12           ; ADD and SUB only
```

where:

op is one of:

ADD Add.

ADC Add with Carry.

SUB Subtract.

SBC Subtract with Carry.

RSB Reverse Subtract.

S is an optional suffix. If S is specified, the condition code flags are updated on the result of the operation, see [“Conditional Execution”](#).

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register. If Rd is omitted, the destination register is Rn.

Rn is the register holding the first operand.

Operand2 is a flexible second operand. See [“Flexible Second Operand”](#) for details of the options.

imm12 is any value in the range 0-4095.

Operation

The ADD instruction adds the value of *Operand2* or *imm12* to the value in *Rn*.

The ADC instruction adds the values in *Rn* and *Operand2*, together with the carry flag.

The SUB instruction subtracts the value of *Operand2* or *imm12* from the value in *Rn*.

The SBC instruction subtracts the value of *Operand2* from the value in *Rn*. If the carry flag is clear, the result is reduced by one.

The RSB instruction subtracts the value in *Rn* from the value of *Operand2*. This is useful because of the wide range of options for *Operand2*.

Use ADC and SBC to synthesize multiword arithmetic, see *Multiword arithmetic examples* on.

See also [“ADR”](#).

Note: ADDW is equivalent to the ADD syntax that uses the *imm12* operand. SUBW is equivalent to the SUB syntax that uses the *imm12* operand.

Restrictions

In these instructions:

- *Operand2* must not be SP and must not be PC
- *Rd* can be SP only in ADD and SUB, and only with the additional restrictions:
  - *Rn* must also be SP
  - Any shift in *Operand2* must be limited to a maximum of 3 bits using LSL
- *Rn* can be SP only in ADD and SUB
- *Rd* can be PC only in the ADD{cond} PC, PC, Rm instruction where:
  - The user must not specify the S suffix
  - *Rm* must not be PC and must not be SP
  - If the instruction is conditional, it must be the last instruction in the IT block

- With the exception of the ADD{cond} PC, PC, Rm instruction, *Rn* can be PC only in ADD and SUB, and only with the additional restrictions:
  - The user must not specify the S suffix
  - The second operand must be a constant in the range 0 to 4095.
  - Note: When using the PC for an addition or a subtraction, bits[1:0] of the PC are rounded to 0b00 before performing the calculation, making the base address for the calculation word-aligned.
  - Note: To generate the address of an instruction, the constant based on the value of the PC must be adjusted. ARM recommends to use the ADR instruction instead of ADD or SUB with *Rn* equal to the PC, because the assembler automatically calculates the correct constant for the ADR instruction.

When *Rd* is PC in the ADD{cond} PC, PC, Rm instruction:

- Bit[0] of the value written to the PC is ignored
- A branch occurs to the address created by forcing bit[0] of that value to 0.

#### Condition Flags

If S is specified, these instructions update the N, Z, C and V flags according to the result.

#### Examples

```
ADD    R2, R1, R3           ; Sets the flags on the result
SUBS   R8, R6, #240         ; Subtracts contents of R4 from 1280
RSB    R4, R4, #1280        ; Only executed if C flag set and Z
ADCHI  R11, R0, R3          ; flag clear.
```

#### Multiword Arithmetic Examples

The example below shows two instructions that add a 64-bit integer contained in R2 and R3 to another 64-bit integer contained in R0 and R1, and place the result in R4 and R5.

#### 64-bit Addition Example

```
ADDS   R4, R0, R2          ; add the least significant words
ADC     R5, R1, R3          ; add the most significant words with carry
```

Multiword values do not have to use consecutive registers. The example below shows instructions that subtract a 96-bit integer contained in R9, R1, and R11 from another contained in R6, R2, and R8. The example stores the result in R6, R9, and R2.

#### 96-bit Subtraction Example

```
SUBS   R6, R6, R9          ; subtract the least significant words
SBCS   R9, R2, R1          ; subtract the middle words with carry
SBC     R2, R8, R11        ; subtract the most significant words with carry
```

### 12.6.5.2 AND, ORR, EOR, BIC, and ORN

Logical AND, OR, Exclusive OR, Bit Clear, and OR NOT.

Syntax

*op*{*S*}{*cond*} {*Rd*,} *Rn*, *Operand2*

where:

*op* is one of:

AND logical AND.

ORR logical OR, or bit set.

EOR logical Exclusive OR.

BIC logical AND NOT, or bit clear.

ORN logical OR NOT.

*S* is an optional suffix. If *S* is specified, the condition code flags are updated on the result of the operation, see [“Conditional Execution”](#).

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn* is the register holding the first operand.

*Operand2* is a flexible second operand. See [“Flexible Second Operand”](#) for details of the options

Operation

The AND, EOR, and ORR instructions perform bitwise AND, Exclusive OR, and OR operations on the values in *Rn* and *Operand2*.

The BIC instruction performs an AND operation on the bits in *Rn* with the complements of the corresponding bits in the value of *Operand2*.

The ORN instruction performs an OR operation on the bits in *Rn* with the complements of the corresponding bits in the value of *Operand2*.

Restrictions

Do not use SP and do not use PC.

Condition Flags

If *S* is specified, these instructions:

- Update the N and Z flags according to the result
- Can update the C flag during the calculation of *Operand2*, see [“Flexible Second Operand”](#)
- Do not affect the V flag.

Examples

```
AND      R9, R2, #0xFF00
ORREQ    R2, R0, R5
ANDS     R9, R8, #0x19
EORS     R7, R11, #0x18181818
BIC      R0, R1, #0xab
ORN      R7, R11, R14, ROR #4
ORNS     R7, R11, R14, ASR #32
```



### 12.6.5.3 ASR, LSL, LSR, ROR, and RRX

Arithmetic Shift Right, Logical Shift Left, Logical Shift Right, Rotate Right, and Rotate Right with Extend.

Syntax

```
op{S}{cond} Rd, Rm, Rs
op{S}{cond} Rd, Rm, #n
RRX{S}{cond} Rd, Rm
```

where:

op is one of:

ASR Arithmetic Shift Right.

LSL Logical Shift Left.

LSR Logical Shift Right.

ROR Rotate Right.

S is an optional suffix. If S is specified, the condition code flags are updated on the result of the operation, see [“Conditional Execution”](#).

Rd is the destination register.

Rm is the register holding the value to be shifted.

Rs is the register holding the shift length to apply to the value in *Rm*. Only the least significant byte is used and can be in the range 0 to 255.

n is the shift length. The range of shift length depends on the instruction:

ASR shift length from 1 to 32

LSL shift length from 0 to 31

LSR shift length from 1 to 32

ROR shift length from 0 to 31

MOVS Rd, Rm is the preferred syntax for LSLS Rd, Rm, #0.

Operation

ASR, LSL, LSR, and ROR move the bits in the register *Rm* to the left or right by the number of places specified by constant *n* or register *Rs*.

RRX moves the bits in register *Rm* to the right by 1.

In all these instructions, the result is written to *Rd*, but the value in register *Rm* remains unchanged. For details on what result is generated by the different instructions, see [“Shift Operations”](#).

Restrictions

Do not use SP and do not use PC.

Condition Flags

If S is specified:

- These instructions update the N and Z flags according to the result
- The C flag is updated to the last bit shifted out, except when the shift length is 0, see [“Shift Operations”](#).

Examples

```
ASR    R7, R8, #9    ; Arithmetic shift right by 9 bits
SLS    R1, R2, #3    ; Logical shift left by 3 bits with flag update
LSR    R4, R5, #6    ; Logical shift right by 6 bits
ROR    R4, R5, R6    ; Rotate right by the value in the bottom byte of R6
RRX    R4, R5        ; Rotate right with extend.
```

#### 12.6.5.4 CLZ

Count Leading Zeros.

Syntax

`CLZ{cond} Rd, Rm`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

`Rd` is the destination register.

`Rm` is the operand register.

Operation

The CLZ instruction counts the number of leading zeros in the value in *Rm* and returns the result in *Rd*. The result value is 32 if no bits are set and zero if bit[31] is set.

Restrictions

Do not use SP and do not use PC.

Condition Flags

This instruction does not change the flags.

Examples

```
CLZ      R4, R9
CLZNE    R2, R3
```

#### 12.6.5.5 CMP and CMN

Compare and Compare Negative.

Syntax

`CMP{cond} Rn, Operand2`

`CMN{cond} Rn, Operand2`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

`Rn` is the register holding the first operand.

`Operand2` is a flexible second operand. See [“Flexible Second Operand”](#) for details of the options

Operation

These instructions compare the value in a register with *Operand2*. They update the condition flags on the result, but do not write the result to a register.

The CMP instruction subtracts the value of *Operand2* from the value in *Rn*. This is the same as a SUBS instruction, except that the result is discarded.

The CMN instruction adds the value of *Operand2* to the value in *Rn*. This is the same as an ADDS instruction, except that the result is discarded.

Restrictions

In these instructions:

- Do not use PC
- *Operand2* must not be SP.

Condition Flags

These instructions update the N, Z, C and V flags according to the result.

Examples

```
CMP      R2, R9
CMN      R0, #6400
CMPGT    SP, R7, LSL #2
```

### 12.6.5.6 MOV and MVN

Move and Move NOT.

Syntax

```
MOV{S}{cond} Rd, Operand2
MOV{cond} Rd, #imm16
MVN{S}{cond} Rd, Operand2
```

where:

S is an optional suffix. If S is specified, the condition code flags are updated on the result of the operation, see [“Conditional Execution”](#).

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Operand2 is a flexible second operand. See [“Flexible Second Operand”](#) for details of the options

imm16 is any value in the range 0-65535.

Operation

The MOV instruction copies the value of *Operand2* into *Rd*.

When *Operand2* in a MOV instruction is a register with a shift other than LSL #0, the preferred syntax is the corresponding shift instruction:

- ASR{S}{cond} Rd, Rm, #n is the preferred syntax for MOV{S}{cond} Rd, Rm, ASR #n
- LSL{S}{cond} Rd, Rm, #n is the preferred syntax for MOV{S}{cond} Rd, Rm, LSL #n if  $n \neq 0$
- LSR{S}{cond} Rd, Rm, #n is the preferred syntax for MOV{S}{cond} Rd, Rm, LSR #n
- ROR{S}{cond} Rd, Rm, #n is the preferred syntax for MOV{S}{cond} Rd, Rm, ROR #n
- RRX{S}{cond} Rd, Rm is the preferred syntax for MOV{S}{cond} Rd, Rm, RRX.

Also, the MOV instruction permits additional forms of *Operand2* as synonyms for shift instructions:

- MOV{S}{cond} Rd, Rm, ASR Rs is a synonym for ASR{S}{cond} Rd, Rm, Rs
- MOV{S}{cond} Rd, Rm, LSL Rs is a synonym for LSL{S}{cond} Rd, Rm, Rs
- MOV{S}{cond} Rd, Rm, LSR Rs is a synonym for LSR{S}{cond} Rd, Rm, Rs
- MOV{S}{cond} Rd, Rm, ROR Rs is a synonym for ROR{S}{cond} Rd, Rm, Rs

See [“ASR, LSL, LSR, ROR, and RRX”](#).

The MVN instruction takes the value of *Operand2*, performs a bitwise logical NOT operation on the value, and places the result into *Rd*.

The MOVW instruction provides the same function as MOV, but is restricted to using the *imm16* operand.

Restrictions

SP and PC only can be used in the MOV instruction, with the following restrictions:

- The second operand must be a register without shift
- The S suffix must not be specified.

When *Rd* is PC in a MOV instruction:

- Bit[0] of the value written to the PC is ignored
- A branch occurs to the address created by forcing bit[0] of that value to 0.

Though it is possible to use MOV as a branch instruction, ARM strongly recommends the use of a BX or BLX instruction to branch for software portability to the ARM instruction set.

Condition Flags

If S is specified, these instructions:

- Update the N and Z flags according to the result

- Can update the C flag during the calculation of *Operand2*, see “Flexible Second Operand”
- Do not affect the V flag.

#### Examples

```

    MOVS    R11, #0x000B           ; Write value of 0x000B to
R11, flags get updated
    MOV     R1, #0xFA05           ; Write value of 0xFA05 to
R1, flags are not updated
    MOVS    R10, R12              ; Write value in R12 to R10,
flags get updated
    MOV     R3, #23               ; Write value of 23 to R3
    MOV     R8, SP                ; Write value of stack pointer to R8
    MVNS    R2, #0xF              ; Write value of 0xFFFFFFFF (bitwise inverse of 0xF)
                                ; to the R2 and update flags.

```

#### 12.6.5.7 MOV<sub>T</sub>

Move Top.

#### Syntax

```
MOVT{cond} Rd, #imm16
```

where:

*cond* is an optional condition code, see “Conditional Execution”.

*Rd* is the destination register.

*imm16* is a 16-bit immediate constant.

#### Operation

MOV<sub>T</sub> writes a 16-bit immediate value, *imm16*, to the top halfword, *Rd*[31:16], of its destination register. The write does not affect *Rd*[15:0].

The MOV, MOV<sub>T</sub> instruction pair enables to generate any 32-bit constant.

#### Restrictions

*Rd* must not be SP and must not be PC.

#### Condition Flags

This instruction does not change the flags.

#### Examples

```

    MOVT    R3, #0xF123 ; Write 0xF123 to upper halfword of R3, lower halfword
                                ; and APSR are unchanged.

```

#### 12.6.5.8 REV, REV16, REVSH, and RBIT

Reverse bytes and Reverse bits.

#### Syntax

```
op{cond} Rd, Rn
```

where:

*op* is any of:

REV Reverse byte order in a word.

REV16 Reverse byte order in each halfword independently.

REVSH Reverse byte order in the bottom halfword, and sign extend to 32 bits.

RBIT Reverse the bit order in a 32-bit word.

*cond* is an optional condition code, see “Conditional Execution”.

Rd is the destination register.  
Rn is the register holding the operand.

#### Operation

Use these instructions to change endianness of data:

REV converts either:

- 32-bit big-endian data into little-endian data
- 32-bit little-endian data into big-endian data.

REV16 converts either:

- 16-bit big-endian data into little-endian data
- 16-bit little-endian data into big-endian data.

REVSH converts either:

- 16-bit signed big-endian data into 32-bit signed little-endian data
- 16-bit signed little-endian data into 32-bit signed big-endian data.

#### Restrictions

Do not use SP and do not use PC.

#### Condition Flags

These instructions do not change the flags.

#### Examples

```
REV    R3, R7; Reverse byte order of value in R7 and write it to R3
REV16  R0, R0; Reverse byte order of each 16-bit halfword in R0
REVSH  R0, R5; Reverse Signed Halfword
REVHS  R3, R7; Reverse with Higher or Same condition
RBIT   R7, R8; Reverse bit order of value in R8 and write the result to R7.
```

### 12.6.5.9 SADD16 and SADD8

Signed Add 16 and Signed Add 8

Syntax

*op*{*cond*}{*Rd*,} *Rn*, *Rm*

where:

*op* is any of:

SADD16 Performs two 16-bit signed integer additions.

SADD8 Performs four 8-bit signed integer additions.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn* is the first register holding the operand.

*Rm* is the second register holding the operand.

Operation

Use these instructions to perform a halfword or byte add in parallel:

The SADD16 instruction:

1. Adds each halfword from the first operand to the corresponding halfword of the second operand.
2. Writes the result in the corresponding halfwords of the destination register.

The SADD8 instruction:

1. Adds each byte of the first operand to the corresponding byte of the second operand.

Writes the result in the corresponding bytes of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
SADD16 R1, R0      ; Adds the halfwords in R0 to the corresponding
                   ; halfwords of R1 and writes to corresponding halfword
                   ; of R1.
SADD8  R4, R0, R5   ; Adds bytes of R0 to the corresponding byte in R5 and
                   ; writes to the corresponding byte in R4.
```

### 12.6.5.10 SHADD16 and SHADD8

Signed Halving Add 16 and Signed Halving Add 8

Syntax

$op\{cond\}\{Rd,\} Rn, Rm$

where:

op is any of:

SHADD16 Signed Halving Add 16.

SHADD8 Signed Halving Add 8.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first operand register.

Rm is the second operand register.

Operation

Use these instructions to add 16-bit and 8-bit data and then to halve the result before writing the result to the destination register:

The SHADD16 instruction:

1. Adds each halfword from the first operand to the corresponding halfword of the second operand.
2. Shuffles the result by one bit to the right, halving the data.
3. Writes the halfword results in the destination register.

The SHADD8 instruction:

1. Adds each byte of the first operand to the corresponding byte of the second operand.
2. Shuffles the result by one bit to the right, halving the data.
3. Writes the byte results in the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
SHADD16 R1, R0      ; Adds halfwords in R0 to corresponding halfword of R1
                    ; and writes halved result to corresponding halfword in
                    ; R1
SHADD8  R4, R0, R5   ; Adds bytes of R0 to corresponding byte in R5 and
                    ; writes halved result to corresponding byte in R4.
```

### 12.6.5.11 SHASX and SHSAX

Signed Halving Add and Subtract with Exchange and Signed Halving Subtract and Add with Exchange.

Syntax

$op\{cond\} \{Rd\}, Rn, Rm$

where:

op is any of:

SHASX Add and Subtract with Exchange and Halving.

SHSAX Subtract and Add with Exchange and Halving.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The SHASX instruction:

1. Adds the top halfword of the first operand with the bottom halfword of the second operand.
2. Writes the halfword result of the addition to the top halfword of the destination register, shifted by one bit to the right causing a divide by two, or halving.
3. Subtracts the top halfword of the second operand from the bottom highword of the first operand.
4. Writes the halfword result of the division in the bottom halfword of the destination register, shifted by one bit to the right causing a divide by two, or halving.

The SHSAX instruction:

1. Subtracts the bottom halfword of the second operand from the top highword of the first operand.
2. Writes the halfword result of the addition to the bottom halfword of the destination register, shifted by one bit to the right causing a divide by two, or halving.
3. Adds the bottom halfword of the first operand with the top halfword of the second operand.
4. Writes the halfword result of the division in the top halfword of the destination register, shifted by one bit to the right causing a divide by two, or halving.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
SHASX    R7, R4, R2    ; Adds top halfword of R4 to bottom halfword of R2
                        ; and writes halved result to top halfword of R7
                        ; Subtracts top halfword of R2 from bottom halfword of
                        ; R4 and writes halved result to bottom halfword of R7
SHSAX    R0, R3, R5    ; Subtracts bottom halfword of R5 from top halfword
                        ; of R3 and writes halved result to top halfword of R0
                        ; Adds top halfword of R5 to bottom halfword of R3 and
                        ; writes halved result to bottom halfword of R0.
```



### 12.6.5.12 SHSUB16 and SHSUB8

Signed Halving Subtract 16 and Signed Halving Subtract 8

Syntax

*op*{*cond*}{*Rd*,} *Rn*, *Rm*

where:

*op* is any of:

SHSUB16 Signed Halving Subtract 16.

SHSUB8 Signed Halving Subtract 8.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn* is the first operand register.

*Rm* is the second operand register.

Operation

Use these instructions to add 16-bit and 8-bit data and then to halve the result before writing the result to the destination register:

The SHSUB16 instruction:

1. Subtracts each halfword of the second operand from the corresponding halfwords of the first operand.
2. Shuffles the result by one bit to the right, halving the data.
3. Writes the halved halfword results in the destination register.

The SHSUB8 instruction:

1. Subtracts each byte of the second operand from the corresponding byte of the first operand,
2. Shuffles the result by one bit to the right, halving the data,
3. Writes the corresponding signed byte results in the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
SHSUB16 R1, R0      ; Subtracts halfwords in R0 from corresponding halfword
                    ; of R1 and writes to corresponding halfword of R1
SHSUB8  R4, R0, R5  ; Subtracts bytes of R0 from corresponding byte in R5,
                    ; and writes to corresponding byte in R4.
```

### 12.6.5.13 SSUB16 and SSUB8

Signed Subtract 16 and Signed Subtract 8

Syntax

$op\{cond\}\{Rd,\} Rn, Rm$

where:

op is any of:

SSUB16 Performs two 16-bit signed integer subtractions.

SSUB8 Performs four 8-bit signed integer subtractions.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first operand register.

Rm is the second operand register.

Operation

Use these instructions to change endianness of data:

The SSUB16 instruction:

1. Subtracts each halfword from the second operand from the corresponding halfword of the first operand
2. Writes the difference result of two signed halfwords in the corresponding halfword of the destination register.

The SSUB8 instruction:

1. Subtracts each byte of the second operand from the corresponding byte of the first operand
2. Writes the difference result of four signed bytes in the corresponding byte of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
SSUB16 R1, R0      ; Subtracts halfwords in R0 from corresponding halfword
                   ; of R1 and writes to corresponding halfword of R1
SSUB8  R4, R0, R5   ; Subtracts bytes of R5 from corresponding byte in
                   ; R0, and writes to corresponding byte of R4.
```

#### 12.6.5.14 SASX and SSAX

Signed Add and Subtract with Exchange and Signed Subtract and Add with Exchange.

Syntax

*op{cond} {Rd}, Rm, Rn*

where:

op is any of:

SASX Signed Add and Subtract with Exchange.

SSAX Signed Subtract and Add with Exchange.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The SASX instruction:

1. Adds the signed top halfword of the first operand with the signed bottom halfword of the second operand.
2. Writes the signed result of the addition to the top halfword of the destination register.
3. Subtracts the signed bottom halfword of the second operand from the top signed highword of the first operand.
4. Writes the signed result of the subtraction to the bottom halfword of the destination register.

The SSAX instruction:

1. Subtracts the signed bottom halfword of the second operand from the top signed highword of the first operand.
2. Writes the signed result of the addition to the bottom halfword of the destination register.
3. Adds the signed top halfword of the first operand with the signed bottom halfword of the second operand.
4. Writes the signed result of the subtraction to the top halfword of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
SASX  R0, R4, R5 ; Adds top halfword of R4 to bottom halfword of R5 and
                ; writes to top halfword of R0
                ; Subtracts bottom halfword of R5 from top halfword of R4
                ; and writes to bottom halfword of R0
SSAX  R7, R3, R2 ; Subtracts top halfword of R2 from bottom halfword of R3
                ; and writes to bottom halfword of R7
                ; Adds top halfword of R3 with bottom halfword of R2 and
                ; writes to top halfword of R7.
```

### 12.6.5.15 TST and TEQ

Test bits and Test Equivalence.

Syntax

```
TST{cond} Rn, Operand2
TEQ{cond} Rn, Operand2
```

where

**cond** is an optional condition code, see [“Conditional Execution”](#).

**Rn** is the register holding the first operand.

**Operand2** is a flexible second operand. See [“Flexible Second Operand”](#) for details of the options

Operation

These instructions test the value in a register against *Operand2*. They update the condition flags based on the result, but do not write the result to a register.

The TST instruction performs a bitwise AND operation on the value in *Rn* and the value of *Operand2*. This is the same as the ANDS instruction, except that it discards the result.

To test whether a bit of *Rn* is 0 or 1, use the TST instruction with an *Operand2* constant that has that bit set to 1 and all other bits cleared to 0.

The TEQ instruction performs a bitwise Exclusive OR operation on the value in *Rn* and the value of *Operand2*. This is the same as the EORS instruction, except that it discards the result.

Use the TEQ instruction to test if two values are equal without affecting the V or C flags.

TEQ is also useful for testing the sign of a value. After the comparison, the N flag is the logical Exclusive OR of the sign bits of the two operands.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions:

- Update the N and Z flags according to the result
- Can update the C flag during the calculation of *Operand2*, see [“Flexible Second Operand”](#)
- Do not affect the V flag.

Examples

```
TST    R0, #0x3F8 ; Perform bitwise AND of R0 value to 0x3F8,
                ; APSR is updated but result is discarded
TEQEQ  R10, R9    ; Conditionally test if value in R10 is equal to
                ; value in R9, APSR is updated but result is discarded.
```

### 12.6.5.16 UADD16 and UADD8

Unsigned Add 16 and Unsigned Add 8

Syntax

$op\{cond\}\{Rd,\} Rn, Rm$

where:

op is any of:

UADD16 Performs two 16-bit unsigned integer additions.

UADD8 Performs four 8-bit unsigned integer additions.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first register holding the operand.

Rm is the second register holding the operand.

Operation

Use these instructions to add 16- and 8-bit unsigned data:

The UADD16 instruction:

1. Adds each halfword from the first operand to the corresponding halfword of the second operand.
2. Writes the unsigned result in the corresponding halfwords of the destination register.

The UADD8 instruction:

1. Adds each byte of the first operand to the corresponding byte of the second operand.
2. Writes the unsigned result in the corresponding byte of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
UADD16 R1, R0      ; Adds halfwords in R0 to corresponding halfword of R1,
                   ; writes to corresponding halfword of R1
UADD8  R4, R0, R5   ; Adds bytes of R0 to corresponding byte in R5 and
                   ; writes to corresponding byte in R4.
```

### 12.6.5.17 UASX and USAX

Add and Subtract with Exchange and Subtract and Add with Exchange.

Syntax

*op{cond} {Rd}, Rn, Rm*

where:

*op* is one of:

UASX Add and Subtract with Exchange.

USAX Subtract and Add with Exchange.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn, Rm* are registers holding the first and second operands.

Operation

The UASX instruction:

1. Subtracts the top halfword of the second operand from the bottom halfword of the first operand.
2. Writes the unsigned result from the subtraction to the bottom halfword of the destination register.
3. Adds the top halfword of the first operand with the bottom halfword of the second operand.
4. Writes the unsigned result of the addition to the top halfword of the destination register.

The USAX instruction:

1. Adds the bottom halfword of the first operand with the top halfword of the second operand.
2. Writes the unsigned result of the addition to the bottom halfword of the destination register.
3. Subtracts the bottom halfword of the second operand from the top halfword of the first operand.
4. Writes the unsigned result from the subtraction to the top halfword of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
UASX  R0, R4, R5 ; Adds top halfword of R4 to bottom halfword of R5 and
                  ; writes to top halfword of R0
                  ; Subtracts bottom halfword of R5 from top halfword of R0
                  ; and writes to bottom halfword of R0
USAX  R7, R3, R2 ; Subtracts top halfword of R2 from bottom halfword of R3
                  ; and writes to bottom halfword of R7
                  ; Adds top halfword of R3 to bottom halfword of R2 and
                  ; writes to top halfword of R7.
```

### 12.6.5.18 UHADD16 and UHADD8

Unsigned Halving Add 16 and Unsigned Halving Add 8

Syntax

*op*{*cond*}{*Rd*,} *Rn*, *Rm*

where:

*op* is any of:

UHADD16 Unsigned Halving Add 16.

UHADD8 Unsigned Halving Add 8.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn* is the register holding the first operand.

*Rm* is the register holding the second operand.

Operation

Use these instructions to add 16- and 8-bit data and then to halve the result before writing the result to the destination register:

The UHADD16 instruction:

1. Adds each halfword from the first operand to the corresponding halfword of the second operand.
2. Shuffles the halfword result by one bit to the right, halving the data.
3. Writes the unsigned results to the corresponding halfword in the destination register.

The UHADD8 instruction:

1. Adds each byte of the first operand to the corresponding byte of the second operand.
2. Shuffles the byte result by one bit to the right, halving the data.
3. Writes the unsigned results in the corresponding byte in the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
UHADD16 R7, R3      ; Adds halfwords in R7 to corresponding halfword of R3
                    ; and writes halved result to corresponding halfword
                    ; in R7
UHADD8  R4, R0, R5   ; Adds bytes of R0 to corresponding byte in R5 and
                    ; writes halved result to corresponding byte in R4.
```

### 12.6.5.19 UHASX and UHSAX

Unsigned Halving Add and Subtract with Exchange and Unsigned Halving Subtract and Add with Exchange.

Syntax

$op\{cond\} \{Rd\}, Rn, Rm$

where:

op is one of:

UHASX Add and Subtract with Exchange and Halving.

UHSAX Subtract and Add with Exchange and Halving.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The UHASX instruction:

1. Adds the top halfword of the first operand with the bottom halfword of the second operand.
2. Shifts the result by one bit to the right causing a divide by two, or halving.
3. Writes the halfword result of the addition to the top halfword of the destination register.
4. Subtracts the top halfword of the second operand from the bottom highword of the first operand.
5. Shifts the result by one bit to the right causing a divide by two, or halving.
6. Writes the halfword result of the division in the bottom halfword of the destination register.

The UHSAX instruction:

1. Subtracts the bottom halfword of the second operand from the top highword of the first operand.
2. Shifts the result by one bit to the right causing a divide by two, or halving.
3. Writes the halfword result of the subtraction in the top halfword of the destination register.
4. Adds the bottom halfword of the first operand with the top halfword of the second operand.
5. Shifts the result by one bit to the right causing a divide by two, or halving.
6. Writes the halfword result of the addition to the bottom halfword of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
UHASX  R7, R4, R2 ; Adds top halfword of R4 with bottom halfword of R2
                ; and writes halved result to top halfword of R7
                ; Subtracts top halfword of R2 from bottom halfword of
                ; R7 and writes halved result to bottom halfword of R7
UHSAX  R0, R3, R5 ; Subtracts bottom halfword of R5 from top halfword of
                ; R3 and writes halved result to top halfword of R0
                ; Adds top halfword of R5 to bottom halfword of R3 and
                ; writes halved result to bottom halfword of R0.
```



### 12.6.5.20 UHSUB16 and UHSUB8

Unsigned Halving Subtract 16 and Unsigned Halving Subtract 8

Syntax

$op\{cond\}\{Rd,\} Rn, Rm$

where:

op is any of:

UHSUB16 Performs two unsigned 16-bit integer additions, halves the results, and writes the results to the destination register.

UHSUB8 Performs four unsigned 8-bit integer additions, halves the results, and writes the results to the destination register.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first register holding the operand.

Rm is the second register holding the operand.

Operation

Use these instructions to add 16-bit and 8-bit data and then to halve the result before writing the result to the destination register:

The UHSUB16 instruction:

1. Subtracts each halfword of the second operand from the corresponding halfword of the first operand.
2. Shuffles each halfword result to the right by one bit, halving the data.
3. Writes each unsigned halfword result to the corresponding halfwords in the destination register.

The UHSUB8 instruction:

1. Subtracts each byte of second operand from the corresponding byte of the first operand.
2. Shuffles each byte result by one bit to the right, halving the data.
3. Writes the unsigned byte results to the corresponding byte of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
UHSUB16 R1, R0      ; Subtracts halfwords in R0 from corresponding halfword of
                    ; R1 and writes halved result to corresponding halfword in R1
UHSUB8  R4, R0, R5   ; Subtracts bytes of R5 from corresponding byte in R0 and
                    ; writes halved result to corresponding byte in R4.
```

### 12.6.5.21 SEL

Select Bytes. Selects each byte of its result from either its first operand or its second operand, according to the values of the GE flags.

Syntax

```
SEL{<c>}{<q>} {<Rd> , } <Rn> , <Rm>
```

where:

c, q are standard assembler syntax fields.

Rd is the destination register.

Rn is the first register holding the operand.

Rm is the second register holding the operand.

Operation

The SEL instruction:

1. Reads the value of each bit of APSR.GE.
2. Depending on the value of APSR.GE, assigns the destination register the value of either the first or second operand register.

Restrictions

None.

Condition Flags

These instructions do not change the flags.

Examples

```
SADD16 R0, R1, R2    ; Set GE bits based on result
SEL     R0, R0, R3    ; Select bytes from R0 or R3, based on GE.
```

### 12.6.5.22 USAD8

Unsigned Sum of Absolute Differences

Syntax

`USAD8 {cond} {Rd}, Rn, Rm`

where:

`cond` is an optional condition code, see “Conditional Execution”.

`Rd` is the destination register.

`Rn` is the first operand register.

`Rm` is the second operand register.

Operation

The USAD8 instruction:

1. Subtracts each byte of the second operand register from the corresponding byte of the first operand register.
2. Adds the absolute values of the differences together.
3. Writes the result to the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
USAD8 R1, R4, R0 ; Subtracts each byte in R0 from corresponding byte of R4
                  ; adds the differences and writes to R1
USAD8 R0, R5     ; Subtracts bytes of R5 from corresponding byte in R0,
                  ; adds the differences and writes to R0.
```

### 12.6.5.23 USADA8

Unsigned Sum of Absolute Differences and Accumulate

Syntax

`USADA8{cond}{Rd}, Rn, Rm, Ra`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

`Rd` is the destination register.

`Rn` is the first operand register.

`Rm` is the second operand register.

`Ra` is the register that contains the accumulation value.

Operation

The USADA8 instruction:

1. Subtracts each byte of the second operand register from the corresponding byte of the first operand register.
2. Adds the unsigned absolute differences together.
3. Adds the accumulation value to the sum of the absolute differences.
4. Writes the result to the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
USADA8 R1, R0, R6      ; Subtracts bytes in R0 from corresponding halfword of R1
                        ; adds differences, adds value of R6, writes to R1
USADA8 R4, R0, R5, R2   ; Subtracts bytes of R5 from corresponding byte in R0
                        ; adds differences, adds value of R2 writes to R4.
```

#### 12.6.5.24 USUB16 and USUB8

Unsigned Subtract 16 and Unsigned Subtract 8

Syntax

$op\{cond\}\{Rd,\} Rn, Rm$

where

op is any of:

USUB16 Unsigned Subtract 16.

USUB8 Unsigned Subtract 8.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first operand register.

Rm is the second operand register.

Operation

Use these instructions to subtract 16-bit and 8-bit data before writing the result to the destination register:

The USUB16 instruction:

1. Subtracts each halfword from the second operand register from the corresponding halfword of the first operand register.
2. Writes the unsigned result in the corresponding halfwords of the destination register.

The USUB8 instruction:

1. Subtracts each byte of the second operand register from the corresponding byte of the first operand register.
2. Writes the unsigned byte result in the corresponding byte of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
USUB16 R1, R0 ; Subtracts halfwords in R0 from corresponding halfword of R1
               ; and writes to corresponding halfword in R1
USUB8 R4, R0, R5
               ; Subtracts bytes of R5 from corresponding byte in R0 and
               ; writes to the corresponding byte in R4.
```

## 12.6.6 Multiply and Divide Instructions

The table below shows the multiply and divide instructions:

**Table 12-21. Multiply and Divide Instructions**

Mnemonic	Description
MLA	Multiply with Accumulate, 32-bit result
MLS	Multiply and Subtract, 32-bit result
MUL	Multiply, 32-bit result
SDIV	Signed Divide
SMLA[B,T]	Signed Multiply Accumulate (halfwords)
SMLAD, SMLADX	Signed Multiply Accumulate Dual
SMLAL	Signed Multiply with Accumulate (32x32+64), 64-bit result
SMLAL[B,T]	Signed Multiply Accumulate Long (halfwords)
SMLALD, SMLALDX	Signed Multiply Accumulate Long Dual
SMLAW[B,T]	Signed Multiply Accumulate (word by halfword)
SMLSD	Signed Multiply Subtract Dual
SMLSLD	Signed Multiply Subtract Long Dual
SMMLA	Signed Most Significant Word Multiply Accumulate
SMMLS, SMMLSR	Signed Most Significant Word Multiply Subtract
SMUAD, SMUADX	Signed Dual Multiply Add
SMUL[B,T]	Signed Multiply (word by halfword)
SMMUL, SMMULR	Signed Most Significant Word Multiply
SMULL	Signed Multiply (32x32), 64-bit result
SMULWB, SMULWT	Signed Multiply (word by halfword)
SMUSD, SMUSDX	Signed Dual Multiply Subtract
UDIV	Unsigned Divide
UMAAL	Unsigned Multiply Accumulate Accumulate Long (32x32+32+32), 64-bit result
UMLAL	Unsigned Multiply with Accumulate (32x32+64), 64-bit result
UMULL	Unsigned Multiply (32x32), 64-bit result

### 12.6.6.1 MUL, MLA, and MLS

Multiply, Multiply with Accumulate, and Multiply with Subtract, using 32-bit operands, and producing a 32-bit result.

Syntax

```
MUL{S}{cond} {Rd}, Rn, Rm ; Multiply
MLA{cond} Rd, Rn, Rm, Ra ; Multiply with accumulate
MLS{cond} Rd, Rn, Rm, Ra ; Multiply with subtract
```

where:

**cond** is an optional condition code, see “Conditional Execution”.

**S** is an optional suffix. If S is specified, the condition code flags are updated on the result of the operation, see “Conditional Execution”.

**Rd** is the destination register. If *Rd* is omitted, the destination register is *Rn*.

**Rn, Rm** are registers holding the values to be multiplied.

**Ra** is a register holding the value to be added or subtracted from.

Operation

The MUL instruction multiplies the values from *Rn* and *Rm*, and places the least significant 32 bits of the result in *Rd*.

The MLA instruction multiplies the values from *Rn* and *Rm*, adds the value from *Ra*, and places the least significant 32 bits of the result in *Rd*.

The MLS instruction multiplies the values from *Rn* and *Rm*, subtracts the product from the value from *Ra*, and places the least significant 32 bits of the result in *Rd*.

The results of these instructions do not depend on whether the operands are signed or unsigned.

Restrictions

In these instructions, do not use SP and do not use PC.

If the S suffix is used with the MUL instruction:

- *Rd*, *Rn*, and *Rm* must all be in the range R0 to R7
- *Rd* must be the same as *Rm*
- The *cond* suffix must not be used.

Condition Flags

If S is specified, the MUL instruction:

- Updates the N and Z flags according to the result
- Does not affect the C and V flags.

Examples

```
MUL    R10, R2, R5      ; Multiply, R10 = R2 x R5
MLA    R10, R2, R1, R5  ; Multiply with accumulate, R10 = (R2 x R1) + R5
MULS   R0, R2, R2       ; Multiply with flag update, R0 = R2 x R2
MULLT  R2, R3, R2       ; Conditionally multiply, R2 = R3 x R2
MLS    R4, R5, R6, R7   ; Multiply with subtract, R4 = R7 - (R5 x R6)
```

### 12.6.6.2 UMULL, UMAAL, UMLAL

Unsigned Long Multiply, with optional Accumulate, using 32-bit operands and producing a 64-bit result.

Syntax

*op*{*cond*} *RdLo*, *RdHi*, *Rn*, *Rm*

where:

*op* is one of:

UMULL Unsigned Long Multiply.

UMAAL Unsigned Long Multiply with Accumulate Accumulate.

UMLAL Unsigned Long Multiply, with Accumulate.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*RdHi*, *RdLo* are the destination registers. For UMAAL, UMLAL and UMLAL they also hold the accumulating value.

*Rn*, *Rm* are registers holding the first and second operands.

Operation

These instructions interpret the values from *Rn* and *Rm* as unsigned 32-bit integers.

The UMULL instruction:

- Multiplies the two unsigned integers in the first and second operands.
- Writes the least significant 32 bits of the result in *RdLo*.
- Writes the most significant 32 bits of the result in *RdHi*.

The UMAAL instruction:

- Multiplies the two unsigned 32-bit integers in the first and second operands.
- Adds the unsigned 32-bit integer in *RdHi* to the 64-bit result of the multiplication.
- Adds the unsigned 32-bit integer in *RdLo* to the 64-bit result of the addition.
- Writes the top 32-bits of the result to *RdHi*.
- Writes the lower 32-bits of the result to *RdLo*.

The UMLAL instruction:

- Multiplies the two unsigned integers in the first and second operands.
- Adds the 64-bit result to the 64-bit unsigned integer contained in *RdHi* and *RdLo*.
- Writes the result back to *RdHi* and *RdLo*.

Restrictions

In these instructions:

- Do not use SP and do not use PC.
- *RdHi* and *RdLo* must be different registers.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
UMULL    R0, R4, R5, R6    ; Multiplies R5 and R6, writes the top 32 bits to R4
                          ; and the bottom 32 bits to R0
UMAAL    R3, R6, R2, R7    ; Multiplies R2 and R7, adds R6, adds R3, writes the
                          ; top 32 bits to R6, and the bottom 32 bits to R3
UMLAL    R2, R1, R3, R5    ; Multiplies R5 and R3, adds R1:R2, writes to R1:R2.
```



### 12.6.6.3 SMLA and SMLAW

Signed Multiply Accumulate (halfwords).

Syntax

```
op{XY}{cond} Rd, Rn, Rm
op{Y}{cond} Rd, Rn, Rm, Ra
```

where:

op

is one of:

SMLA Signed Multiply Accumulate Long (halfwords).

X and Y specifies which half of the source registers *Rn* and *Rm* are used as the first and second multiply operand.

If X is B, then the bottom halfword, bits [15:0], of *Rn* is used.

If X is T, then the top halfword, bits [31:16], of *Rn* is used.

If Y is B, then the bottom halfword, bits [15:0], of *Rm* is used.

If Y is T, then the top halfword, bits [31:16], of *Rm* is used

SMLAW Signed Multiply Accumulate (word by halfword).

Y specifies which half of the source register *Rm* is used as the second multiply operand.

If Y is T, then the top halfword, bits [31:16] of *Rm* is used.

If Y is B, then the bottom halfword, bits [15:0] of *Rm* is used.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register. If *Rd* is omitted, the destination register is *Rn*.

Rn, Rm are registers holding the values to be multiplied.

Ra is a register holding the value to be added or subtracted from.

Operation

The SMALBB, SMLABT, SMLATB, SMLATT instructions:

- Multiplies the specified signed halfword, top or bottom, values from *Rn* and *Rm*.
- Adds the value in *Ra* to the resulting 32-bit product.
- Writes the result of the multiplication and addition in *Rd*.

The non-specified halfwords of the source registers are ignored.

The SMLAWB and SMLAWT instructions:

- Multiply the 32-bit signed values in *Rn* with:
  - The top signed halfword of *Rm*, T instruction suffix.
  - The bottom signed halfword of *Rm*, B instruction suffix.
- Add the 32-bit signed value in *Ra* to the top 32 bits of the 48-bit product
- Writes the result of the multiplication and addition in *Rd*.

The bottom 16 bits of the 48-bit product are ignored.

If overflow occurs during the addition of the accumulate value, the instruction sets the Q flag in the APSR. No overflow can occur during the multiplication.

Restrictions

In these instructions, do not use SP and do not use PC.

Condition Flags

If an overflow is detected, the Q flag is set.

Examples

```
SMLABB R5, R6, R4, R1 ; Multiplies bottom halfwords of R6 and R4, adds
                        ; R1 and writes to R5
SMLATB R5, R6, R4, R1 ; Multiplies top halfword of R6 with bottom halfword
                        ; of R4, adds R1 and writes to R5
```

```

SMLATT  R5, R6, R4, R1 ; Multiplies top halfwords of R6 and R4, adds
                        ; R1 and writes the sum to R5
SMLABT  R5, R6, R4, R1 ; Multiplies bottom halfword of R6 with top halfword
                        ; of R4, adds R1 and writes to R5
SMLABT  R4, R3, R2      ; Multiplies bottom halfword of R4 with top halfword of
                        ; R3, adds R2 and writes to R4
SMLAWB  R10, R2, R5, R3 ; Multiplies R2 with bottom halfword of R5, adds
                        ; R3 to the result and writes top 32-bits to R10
SMLAWT  R10, R2, R1, R5 ; Multiplies R2 with top halfword of R1, adds R5
                        ; and writes top 32-bits to R10.

```

#### 12.6.6.4 SMLAD

Signed Multiply Accumulate Long Dual

Syntax

```
op{X}{cond} Rd, Rn, Rm, Ra ;
```

where:

op is one of:

SMLAD Signed Multiply Accumulate Dual.

SMLADX Signed Multiply Accumulate Dual Reverse.

X specifies which halfword of the source register *Rn* is used as the multiply operand.

If X is omitted, the multiplications are bottom × bottom and top × top.

If X is present, the multiplications are bottom × top and top × bottom.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first operand register holding the values to be multiplied.

Rm the second operand register.

Ra is the accumulate value.

Operation

The SMLAD and SMLADX instructions regard the two operands as four halfword 16-bit values. The SMLAD and SMLADX instructions:

- If X is not present, multiply the top signed halfword value in *Rn* with the top signed halfword of *Rm* and the bottom signed halfword values in *Rn* with the bottom signed halfword of *Rm*.
- Or if X is present, multiply the top signed halfword value in *Rn* with the bottom signed halfword of *Rm* and the bottom signed halfword values in *Rn* with the top signed halfword of *Rm*.
- Add both multiplication results to the signed 32-bit value in *Ra*.
- Writes the 32-bit signed result of the multiplication and addition to *Rd*.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```

SMLAD   R10, R2, R1, R5 ; Multiplies two halfword values in R2 with
                        ; corresponding halfwords in R1, adds R5 and
                        ; writes to R10
SMLALDX R0, R2, R4, R6 ; Multiplies top halfword of R2 with bottom
                        ; halfword of R4, multiplies bottom halfword of R2
                        ; with top halfword of R4, adds R6 and writes to
                        ; R0.

```

### 12.6.6.5 SMLAL and SMLALD

Signed Multiply Accumulate Long, Signed Multiply Accumulate Long (halfwords) and Signed Multiply Accumulate Long Dual.

Syntax

```
op{cond} RdLo, RdHi, Rn, Rm
op{XY}{cond} RdLo, RdHi, Rn, Rm
op{X}{cond} RdLo, RdHi, Rn, Rm
```

where:

op is one of:

MLAL Signed Multiply Accumulate Long.

SMLAL Signed Multiply Accumulate Long (halfwords, X and Y).

X and Y specify which halfword of the source registers *Rn* and *Rm* are used as the first and second multiply operand:

If X is B, then the bottom halfword, bits [15:0], of *Rn* is used.

If X is T, then the top halfword, bits [31:16], of *Rn* is used.

If Y is B, then the bottom halfword, bits [15:0], of *Rm* is used.

If Y is T, then the top halfword, bits [31:16], of *Rm* is used.

SMLALD Signed Multiply Accumulate Long Dual.

SMLALDX Signed Multiply Accumulate Long Dual Reversed.

If the X is omitted, the multiplications are bottom × bottom and top × top.

If X is present, the multiplications are bottom × top and top × bottom.

cond is an optional condition code, see [“Conditional Execution”](#).

RdHi, RdLo are the destination registers.

*RdLo* is the lower 32 bits and *RdHi* is the upper 32 bits of the 64-bit integer.

For SMLAL, SMLALBB, SMLALBT, SMLALTB, SMLALTT, SMLALD and SMLALDX, they also hold the accumulating value.

Rn, Rm are registers holding the first and second operands.

Operation

The SMLAL instruction:

- Multiplies the two's complement signed word values from *Rn* and *Rm*.
- Adds the 64-bit value in *RdLo* and *RdHi* to the resulting 64-bit product.
- Writes the 64-bit result of the multiplication and addition in *RdLo* and *RdHi*.

The SMLALBB, SMLALBT, SMLALTB and SMLALTT instructions:

- Multiplies the specified signed halfword, Top or Bottom, values from *Rn* and *Rm*.
- Adds the resulting sign-extended 32-bit product to the 64-bit value in *RdLo* and *RdHi*.
- Writes the 64-bit result of the multiplication and addition in *RdLo* and *RdHi*.

The non-specified halfwords of the source registers are ignored.

The SMLALD and SMLALDX instructions interpret the values from *Rn* and *Rm* as four halfword two's complement signed 16-bit integers. These instructions:

- If X is not present, multiply the top signed halfword value of *Rn* with the top signed halfword of *Rm* and the bottom signed halfword values of *Rn* with the bottom signed halfword of *Rm*.
- Or if X is present, multiply the top signed halfword value of *Rn* with the bottom signed halfword of *Rm* and the bottom signed halfword values of *Rn* with the top signed halfword of *Rm*.
- Add the two multiplication results to the signed 64-bit value in *RdLo* and *RdHi* to create the resulting 64-bit product.
- Write the 64-bit product in *RdLo* and *RdHi*.

## Restrictions

In these instructions:

- Do not use SP and do not use PC.
- *RdHi* and *RdLo* must be different registers.

## Condition Flags

These instructions do not affect the condition code flags.

## Examples

```
SMLAL    R4, R5, R3, R8 ; Multiplies R3 and R8, adds R5:R4 and writes to
                        ; R5:R4
SMLALBT  R2, R1, R6, R7 ; Multiplies bottom halfword of R6 with top
                        ; halfword of R7, sign extends to 32-bit, adds
                        ; R1:R2 and writes to R1:R2
SMLALTB  R2, R1, R6, R7 ; Multiplies top halfword of R6 with bottom
                        ; halfword of R7, sign extends to 32-bit, adds R1:R2
                        ; and writes to R1:R2
SMLALD   R6, R8, R5, R1 ; Multiplies top halfwords in R5 and R1 and bottom
                        ; halfwords of R5 and R1, adds R8:R6 and writes to
                        ; R8:R6
SMLALDX  R6, R8, R5, R1 ; Multiplies top halfword in R5 with bottom
                        ; halfword of R1, and bottom halfword of R5 with
                        ; top halfword of R1, adds R8:R6 and writes to
                        ; R8:R6.
```

### 12.6.6.6 SMLSD and SMLSLD

Signed Multiply Subtract Dual and Signed Multiply Subtract Long Dual

#### Syntax

*op*{*X*}{*cond*} *Rd*, *Rn*, *Rm*, *Ra*

where:

*op* is one of:

SMLSD Signed Multiply Subtract Dual.

SMLSDX Signed Multiply Subtract Dual Reversed.

SMLSLD Signed Multiply Subtract Long Dual.

SMLSLDX Signed Multiply Subtract Long Dual Reversed.

SMLAW Signed Multiply Accumulate (word by halfword).

If *X* is present, the multiplications are bottom × top and top × bottom.

If the *X* is omitted, the multiplications are bottom × bottom and top × top.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn*, *Rm* are registers holding the first and second operands.

*Ra* is the register holding the accumulate value.

#### Operation

The SMLSD instruction interprets the values from the first and second operands as four signed halfwords. This instruction:

- Optionally rotates the halfwords of the second operand.
- Performs two signed 16 × 16-bit halfword multiplications.
- Subtracts the result of the upper halfword multiplication from the result of the lower halfword multiplication.
- Adds the signed accumulate value to the result of the subtraction.

- Writes the result of the addition to the destination register.

The SMLSXD instruction interprets the values from *Rn* and *Rm* as four signed halfwords.

This instruction:

- Optionally rotates the halfwords of the second operand.
- Performs two signed 16 × 16-bit halfword multiplications.
- Subtracts the result of the upper halfword multiplication from the result of the lower halfword multiplication.
- Adds the 64-bit value in *RdHi* and *RdLo* to the result of the subtraction.
- Writes the 64-bit result of the addition to the *RdHi* and *RdLo*.

Restrictions

In these instructions:

- Do not use SP and do not use PC.

Condition Flags

This instruction sets the Q flag if the accumulate operation overflows. Overflow cannot occur during the multiplications or subtraction.

For the Thumb instruction set, these instructions do not affect the condition code flags.

Examples

```
SMLSXD  R0, R4, R5, R6 ; Multiplies bottom halfword of R4 with bottom
                        ; halfword of R5, multiplies top halfword of R4
                        ; with top halfword of R5, subtracts second from
                        ; first, adds R6, writes to R0
SMLSXD  R1, R3, R2, R0 ; Multiplies bottom halfword of R3 with top
                        ; halfword of R2, multiplies top halfword of R3
                        ; with bottom halfword of R2, subtracts second from
                        ; first, adds R0, writes to R1
SMLSXD  R3, R6, R2, R7 ; Multiplies bottom halfword of R6 with bottom
                        ; halfword of R2, multiplies top halfword of R6
                        ; with top halfword of R2, subtracts second from
                        ; first, adds R6:R3, writes to R6:R3
SMLSXD  R3, R6, R2, R7 ; Multiplies bottom halfword of R6 with top
                        ; halfword of R2, multiplies top halfword of R6
                        ; with bottom halfword of R2, subtracts second from
                        ; first, adds R6:R3, writes to R6:R3.
```

### 12.6.6.7 SMMLA and SMMLS

Signed Most Significant Word Multiply Accumulate and Signed Most Significant Word Multiply Subtract

Syntax

$op\{R\}\{cond\} Rd, Rn, Rm, Ra$

where:

*op* is one of:

SMMLA Signed Most Significant Word Multiply Accumulate.

SMMLS Signed Most Significant Word Multiply Subtract.

If the *X* is omitted, the multiplications are bottom  $\times$  bottom and top  $\times$  top.

*R* is a rounding error flag. If *R* is specified, the result is rounded instead of being truncated. In this case the constant 0x80000000 is added to the product before the high word is extracted.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn, Rm* are registers holding the first and second multiply operands.

*Ra* is the register holding the accumulate value.

Operation

The SMMLA instruction interprets the values from *Rn* and *Rm* as signed 32-bit words.

The SMMLA instruction:

- Multiplies the values in *Rn* and *Rm*.
- Optionally rounds the result by adding 0x80000000.
- Extracts the most significant 32 bits of the result.
- Adds the value of *Ra* to the signed extracted value.
- Writes the result of the addition in *Rd*.

The SMMLS instruction interprets the values from *Rn* and *Rm* as signed 32-bit words.

The SMMLS instruction:

- Multiplies the values in *Rn* and *Rm*.
- Optionally rounds the result by adding 0x80000000.
- Extracts the most significant 32 bits of the result.
- Subtracts the extracted value of the result from the value in *Ra*.
- Writes the result of the subtraction in *Rd*.

Restrictions

In these instructions:

- Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
SMMLA  R0, R4, R5, R6 ; Multiplies R4 and R5, extracts top 32 bits, adds
                        ; R6, truncates and writes to R0
SMMLAR R6, R2, R1, R4 ; Multiplies R2 and R1, extracts top 32 bits, adds
                        ; R4, rounds and writes to R6
SMMLSR R3, R6, R2, R7 ; Multiplies R6 and R2, extracts top 32 bits,
                        ; subtracts R7, rounds and writes to R3
SMMLS  R4, R5, R3, R8 ; Multiplies R5 and R3, extracts top 32 bits,
                        ; subtracts R8, truncates and writes to R4.
```

### 12.6.6.8 SMMUL

Signed Most Significant Word Multiply

Syntax

$op\{R\}\{cond\} Rd, Rn, Rm$

where:

*op* is one of:

SMMUL Signed Most Significant Word Multiply.

*R* is a rounding error flag. If *R* is specified, the result is rounded instead of being truncated. In this case the constant 0x80000000 is added to the product before the high word is extracted.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn, Rm* are registers holding the first and second operands.

Operation

The SMMUL instruction interprets the values from *Rn* and *Rm* as two's complement 32-bit signed integers. The SMMUL instruction:

- Multiplies the values from *Rn* and *Rm*.
- Optionally rounds the result, otherwise truncates the result.
- Writes the most significant signed 32 bits of the result in *Rd*.

Restrictions

In this instruction:

- do not use SP and do not use PC.

Condition Flags

This instruction does not affect the condition code flags.

Examples

```
SMULL    R0, R4, R5    ; Multiplies R4 and R5, truncates top 32 bits
                        ; and writes to R0
SMULLR   R6, R2        ; Multiplies R6 and R2, rounds the top 32 bits
                        ; and writes to R6.
```

### 12.6.6.9 SMUAD and SMUSD

Signed Dual Multiply Add and Signed Dual Multiply Subtract

Syntax

$op\{X\}\{cond\} Rd, Rn, Rm$

where:

op is one of:

SMUAD Signed Dual Multiply Add.

SMUADX Signed Dual Multiply Add Reversed.

SMUSD Signed Dual Multiply Subtract.

SMUSDX Signed Dual Multiply Subtract Reversed.

If X is present, the multiplications are bottom × top and top × bottom.

If the X is omitted, the multiplications are bottom × bottom and top × top.

cond is an optional condition code, see “Conditional Execution”.

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The SMUAD instruction interprets the values from the first and second operands as two signed halfwords in each operand. This instruction:

- Optionally rotates the halfwords of the second operand.
- Performs two signed 16 × 16-bit multiplications.
- Adds the two multiplication results together.
- Writes the result of the addition to the destination register.

The SMUSD instruction interprets the values from the first and second operands as two's complement signed integers. This instruction:

- Optionally rotates the halfwords of the second operand.
- Performs two signed 16 × 16-bit multiplications.
- Subtracts the result of the top halfword multiplication from the result of the bottom halfword multiplication.
- Writes the result of the subtraction to the destination register.

Restrictions

In these instructions:

- Do not use SP and do not use PC.

Condition Flags

Sets the Q flag if the addition overflows. The multiplications cannot overflow.

Examples

```
SMUAD    R0, R4, R5    ; Multiplies bottom halfword of R4 with the bottom
                        ; halfword of R5, adds multiplication of top halfword
                        ; of R4 with top halfword of R5, writes to R0
SMUADX   R3, R7, R4    ; Multiplies bottom halfword of R7 with top halfword
                        ; of R4, adds multiplication of top halfword of R7
                        ; with bottom halfword of R4, writes to R3
SMUSD    R3, R6, R2    ; Multiplies bottom halfword of R4 with bottom halfword
                        ; of R6, subtracts multiplication of top halfword of R6
                        ; with top halfword of R3, writes to R3
SMUSDX   R4, R5, R3    ; Multiplies bottom halfword of R5 with top halfword of
                        ; R3, subtracts multiplication of top halfword of R5
                        ; with bottom halfword of R3, writes to R4.
```



### 12.6.6.10 SMUL and SMULW

Signed Multiply (halfwords) and Signed Multiply (word by halfword)

Syntax

```
op{XY}{cond} Rd, Rn, Rm
op{Y}{cond} Rd, Rn, Rm
```

For *SMULXY* only:

op is one of:

**SMUL{XY}** Signed Multiply (halfwords).

X and Y specify which halfword of the source registers *Rn* and *Rm* is used as the first and second multiply operand.

If X is B, then the bottom halfword, bits [15:0] of *Rn* is used.

If X is T, then the top halfword, bits [31:16] of *Rn* is used. If Y is B, then the bottom halfword, bits [15:0], of *Rm* is used.

If Y is T, then the top halfword, bits [31:16], of *Rm* is used.

**SMULW{Y}** Signed Multiply (word by halfword).

Y specifies which halfword of the source register *Rm* is used as the second multiply operand.

If Y is B, then the bottom halfword (bits [15:0]) of *Rm* is used.

If Y is T, then the top halfword (bits [31:16]) of *Rm* is used.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The *SMULBB*, *SMULTB*, *SMULBT* and *SMULTT* instructions interpret the values from *Rn* and *Rm* as four signed 16-bit integers. These instructions:

- Multiplies the specified signed halfword, Top or Bottom, values from *Rn* and *Rm*.
- Writes the 32-bit result of the multiplication in *Rd*.

The *SMULWT* and *SMULWB* instructions interpret the values from *Rn* as a 32-bit signed integer and *Rm* as two halfword 16-bit signed integers. These instructions:

- Multiplies the first operand and the top, T suffix, or the bottom, B suffix, halfword of the second operand.
- Writes the signed most significant 32 bits of the 48-bit result in the destination register.

Restrictions

In these instructions:

- Do not use SP and do not use PC.
- *RdHi* and *RdLo* must be different registers.

Examples

```
SMULBT    R0, R4, R5    ; Multiplies the bottom halfword of R4 with the
                        ; top halfword of R5, multiplies results and
                        ; writes to R0
SMULBB    R0, R4, R5    ; Multiplies the bottom halfword of R4 with the
                        ; bottom halfword of R5, multiplies results and
                        ; writes to R0
SMULTT    R0, R4, R5    ; Multiplies the top halfword of R4 with the top
                        ; halfword of R5, multiplies results and writes
                        ; to R0
SMULTB    R0, R4, R5    ; Multiplies the top halfword of R4 with the
                        ; bottom halfword of R5, multiplies results and
```

			; and writes to R0
SMULWT	R4, R5, R3		; Multiplies R5 with the top halfword of R3,
			; extracts top 32 bits and writes to R4
SMULWB	R4, R5, R3		; Multiplies R5 with the bottom halfword of R3,
			; extracts top 32 bits and writes to R4.

#### 12.6.6.11 UMULL, UMLAL, SMULL, and SMLAL

Signed and Unsigned Long Multiply, with optional Accumulate, using 32-bit operands and producing a 64-bit result.

Syntax

```
op{cond} RdLo, RdHi, Rn, Rm
```

where:

op is one of:

UMULL Unsigned Long Multiply.

UMLAL Unsigned Long Multiply, with Accumulate.

SMULL Signed Long Multiply.

SMLAL Signed Long Multiply, with Accumulate.

cond is an optional condition code, see [“Conditional Execution”](#).

RdHi, RdLo are the destination registers. For UMLAL and SMLAL they also hold the accumulating value.

Rn, Rm are registers holding the operands.

Operation

The UMULL instruction interprets the values from *Rn* and *Rm* as unsigned integers. It multiplies these integers and places the least significant 32 bits of the result in *RdLo*, and the most significant 32 bits of the result in *RdHi*.

The UMLAL instruction interprets the values from *Rn* and *Rm* as unsigned integers. It multiplies these integers, adds the 64-bit result to the 64-bit unsigned integer contained in *RdHi* and *RdLo*, and writes the result back to *RdHi* and *RdLo*.

The SMULL instruction interprets the values from *Rn* and *Rm* as two's complement signed integers. It multiplies these integers and places the least significant 32 bits of the result in *RdLo*, and the most significant 32 bits of the result in *RdHi*.

The SMLAL instruction interprets the values from *Rn* and *Rm* as two's complement signed integers. It multiplies these integers, adds the 64-bit result to the 64-bit signed integer contained in *RdHi* and *RdLo*, and writes the result back to *RdHi* and *RdLo*.

Restrictions

In these instructions:

- Do not use SP and do not use PC
- *RdHi* and *RdLo* must be different registers.

Condition Flags

These instructions do not affect the condition code flags.

Examples

UMULL	R0, R4, R5, R6	; Unsigned (R4,R0) = R5 x R6
SMLAL	R4, R5, R3, R8	; Signed (R5,R4) = (R5,R4) + R3 x R8

### 12.6.6.12 SDIV and UDIV

Signed Divide and Unsigned Divide.

Syntax

```
SDIV{cond} {Rd}, Rn, Rm
UDIV{cond} {Rd}, Rn, Rm
```

where:

cond is an optional condition code, see “Conditional Execution”.

Rd is the destination register. If Rd is omitted, the destination register is Rn.

Rn is the register holding the value to be divided.

Rm is a register holding the divisor.

Operation

SDIV performs a signed integer division of the value in Rn by the value in Rm.

UDIV performs an unsigned integer division of the value in Rn by the value in Rm.

For both instructions, if the value in Rn is not divisible by the value in Rm, the result is rounded towards zero.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not change the flags.

Examples

```
SDIV R0, R2, R4 ; Signed divide, R0 = R2/R4
UDIV R8, R8, R1 ; Unsigned divide, R8 = R8/R1
```

## 12.6.7 Saturating Instructions

The table below shows the saturating instructions:

**Table 12-22. Saturating Instructions**

Mnemonic	Description
SSAT	Signed Saturate
SSAT16	Signed Saturate Halfword
USAT	Unsigned Saturate
USAT16	Unsigned Saturate Halfword
QADD	Saturating Add
QSUB	Saturating Subtract
QSUB16	Saturating Subtract 16
QASX	Saturating Add and Subtract with Exchange
QSAX	Saturating Subtract and Add with Exchange
QDADD	Saturating Double and Add
QDSUB	Saturating Double and Subtract
UQADD16	Unsigned Saturating Add 16
UQADD8	Unsigned Saturating Add 8
UQASX	Unsigned Saturating Add and Subtract with Exchange
UQSAX	Unsigned Saturating Subtract and Add with Exchange
UQSUB16	Unsigned Saturating Subtract 16
UQSUB8	Unsigned Saturating Subtract 8

For signed  $n$ -bit saturation, this means that:

- If the value to be saturated is less than  $-2^{n-1}$ , the result returned is  $-2^{n-1}$
- If the value to be saturated is greater than  $2^{n-1}-1$ , the result returned is  $2^{n-1}-1$
- Otherwise, the result returned is the same as the value to be saturated.

For unsigned  $n$ -bit saturation, this means that:

- If the value to be saturated is less than 0, the result returned is 0
- If the value to be saturated is greater than  $2^n-1$ , the result returned is  $2^n-1$
- Otherwise, the result returned is the same as the value to be saturated.

If the returned result is different from the value to be saturated, it is called *saturation*. If saturation occurs, the instruction sets the Q flag to 1 in the APSR. Otherwise, it leaves the Q flag unchanged. To clear the Q flag to 0, the MSR instruction must be used; see “MSR”.

To read the state of the Q flag, the MRS instruction must be used; see “MRS”.

### 12.6.7.1 SSAT and USAT

Signed Saturate and Unsigned Saturate to any bit position, with optional shift before saturating.

Syntax

`op{cond} Rd, #n, Rm {, shift #s}`

where:

op is one of:

SSAT Saturates a signed value to a signed range.

USAT Saturates a signed value to an unsigned range.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

n specifies the bit position to saturate to:

n ranges from 1 to 32 for SSAT  
n ranges from 0 to 31 for USAT.

Rm is the register containing the value to saturate.

shift #s is an optional shift applied to Rm before saturating. It must be one of the following:

ASR #s where s is in the range 1 to 31.

LSL #s where s is in the range 0 to 31.

Operation

These instructions saturate to a signed or unsigned  $n$ -bit value.

The SSAT instruction applies the specified shift, then saturates to the signed range  $-2^{n-1} \leq x \leq 2^{n-1}-1$ .

The USAT instruction applies the specified shift, then saturates to the unsigned range  $0 \leq x \leq 2^n-1$ .

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

If saturation occurs, these instructions set the Q flag to 1.

Examples

```
SSAT    R7, #16, R7, LSL #4    ; Logical shift left value in R7 by 4, then
                                ; saturate it as a signed 16-bit value and
                                ; write it back to R7
USATNE  R0, #7, R5             ; Conditionally saturate value in R5 as an
                                ; unsigned 7 bit value and write it to R0.
```

### 12.6.7.2 SSAT16 and USAT16

Signed Saturate and Unsigned Saturate to any bit position for two halfwords.

Syntax

*op*{*cond*} *Rd*, #*n*, *Rm*

where:

*op* is one of:

SSAT16 Saturates a signed halfword value to a signed range.

USAT16 Saturates a signed halfword value to an unsigned range.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*n* specifies the bit position to saturate to:

*n* ranges from 1 to 16 for SSAT

*n* ranges from 0 to 15 for USAT.

*Rm* is the register containing the value to saturate.

Operation

The SSAT16 instruction:

Saturates two signed 16-bit halfword values of the register with the value to saturate from selected by the bit position in *n*.

Writes the results as two signed 16-bit halfwords to the destination register.

The USAT16 instruction:

Saturates two unsigned 16-bit halfword values of the register with the value to saturate from selected by the bit position in *n*.

Writes the results as two unsigned halfwords in the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

If saturation occurs, these instructions set the Q flag to 1.

Examples

```
SSAT16    R7, #9, R2    ; Saturates the top and bottom highwords of R2
                        ; as 9-bit values, writes to corresponding halfword
                        ; of R7
USAT16NE  R0, #13, R5   ; Conditionally saturates the top and bottom
                        ; halfwords of R5 as 13-bit values, writes to
                        ; corresponding halfword of R0.
```

### 12.6.7.3 QADD and QSUB

Saturating Add and Saturating Subtract, signed.

Syntax

```
op{cond} {Rd}, Rn, Rm
op{cond} {Rd}, Rn, Rm
```

where:

op is one of:

QADD Saturating 32-bit add.

QADD8 Saturating four 8-bit integer additions.

QADD16 Saturating two 16-bit integer additions.

QSUB Saturating 32-bit subtraction.

QSUB8 Saturating four 8-bit integer subtraction.

QSUB16 Saturating two 16-bit integer subtraction.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

These instructions add or subtract two, four or eight values from the first and second operands and then writes a signed saturated value in the destination register.

The QADD and QSUB instructions apply the specified add or subtract, and then saturate the result to the signed range -  $2^{n-1} \times x$  to  $2^{n-1}-1$ , where  $x$  is given by the number of bits applied in the instruction, 32, 16 or 8.

If the returned result is different from the value to be saturated, it is called *saturation*. If saturation occurs, the QADD and QSUB instructions set the Q flag to 1 in the APSR. Otherwise, it leaves the Q flag unchanged. The 8-bit and 16-bit QADD and QSUB instructions always leave the Q flag unchanged.

To clear the Q flag to 0, the MSR instruction must be used; see [“MSR”](#).

To read the state of the Q flag, the MRS instruction must be used; see [“MRS”](#).

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

If saturation occurs, these instructions set the Q flag to 1.

Examples

```
QADD16    R7, R4, R2    ; Adds halfwords of R4 with corresponding halfword of
                        ; R2, saturates to 16 bits and writes to
                        ; corresponding halfword of R7
QADD8      R3, R1, R6    ; Adds bytes of R1 to the corresponding bytes of R6,
                        ; saturates to 8 bits and writes to corresponding
                        ; byte of R3
QSUB16     R4, R2, R3    ; Subtracts halfwords of R3 from corresponding
                        ; halfword of R2, saturates to 16 bits, writes to
                        ; corresponding halfword of R4
QSUB8      R4, R2, R5    ; Subtracts bytes of R5 from the corresponding byte
                        ; in R2, saturates to 8 bits, writes to corresponding
                        ; byte of R4.
```

#### 12.6.7.4 QASX and QSAX

Saturating Add and Subtract with Exchange and Saturating Subtract and Add with Exchange, signed.

Syntax

*op*{*cond*} {*Rd*}, *Rm*, *Rn*

where:

*op* is one of:

QASX Add and Subtract with Exchange and Saturate.

QSAX Subtract and Add with Exchange and Saturate.

*cond* is an optional condition code, see [“Conditional Execution”](#).

*Rd* is the destination register.

*Rn*, *Rm* are registers holding the first and second operands.

Operation

The QASX instruction:

1. Adds the top halfword of the source operand with the bottom halfword of the second operand.
2. Subtracts the top halfword of the second operand from the bottom highword of the first operand.
3. Saturates the result of the subtraction and writes a 16-bit signed integer in the range  $-2^{15} \leq x \leq 2^{15} - 1$ , where *x* equals 16, to the bottom halfword of the destination register.
4. Saturates the results of the sum and writes a 16-bit signed integer in the range  $-2^{15} \leq x \leq 2^{15} - 1$ , where *x* equals 16, to the top halfword of the destination register.

The QSAX instruction:

1. Subtracts the bottom halfword of the second operand from the top highword of the first operand.
2. Adds the bottom halfword of the source operand with the top halfword of the second operand.
3. Saturates the results of the sum and writes a 16-bit signed integer in the range  $-2^{15} \leq x \leq 2^{15} - 1$ , where *x* equals 16, to the bottom halfword of the destination register.
4. Saturates the result of the subtraction and writes a 16-bit signed integer in the range  $-2^{15} \leq x \leq 2^{15} - 1$ , where *x* equals 16, to the top halfword of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
QASX    R7, R4, R2 ; Adds top halfword of R4 to bottom halfword of R2,
                ; saturates to 16 bits, writes to top halfword of R7
                ; Subtracts top highword of R2 from bottom halfword of
                ; R4, saturates to 16 bits and writes to bottom halfword
                ; of R7
QSAX    R0, R3, R5 ; Subtracts bottom halfword of R5 from top halfword of
                ; R3, saturates to 16 bits, writes to top halfword of R0
                ; Adds bottom halfword of R3 to top halfword of R5,
                ; saturates to 16 bits, writes to bottom halfword of R0.
```



### 12.6.7.5 QDADD and QDSUB

Saturating Double and Add and Saturating Double and Subtract, signed.

Syntax

$op\{cond\} \{Rd\}, Rm, Rn$

where:

op is one of:

QDADD Saturating Double and Add.

QDSUB Saturating Double and Subtract.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rm, Rn are registers holding the first and second operands.

Operation

The QDADD instruction:

- Doubles the second operand value.
- Adds the result of the doubling to the signed saturated value in the first operand.
- Writes the result to the destination register.

The QDSUB instruction:

- Doubles the second operand value.
- Subtracts the doubled value from the signed saturated value in the first operand.
- Writes the result to the destination register.

Both the doubling and the addition or subtraction have their results saturated to the 32-bit signed integer range  $-2^{31} \leq x \leq 2^{31} - 1$ . If saturation occurs in either operation, it sets the Q flag in the APSR.

Restrictions

Do not use SP and do not use PC.

Condition Flags

If saturation occurs, these instructions set the Q flag to 1.

Examples

```
QDADD    R7, R4, R2    ; Doubles and saturates R4 to 32 bits, adds R2,
                        ; saturates to 32 bits, writes to R7
QDSUB    R0, R3, R5    ; Subtracts R3 doubled and saturated to 32 bits
                        ; from R5, saturates to 32 bits, writes to R0.
```

### 12.6.7.6 UQASX and UQSAX

Saturating Add and Subtract with Exchange and Saturating Subtract and Add with Exchange, unsigned.

Syntax

$op\{cond\} \{Rd\}, Rm, Rn$

where:

type is one of:

UQASX Add and Subtract with Exchange and Saturate.

UQSAX Subtract and Add with Exchange and Saturate.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

The UQASX instruction:

1. Adds the bottom halfword of the source operand with the top halfword of the second operand.
2. Subtracts the bottom halfword of the second operand from the top halfword of the first operand.
3. Saturates the results of the sum and writes a 16-bit unsigned integer in the range  $0 \leq x \leq 2^{16} - 1$ , where  $x$  equals 16, to the top halfword of the destination register.
4. Saturates the result of the subtraction and writes a 16-bit unsigned integer in the range  $0 \leq x \leq 2^{16} - 1$ , where  $x$  equals 16, to the bottom halfword of the destination register.

The UQSAX instruction:

1. Subtracts the bottom halfword of the second operand from the top halfword of the first operand.
2. Adds the bottom halfword of the first operand with the top halfword of the second operand.
3. Saturates the result of the subtraction and writes a 16-bit unsigned integer in the range  $0 \leq x \leq 2^{16} - 1$ , where  $x$  equals 16, to the top halfword of the destination register.
4. Saturates the results of the addition and writes a 16-bit unsigned integer in the range  $0 \leq x \leq 2^{16} - 1$ , where  $x$  equals 16, to the bottom halfword of the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
UQASX    R7, R4, R2    ; Adds top halfword of R4 with bottom halfword of R2,
                        ; saturates to 16 bits, writes to top halfword of R7
                        ; Subtracts top halfword of R2 from bottom halfword of
                        ; R4, saturates to 16 bits, writes to bottom halfword of R7
UQSAX    R0, R3, R5    ; Subtracts bottom halfword of R5 from top halfword of R3,
                        ; saturates to 16 bits, writes to top halfword of R0
                        ; Adds bottom halfword of R4 to top halfword of R5
                        ; saturates to 16 bits, writes to bottom halfword of R0.
```

### 12.6.7.7 UQADD and UQSUB

Saturating Add and Saturating Subtract Unsigned.

Syntax

```
op{cond} {Rd}, Rn, Rm
op{cond} {Rd}, Rn, Rm
```

where:

op is one of:

UQADD8 Saturating four unsigned 8-bit integer additions.

UQADD16 Saturating two unsigned 16-bit integer additions.

UDSUB8 Saturating four unsigned 8-bit integer subtractions.

UQSUB16 Saturating two unsigned 16-bit integer subtractions.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn, Rm are registers holding the first and second operands.

Operation

These instructions add or subtract two or four values and then writes an unsigned saturated value in the destination register.

The UQADD16 instruction:

- Adds the respective top and bottom halfwords of the first and second operands.
- Saturates the result of the additions for each halfword in the destination register to the unsigned range  $0 \leq x \leq 2^{16}-1$ , where x is 16.

The UQADD8 instruction:

- Adds each respective byte of the first and second operands.
- Saturates the result of the addition for each byte in the destination register to the unsigned range  $0 \leq x \leq 2^8-1$ , where x is 8.

The UQSUB16 instruction:

- Subtracts both halfwords of the second operand from the respective halfwords of the first operand.
- Saturates the result of the differences in the destination register to the unsigned range  $0 \leq x \leq 2^{16}-1$ , where x is 16.

The UQSUB8 instructions:

- Subtracts the respective bytes of the second operand from the respective bytes of the first operand.
- Saturates the results of the differences for each byte in the destination register to the unsigned range  $0 \leq x \leq 2^8-1$ , where x is 8.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the condition code flags.

Examples

```
UQADD16 R7, R4, R2 ; Adds halfwords in R4 to corresponding halfword in R2,
                   ; saturates to 16 bits, writes to corresponding halfword of R7
UQADD8  R4, R2, R5 ; Adds bytes of R2 to corresponding byte of R5, saturates
                   ; to 8 bits, writes to corresponding bytes of R4
UQSUB16 R6, R3, R0 ; Subtracts halfwords in R0 from corresponding halfword
                   ; in R3, saturates to 16 bits, writes to corresponding
                   ; halfword in R6
UQSUB8  R1, R5, R6 ; Subtracts bytes in R6 from corresponding byte of R5,
                   ; saturates to 8 bits, writes to corresponding byte of R1.
```

### 12.6.8 Packing and Unpacking Instructions

The table below shows the instructions that operate on packing and unpacking data:

**Table 12-23. Packing and Unpacking Instructions**

Mnemonic	Description
PKH	Pack Halfword
SXTAB	Extend 8 bits to 32 and add
SXTAB16	Dual extend 8 bits to 16 and add
SXTAH	Extend 16 bits to 32 and add
SXTB	Sign extend a byte
SXTB16	Dual extend 8 bits to 16 and add
SXTH	Sign extend a halfword
UXTAB	Extend 8 bits to 32 and add
UXTAB16	Dual extend 8 bits to 16 and add
UXTAH	Extend 16 bits to 32 and add
UXTB	Zero extend a byte
UXTB16	Dual zero extend 8 bits to 16 and add
UXTH	Zero extend a halfword

### 12.6.8.1 PKHBT and PKHTB

#### Pack Halfword

##### Syntax

$op\{cond\} \{Rd\}, Rn, Rm \{, LSL \#imm\}$   
 $op\{cond\} \{Rd\}, Rn, Rm \{, ASR \#imm\}$

where:

**op** is one of:  
PKHBT Pack Halfword, bottom and top with shift.  
PKHTB Pack Halfword, top and bottom with shift.

**cond** is an optional condition code, see [“Conditional Execution”](#).

**Rd** is the destination register.

**Rn** is the first operand register

**Rm** is the second operand register holding the value to be optionally shifted.

**imm** is the shift length. The type of shift length depends on the instruction:  
For PKHBT  
LSL a left shift with a shift length from 1 to 31, 0 means no shift.  
For PKHTB  
ASR an arithmetic shift right with a shift length from 1 to 32,  
a shift of 32-bits is encoded as 0b00000.

##### Operation

The PKHBT instruction:

1. Writes the value of the bottom halfword of the first operand to the bottom halfword of the destination register.
2. If shifted, the shifted value of the second operand is written to the top halfword of the destination register.

The PKHTB instruction:

1. Writes the value of the top halfword of the first operand to the top halfword of the destination register.
2. If shifted, the shifted value of the second operand is written to the bottom halfword of the destination register.

##### Restrictions

*Rd* must not be SP and must not be PC.

##### Condition Flags

This instruction does not change the flags.

##### Examples

```
PKHBT    R3, R4, R5 LSL #0    ; Writes bottom halfword of R4 to bottom halfword of
                                ; R3, writes top halfword of R5, unshifted, to top
                                ; halfword of R3
PKHTB    R4, R0, R2 ASR #1    ; Writes R2 shifted right by 1 bit to bottom halfword
                                ; of R4, and writes top halfword of R0 to top
                                ; halfword of R4.
```

### 12.6.8.2 SXT and UXT

Sign extend and Zero extend.

Syntax

```
op{cond} {Rd}, Rm {, ROR #n}
op{cond} {Rd}, Rm {, ROR #n}
```

where:

op is one of:

SXTB Sign extends an 8-bit value to a 32-bit value.

SXTH Sign extends a 16-bit value to a 32-bit value.

SXTB16 Sign extends two 8-bit values to two 16-bit values.

UXTB Zero extends an 8-bit value to a 32-bit value.

UXTH Zero extends a 16-bit value to a 32-bit value.

UXTB16 Zero extends two 8-bit values to two 16-bit values.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rm is the register holding the value to extend.

ROR #n is one of:

ROR #8 Value from *Rm* is rotated right 8 bits.

Operation

These instructions do the following:

1. Rotate the value from *Rm* right by 0, 8, 16 or 24 bits.
2. Extract bits from the resulting value:
  - SXTB extracts bits[7:0] and sign extends to 32 bits.
  - UXTB extracts bits[7:0] and zero extends to 32 bits.
  - SXTH extracts bits[15:0] and sign extends to 32 bits.
  - UXTH extracts bits[15:0] and zero extends to 32 bits.
  - SXTB16 extracts bits[7:0] and sign extends to 16 bits, and extracts bits [23:16] and sign extends to 16 bits.
  - UXTB16 extracts bits[7:0] and zero extends to 16 bits, and extracts bits [23:16] and zero extends to 16 bits.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the flags.

Examples

```
SXTH R4, R6, ROR #16 ; Rotates R6 right by 16 bits, obtains bottom halfword of
                    ; of result, sign extends to 32 bits and writes to R4
UXTB R3, R10         ; Extracts lowest byte of value in R10, zero extends, and
                    ; writes to R3.
```

### 12.6.8.3 SXTA and UXTA

#### Signed and Unsigned Extend and Add

##### Syntax

$op\{cond\} \{Rd,\} Rn, Rm \{, ROR \#n\}$   
 $op\{cond\} \{Rd,\} Rn, Rm \{, ROR \#n\}$

where:

op is one of:

SXTAB Sign extends an 8-bit value to a 32-bit value and add.

SXTAH Sign extends a 16-bit value to a 32-bit value and add.

SXTAB16 Sign extends two 8-bit values to two 16-bit values and add.

UXTAB Zero extends an 8-bit value to a 32-bit value and add.

UXTAH Zero extends a 16-bit value to a 32-bit value and add.

UXTAB16 Zero extends two 8-bit values to two 16-bit values and add.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the first operand register.

Rm is the register holding the value to rotate and extend.

ROR #n is one of:

ROR #8 Value from *Rm* is rotated right 8 bits.

ROR #16 Value from *Rm* is rotated right 16 bits.

ROR #24 Value from *Rm* is rotated right 24 bits.

If ROR #n is omitted, no rotation is performed.

##### Operation

These instructions do the following:

1. Rotate the value from *Rm* right by 0, 8, 16 or 24 bits.
2. Extract bits from the resulting value:
  - SXTAB extracts bits[7:0] from *Rm* and sign extends to 32 bits.
  - UXTAB extracts bits[7:0] from *Rm* and zero extends to 32 bits.
  - SXTAH extracts bits[15:0] from *Rm* and sign extends to 32 bits.
  - UXTAH extracts bits[15:0] from *Rm* and zero extends to 32 bits.
  - SXTAB16 extracts bits[7:0] from *Rm* and sign extends to 16 bits, and extracts bits [23:16] from *Rm* and sign extends to 16 bits.
  - UXTAB16 extracts bits[7:0] from *Rm* and zero extends to 16 bits, and extracts bits [23:16] from *Rm* and zero extends to 16 bits.
3. Adds the signed or zero extended value to the word or corresponding halfword of *Rn* and writes the result in *Rd*.

##### Restrictions

Do not use SP and do not use PC.

##### Condition Flags

These instructions do not affect the flags.

##### Examples

```
SXTAH  R4, R8, R6, ROR #16 ; Rotates R6 right by 16 bits, obtains bottom
                             ; halfword, sign extends to 32 bits, adds
                             ; R8, and writes to R4
UXTAB  R3, R4, R10          ; Extracts bottom byte of R10 and zero extends
                             ; to 32 bits, adds R4, and writes to R3.
```

### 12.6.9 Bitfield Instructions

The table below shows the instructions that operate on adjacent sets of bits in registers or bitfields:

**Table 12-24. Packing and Unpacking Instructions**

Mnemonic	Description
BFC	Bit Field Clear
BFI	Bit Field Insert
SBFX	Signed Bit Field Extract
SXTB	Sign extend a byte
SXTH	Sign extend a halfword
UBFX	Unsigned Bit Field Extract
UXTB	Zero extend a byte
UXTH	Zero extend a halfword



### 12.6.9.1 BFC and BFI

Bit Field Clear and Bit Field Insert.

Syntax

```
BFC{cond} Rd, #lsb, #width
BFI{cond} Rd, Rn, #lsb, #width
```

where:

**cond** is an optional condition code, see [“Conditional Execution”](#).

**Rd** is the destination register.

**Rn** is the source register.

**lsb** is the position of the least significant bit of the bitfield. *lsb* must be in the range 0 to 31.

**width** is the width of the bitfield and must be in the range 1 to 32-*lsb*.

Operation

BFC clears a bitfield in a register. It clears *width* bits in *Rd*, starting at the low bit position *lsb*. Other bits in *Rd* are unchanged.

BFI copies a bitfield into one register from another register. It replaces *width* bits in *Rd* starting at the low bit position *lsb*, with *width* bits from *Rn* starting at bit[0]. Other bits in *Rd* are unchanged.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the flags.

Examples

```
BFC    R4, #8, #12    ; Clear bit 8 to bit 19 (12 bits) of R4 to 0
BFI    R9, R2, #8, #12 ; Replace bit 8 to bit 19 (12 bits) of R9 with
                        ; bit 0 to bit 11 from R2.
```

### 12.6.9.2 SBFX and UBFX

Signed Bit Field Extract and Unsigned Bit Field Extract.

Syntax

```
SBFX{cond} Rd, Rn, #lsb, #width
UBFX{cond} Rd, Rn, #lsb, #width
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rn is the source register.

lsb is the position of the least significant bit of the bitfield. *lsb* must be in the range 0 to 31.

width is the width of the bitfield and must be in the range 1 to 32-*lsb*.

Operation

SBFX extracts a bitfield from one register, sign extends it to 32 bits, and writes the result to the destination register.

UBFX extracts a bitfield from one register, zero extends it to 32 bits, and writes the result to the destination register.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the flags.

Examples

```
SBFX R0, R1, #20, #4 ; Extract bit 20 to bit 23 (4 bits) from R1 and sign
                        ; extend to 32 bits and then write the result to R0.
UBFX R8, R11, #9, #10 ; Extract bit 9 to bit 18 (10 bits) from R11 and zero
                        ; extend to 32 bits and then write the result to R8.
```

### 12.6.9.3 SXT and UXT

Sign extend and Zero extend.

Syntax

```
SXTextend{cond} {Rd}, Rm {, ROR #n}
UXTextend{cond} {Rd}, Rm {, ROR #n}
```

where:

extend is one of:

B Extends an 8-bit value to a 32-bit value.

H Extends a 16-bit value to a 32-bit value.

cond is an optional condition code, see [“Conditional Execution”](#).

Rd is the destination register.

Rm is the register holding the value to extend.

ROR #n is one of:

ROR #8 Value from *Rm* is rotated right 8 bits.

ROR #16 Value from *Rm* is rotated right 16 bits.

ROR #24 Value from *Rm* is rotated right 24 bits.

If ROR #n is omitted, no rotation is performed.

Operation

These instructions do the following:

1. Rotate the value from *Rm* right by 0, 8, 16 or 24 bits.
2. Extract bits from the resulting value:
  - SXTB extracts bits[7:0] and sign extends to 32 bits.
  - UXTB extracts bits[7:0] and zero extends to 32 bits.
  - SXTH extracts bits[15:0] and sign extends to 32 bits.
  - UXTH extracts bits[15:0] and zero extends to 32 bits.

Restrictions

Do not use SP and do not use PC.

Condition Flags

These instructions do not affect the flags.

Examples

```
SXTH R4, R6, ROR #16 ; Rotate R6 right by 16 bits, then obtain the lower
                     ; halfword of the result and then sign extend to
                     ; 32 bits and write the result to R4.
UXTB R3, R10         ; Extract lowest byte of the value in R10 and zero
                     ; extend it, and write the result to R3.
```

### 12.6.10 Branch and Control Instructions

The table below shows the branch and control instructions:

**Table 12-25. Branch and Control Instructions**

Mnemonic	Description
B	Branch
BL	Branch with Link
BLX	Branch indirect with Link
BX	Branch indirect
CBNZ	Compare and Branch if Non Zero
CBZ	Compare and Branch if Zero
IT	If-Then
TBB	Table Branch Byte
TBH	Table Branch Halfword

### 12.6.10.1 B, BL, BX, and BLX

Branch instructions.

Syntax

```
B{cond} label
BL{cond} label
BX{cond} Rm
BLX{cond} Rm
```

where:

B is branch (immediate).  
BL is branch with link (immediate).  
BX is branch indirect (register).  
BLX is branch indirect with link (register).  
cond is an optional condition code, see [“Conditional Execution”](#).  
label is a PC-relative expression. See [“PC-relative Expressions”](#).  
Rm is a register that indicates an address to branch to. Bit[0] of the value in *Rm* must be 1, but the address to branch to is created by changing bit[0] to 0.

Operation

All these instructions cause a branch to *label*, or to the address indicated in *Rm*. In addition:

- The BL and BLX instructions write the address of the next instruction to LR (the link register, R14).
- The BX and BLX instructions result in a UsageFault exception if bit[0] of *Rm* is 0.

*Bcond* label is the only conditional instruction that can be either inside or outside an IT block. All other branch instructions must be conditional inside an IT block, and must be unconditional outside the IT block, see [“IT”](#).

The table below shows the ranges for the various branch instructions.

**Table 12-26. Branch Ranges**

Instruction	Branch Range
B label	–16 MB to +16 MB
<i>Bcond</i> label (outside IT block)	–1 MB to +1 MB
<i>Bcond</i> label (inside IT block)	–16 MB to +16 MB
BL{cond} label	–16 MB to +16 MB
BX{cond} Rm	Any value in register
BLX{cond} Rm	Any value in register

The .W suffix might be used to get the maximum branch range. See [“Instruction Width Selection”](#).

Restrictions

The restrictions are:

- Do not use PC in the BLX instruction
- For BX and BLX, bit[0] of *Rm* must be 1 for correct execution but a branch occurs to the target address created by changing bit[0] to 0
- When any of these instructions is inside an IT block, it must be the last instruction of the IT block.

*Bcond* is the only conditional instruction that is not required to be inside an IT block. However, it has a longer branch range when it is inside an IT block.

## Condition Flags

These instructions do not change the flags.

### Examples

```
B      loopA      ; Branch to loopA
BLE    ng         ; Conditionally branch to label ng
B.W    target     ; Branch to target within 16MB range
BEQ    target     ; Conditionally branch to target
BEQ.W  target     ; Conditionally branch to target within 1MB
BL     funC       ; Branch with link (Call) to function funC, return address
                     ; stored in LR
BX     LR         ; Return from function call
BXNE   R0         ; Conditionally branch to address stored in R0
BLX    R0         ; Branch with link and exchange (Call) to a address stored in R0.
```

### 12.6.10.2 CBZ and CBNZ

Compare and Branch on Zero, Compare and Branch on Non-Zero.

#### Syntax

```
CBZ Rn, label
CBNZ Rn, label
```

where:

Rn is the register holding the operand.

label is the branch destination.

#### Operation

Use the CBZ or CBNZ instructions to avoid changing the condition code flags and to reduce the number of instructions.

CBZ Rn, label does not change condition flags but is otherwise equivalent to:

```
CMP     Rn, #0
BEQ     label
```

CBNZ Rn, label does not change condition flags but is otherwise equivalent to:

```
CMP     Rn, #0
BNE     label
```

#### Restrictions

The restrictions are:

- Rn must be in the range of R0 to R7
- The branch destination must be within 4 to 130 bytes after the instruction
- These instructions must not be used inside an IT block.

## Condition Flags

These instructions do not change the flags.

### Examples

```
CBZ     R5, target ; Forward branch if R5 is zero
CBNZ    R0, target ; Forward branch if R0 is not zero
```

### 12.6.10.3 IT

If-Then condition instruction.

Syntax

$IT\{x\{y\{z\}\}\} \text{ cond}$

where:

- x specifies the condition switch for the second instruction in the IT block.
- y specifies the condition switch for the third instruction in the IT block.
- z specifies the condition switch for the fourth instruction in the IT block.
- cond specifies the condition for the first instruction in the IT block.

The condition switch for the second, third and fourth instruction in the IT block can be either:

- T Then. Applies the condition *cond* to the instruction.
- E Else. Applies the inverse condition of *cond* to the instruction.

It is possible to use AL (the *always* condition) for *cond* in an IT instruction. If this is done, all of the instructions in the IT block must be unconditional, and each of x, y, and z must be T or omitted but not E.

Operation

The IT instruction makes up to four following instructions conditional. The conditions can be all the same, or some of them can be the logical inverse of the others. The conditional instructions following the IT instruction form the *IT block*.

The instructions in the IT block, including any branches, must specify the condition in the {*cond*} part of their syntax.

The assembler might be able to generate the required IT instructions for conditional instructions automatically, so that the user does not have to write them. See the assembler documentation for details.

A BKPT instruction in an IT block is always executed, even if its condition fails.

Exceptions can be taken between an IT instruction and the corresponding IT block, or within an IT block. Such an exception results in entry to the appropriate exception handler, with suitable return information in LR and stacked PSR.

Instructions designed for use for exception returns can be used as normal to return from the exception, and execution of the IT block resumes correctly. This is the only way that a PC-modifying instruction is permitted to branch to an instruction in an IT block.

Restrictions

The following instructions are not permitted in an IT block:

- IT
- CBZ and CBNZ
- CPSID and CPSIE.

Other restrictions when using an IT block are:

- A branch or any instruction that modifies the PC must either be outside an IT block or must be the last instruction inside the IT block. These are:
  - ADD PC, PC, Rm
  - MOV PC, Rm
  - B, BL, BX, BLX
  - Any LDM, LDR, or POP instruction that writes to the PC
  - TBB and TBH
- Do not branch to any instruction inside an IT block, except when returning from an exception handler
- All conditional instructions except *Bcond* must be inside an IT block. *Bcond* can be either outside or inside an IT block but has a larger branch range if it is inside one
- Each instruction inside the IT block must specify a condition code suffix that is either the same or logical inverse as for the other instructions in the block.

Your assembler might place extra restrictions on the use of IT blocks, such as prohibiting the use of assembler directives within them.

### Condition Flags

This instruction does not change the flags.

### Example

```
ITTE    NE                ; Next 3 instructions are conditional
ANDNE   R0, R0, R1        ; ANDNE does not update condition flags
ADDSNE  R2, R2, #1        ; ADDSNE updates condition flags
MOVEQ   R2, R3            ; Conditional move

CMP      R0, #9           ; Convert R0 hex value (0 to 15) into ASCII
                        ; ('0'-'9', 'A'-'F')
ITE      GT              ; Next 2 instructions are conditional
ADDGT   R1, R0, #55       ; Convert 0xA -> 'A'
ADDLE   R1, R0, #48       ; Convert 0x0 -> '0'

IT       GT              ; IT block with only one conditional instruction
ADDGT   R1, R1, #1        ; Increment R1 conditionally

ITTEE   EQ              ; Next 4 instructions are conditional
MOVEQ   R0, R1            ; Conditional move
ADDEQ   R2, R2, #10       ; Conditional add
ANDNE   R3, R3, #1        ; Conditional AND
BNE.W   dloop            ; Branch instruction can only be used in the last
                        ; instruction of an IT block

IT       NE              ; Next instruction is conditional
ADD     R0, R0, R1        ; Syntax error: no condition code used in IT block
```



#### 12.6.10.4 TBB and TBH

Table Branch Byte and Table Branch Halfword.

Syntax

```
TBB [Rn, Rm]
TBH [Rn, Rm, LSL #1]
```

where:

**Rn** is the register containing the address of the table of branch lengths.  
If *Rn* is PC, then the address of the table is the address of the byte immediately following the TBB or TBH instruction.

**Rm** is the index register. This contains an index into the table. For halfword tables, LSL #1 doubles the value in *Rm* to form the right offset into the table.

Operation

These instructions cause a PC-relative forward branch using a table of single byte offsets for TBB, or halfword offsets for TBH. *Rn* provides a pointer to the table, and *Rm* supplies an index into the table. For TBB the branch offset is twice the unsigned value of the byte returned from the table. and for TBH the branch offset is twice the unsigned value of the halfword returned from the table. The branch occurs to the address at that offset from the address of the byte immediately after the TBB or TBH instruction.

Restrictions

The restrictions are:

- *Rn* must not be SP
- *Rm* must not be SP and must not be PC
- When any of these instructions is used inside an IT block, it must be the last instruction of the IT block.

Condition Flags

These instructions do not change the flags.

Examples

```
ADR.W R0, BranchTable_Byte
TBB [R0, R1] ; R1 is the index, R0 is the base address of the
              ; branch table

Case1
; an instruction sequence follows
Case2
; an instruction sequence follows
Case3
; an instruction sequence follows
BranchTable_Byte
DCB 0 ; Case1 offset calculation
DCB ((Case2-Case1)/2) ; Case2 offset calculation
DCB ((Case3-Case1)/2) ; Case3 offset calculation

TBH [PC, R1, LSL #1] ; R1 is the index, PC is used as base of the
                     ; branch table

BranchTable_H
DCI ((CaseA - BranchTable_H)/2) ; CaseA offset calculation
DCI ((CaseB - BranchTable_H)/2) ; CaseB offset calculation
DCI ((CaseC - BranchTable_H)/2) ; CaseC offset calculation

CaseA
; an instruction sequence follows
CaseB
; an instruction sequence follows
CaseC
; an instruction sequence follows
```

## 12.6.11 Floating-point Instructions

The table below shows the floating-point instructions.

These instructions are only available if the FPU is included, and enabled, in the system. See [“Enabling the FPU”](#) for information about enabling the floating-point unit.

**Table 12-27. Floating-point Instructions**

Mnemonic	Description
VABS	Floating-point Absolute
VADD	Floating-point Add
VCMP	Compare two floating-point registers, or one floating-point register and zero
VCMPPE	Compare two floating-point registers, or one floating-point register and zero with Invalid Operation check
VCVT	Convert between floating-point and integer
VCVT	Convert between floating-point and fixed point
VCVTR	Convert between floating-point and integer with rounding
VCVTB	Converts half-precision value to single-precision
VCVTT	Converts single-precision register to half-precision
VDIV	Floating-point Divide
VFMA	Floating-point Fused Multiply Accumulate
VFNMA	Floating-point Fused Negate Multiply Accumulate
VFMS	Floating-point Fused Multiply Subtract
VFNMS	Floating-point Fused Negate Multiply Subtract
VLDM	Load Multiple extension registers
VLDR	Loads an extension register from memory
VLMA	Floating-point Multiply Accumulate
VLMS	Floating-point Multiply Subtract
VMOV	Floating-point Move Immediate
VMOV	Floating-point Move Register
VMOV	Copy ARM core register to single precision
VMOV	Copy 2 ARM core registers to 2 single precision
VMOV	Copies between ARM core register to scalar
VMOV	Copies between Scalar to ARM core register
VMRS	Move to ARM core register from floating-point System Register
VMSR	Move to floating-point System Register from ARM Core register
VMUL	Multiply floating-point
VNEG	Floating-point negate
VNMLA	Floating-point multiply and add
VNMLS	Floating-point multiply and subtract
VNMUL	Floating-point multiply
VPOP	Pop extension registers

**Table 12-27. Floating-point Instructions (Continued)**

Mnemonic	Description
VPUSH	Push extension registers
VSQRT	Floating-point square root
VSTM	Store Multiple extension registers
VSTR	Stores an extension register to memory
VSUB	Floating-point Subtract

### 12.6.11.1 VABS

Floating-point Absolute.

Syntax

`VABS{cond}.F32 Sd, Sm`

where:

`cond` is an optional condition code, see “[Conditional Execution](#)”.

`Sd, Sm` are the destination floating-point value and the operand floating-point value.

Operation

This instruction:

1. Takes the absolute value of the operand floating-point register.
2. Places the results in the destination floating-point register.

Restrictions

There are no restrictions.

Condition Flags

The floating-point instruction clears the sign bit.

Examples

`VABS.F32 S4, S6`

### 12.6.11.2 VADD

Floating-point Add

Syntax

`VADD{cond}.F32 {Sd,} Sn, Sm`

where:

`cond` is an optional condition code, see “[Conditional Execution](#)”.

`Sd,` is the destination floating-point value.

`Sn, Sm` are the operand floating-point values.

Operation

This instruction:

1. Adds the values in the two floating-point operand registers.
2. Places the results in the destination floating-point register.

Restrictions

There are no restrictions.

Condition Flags

This instruction does not change the flags.

Examples

`VADD.F32 S4, S6, S7`

### 12.6.11.3 VCMP, VCMPE

Compares two floating-point registers, or one floating-point register and zero.

Syntax

```
VCMP{E}{cond}.F32 Sd, Sm
VCMP{E}{cond}.F32 Sd, #0.0
```

where:

- cond is an optional condition code, see “Conditional Execution”.
- E If present, any NaN operand causes an Invalid Operation exception. Otherwise, only a signaling NaN causes the exception.
- Sd is the floating-point operand to compare.
- Sm is the floating-point operand that is compared with.

Operation

This instruction:

1. Compares:
  - Two floating-point registers.
  - One floating-point register and zero.
2. Writes the result to the FPSCR flags.

Restrictions

This instruction can optionally raise an Invalid Operation exception if either operand is any type of NaN. It always raises an Invalid Operation exception if either operand is a signaling NaN.

Condition Flags

When this instruction writes the result to the FPSCR flags, the values are normally transferred to the ARM flags by a subsequent VMRS instruction, see “”.

Examples

```
VCMP.F32 S4, #0.0
VCMP.F32 S4, S2
```

#### 12.6.11.4 VCVT, VCVTR between Floating-point and Integer

Converts a value in a register from floating-point to a 32-bit integer.

Syntax

$$\text{VCVT}\{R\}\{cond\}.Tm.F32\ Sd, Sm$$
$$\text{VCVT}\{cond\}.F32.Tm\ Sd, Sm$$

where:

**R** If *R* is specified, the operation uses the rounding mode specified by the FPSCR.

If *R* is omitted, the operation uses the Round towards Zero rounding mode.

**cond** is an optional condition code, see “Conditional Execution” .

**Tm** is the data type for the operand. It must be one of:

*S32 signed 32-bit value.*      *U32 unsigned 32-bit value.*

**Sd, Sm** are the destination register and the operand register.

Operation

These instructions:

1. Either
  - Converts a value in a register from floating-point value to a 32-bit integer.
  - Converts from a 32-bit integer to floating-point value.
2. Places the result in a second register.

The floating-point to integer operation normally uses the *Round towards Zero* rounding mode, but can optionally use the rounding mode specified by the FPSCR.

The integer to floating-point operation uses the rounding mode specified by the FPSCR.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.5 VCVT between Floating-point and Fixed-point

Converts a value in a register from floating-point to and from fixed-point.

Syntax

```
VCVT{cond}.Td.F32 Sd, Sd, #fbits  
VCVT{cond}.F32.Td Sd, Sd, #fbits
```

where:

**cond** is an optional condition code, see “Conditional Execution”.

**Td** is the data type for the fixed-point number. It must be one of:

- S16 signed 16-bit value.
- U16 unsigned 16-bit value.
- S32 signed 32-bit value.
- U32 unsigned 32-bit value.

**Sd** is the destination register and the operand register.

**fbits** is the number of fraction bits in the fixed-point number:

- If *Td* is S16 or U16, *fbits* must be in the range 0-16.
- If *Td* is S32 or U32, *fbits* must be in the range 1-32.

Operation

These instructions:

1. Either
  - Converts a value in a register from floating-point to fixed-point.
  - Converts a value in a register from fixed-point to floating-point.
2. Places the result in a second register.

The floating-point values are single-precision.

The fixed-point value can be 16-bit or 32-bit. Conversions from fixed-point values take their operand from the low-order bits of the source register and ignore any remaining bits.

Signed conversions to fixed-point values sign-extend the result value to the destination register width.

Unsigned conversions to fixed-point values zero-extend the result value to the destination register width.

The floating-point to fixed-point operation uses the *Round towards Zero* rounding mode. The fixed-point to floating-point operation uses the *Round to Nearest* rounding mode.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.6 VCVTB, VCVTT

Converts between a half-precision value and a single-precision value.

Syntax

```
VCVT{y}{cond}.F32.F16 Sd, Sm
VCVT{y}{cond}.F16.F32 Sd, Sm
```

where:

y Specifies which half of the operand register *Sm* or destination register *Sd* is used for the operand or destination:

- If y is B, then the bottom half, bits [15:0], of *Sm* or *Sd* is used.
- If y is T, then the top half, bits [31:16], of *Sm* or *Sd* is used.

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination register.

Sm is the operand register.

Operation

This instruction with the.F16.32 suffix:

1. Converts the half-precision value in the top or bottom half of a single-precision. register to single-precision.
2. Writes the result to a single-precision register.

This instruction with the.F32.F16 suffix:

1. Converts the value in a single-precision register to half-precision.
2. Writes the result into the top or bottom half of a single-precision register, preserving the other half of the target register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.7 VDIV

Divides floating-point values.

Syntax

```
VDIV{cond}.F32 {Sd,} Sn, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination register.

Sn, Sm are the operand registers.

Operation

This instruction:

1. Divides one floating-point value by another floating-point value.
2. Writes the result to the floating-point destination register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.



### 12.6.11.8 VFMA, VFMS

Floating-point Fused Multiply Accumulate and Subtract.

Syntax

```
VFMA{cond}.F32 {Sd}, Sn, Sm  
VFMS{cond}.F32 {Sd}, Sn, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination register.

Sn, Sm are the operand registers.

Operation

The VFMA instruction:

1. Multiplies the floating-point values in the operand registers.
2. Accumulates the results into the destination register.

The result of the multiply is not rounded before the accumulation.

The VFMS instruction:

1. Negates the first operand register.
2. Multiplies the floating-point values of the first and second operand registers.
3. Adds the products to the destination register.
4. Places the results in the destination register.

The result of the multiply is not rounded before the addition.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.9 VFNMA, VFNMS

Floating-point Fused Negate Multiply Accumulate and Subtract.

Syntax

```
VFNMA{cond}.F32 {Sd}, Sn, Sm  
VFNMS{cond}.F32 {Sd}, Sn, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination register.

Sn, Sm are the operand registers.

Operation

The VFNMA instruction:

1. Negates the first floating-point operand register.
2. Multiplies the first floating-point operand with second floating-point operand.
3. Adds the negation of the floating -point destination register to the product
4. Places the result into the destination register.

The result of the multiply is not rounded before the addition.

The VFNMS instruction:

1. Multiplies the first floating-point operand with second floating-point operand.
2. Adds the negation of the floating-point value in the destination register to the product.
3. Places the result in the destination register.

The result of the multiply is not rounded before the addition.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.10 VLDM

Floating-point Load Multiple

Syntax

`VLDM{mode}{cond}{.size} Rn{!}, list`

where:

mode	is the addressing mode: <ul style="list-style-type: none"><li>- <i>IA</i> Increment After. The consecutive addresses start at the address specified in <i>Rn</i>.</li><li>- <i>DB</i> Decrement Before. The consecutive addresses end just before the address specified in <i>Rn</i>.</li></ul>
cond	is an optional condition code, see <a href="#">“Conditional Execution”</a> .
size	is an optional data size specifier.
Rn	is the base register. The SP can be used
!	is the command to the instruction to write a modified value back to <i>Rn</i> . This is required if mode == DB, and is optional if mode == IA.
list	is the list of extension registers to be loaded, as a list of consecutively numbered doubleword or singleword registers, separated by commas and surrounded by brackets.

Operation

This instruction loads:

- Multiple extension registers from consecutive memory locations using an address from an ARM core register as the base address.

Restrictions

The restrictions are:

- If *size* is present, it must be equal to the size in bits, 32 or 64, of the registers in *list*.
- For the base address, the SP can be used.  
In the ARM instruction set, if *!* is not specified the PC can be used.
- *list* must contain at least one register. If it contains doubleword registers, it must not contain more than 16 registers.
- If using the *Decrement Before addressing* mode, the write back flag, *!*, must be appended to the base register specification.

Condition Flags

These instructions do not change the flags.

### 12.6.11.11 VLDR

Loads a single extension register from memory

Syntax

```
VLDR{cond}{.64} Dd, [Rn{#imm}]
VLDR{cond}{.64} Dd, label
VLDR{cond}{.64} Dd, [PC, #imm]
VLDR{cond}{.32} Sd, [Rn{, #imm}]
VLDR{cond}{.32} Sd, label
VLDR{cond}{.32} Sd, [PC, #imm]
```

where:

**cond** is an optional condition code, see [“Conditional Execution”](#).

**64, 32** are the optional data size specifiers.

**Dd** is the destination register for a doubleword load.

**Sd** is the destination register for a singleword load.

**Rn** is the base register. The SP can be used.

**imm** is the + or - immediate offset used to form the address.  
Permitted address values are multiples of 4 in the range 0 to 1020.

**label** is the label of the literal data item to be loaded.

Operation

This instruction:

- Loads a single extension register from memory, using a base address from an ARM core register, with an optional offset.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.12 VLMA, VLMS

Multiplies two floating-point values, and accumulates or subtracts the results.

Syntax

```
VLMA{cond}.F32 Sd, Sn, Sm  
VLMS{cond}.F32 Sd, Sn, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination floating-point value.

Sn, Sm are the operand floating-point values.

Operation

The floating-point Multiply Accumulate instruction:

1. Multiplies two floating-point values.
2. Adds the results to the destination floating-point value.

The floating-point Multiply Subtract instruction:

1. Multiplies two floating-point values.
2. Subtracts the products from the destination floating-point value.
3. Places the results in the destination register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.13 VMOV Immediate

Move floating-point Immediate

Syntax

```
VMOV{cond}.F32 Sd, #imm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the branch destination.

imm is a floating-point constant.

Operation

This instruction copies a constant value to a floating-point register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.14 VMOV Register

Copies the contents of one register to another.

Syntax

```
VMOV{cond}.F64 Dd, Dm  
VMOV{cond}.F32 Sd, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Dd is the destination register, for a doubleword operation.

Dm is the source register, for a doubleword operation.

Sd is the destination register, for a singleword operation.

Sm is the source register, for a singleword operation.

Operation

This instruction copies the contents of one floating-point register to another.

Restrictions

There are no restrictions

Condition Flags

These instructions do not change the flags.

### 12.6.11.15 VMOV Scalar to ARM Core Register

Transfers one word of a doubleword floating-point register to an ARM core register.

Syntax

`VMOV{cond} Rt, Dn[x]`

where:

**cond** is an optional condition code, see [“Conditional Execution”](#).

**Rt** is the destination ARM core register.

**Dn** is the 64-bit doubleword register.

**x** Specifies which half of the doubleword register to use:  
- If x is 0, use lower half of doubleword register  
- If x is 1, use upper half of doubleword register.

Operation

This instruction transfers:

- One word from the upper or lower half of a doubleword floating-point register to an ARM core register.

Restrictions

*Rt* cannot be PC or SP.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.16 VMOV ARM Core Register to Single Precision

Transfers a single-precision register to and from an ARM core register.

Syntax

```
VMOV{cond} Sn, Rt  
VMOV{cond} Rt, Sn
```

where:

cond is an optional condition code, see “[Conditional Execution](#)”.

Sn is the single-precision floating-point register.

Rt is the ARM core register.

Operation

This instruction transfers:

- The contents of a single-precision register to an ARM core register.
- The contents of an ARM core register to a single-precision register.

Restrictions

Rt cannot be PC or SP.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.17 VMOV Two ARM Core Registers to Two Single Precision

Transfers two consecutively numbered single-precision registers to and from two ARM core registers.

Syntax

```
VMOV{cond} Sm, Sm1, Rt, Rt2  
VMOV{cond} Rt, Rt2, Sm, Sm
```

where:

cond is an optional condition code, see “[Conditional Execution](#)”.

Sm is the first single-precision register.

Sm1 is the second single-precision register.  
This is the next single-precision register after Sm.

Rt is the ARM core register that Sm is transferred to or from.

Rt2 is the The ARM core register that Sm1 is transferred to or from.

Operation

This instruction transfers:

- The contents of two consecutively numbered single-precision registers to two ARM core registers.
- The contents of two ARM core registers to a pair of single-precision registers.

Restrictions

- The restrictions are:
- The floating-point registers must be contiguous, one after the other.
- The ARM core registers do not have to be contiguous.
- Rt cannot be PC or SP.

Condition Flags

These instructions do not change the flags.



#### 12.6.11.18 VMOV ARM Core Register to Scalar

Transfers one word to a floating-point register from an ARM core register.

##### Syntax

$VMOV\{cond\}\{.32\} Dd[x], Rt$

where:

*cond* is an optional condition code, see [“Conditional Execution”](#).

*.32* is an optional data size specifier.

*Dd[x]* is the destination, where *[x]* defines which half of the doubleword is transferred, as follows:

If *x* is 0, the lower half is extracted

If *x* is 1, the upper half is extracted.

*Rt* is the source ARM core register.

##### Operation

This instruction transfers one word to the upper or lower half of a doubleword floating-point register from an ARM core register.

##### Restrictions

*Rt* cannot be PC or SP.

##### Condition Flags

These instructions do not change the flags.

### 12.6.11.19 VMRS

Move to ARM Core register from floating-point System Register.

Syntax

```
VMRS{cond} Rt, FPSCR
VMRS{cond} APSR_nzcv, FPSCR
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Rt is the destination ARM core register. This register can be R0-R14.

APSR\_nzcv Transfer floating-point flags to the APSR flags.

Operation

This instruction performs one of the following actions:

- Copies the value of the FPSCR to a general-purpose register.
- Copies the value of the FPSCR flag bits to the APSR N, Z, C, and V flags.

Restrictions

Rt cannot be PC or SP.

Condition Flags

These instructions optionally change the flags: N, Z, C, V.

### 12.6.11.20 VMSR

Move to floating-point System Register from ARM Core register.

Syntax

```
VMSR{cond} FPSCR, Rt
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Rt is the general-purpose register to be transferred to the FPSCR.

Operation

This instruction moves the value of a general-purpose register to the FPSCR. See [“Floating-point Status Control Register”](#) for more information.

Restrictions

The restrictions are:

- Rt cannot be PC or SP.

Condition Flags

This instruction updates the FPSCR.

### 12.6.11.21 VMUL

Floating-point Multiply.

Syntax

`VMUL{cond}.F32 {Sd}, Sn, Sm`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

`Sd` is the destination floating-point value.

`Sn, Sm` are the operand floating-point values.

Operation

This instruction:

1. Multiplies two floating-point values.
2. Places the results in the destination register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.22 VNEG

Floating-point Negate.

Syntax

`VNEG{cond}.F32 Sd, Sm`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

`Sd` is the destination floating-point value.

`Sm` is the operand floating-point value.

Operation

This instruction:

1. Negates a floating-point value.
2. Places the results in a second floating-point register.

The floating-point instruction inverts the sign bit.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.23 VNMLA, VNMLS, VNMUL

Floating-point multiply with negation followed by add or subtract.

Syntax

```
VNMLA{cond}.F32 Sd, Sn, Sm
VNMLS{cond}.F32 Sd, Sn, Sm
VNMUL{cond}.F32 {Sd,} Sn, Sm
```

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination floating-point register.

Sn, Sm are the operand floating-point registers.

Operation

The VNMLA instruction:

1. Multiplies two floating-point register values.
2. Adds the negation of the floating-point value in the destination register to the negation of the product.
3. Writes the result back to the destination register.

The VNMLS instruction:

1. Multiplies two floating-point register values.
2. Adds the negation of the floating-point value in the destination register to the product.
3. Writes the result back to the destination register.

The VNMUL instruction:

1. Multiplies together two floating-point register values.
2. Writes the negation of the result to the destination register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.24 VPOP

Floating-point extension register Pop.

Syntax

`VPOP{cond}{.size} list`

where:

- cond** is an optional condition code, see “[Conditional Execution](#)”.
- size** is an optional data size specifier.  
If present, it must be equal to the size in bits, 32 or 64, of the registers in *list*.
- list** is the list of extension registers to be loaded, as a list of consecutively numbered doubleword or singleword registers, separated by commas and surrounded by brackets.

Operation

This instruction loads multiple consecutive extension registers from the stack.

Restrictions

The list must contain at least one register, and not more than sixteen registers.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.25 VPUSH

Floating-point extension register Push.

Syntax

`VPUSH{cond}{.size} list`

where:

- cond** is an optional condition code, see “[Conditional Execution](#)”.
- size** is an optional data size specifier.  
If present, it must be equal to the size in bits, 32 or 64, of the registers in *list*.
- list** is a list of the extension registers to be stored, as a list of consecutively numbered doubleword or singleword registers, separated by commas and surrounded by brackets.

Operation

This instruction:

- Stores multiple consecutive extension registers to the stack.

Restrictions

The restrictions are:

- *list* must contain at least one register, and not more than sixteen.

Condition Flags

These instructions do not change the flags.

#### 12.6.11.26 VSQRT

Floating-point Square Root.

Syntax

`VSQRT{cond}.F32 Sd, Sm`

where:

cond is an optional condition code, see [“Conditional Execution”](#).

Sd is the destination floating-point value.

Sm is the operand floating-point value.

Operation

This instruction:

- Calculates the square root of the value in a floating-point register.
- Writes the result to another floating-point register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.

### 12.6.11.27 VSTM

Floating-point Store Multiple.

Syntax

`VSTM{mode}{cond}{.size} Rn{!}, list`

where:

mode	is the addressing mode: <ul style="list-style-type: none"><li>- <i>IA</i> Increment After. The consecutive addresses start at the address specified in <i>Rn</i>. This is the default and can be omitted.</li><li>- <i>DB</i> Decrement Before. The consecutive addresses end just before the address specified in <i>Rn</i>.</li></ul>
cond	is an optional condition code, see <a href="#">“Conditional Execution”</a> .
size	is an optional data size specifier. If present, it must be equal to the size in bits, 32 or 64, of the registers in <i>list</i> .
Rn	is the base register. The SP can be used
!	is the function that causes the instruction to write a modified value back to <i>Rn</i> . Required if mode == DB.
list	is a list of the extension registers to be stored, as a list of consecutively numbered doubleword or singleword registers, separated by commas and surrounded by brackets.

Operation

This instruction:

- Stores multiple extension registers to consecutive memory locations using a base address from an ARM core register.

Restrictions

The restrictions are:

- list must contain at least one register.  
If it contains doubleword registers it must not contain more than 16 registers.
- Use of the PC as *Rn* is deprecated.

Condition Flags

These instructions do not change the flags.

### 12.6.11.28 VSTR

Floating-point Store.

Syntax

$$\text{VSTR}\{\text{cond}\}\{.32\} \text{ Sd}, [\text{Rn}\{, \#imm\}]$$
$$\text{VSTR}\{\text{cond}\}\{.64\} \text{ Dd}, [\text{Rn}\{, \#imm\}]$$

where

cond is an optional condition code, see “Conditional Execution”.

32, 64 are the optional data size specifiers.

Sd is the source register for a singleword store.

Dd is the source register for a doubleword store.

Rn is the base register. The SP can be used.

imm is the + or - immediate offset used to form the address. Values are multiples of 4 in the range 0-1020. *imm* can be omitted, meaning an offset of +0.

Operation

This instruction:

- Stores a single extension register to memory, using an address from an ARM core register, with an optional offset, defined in *imm*.

Restrictions

The restrictions are:

- The use of PC for *Rn* is deprecated.

Condition Flags

These instructions do not change the flags.

### 12.6.11.29 VSUB

Floating-point Subtract.

Syntax

$$\text{VSUB}\{\text{cond}\}.F32 \{Sd,\} Sn, Sm$$

where:

cond is an optional condition code, see “Conditional Execution”.

Sd is the destination floating-point value.

Sn, Sm are the operand floating-point value.

Operation

This instruction:

1. Subtracts one floating-point value from another floating-point value.
2. Places the results in the destination floating-point register.

Restrictions

There are no restrictions.

Condition Flags

These instructions do not change the flags.



### 12.6.12 Miscellaneous Instructions

The table below shows the remaining Cortex-M4 instructions:

**Table 12-28. Miscellaneous Instructions**

Mnemonic	Description
BKPT	Breakpoint
CPSID	Change Processor State, Disable Interrupts
CPSIE	Change Processor State, Enable Interrupts
DMB	Data Memory Barrier
DSB	Data Synchronization Barrier
ISB	Instruction Synchronization Barrier
MRS	Move from special register to register
MSR	Move from register to special register
NOP	No Operation
SEV	Send Event
SVC	Supervisor Call
WFE	Wait For Event
WFI	Wait For Interrupt

### 12.6.12.1 BKPT

Breakpoint.

Syntax

```
BKPT #imm
```

where:

*imm* is an expression evaluating to an integer in the range 0-255 (8-bit value).

Operation

The BKPT instruction causes the processor to enter Debug state. Debug tools can use this to investigate system state when the instruction at a particular address is reached.

*imm* is ignored by the processor. If required, a debugger can use it to store additional information about the breakpoint.

The BKPT instruction can be placed inside an IT block, but it executes unconditionally, unaffected by the condition specified by the IT instruction.

Condition Flags

This instruction does not change the flags.

Examples

```
BKPT 0xAB ; Breakpoint with immediate value set to 0xAB (debugger can  
          ; extract the immediate value by locating it using the PC)
```

Note: ARM does not recommend the use of the BKPT instruction with an immediate value set to 0xAB for any purpose other than Semi-hosting.

### 12.6.12.2 CPS

Change Processor State.

Syntax

*CPSeffect iflags*

where:

effect is one of:

- |    |                                      |
|----|--------------------------------------|
| IE | Clears the special purpose register. |
| ID | Sets the special purpose register.   |

iflags is a sequence of one or more flags:

- |   |                         |
|---|-------------------------|
| i | Set or clear PRIMASK.   |
| f | Set or clear FAULTMASK. |

Operation

CPS changes the PRIMASK and FAULTMASK special register values. See [“Exception Mask Registers”](#) for more information about these registers.

Restrictions

The restrictions are:

- Use CPS only from privileged software, it has no effect if used in unprivileged software
- CPS cannot be conditional and so must not be used inside an IT block.

Condition Flags

This instruction does not change the condition flags.

Examples

```
CPSID i ; Disable interrupts and configurable fault handlers (set PRIMASK)
CPSID f ; Disable interrupts and all fault handlers (set FAULTMASK)
CPSIE i ; Enable interrupts and configurable fault handlers (clear PRIMASK)
CPSIE f ; Enable interrupts and fault handlers (clear FAULTMASK)
```

### 12.6.12.3 DMB

Data Memory Barrier.

Syntax

`DMB{cond}`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

Operation

DMB acts as a data memory barrier. It ensures that all explicit memory accesses that appear, in program order, before the DMB instruction are completed before any explicit memory accesses that appear, in program order, after the DMB instruction. DMB does not affect the ordering or execution of instructions that do not access memory.

Condition Flags

This instruction does not change the flags.

Examples

```
DMB ; Data Memory Barrier
```

### 12.6.12.4 DSB

Data Synchronization Barrier.

Syntax

`DSB{cond}`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

Operation

DSB acts as a special data synchronization memory barrier. Instructions that come after the DSB, in program order, do not execute until the DSB instruction completes. The DSB instruction completes when all explicit memory accesses before it complete.

Condition Flags

This instruction does not change the flags.

Examples

```
DSB ; Data Synchronisation Barrier
```

### 12.6.12.5 ISB

Instruction Synchronization Barrier.

Syntax

`ISB{cond}`

where:

`cond` is an optional condition code, see “Conditional Execution”.

Operation

ISB acts as an instruction synchronization barrier. It flushes the pipeline of the processor, so that all instructions following the ISB are fetched from memory again, after the ISB instruction has been completed.

Condition Flags

This instruction does not change the flags.

Examples

```
ISB ; Instruction Synchronisation Barrier
```

### 12.6.12.6 MRS

Move the contents of a special register to a general-purpose register.

Syntax

`MRS{cond} Rd, spec_reg`

where:

`cond` is an optional condition code, see “Conditional Execution”.

`Rd` is the destination register.

`spec_reg` can be any of: APSR, IPSR, EPSR, IEPSR, IAPSR, EAPSR, PSR, MSP, PSP, PRIMASK, BASEPRI, BASEPRI\_MAX, FAULTMASK, or CONTROL.

Operation

Use MRS in combination with MSR as part of a read-modify-write sequence for updating a PSR, for example to clear the Q flag.

In process swap code, the programmers model state of the process being swapped out must be saved, including relevant PSR contents. Similarly, the state of the process being swapped in must also be restored. These operations use MRS in the state-saving instruction sequence and MSR in the state-restoring instruction sequence.

Note: BASEPRI\_MAX is an alias of BASEPRI when used with the MRS instruction.

See “MSR”.

Restrictions

`Rd` must not be SP and must not be PC.

Condition Flags

This instruction does not change the flags.

Examples

```
MRS R0, PRIMASK ; Read PRIMASK value and write it to R0
```

### 12.6.12.7 MSR

Move the contents of a general-purpose register into the specified special register.

Syntax

```
MSR{cond} spec_reg, Rn
```

where:

cond is an optional condition code, see “Conditional Execution” .

Rn is the source register.

spec\_reg can be any of: APSR, IPSR, EPSR, IEPSR, IAPSR, EAPSR, PSR, MSP, PSP, PRIMASK, BASEPRI, BASEPRI\_MAX, FAULTMASK, or CONTROL.

Operation

The register access operation in MSR depends on the privilege level. Unprivileged software can only access the APSR. See “Application Program Status Register” . Privileged software can access all special registers.

In unprivileged software writes to unallocated or execution state bits in the PSR are ignored.

Note: When the user writes to BASEPRI\_MAX, the instruction writes to BASEPRI only if either:  
Rn is non-zero and the current BASEPRI value is 0  
Rn is non-zero and less than the current BASEPRI value.

See “MRS” .

Restrictions

Rn must not be SP and must not be PC.

Condition Flags

This instruction updates the flags explicitly based on the value in Rn.

Examples

```
MSR CONTROL, R1 ; Read R1 value and write it to the CONTROL register
```

### 12.6.12.8 NOP

No Operation.

Syntax

```
NOP{cond}
```

where:

cond is an optional condition code, see “Conditional Execution” .

Operation

NOP does nothing. NOP is not necessarily a time-consuming NOP. The processor might remove it from the pipeline before it reaches the execution stage.

Use NOP for padding, for example to place the following instruction on a 64-bit boundary.

Condition Flags

This instruction does not change the flags.

Examples

```
NOP ; No operation
```

### 12.6.12.9 SEV

Send Event.

Syntax

`SEV{cond}`

where:

`cond` is an optional condition code, see “Conditional Execution”.

Operation

SEV is a hint instruction that causes an event to be signaled to all processors within a multiprocessor system. It also sets the local event register to 1, see “Power Management”.

Condition Flags

This instruction does not change the flags.

Examples

```
SEV ; Send Event
```

### 12.6.12.10 SVC

Supervisor Call.

Syntax

`SVC{cond} #imm`

where:

`cond` is an optional condition code, see “Conditional Execution”.

`imm` is an expression evaluating to an integer in the range 0-255 (8-bit value).

Operation

The SVC instruction causes the SVC exception.

`imm` is ignored by the processor. If required, it can be retrieved by the exception handler to determine what service is being requested.

Condition Flags

This instruction does not change the flags.

Examples

```
SVC 0x32 ; Supervisor Call (SVC handler can extract the immediate value  
; by locating it via the stacked PC)
```

### 12.6.12.11 WFE

Wait For Event.

Syntax

`WFE{cond}`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

Operation

WFE is a hint instruction.

If the event register is 0, WFE suspends execution until one of the following events occurs:

- An exception, unless masked by the exception mask registers or the current priority level
- An exception enters the Pending state, if SEVONPEND in the System Control Register is set
- A Debug Entry request, if Debug is enabled
- An event signaled by a peripheral or another processor in a multiprocessor system using the SEV instruction.

If the event register is 1, WFE clears it to 0 and returns immediately.

For more information, see [“Power Management”](#).

Condition Flags

This instruction does not change the flags.

Examples

```
WFE ; Wait for event
```

### 12.6.12.12 WFI

Wait for Interrupt.

Syntax

`WFI{cond}`

where:

`cond` is an optional condition code, see [“Conditional Execution”](#).

Operation

WFI is a hint instruction that suspends execution until one of the following events occurs:

- An exception
- A Debug Entry request, regardless of whether Debug is enabled.

Condition Flags

This instruction does not change the flags.

Examples

```
WFI ; Wait for interrupt
```



## 12.7 Cortex-M4 Core Peripherals

### 12.7.1 Peripherals

- **Nested Vectored Interrupt Controller (NVIC)**  
The Nested Vectored Interrupt Controller (NVIC) is an embedded interrupt controller that supports low latency interrupt processing. See [Section 12.8 "Nested Vectored Interrupt Controller \(NVIC\)"](#)
- **System Control Block (SCB)**  
The System Control Block (SCB) is the programmers model interface to the processor. It provides system implementation information and system control, including configuration, control, and reporting of system exceptions. See [Section 12.9 "System Control Block \(SCB\)"](#)
- **System Timer (SysTick)**  
The System Timer, SysTick, is a 24-bit count-down timer. Use this as a Real Time Operating System (RTOS) tick timer or as a simple counter. See [Section 12.10 "System Timer \(SysTick\)"](#)
- **Memory Protection Unit (MPU)**  
The Memory Protection Unit (MPU) improves system reliability by defining the memory attributes for different memory regions. It provides up to eight different regions, and an optional predefined background region. See [Section 12.11 "Memory Protection Unit \(MPU\)"](#)
- **Floating-point Unit (FPU)**  
The Floating-point Unit (FPU) provides IEEE754-compliant operations on single-precision, 32-bit, floating-point values. See [Section 12.12 "Floating Point Unit \(FPU\)"](#)

### 12.7.2 Address Map

The address map of the *Private peripheral bus* (PPB) is:

**Table 12-29. Core Peripheral Register Regions**

Address	Core Peripheral
0xE000E008-0xE000E00F	System Control Block
0xE000E010-0xE000E01F	System Timer
0xE000E100-0xE000E4EF	Nested Vectored Interrupt Controller
0xE000ED00-0xE000ED3F	System control block
0xE000ED90-0xE000EDB8	Memory Protection Unit
0xE000EF00-0xE000EF03	Nested Vectored Interrupt Controller
0xE000EF30-0xE000EF44	Floating-point Unit

In register descriptions:

- The *required privilege* gives the privilege level required to access the register, as follows:
  - Privileged: Only privileged software can access the register.
  - Unprivileged: Both unprivileged and privileged software can access the register.

## 12.8 Nested Vectored Interrupt Controller (NVIC)

This section describes the NVIC and the registers it uses. The NVIC supports:

- 1 to 38 interrupts.
- A programmable priority level of 0-15 for each interrupt. A higher level corresponds to a lower priority, so level 0 is the highest interrupt priority.
- Level detection of interrupt signals.
- Dynamic reprioritization of interrupts.
- Grouping of priority values into group priority and subpriority fields.
- Interrupt tail-chaining.
- An external *Non-maskable interrupt* (NMI)

The processor automatically stacks its state on exception entry and unstacks this state on exception exit, with no instruction overhead. This provides low latency exception handling.

### 12.8.1 Level-sensitive Interrupts

The processor supports level-sensitive interrupts. A level-sensitive interrupt is held asserted until the peripheral deasserts the interrupt signal. Typically, this happens because the ISR accesses the peripheral, causing it to clear the interrupt request.

When the processor enters the ISR, it automatically removes the pending state from the interrupt (see [“Hardware and Software Control of Interrupts”](#)). For a level-sensitive interrupt, if the signal is not deasserted before the processor returns from the ISR, the interrupt becomes pending again, and the processor must execute its ISR again. This means that the peripheral can hold the interrupt signal asserted until it no longer requires servicing.

#### 12.8.1.1 Hardware and Software Control of Interrupts

The Cortex-M4 latches all interrupts. A peripheral interrupt becomes pending for one of the following reasons:

- The NVIC detects that the interrupt signal is HIGH and the interrupt is not active
- The NVIC detects a rising edge on the interrupt signal
- A software writes to the corresponding interrupt set-pending register bit, see [“Interrupt Set-pending Registers”](#), or to the NVIC\_STIR register to make an interrupt pending, see [“Software Trigger Interrupt Register”](#).

A pending interrupt remains pending until one of the following:

- The processor enters the ISR for the interrupt. This changes the state of the interrupt from pending to active. Then:
  - For a level-sensitive interrupt, when the processor returns from the ISR, the NVIC samples the interrupt signal. If the signal is asserted, the state of the interrupt changes to pending, which might cause the processor to immediately re-enter the ISR. Otherwise, the state of the interrupt changes to inactive.
- Software writes to the corresponding interrupt clear-pending register bit.  
For a level-sensitive interrupt, if the interrupt signal is still asserted, the state of the interrupt does not change. Otherwise, the state of the interrupt changes to inactive.

### 12.8.2 NVIC Design Hints and Tips

Ensure that the software uses correctly aligned register accesses. The processor does not support unaligned accesses to NVIC registers. See the individual register descriptions for the supported access sizes.

An interrupt can enter a pending state even if it is disabled. Disabling an interrupt only prevents the processor from taking that interrupt.

Before programming SCB\_VTOR to relocate the vector table, ensure that the vector table entries of the new vector table are set up for fault handlers, NMI and all enabled exception like interrupts. For more information, see the [“Vector Table Offset Register”](#).

### 12.8.2.1 NVIC Programming Hints

The software uses the CPSIE I and CPSID I instructions to enable and disable the interrupts. The CMSIS provides the following intrinsic functions for these instructions:

```
void __disable_irq(void) // Disable Interrupts
```

```
void __enable_irq(void) // Enable Interrupts
```

In addition, the CMSIS provides a number of functions for NVIC control, including:

**Table 12-30. CMSIS Functions for NVIC Control**

CMSIS Interrupt Control Function	Description
void NVIC_SetPriorityGrouping(uint32_t priority_grouping)	Set the priority grouping
void NVIC_EnableIRQ(IRQn_t IRQn)	Enable IRQn
void NVIC_DisableIRQ(IRQn_t IRQn)	Disable IRQn
uint32_t NVIC_GetPendingIRQ (IRQn_t IRQn)	Return true (IRQ-Number) if IRQn is pending
void NVIC_SetPendingIRQ (IRQn_t IRQn)	Set IRQn pending
void NVIC_ClearPendingIRQ (IRQn_t IRQn)	Clear IRQn pending status
uint32_t NVIC_GetActive (IRQn_t IRQn)	Return the IRQ number of the active interrupt
void NVIC_SetPriority (IRQn_t IRQn, uint32_t priority)	Set priority for IRQn
uint32_t NVIC_GetPriority (IRQn_t IRQn)	Read priority of IRQn
void NVIC_SystemReset (void)	Reset the system

The input parameter IRQn is the IRQ number. For more information about these functions, see the CMSIS documentation.

To improve software efficiency, the CMSIS simplifies the NVIC register presentation. In the CMSIS:

- The Set-enable, Clear-enable, Set-pending, Clear-pending and Active Bit registers map to arrays of 32-bit integers, so that:
  - The array ISER[0] corresponds to the registers ISER0
  - The array ICER[0] corresponds to the registers ICER0
  - The array ISPR[0] corresponds to the registers ISPR0
  - The array ICPR[0] corresponds to the registers ICPR0
  - The array IABR[0] corresponds to the registers IABR0
- The 4-bit fields of the Interrupt Priority Registers map to an array of 4-bit integers, so that the array IP[0] to IP[37] corresponds to the registers IPR0-IPR9, and the array entry IP[n] holds the interrupt priority for interrupt *n*.

The CMSIS provides thread-safe code that gives atomic access to the Interrupt Priority Registers. [Table 12-31](#) shows how the interrupts, or IRQ numbers, map onto the interrupt registers and corresponding CMSIS variables that have one bit per interrupt.

**Table 12-31. Mapping of Interrupts to the Interrupt Variables**

Interrupts	CMSIS Array Elements <sup>(1)</sup>				
	Set-enable	Clear-enable	Set-pending	Clear-pending	Active Bit
0-37	ISER[0]	ICER[0]	ISPR[0]	ICPR[0]	IABR[0]

Note: 1. Each array element corresponds to a single NVIC register, for example the ICER[0] element corresponds to the ICER0 register.

### 12.8.3 Nested Vectored Interrupt Controller (NVIC) User Interface

Table 12-32. Nested Vectored Interrupt Controller (NVIC) Register Mapping

Offset	Register	Name	Access	Reset
0xE000E100	Interrupt Set-enable Register 0	NVIC_ISER0	Read-write	0x00000000
...	...	...	...	...
0xE000E11C	Interrupt Set-enable Register 7	NVIC_ISER7	Read-write	0x00000000
0xE000E180	Interrupt Clear-enable Register 0	NVIC_ICER0	Read-write	0x00000000
...	...	...	...	...
0xE000E19C	Interrupt Clear-enable Register 7	NVIC_ICER7	Read-write	0x00000000
0xE000E200	Interrupt Set-pending Register 0	NVIC_ISPR0	Read-write	0x00000000
...	...	...	...	...
0xE000E21C	Interrupt Set-pending Register 7	NVIC_ISPR7	Read-write	0x00000000
0xE000E280	Interrupt Clear-pending Register 0	NVIC_ICPR0	Read-write	0x00000000
...	...	...	...	...
0xE000E29C	Interrupt Clear-pending Register 7	NVIC_ICPR7	Read-write	0x00000000
0xE000E300	Interrupt Active Bit Register 0	NVIC_IABR0	Read-write	0x00000000
...	...	...	...	...
0xE000E31C	Interrupt Active Bit Register 7	NVIC_IABR7	Read-write	0x00000000
0xE000E400	Interrupt Priority Register 0	NVIC_IPR0	Read-write	0x00000000
...	...	...	...	...
24	Interrupt Priority Register 9	NVIC_IPR9	Read-write	0x00000000
0xE000EF00	Software Trigger Interrupt Register	NVIC_STIR	Write-only	0x00000000

### 12.8.3.1 Interrupt Set-enable Registers

**Name:** NVIC\_ISERx [x=0..7]

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
SETENA							
23	22	21	20	19	18	17	16
SETENA							
15	14	13	12	11	10	9	8
SETENA							
7	6	5	4	3	2	1	0
SETENA							

These registers enable interrupts and show which interrupts are enabled.

- **SETENA: Interrupt Set-enable**

Write:

0: No effect.

1: Enables the interrupt.

Read:

0: Interrupt disabled.

1: Interrupt enabled.

Notes: 1. If a pending interrupt is enabled, the NVIC activates the interrupt based on its priority.

2. If an interrupt is not enabled, asserting its interrupt signal changes the interrupt state to pending, the NVIC never activates the interrupt, regardless of its priority.

### 12.8.3.2 Interrupt Clear-enable Registers

**Name:** NVIC\_ICERx [x=0..7]

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
CLRENA							
23	22	21	20	19	18	17	16
CLRENA							
15	14	13	12	11	10	9	8
CLRENA							
7	6	5	4	3	2	1	0
CLRENA							

These registers disable interrupts, and show which interrupts are enabled.

- **CLRENA: Interrupt Clear-enable**

Write:

0: No effect.

1: Disables the interrupt.

Read:

0: Interrupt disabled.

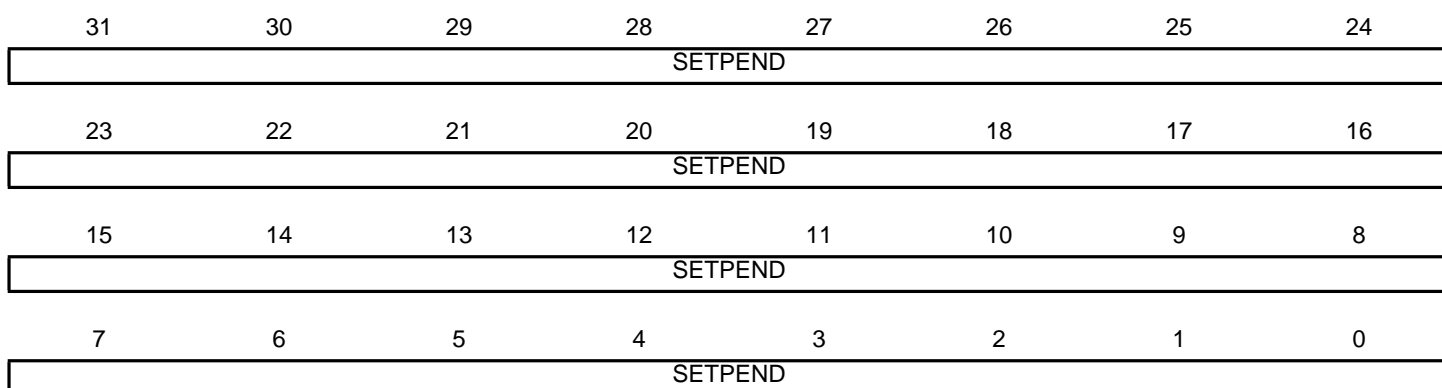
1: Interrupt enabled.

### 12.8.3.3 Interrupt Set-pending Registers

**Name:** NVIC\_ISPRx [x=0..7]

**Access:** Read-write

**Reset:** 0x00000000



These registers force interrupts into the pending state, and show which interrupts are pending.

- **SETPEND: Interrupt Set-pending**

Write:

0: No effect.

1: Changes the interrupt state to pending.

Read:

0: Interrupt is not pending.

1: Interrupt is pending.

Notes: 1. Writing 1 to an ISPR bit corresponding to an interrupt that is pending has no effect.

2. Writing 1 to an ISPR bit corresponding to a disabled interrupt sets the state of that interrupt to pending.

#### 12.8.3.4 Interrupt Clear-pending Registers

**Name:** NVIC\_ICPRx [x=0..7]

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
CLRPEND							
23	22	21	20	19	18	17	16
CLRPEND							
15	14	13	12	11	10	9	8
CLRPEND							
7	6	5	4	3	2	1	0
CLRPEND							

These registers remove the pending state from interrupts, and show which interrupts are pending.

- **CLRPEND: Interrupt Clear-pending**

Write:

0: No effect.

1: Removes the pending state from an interrupt.

Read:

0: Interrupt is not pending.

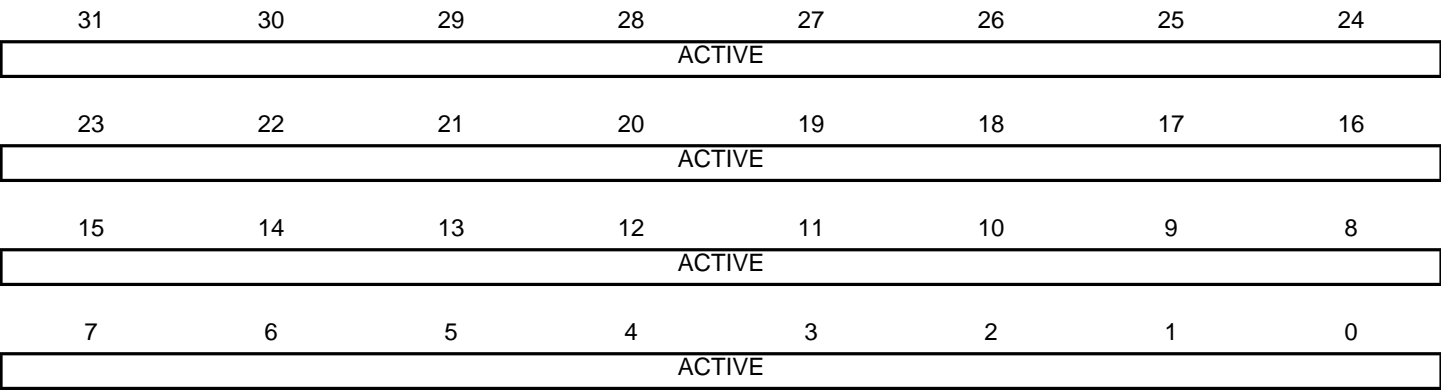
1: Interrupt is pending.

**Note:** Writing 1 to an ICPR bit does not affect the active state of the corresponding interrupt.



12.8.3.5 Interrupt Active Bit Registers

**Name:** NVIC\_IABRx [x=0..7]  
**Access:** Read-write  
**Reset:** 0x00000000



These registers indicate which interrupts are active.

• **ACTIVE: Interrupt Active Flags**

- 0: Interrupt is not active.
- 1: Interrupt is active.

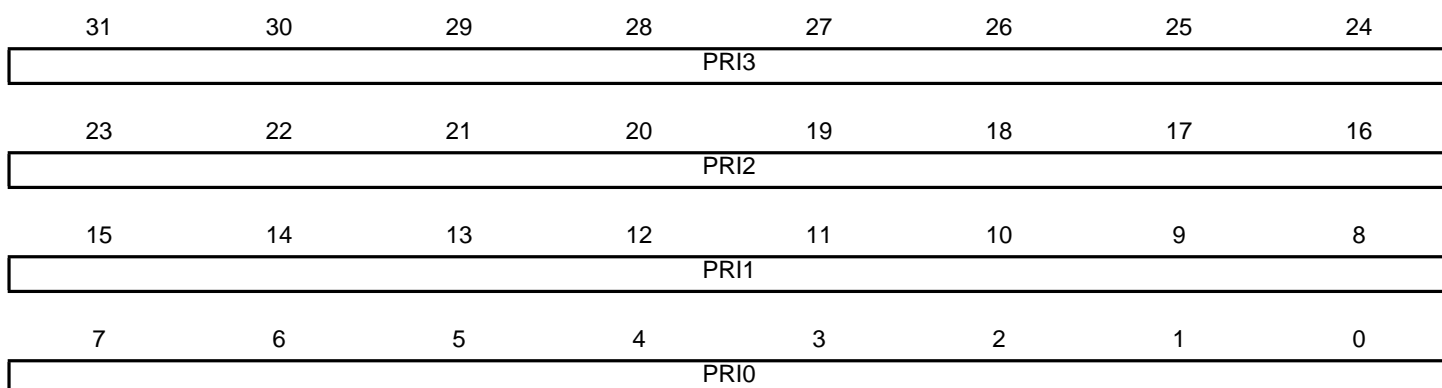
**Note:** A bit reads as one if the status of the corresponding interrupt is active, or active and pending.

### 12.8.3.6 Interrupt Priority Registers

**Name:** NVIC\_IPRx [x=0..9]

**Access:** Read-write

**Reset:** 0x00000000



The NVIC\_IPR0-NVIC\_IPR9 registers provide a 4-bit priority field for each interrupt. These registers are byte-accessible. Each register holds four priority fields, that map up to four elements in the CMSIS interrupt priority array IP[0] to IP[37]

- **PRI3: Priority (4m+3)**

Priority, Byte Offset 3, refers to register bits [31:24].

- **PRI2: Priority (4m+2)**

Priority, Byte Offset 2, refers to register bits [23:16].

- **PRI1: Priority (4m+1)**

Priority, Byte Offset 1, refers to register bits [15:8].

- **PRI0: Priority (4m)**

Priority, Byte Offset 0, refers to register bits [7:0].

- Notes:
1. Each priority field holds a priority value, 0-15. The lower the value, the greater the priority of the corresponding interrupt. The processor implements only bits[7:4] of each field; bits[3:0] read as zero and ignore writes.
  2. for more information about the IP[0] to IP[37] interrupt priority array, that provides the software view of the interrupt priorities, see [Table 12-30, "CMSIS Functions for NVIC Control"](#).
  3. The corresponding IPR number  $n$  is given by  $n = m \text{ DIV } 4$ .
  4. The byte offset of the required Priority field in this register is  $m \text{ MOD } 4$ .

### 12.8.3.7 Software Trigger Interrupt Register

**Name:** NVIC\_STIR

**Access:** Write-only

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	INTID
7	6	5	4	3	2	1	0
INTID							

Write to this register to generate an interrupt from the software.

- **INTID: Interrupt ID**

Interrupt ID of the interrupt to trigger, in the range 0-239. For example, a value of 0x03 specifies interrupt IRQ3.

## 12.9 System Control Block (SCB)

The System Control Block (SCB) provides system implementation information, and system control. This includes configuration, control, and reporting of the system exceptions.

Ensure that the software uses aligned accesses of the correct size to access the system control block registers:

- Except for the SCB\_CFSR and SCB\_SHPR1-SCB\_SHPR3 registers, it must use aligned word accesses
- For the SCB\_CFSR and SCB\_SHPR1-SCB\_SHPR3 registers, it can use byte or aligned halfword or word accesses.

The processor does not support unaligned accesses to system control block registers.

In a fault handler, to determine the true faulting address:

1. Read and save the MMFAR or SCB\_BFAR value.
2. Read the MMARVALID bit in the MMFSR subregister, or the BFARVALID bit in the BFSR subregister. The SCB\_MMFAR or SCB\_BFAR address is valid only if this bit is 1.

The software must follow this sequence because another higher priority exception might change the SCB\_MMFAR or SCB\_BFAR value. For example, if a higher priority handler preempts the current fault handler, the other fault might change the SCB\_MMFAR or SCB\_BFAR value.

## 12.9.1 System Control Block (SCB) User Interface

**Table 12-33. System Control Block (SCB) Register Mapping**

Offset	Register	Name	Access	Reset
0xE000E008	Auxiliary Control Register	SCB_ACTLR	Read-write	0x00000000
0xE000ED00	CPUID Base Register	SCB_CPUID	Read-only	0x410FC240
0xE000ED04	Interrupt Control and State Register	SCB_ICSR	Read-write <sup>(1)</sup>	0x00000000
0xE000ED08	Vector Table Offset Register	SCB_VTOR	Read-write	0x00000000
0xE000ED0C	Application Interrupt and Reset Control Register	SCB_AIRCR	Read-write	0xFA050000
0xE000ED10	System Control Register	SCB_SCR	Read-write	0x00000000
0xE000ED14	Configuration and Control Register	SCB_CCR	Read-write	0x00000200
0xE000ED18	System Handler Priority Register 1	SCB_SHPR1	Read-write	0x00000000
0xE000ED1C	System Handler Priority Register 2	SCB_SHPR2	Read-write	0x00000000
0xE000ED20	System Handler Priority Register 3	SCB_SHPR3	Read-write	0x00000000
0xE000ED24	System Handler Control and State Register	SCB_SHCSR	Read-write	0x00000000
0xE000ED28	Configurable Fault Status Register	SCB_CFSR <sup>(2)</sup>	Read-write	0x00000000
0xE000ED2C	HardFault Status Register	SCB_HFSR	Read-write	0x00000000
0xE000ED34	MemManage Fault Address Register	SCB_MM FAR	Read-write	Unknown
0xE000ED38	BusFault Address Register	SCB_BFAR	Read-write	Unknown
0xE000ED3C	Auxiliary Fault Status Register	SCB_AFSR	Read-write	0x00000000

Notes: 1. See the register description for more information.

2. This register contains the subregisters: “[MMFSR: Memory Management Fault Status Subregister](#)” (0xE000ED28 - 8 bits), “[BFSR: Bus Fault Status Subregister](#)” (0xE000ED29 - 8 bits), “[UFSR: Usage Fault Status Subregister](#)” (0xE000ED2A - 16 bits).

### 12.9.1.1 Auxiliary Control Register

**Name:** SCB\_ACTLR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–							
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
–						DISOFP	DISFPCA
7	6	5	4	3	2	1	0
–					DISFOLD	DISDEFWBUF	DISMCYCINT

The SCB\_ACTLR register provides disable bits for the following processor functions:

- IT folding
- Write buffer use for accesses to the default memory map
- Interruption of multi-cycle instructions.

By default, this register is set to provide optimum performance from the Cortex-M4 processor, and does not normally require modification.

- **DISOFP: Disable Out Of Order Floating Point**

Disables floating point instructions that complete out of order with respect to integer instructions.

- **DISFPCA: Disable FPCA**

Disables an automatic update of CONTROL.FPCA.

- **DISFOLD: Disable Folding**

When set to 1, disables the IT folding.

**Note:** In some situations, the processor can start executing the first instruction in an IT block while it is still executing the IT instruction. This behavior is called IT folding, and it improves the performance. However, IT folding can cause jitter in looping. If a task must avoid jitter, set the DISFOLD bit to 1 before executing the task, to disable the IT folding.

- **DISDEFWBUF: Disable Default Write Buffer**

When set to 1, it disables the write buffer use during default memory map accesses. This causes BusFault to be precise but decreases the performance, as any store to memory must complete before the processor can execute the next instruction.

This bit only affects write buffers implemented in the Cortex-M4 processor.

- **DISMCYCINT: Disable Multiple Cycle Interruption**

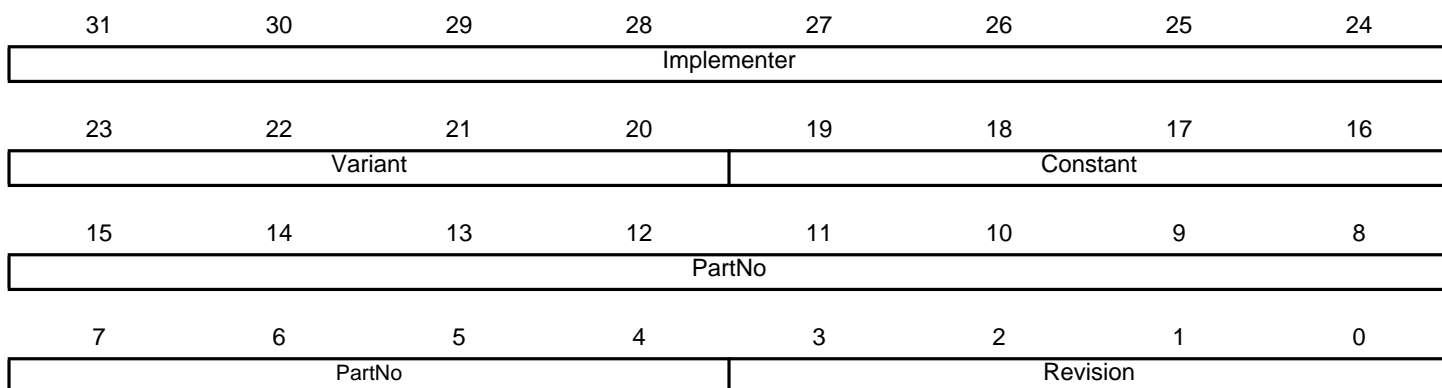
When set to 1, it disables the interruption of load multiple and store multiple instructions. This increases the interrupt latency of the processor, as any LDM or STM must complete before the processor can stack the current state and enter the interrupt handler.

### 12.9.1.2 CPUID Base Register

**Name:** SCB\_CPUID

**Access:** Read-write

**Reset:** 0x00000000



The SCB\_CPUID register contains the processor part number, version, and implementation information.

- **Implementer: Implementer Code**

0x41: ARM.

- **Variant: Variant Number**

It is the r value in the rnpn product revision identifier:

0x0: Revision 0.

- **Constant**

Reads as 0xF.

- **PartNo: Part Number of the Processor**

0xC24 = Cortex-M4.

- **Revision: Revision Number**

It is the p value in the rnpn product revision identifier:

0x0: Patch 0.

### 12.9.1.3 Interrupt Control and State Register

**Name:** SCB\_ICSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
NMIPENDSET	–		PENDSVSET	PENDSVCLR	PENDSTSET	PENDSTCLR	–
23	22	21	20	19	18	17	16
–	ISRPENDING		VECTPENDING				
15	14	13	12	11	10	9	8
VECTPENDING				RETTOBASE	–		VECTACTIVE
7	6	5	4	3	2	1	0
VECTACTIVE							

The SCB\_ICSR register provides a set-pending bit for the Non-Maskable Interrupt (NMI) exception, and set-pending and clear-pending bits for the PendSV and SysTick exceptions.

It indicates:

- The exception number of the exception being processed, and whether there are preempted active exceptions,
- The exception number of the highest priority pending exception, and whether any interrupts are pending.

#### • NMIPENDSET: NMI Set-pending

Write:

PendSV set-pending bit.

Write:

0: No effect.

1: Changes NMI exception state to pending.

Read:

0: NMI exception is not pending.

1: NMI exception is pending.

As NMI is the highest-priority exception, the processor normally enters the NMI exception handler as soon as it registers a write of 1 to this bit. Entering the handler clears this bit to 0. A read of this bit by the NMI exception handler returns 1 only if the NMI signal is reasserted while the processor is executing that handler.

#### • PENDSVSET: PendSV Set-pending

Write:

0: No effect.

1: Changes PendSV exception state to pending.

Read:

0: PendSV exception is not pending.

1: PendSV exception is pending.

Writing 1 to this bit is the only way to set the PendSV exception state to pending.



- **PENDSVCLR: PendSV Clear-pending**

Write:

0: No effect.

1: Removes the pending state from the PendSV exception.

- **PENDSTSET: SysTick Exception Set-pending**

Write:

0: No effect.

1: Changes SysTick exception state to pending.

Read:

0: SysTick exception is not pending.

1: SysTick exception is pending.

- **PENDSTCLR: SysTick Exception Clear-pending**

Write:

0: No effect.

1: Removes the pending state from the SysTick exception.

This bit is Write-only. On a register read, its value is Unknown.

- **ISRPENDING: Interrupt Pending Flag (Excluding NMI and Faults)**

0: Interrupt not pending.

1: Interrupt pending.

- **VECTPENDING: Exception Number of the Highest Priority Pending Enabled Exception**

0: No pending exceptions.

Nonzero: The exception number of the highest priority pending enabled exception.

The value indicated by this field includes the effect of the BASEPRI and FAULTMASK registers, but not any effect of the PRIMASK register.

- **RETTOBASE: Preempted Active Exceptions Present or Not**

0: There are preempted active exceptions to execute.

1: There are no active exceptions, or the currently-executing exception is the only active exception.

- **VECTACTIVE: Active Exception Number Contained**

0: Thread mode.

Nonzero: The exception number of the currently active exception. The value is the same as IPSR bits [8:0]. See [“Interrupt Program Status Register”](#).

Subtract 16 from this value to obtain the IRQ number required to index into the Interrupt Clear-Enable, Set-Enable, Clear-Pending, Set-Pending, or Priority Registers, see [“Interrupt Program Status Register”](#).

Note: When the user writes to the SCB\_ICSR register, the effect is unpredictable if:

- Writing 1 to the PENDSVSET bit and writing 1 to the PENDSVCLR bit
- Writing 1 to the PENDSTSET bit and writing 1 to the PENDSTCLR bit.

### 12.9.1.4 Vector Table Offset Register

**Name:** SCB\_VTOR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
TBLOFF							
23	22	21	20	19	18	17	16
TBLOFF							
15	14	13	12	11	10	9	8
TBLOFF							
7	6	5	4	3	2	1	0
TBLOFF	–						

The SCB\_VTOR register indicates the offset of the vector table base address from memory address 0x00000000.

- **TBLOFF: Vector Table Base Offset**

It contains bits [29:7] of the offset of the table base from the bottom of the memory map.

Bit [29] determines whether the vector table is in the code or SRAM memory region:

0: Code.

1: SRAM.

It is sometimes called the TBLBASE bit.

**Note:** When setting TBLOFF, the offset must be aligned to the number of exception entries in the vector table. Configure the next statement to give the information required for your implementation; the statement reminds the user of how to determine the alignment requirement. The minimum alignment is 32 words, enough for up to 16 interrupts. For more interrupts, adjust the alignment by rounding up to the next power of two. For example, if 21 interrupts are required, the alignment must be on a 64-word boundary because the required table size is 37 words, and the next power of two is 64.

Table alignment requirements mean that bits[6:0] of the table offset are always zero.

### 12.9.1.5 Application Interrupt and Reset Control Register

**Name:** SCB\_AIRCR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
VECTKEYSTAT/VECTKEY							
23	22	21	20	19	18	17	16
VECTKEYSTAT/VECTKEY							
15	14	13	12	11	10	9	8
ENDIANNESS	–				PRIGROUP		
7	6	5	4	3	2	1	0
–					SYSRESETREQ	VECTCLRACTI VE	VECTRESET

The SCB\_AIRCR register provides priority grouping control for the exception model, endian status for data accesses, and reset control of the system. To write to this register, write 0x5FA to the VECTKEY field, otherwise the processor ignores the write.

- **VECTKEYSTAT: Register Key**

Read:

Reads as 0xFA05.

- **VECTKEY: Register Key**

Write:

Writes 0x5FA to VECTKEY, otherwise the write is ignored.

- **ENDIANNESS: Data Endianness**

0: Little-endian.

1: Big-endian.

- **PRIGROUP: Interrupt Priority Grouping**

This field determines the split of group priority from subpriority. It shows the position of the binary point that splits the PRI\_n fields in the Interrupt Priority Registers into separate *group priority* and *subpriority* fields. The table below shows how the PRIGROUP value controls this split:

PRIGROUP	Interrupt Priority Level Value, PRI_M[7:0]			Number of	
	Binary Point <sup>(1)</sup>	Group Priority Bits	Subpriority Bits	Group Priorities	Subpriorities
0b000	bxxxxxxx.y	[7:1]	None	128	2
0b001	bxxxxxx.yy	[7:2]	[4:0]	64	4
0b010	bxxxxx.yyy	[7:3]	[4:0]	32	8
0b011	bxxxx.yyyy	[7:4]	[4:0]	16	16
0b100	bxxx.yyyyy	[7:5]	[4:0]	8	32

PRIGROUP	Interrupt Priority Level Value, PRI_M[7:0]			Number of	
	Binary Point <sup>(1)</sup>	Group Priority Bits	Subpriority Bits	Group Priorities	Subpriorities
0b101	bxx.yyyyyy	[7:6]	[5:0]	4	64
0b110	bx.yyyyyyy	[7]	[6:0]	2	128
0b111	b.yyyyyyy	None	[7:0]	1	256

Note: 1. PRI\_n[7:0] field showing the binary point. x denotes a group priority field bit, and y denotes a subpriority field bit. Determining preemption of an exception uses only the group priority field.

- **SYSRESETREQ: System Reset Request**

0: No system reset request.

1: Asserts a signal to the outer system that requests a reset.

This is intended to force a large system reset of all major components except for debug. This bit reads as 0.

- **VECTCLRACTIVE**

Reserved for Debug use. This bit reads as 0. When writing to the register, write 0 to this bit, otherwise the behavior is unpredictable.

- **VECTRESET**

Reserved for Debug use. This bit reads as 0. When writing to the register, write 0 to this bit, otherwise the behavior is unpredictable.

### 12.9.1.6 System Control Register

**Name:** SCB\_SCR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–							
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
–			SEVONPEND	–	SLEEPDEEP	SLEEPONEXIT	–

- **SEVONPEND: Send Event on Pending Bit**

0: Only enabled interrupts or events can wake up the processor; disabled interrupts are excluded.

1: Enabled events and all interrupts, including disabled interrupts, can wake up the processor.

When an event or an interrupt enters the pending state, the event signal wakes up the processor from WFE. If the processor is not waiting for an event, the event is registered and affects the next WFE.

The processor also wakes up on execution of an SEV instruction or an external event.

- **SLEEPDEEP: Sleep or Deep Sleep**

Controls whether the processor uses sleep or deep sleep as its low power mode:

0: Sleep.

1: Deep sleep.

- **SLEEPONEXIT: Sleep-on-exit**

Indicates sleep-on-exit when returning from the Handler mode to the Thread mode:

0: Do not sleep when returning to Thread mode.

1: Enter sleep, or deep sleep, on return from an ISR.

Setting this bit to 1 enables an interrupt-driven application to avoid returning to an empty main application.

### 12.9.1.7 Configuration and Control Register

**Name:** SCB\_CCR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–							
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
–						STKALIGN	BFHFNMIGN
7	6	5	4	3	2	1	0
–			DIV_0_TRP	UNALIGN_TRP	–	USERSETMPE ND	NONBASETHR DENA

The SCB\_CCR register controls the entry to the Thread mode and enables the handlers for NMI, hard fault and faults escalated by FAULTMASK to ignore BusFaults. It also enables the division by zero and unaligned access trapping, and the access to the NVIC\_STIR register by unprivileged software (see [“Software Trigger Interrupt Register”](#) ).

- **STKALIGN: Stack Alignment**

Indicates the stack alignment on exception entry:

0: 4-byte aligned.

1: 8-byte aligned.

On exception entry, the processor uses bit [9] of the stacked PSR to indicate the stack alignment. On return from the exception, it uses this stacked bit to restore the correct stack alignment.

- **BFHFNMIGN: Bus Faults Ignored**

Enables handlers with priority -1 or -2 to ignore data bus faults caused by load and store instructions. This applies to the hard fault and FAULTMASK escalated handlers:

0: Data bus faults caused by load and store instructions cause a lock-up.

1: Handlers running at priority -1 and -2 ignore data bus faults caused by load and store instructions.

Set this bit to 1 only when the handler and its data are in absolutely safe memory. The normal use of this bit is to probe system devices and bridges to detect control path problems and fix them.

- **DIV\_0\_TRP: Division by Zero Trap**

Enables faulting or halting when the processor executes an SDIV or UDIV instruction with a divisor of 0:

0: Do not trap divide by 0.

1: Trap divide by 0.

When this bit is set to 0, a divide by zero returns a quotient of 0.

- **UNALIGN\_TRP: Unaligned Access Trap**

Enables unaligned access traps:

0: Do not trap unaligned halfword and word accesses.

1: Trap unaligned halfword and word accesses.

If this bit is set to 1, an unaligned access generates a usage fault.

Unaligned LDM, STM, LDRD, and STRD instructions always fault irrespective of whether UNALIGN\_TRP is set to 1.

- **USERSETMPEND**

Enables unprivileged software access to the NVIC\_STIR register, see [“Software Trigger Interrupt Register”](#) :

0: Disable.

1: Enable.

- **NONEBASETHRDENA: Thread Mode Enable**

Indicates how the processor enters Thread mode:

0: The processor can enter the Thread mode only when no exception is active.

1: The processor can enter the Thread mode from any level under the control of an EXC\_RETURN value, see [“Exception Return”](#)

.

### 12.9.1.8 System Handler Priority Registers

The SCB\_SHPR1-SCB\_SHPR3 registers set the priority level, 0 to 15 of the exception handlers that have configurable priority. They are byte-accessible.

The system fault handlers and the priority field and register for each handler are:

**Table 12-34. System Fault Handler Priority Fields**

Handler	Field	Register Description
Memory management fault (MemManage)	PRI_4	"System Handler Priority Register 1"
Bus fault (BusFault)	PRI_5	
Usage fault (UsageFault)	PRI_6	
SVCall	PRI_11	"System Handler Priority Register 2"
PendSV	PRI_14	"System Handler Priority Register 3"
SysTick	PRI_15	

Each PRI\_N field is 8 bits wide, but the processor implements only bits [7:4] of each field, and bits [3:0] read as zero and ignore writes.

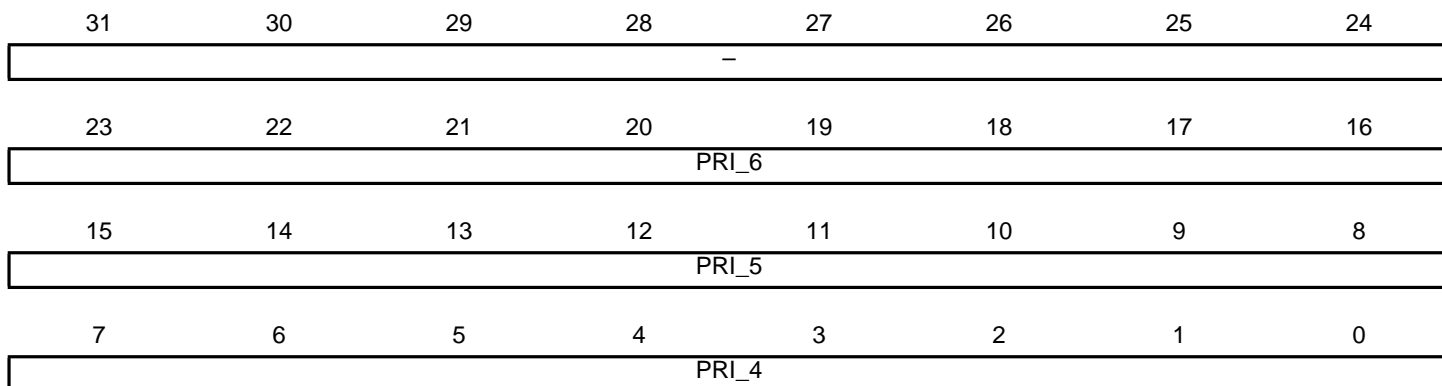


### 12.9.1.9 System Handler Priority Register 1

**Name:** SCB\_SHPR1

**Access:** Read-write

**Reset:** 0x00000000



- **PRI\_6: Priority**

Priority of system handler 6, UsageFault.

- **PRI\_5: Priority**

Priority of system handler 5, BusFault.

- **PRI\_4: Priority**

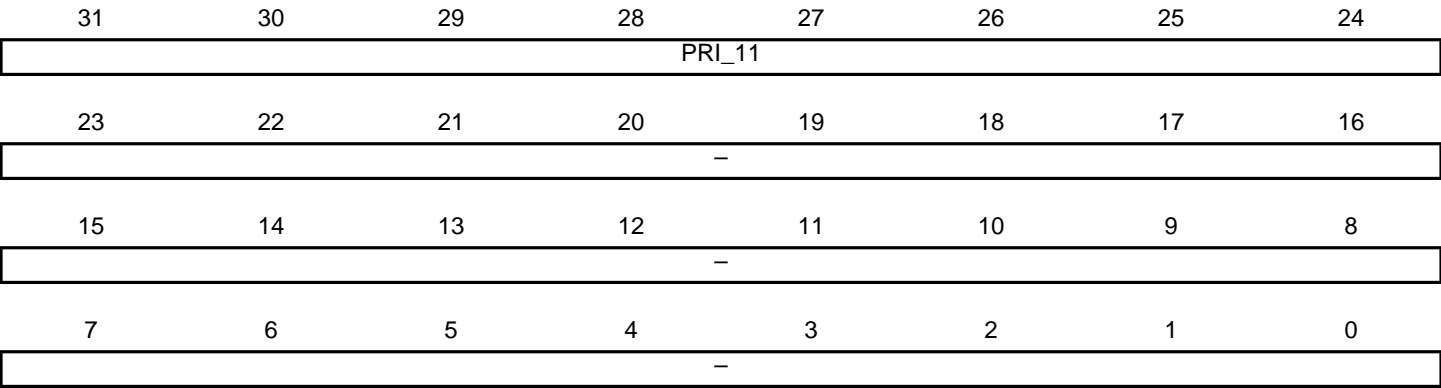
Priority of system handler 4, MemManage.

12.9.1.10 System Handler Priority Register 2

Name: SCB\_SHPR2

Access: Read-write

Reset: 0x00000000



- **PRI\_11: Priority**  
Priority of system handler 11, SVCall.

### 12.9.1.11 System Handler Priority Register 3

**Name:** SCB\_SHPR3

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
PRI_15							
23	22	21	20	19	18	17	16
PRI_14							
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
–							

- **PRI\_15: Priority**

Priority of system handler 15, SysTick exception.

- **PRI\_14: Priority**

Priority of system handler 14, PendSV.

### 12.9.1.12 System Handler Control and State Register

**Name:** SCB\_SHCSR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–							
23	22	21	20	19	18	17	16
–					USGFAULTENA	BUSFAULTENA	MEMFAULTENA
15	14	13	12	11	10	9	8
SVCALLPENDE D	BUSFAULTPEN DED	MEMFAULTPEN DED	USGFAULTPEN DED	SYSTICKACT	PENDSVACT	–	MONITORACT
7	6	5	4	3	2	1	0
SVCALLAVCT	–			USGFAULTACT	–	BUSFAULTACT	MEMFAULTACT

The SHCSR register enables the system handlers, and indicates the pending status of the bus fault, memory management fault, and SVC exceptions; it also indicates the active status of the system handlers.

- **USGFAULTENA: Usage Fault Enable**

0: Disables the exception.

1: Enables the exception.

- **BUSFAULTENA: Bus Fault Enable**

0: Disables the exception.

1: Enables the exception.

- **MEMFAULTENA: Memory Management Fault Enable**

0: Disables the exception.

1: Enables the exception.

- **SVCALLPENDE: SVC Call Pending**

Read:

0: The exception is not pending.

1: The exception is pending.

Note: The user can write to these bits to change the pending status of the exceptions.

- **BUSFAULTPENDE: Bus Fault Exception Pending**

Read:

0: The exception is not pending.

1: The exception is pending.

Note: The user can write to these bits to change the pending status of the exceptions.

- **MEMFAULTPENDE: Memory Management Fault Exception Pending**

Read:

0: The exception is not pending.

1: The exception is pending.

Note: The user can write to these bits to change the pending status of the exceptions.

- **USGFAULTPENDEd: Usage Fault Exception Pending**

Read:

0: The exception is not pending.

1: The exception is pending.

Note: The user can write to these bits to change the pending status of the exceptions.

- **SYSTICKACT: SysTick Exception Active**

Read:

0: The exception is not active.

1: The exception is active.

Note: The user can write to these bits to change the active status of the exceptions.

- Caution: A software that changes the value of an active bit in this register without a correct adjustment to the stacked content can cause the processor to generate a fault exception. Ensure that the software writing to this register retains and subsequently restores the current active status.

- Caution: After enabling the system handlers, to change the value of a bit in this register, the user must use a read-modify-write procedure to ensure that only the required bit is changed.

- **PENDSVACT: PendSV Exception Active**

0: The exception is not active.

1: The exception is active.

- **MONITORACT: Debug Monitor Active**

0: Debug monitor is not active.

1: Debug monitor is active.

- **SVCALLACT: SVC Call Active**

0: SVC call is not active.

1: SVC call is active.

- **USGFAULTACT: Usage Fault Exception Active**

0: Usage fault exception is not active.

1: Usage fault exception is active.

- **BUSFAULTACT: Bus Fault Exception Active**

0: Bus fault exception is not active.

1: Bus fault exception is active.

- **MEMFAULTACT: Memory Management Fault Exception Active**

0: Memory management fault exception is not active.

1: Memory management fault exception is active.

If the user disables a system handler and the corresponding fault occurs, the processor treats the fault as a hard fault.

The user can write to this register to change the pending or active status of system exceptions. An OS kernel can write to the active bits to perform a context switch that changes the current exception type.

### 12.9.1.13 Configurable Fault Status Register

**Name:** SCB\_CFSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–						DIVBYZERO	UNALIGNED
23	22	21	20	19	18	17	16
–				NOCF	INVPC	INVSTATE	UNDEFINTR
15	14	13	12	11	10	9	8
BFRVALID	–		STKERR	UNSTKERR	IMPRECISERR	PRECISERR	IBUSERR
7	6	5	4	3	2	1	0
MMARVALID	–	MLSPERR	MSTKERR	MUNSTKERR	–	DACCVIOL	IACCVIOL

- **IACCVIOL: Instruction Access Violation Flag**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: No instruction access violation fault.

1: The processor attempted an instruction fetch from a location that does not permit execution.

This fault occurs on any access to an XN region, even when the MPU is disabled or not present.

When this bit is 1, the PC value stacked for the exception return points to the faulting instruction. The processor has not written a fault address to the SCB\_MMFAR register.

- **DACCVIOL: Data Access Violation Flag**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: No data access violation fault.

1: The processor attempted a load or store at a location that does not permit the operation.

When this bit is 1, the PC value stacked for the exception return points to the faulting instruction. The processor has loaded the SCB\_MMFAR register with the address of the attempted access.

- **MUNSTKERR: Memory Manager Fault on Unstacking for a Return From Exception**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: No unstacking fault.

1: Unstack for an exception return has caused one or more access violations.

This fault is chained to the handler. This means that when this bit is 1, the original return stack is still present. The processor has not adjusted the SP from the failing return, and has not performed a new save. The processor has not written a fault address to the SCB\_MMFAR register.

- **MSTKERR: Memory Manager Fault on Stacking for Exception Entry**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: No stacking fault.

1: Stacking for an exception entry has caused one or more access violations.

When this bit is 1, the SP is still adjusted but the values in the context area on the stack might be incorrect. The processor has not written a fault address to SCB\_MMFAR register.

- **MLSPERR: MemManage during Lazy State Preservation**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: No MemManage fault occurred during the floating-point lazy state preservation.

1: A MemManage fault occurred during the floating-point lazy state preservation.

- **MMARVALID: Memory Management Fault Address Register (SCB\_MMFAR) Valid Flag**

This is part of “[MMFSR: Memory Management Fault Status Subregister](#)” .

0: The value in SCB\_MMFAR is not a valid fault address.

1: SCB\_MMFAR register holds a valid fault address.

If a memory management fault occurs and is escalated to a hard fault because of priority, the hard fault handler must set this bit to 0. This prevents problems on return to a stacked active memory management fault handler whose SCB\_MMFAR value has been overwritten.

- **IBUSERR: Instruction Bus Error**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: No instruction bus error.

1: Instruction bus error.

The processor detects the instruction bus error on prefetching an instruction, but it sets the IBUSERR flag to 1 only if it attempts to issue the faulting instruction.

When the processor sets this bit to 1, it does not write a fault address to the BFAR register.

- **PRECISERR: Precise Data Bus Error**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: No precise data bus error.

1: A data bus error has occurred, and the PC value stacked for the exception return points to the instruction that caused the fault.

When the processor sets this bit to 1, it writes the faulting address to the SCB\_BFAR register.

- **IMPRECISERR: Imprecise Data Bus Error**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: No imprecise data bus error.

1: A data bus error has occurred, but the return address in the stack frame is not related to the instruction that caused the error.

When the processor sets this bit to 1, it does not write a fault address to the SCB\_BFAR register.

This is an asynchronous fault. Therefore, if it is detected when the priority of the current process is higher than the bus fault priority, the bus fault becomes pending and becomes active only when the processor returns from all higher priority processes. If a precise fault occurs before the processor enters the handler for the imprecise bus fault, the handler detects that both this bit and one of the precise fault status bits are set to 1.

- **UNSTKERR: Bus Fault on Unstacking for a Return From Exception**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: No unstacking fault.

1: Unstack for an exception return has caused one or more bus faults.

This fault is chained to the handler. This means that when the processor sets this bit to 1, the original return stack is still present. The processor does not adjust the SP from the failing return, does not performed a new save, and does not write a fault address to the BFAR.

- **STKERR: Bus Fault on Stacking for Exception Entry**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: No stacking fault.

1: Stacking for an exception entry has caused one or more bus faults.

When the processor sets this bit to 1, the SP is still adjusted but the values in the context area on the stack might be incorrect. The processor does not write a fault address to the SCB\_BFAR register.

- **BFARVALID: Bus Fault Address Register (BFAR) Valid flag**

This is part of “[BFSR: Bus Fault Status Subregister](#)” .

0: The value in SCB\_BFAR is not a valid fault address.

1: SCB\_BFAR holds a valid fault address.

The processor sets this bit to 1 after a bus fault where the address is known. Other faults can set this bit to 0, such as a memory management fault occurring later.

If a bus fault occurs and is escalated to a hard fault because of priority, the hard fault handler must set this bit to 0. This prevents problems if returning to a stacked active bus fault handler whose SCB\_BFAR value has been overwritten.

- **UNDEFINSTR: Undefined Instruction Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” .

0: No undefined instruction usage fault.

1: The processor has attempted to execute an undefined instruction.

When this bit is set to 1, the PC value stacked for the exception return points to the undefined instruction.

An undefined instruction is an instruction that the processor cannot decode.

- **INVSTATE: Invalid State Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” .

0: No invalid state usage fault.

1: The processor has attempted to execute an instruction that makes illegal use of the EPSR.

When this bit is set to 1, the PC value stacked for the exception return points to the instruction that attempted the illegal use of the EPSR.

This bit is not set to 1 if an undefined instruction uses the EPSR.

- **INVPC: Invalid PC Load Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” . It is caused by an invalid PC load by EXC\_RETURN:

0: No invalid PC load usage fault.

1: The processor has attempted an illegal load of EXC\_RETURN to the PC, as a result of an invalid context, or an invalid EXC\_RETURN value.

When this bit is set to 1, the PC value stacked for the exception return points to the instruction that tried to perform the illegal load of the PC.

- **NOCP: No Coprocessor Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” . The processor does not support coprocessor instructions:

0: No usage fault caused by attempting to access a coprocessor.

1: The processor has attempted to access a coprocessor.



- **UNALIGNED: Unaligned Access Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” .

0: No unaligned access fault, or unaligned access trapping not enabled.

1: The processor has made an unaligned memory access.

Enable trapping of unaligned accesses by setting the UNALIGN\_TRP bit in the SCB\_CCR register to 1. See “[Configuration and Control Register](#)” . Unaligned LDM, STM, LDRD, and STRD instructions always fault irrespective of the setting of UNALIGN\_TRP.

- **DIVBYZERO: Divide by Zero Usage Fault**

This is part of “[UFSR: Usage Fault Status Subregister](#)” .

0: No divide by zero fault, or divide by zero trapping not enabled.

1: The processor has executed an SDIV or UDIV instruction with a divisor of 0.

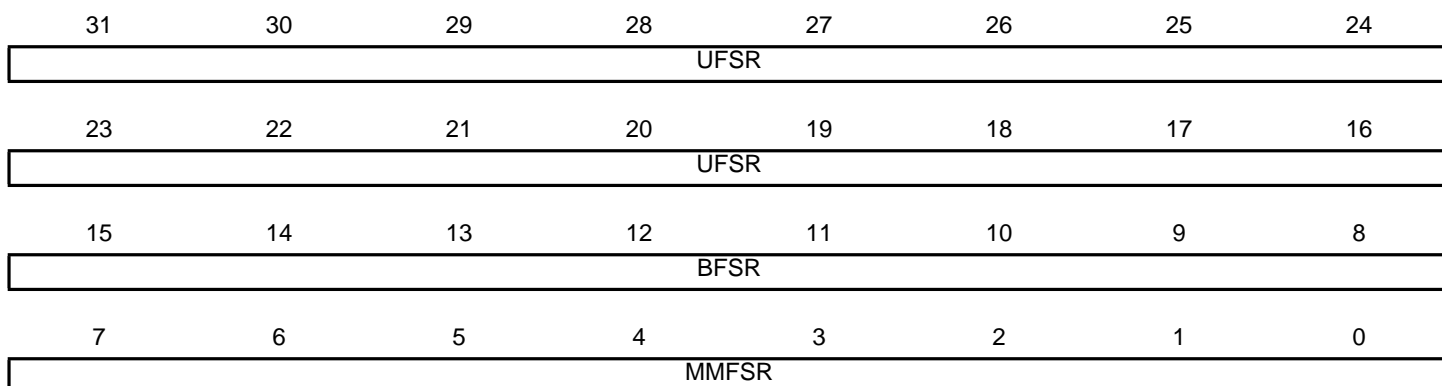
When the processor sets this bit to 1, the PC value stacked for the exception return points to the instruction that performed the divide by zero. Enable trapping of divide by zero by setting the DIV\_0\_TRP bit in the SCB\_CCR register to 1. See “[Configuration and Control Register](#)” .

#### 12.9.1.14 Configurable Fault Status Register (Byte Access)

**Name:** SCB\_CFSR (BYTE)

**Access:** Read-write

**Reset:** 0x00000000



- **MMFSR: Memory Management Fault Status Subregister**

The flags in the MMFSR subregister indicate the cause of memory access faults. See bitfield [7..0] description in [Section 12.9.1.13](#).

- **BFSR: Bus Fault Status Subregister**

The flags in the BFSR subregister indicate the cause of a bus access fault. See bitfield [14..8] description in [Section 12.9.1.13](#).

- **UFSR: Usage Fault Status Subregister**

The flags in the UFSR subregister indicate the cause of a usage fault. See bitfield [31..15] description in [Section 12.9.1.13](#).

**Note:** The UFSR bits are sticky. This means that as one or more fault occurs, the associated bits are set to 1. A bit that is set to 1 is cleared to 0 only by writing 1 to that bit, or by a reset.

The SCB\_CFSR register indicates the cause of a memory management fault, bus fault, or usage fault. It is byte accessible. The user can access the SCB\_CFSR register or its subregisters as follows:

- Access complete SCB\_CFSR with a word access to 0xE000ED28
- Access MMFSR with a byte access to 0xE000ED28
- Access MMFSR and BFSR with a halfword access to 0xE000ED28
- Access BFSR with a byte access to 0xE000ED29
- Access UFSR with a halfword access to 0xE000ED2A.

### 12.9.1.15 Hard Fault Status Register

**Name:** SCB\_HFSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
DEBUGEVT	FORCED	–					
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
–						VECTTBL	–

The HFSR register gives information about events that activate the hard fault handler. This register is read, write to clear. This means that bits in the register read normally, but writing 1 to any bit clears that bit to 0.

- **DEBUGEVT: Reserved for Debug Use**

When writing to the register, write 0 to this bit, otherwise the behavior is unpredictable.

- **FORCED: Forced Hard Fault**

It indicates a forced hard fault, generated by escalation of a fault with configurable priority that cannot be handles, either because of priority or because it is disabled:

0: No forced hard fault.

1: Forced hard fault.

When this bit is set to 1, the hard fault handler must read the other fault status registers to find the cause of the fault.

- **VECTTBL: Bus Fault on a Vector Table**

It indicates a bus fault on a vector table read during an exception processing:

0: No bus fault on vector table read.

1: Bus fault on vector table read.

This error is always handled by the hard fault handler.

When this bit is set to 1, the PC value stacked for the exception return points to the instruction that was preempted by the exception.

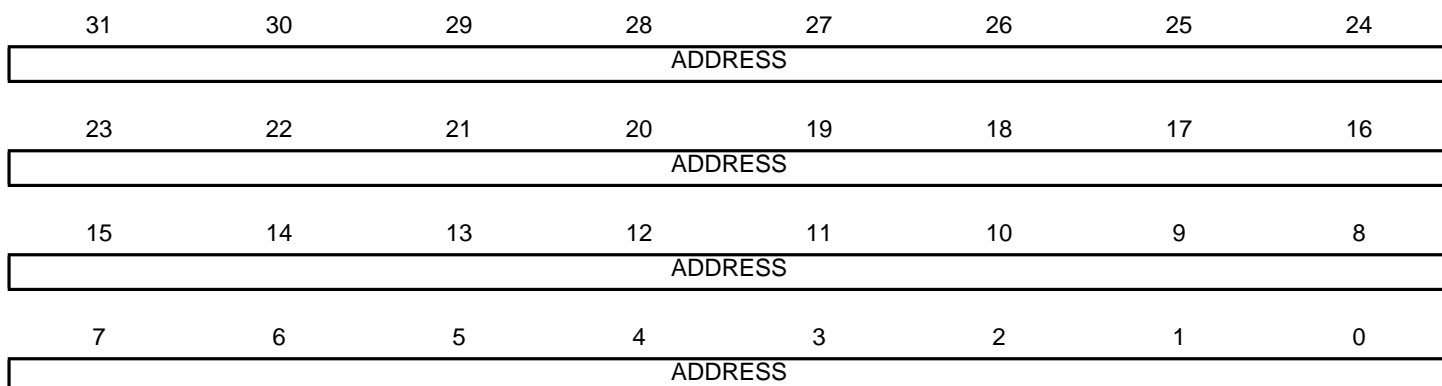
**Note:** The HFSR bits are sticky. This means that, as one or more fault occurs, the associated bits are set to 1. A bit that is set to 1 is cleared to 0 only by writing 1 to that bit, or by a reset.

### 12.9.1.16 MemManage Fault Address Register

**Name:** SCB\_MMFAR

**Access:** Read-write

**Reset:** 0x00000000



The MMFAR register contains the address of the location that generated a memory management fault.

#### • ADDRESS

When the MMARVALID bit of the MMFSR subregister is set to 1, this field holds the address of the location that generated the memory management fault.

- Notes:
1. When an unaligned access faults, the address is the actual address that faulted. Because a single read or write instruction can be split into multiple aligned accesses, the fault address can be any address in the range of the requested access size.
  2. Flags in the MMFSR subregister indicate the cause of the fault, and whether the value in the SCB\_MMFAR register is valid. See [“MMFSR: Memory Management Fault Status Subregister”](#).

### 12.9.1.17 Bus Fault Address Register

**Name:** SCB\_BFAR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
ADDRESS							
23	22	21	20	19	18	17	16
ADDRESS							
15	14	13	12	11	10	9	8
ADDRESS							
7	6	5	4	3	2	1	0
ADDRESS							

The BFAR register contains the address of the location that generated a bus fault.

#### • ADDRESS

When the BFARVALID bit of the BFSR subregister is set to 1, this field holds the address of the location that generated the bus fault.

- Notes:
1. When an unaligned access faults, the address in the SCB\_BFAR register is the one requested by the instruction, even if it is not the address of the fault.
  2. Flags in the BFSR indicate the cause of the fault, and whether the value in the SCB\_BFAR register is valid. See [“BFSR: Bus Fault Status Subregister”](#).

## 12.10 System Timer (SysTick)

The processor has a 24-bit system timer, SysTick, that counts down from the reload value to zero, reloads (wraps to) the value in the SYST\_RVR register on the next clock edge, then counts down on subsequent clocks.

When the processor is halted for debugging, the counter does not decrement.

The SysTick counter runs on the processor clock. If this clock signal is stopped for low power mode, the SysTick counter stops.

Ensure that the software uses aligned word accesses to access the SysTick registers.

The SysTick counter reload and current value are undefined at reset; the correct initialization sequence for the SysTick counter is:

1. Program the reload value.
2. Clear the current value.
3. Program the Control and Status register.

### 12.10.1 System Timer (SysTick) User Interface

Table 12-35. System Timer (SYST) Register Mapping

Offset	Register	Name	Access	Reset
0xE000E010	SysTick Control and Status Register	SYST_CSR	Read-write	0x00000004
0xE000E014	SysTick Reload Value Register	SYST_RVR	Read-write	Unknown
0xE000E018	SysTick Current Value Register	SYST_CVR	Read-write	Unknown
0xE000E01C	SysTick Calibration Value Register	SYST_CALIB	Read-only	0xC0000000

### 12.10.1.1 SysTick Control and Status

**Name:** SYST\_CSR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							COUNTFLAG
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
					CLKSOURCE	TICKINT	ENABLE

The SysTick SYST\_CSR register enables the SysTick features.

- **COUNTFLAG: Count Flag**

Returns 1 if the timer counted to 0 since the last time this was read.

- **CLKSOURCE: Clock Source**

Indicates the clock source:

0: External Clock.

1: Processor Clock.

- **TICKINT**

Enables a SysTick exception request:

0: Counting down to zero does not assert the SysTick exception request.

1: Counting down to zero asserts the SysTick exception request.

The software can use COUNTFLAG to determine if SysTick has ever counted to zero.

- **ENABLE**

Enables the counter:

0: Counter disabled.

1: Counter enabled.

When ENABLE is set to 1, the counter loads the RELOAD value from the SYST\_RVR register and then counts down. On reaching 0, it sets the COUNTFLAG to 1 and optionally asserts the SysTick depending on the value of TICKINT. It then loads the RELOAD value again, and begins counting.

### 12.10.1.2 SysTick Reload Value Registers

**Name:** SYST\_RVR  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
RELOAD							
15	14	13	12	11	10	9	8
RELOAD							
7	6	5	4	3	2	1	0
RELOAD							

The SYST\_RVR register specifies the start value to load into the SYST\_CVR register.

#### • RELOAD

Value to load into the SYST\_CVR register when the counter is enabled and when it reaches 0.

The RELOAD value can be any value in the range 0x00000001-0x00FFFFFF. A start value of 0 is possible, but has no effect because the SysTick exception request and COUNTFLAG are activated when counting from 1 to 0.

The RELOAD value is calculated according to its use: For example, to generate a multi-shot timer with a period of N processor clock cycles, use a RELOAD value of N-1. If the SysTick interrupt is required every 100 clock pulses, set RELOAD to 99.



### 12.10.1.3 SysTick Current Value Register

**Name:** SYST\_CVR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
CURRENT							
15	14	13	12	11	10	9	8
CURRENT							
7	6	5	4	3	2	1	0
CURRENT							

The SysTick SYST\_CVR register contains the current value of the SysTick counter.

- **CURRENT**

Reads return the current value of the SysTick counter.

A write of any value clears the field to 0, and also clears the SYST\_CSR.COUNTFLAG bit to 0.

#### 12.10.1.4 SysTick Calibration Value Register

**Name:** SYST\_CALIB

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
NOREF	SKEW	–					
23	22	21	20	19	18	17	16
TENMS							
15	14	13	12	11	10	9	8
TENMS							
7	6	5	4	3	2	1	0
TENMS							

The SysTick SYST\_CSR register indicates the SysTick calibration properties.

- **NOREF: No Reference Clock**

It indicates whether the device provides a reference clock to the processor:

0: Reference clock provided.

1: No reference clock provided.

If your device does not provide a reference clock, the SYST\_CSR.CLKSOURCE bit reads-as-one and ignores writes.

- **SKEW**

It indicates whether the TENMS value is exact:

0: TENMS value is exact.

1: TENMS value is inexact, or not given.

An inexact TENMS value can affect the suitability of SysTick as a software real time clock.

- **TENMS: Ten Milliseconds**

The reload value for 10 ms (100 Hz) timing is subject to system clock skew errors. If the value reads as zero, the calibration value is not known.

Read as 0x000030D4. The SysTick calibration value is fixed at 0x000030D4 (8000), which allows the generation of a time base of 1 ms with SysTick clock at 12.5 MHz ( $100/8 = 12.5$  MHz).

## 12.11 Memory Protection Unit (MPU)

The MPU divides the memory map into a number of regions, and defines the location, size, access permissions, and memory attributes of each region. It supports:

- Independent attribute settings for each region
- Overlapping regions
- Export of memory attributes to the system.

The memory attributes affect the behavior of memory accesses to the region. The Cortex-M4 MPU defines:

- Eight separate memory regions, 0-7
- A background region.

When memory regions overlap, a memory access is affected by the attributes of the region with the highest number. For example, the attributes for region 7 take precedence over the attributes of any region that overlaps region 7.

The background region has the same memory access attributes as the default memory map, but is accessible from privileged software only.

The Cortex-M4 MPU memory map is unified. This means that instruction accesses and data accesses have the same region settings.

If a program accesses a memory location that is prohibited by the MPU, the processor generates a memory management fault. This causes a fault exception, and might cause the termination of the process in an OS environment.

In an OS environment, the kernel can update the MPU region setting dynamically based on the process to be executed. Typically, an embedded OS uses the MPU for memory protection.

The configuration of MPU regions is based on memory types (see [“Memory Regions, Types and Attributes”](#) ).

[Table 12-36](#) shows the possible MPU region attributes. These include Share ability and cache behavior attributes that are not relevant to most microcontroller implementations. See [“MPU Configuration for a Microcontroller”](#) for guidelines for programming such an implementation.

**Table 12-36. Memory Attributes Summary**

Memory Type	Shareability	Other Attributes	Description
Strongly- ordered	-	-	All accesses to Strongly-ordered memory occur in program order. All Strongly-ordered regions are assumed to be shared.
Device	Shared	-	Memory-mapped peripherals that several processors share.
	Non-shared	-	Memory-mapped peripherals that only a single processor uses.
Normal	Shared		Normal memory that is shared between several processors.
	Non-shared		Normal memory that only a single processor uses.

### 12.11.1 MPU Access Permission Attributes

This section describes the MPU access permission attributes. The access permission bits (TEX, C, B, S, AP, and XN) of the MPU\_RASR control the access to the corresponding memory region. If an access is made to an area of memory without the required permissions, then the MPU generates a permission fault.

The table below shows the encodings for the TEX, C, B, and S access permission bits.

**Table 12-37. TEX, C, B, and S Encoding**

TEX	C	B	S	Memory Type	Shareability	Other Attributes
b000	0	0	x <sup>(1)</sup>	Strongly-ordered	Shareable	-
		1	x <sup>(1)</sup>	Device	Shareable	-
	1	0	0	Normal	Not shareable	Outer and inner write-through. No write allocate.
			1		Shareable	
		1	0	Normal	Not shareable	Outer and inner write-back. No write allocate.
			1		Shareable	
b001	0	0	0	Normal	Not shareable	
			1		Shareable	
		1	x <sup>(1)</sup>	Reserved encoding		-
	1	0	x <sup>(1)</sup>	Implementation defined attributes.		-
		1	0	Normal	Not shareable	Outer and inner write-back. Write and read allocate.
			1		Shareable	
b010	0	0	x <sup>(1)</sup>	Device	Not shareable	Nonshared Device.
		1	x <sup>(1)</sup>	Reserved encoding		-
	1	x <sup>(1)</sup>	x <sup>(1)</sup>	Reserved encoding		-
b1B B	A	A	0	Normal	Not shareable	
			1		Shareable	

Note: 1. The MPU ignores the value of this bit.

Table 12-38 shows the cache policy for memory attribute encodings with a TEX value is in the range 4-7.

**Table 12-38. Cache Policy for Memory Attribute Encoding**

Encoding, AA or BB	Corresponding Cache Policy
00	Non-cacheable
01	Write back, write and read allocate
10	Write through, no write allocate
11	Write back, no write allocate

Table 12-39 shows the AP encodings that define the access permissions for privileged and unprivileged software.

**Table 12-39. AP Encoding**

AP[2:0]	Privileged Permissions	Unprivileged Permissions	Description
000	No access	No access	All accesses generate a permission fault
001	RW	No access	Access from privileged software only
010	RW	RO	Writes by unprivileged software generate a permission fault
011	RW	RW	Full access
100	Unpredictable	Unpredictable	Reserved
101	RO	No access	Reads by privileged software only
110	RO	RO	Read only, by privileged or unprivileged software
111	RO	RO	Read only, by privileged or unprivileged software

#### 12.11.1.1 MPU Mismatch

When an access violates the MPU permissions, the processor generates a memory management fault, see “[Exceptions and Interrupts](#)”. The MMFSR indicates the cause of the fault. See “[MMFSR: Memory Management Fault Status Subregister](#)” for more information.

#### 12.11.1.2 Updating an MPU Region

To update the attributes for an MPU region, update the MPU\_RNR, MPU\_RBAR and MPU\_RASR registers. Each register can be programmed separately, or a multiple-word write can be used to program all of these registers. MPU\_RBAR and MPU\_RASR aliases can be used to program up to four regions simultaneously using an STM instruction.

#### 12.11.1.3 Updating an MPU Region Using Separate Words

Simple code to configure one region:

```

; R1 = region number
; R2 = size/enable
; R3 = attributes
; R4 = address
LDR R0,=MPU_RNR          ; 0xE000ED98, MPU region number register
STR R1, [R0, #0x0]       ; Region Number
STR R4, [R0, #0x4]       ; Region Base Address
STRH R2, [R0, #0x8]      ; Region Size and Enable
STRH R3, [R0, #0xA]      ; Region Attribute

```

Disable a region before writing new region settings to the MPU, if the region being changed was previously enabled. For example:

```

; R1 = region number
; R2 = size/enable
; R3 = attributes
; R4 = address
LDR R0,=MPU_RNR          ; 0xE000ED98, MPU region number register
STR R1, [R0, #0x0]       ; Region Number
BIC R2, R2, #1           ; Disable
STRH R2, [R0, #0x8]      ; Region Size and Enable
STR R4, [R0, #0x4]       ; Region Base Address
STRH R3, [R0, #0xA]      ; Region Attribute
ORR R2, #1               ; Enable
STRH R2, [R0, #0x8]      ; Region Size and Enable

```

The software must use memory barrier instructions:

- Before the MPU setup, if there might be outstanding memory transfers, such as buffered writes, that might be affected by the change in MPU settings
- After the MPU setup, if it includes memory transfers that must use the new MPU settings.

However, memory barrier instructions are not required if the MPU setup process starts by entering an exception handler, or is followed by an exception return, because the exception entry and exception return mechanisms cause memory barrier behavior.

The software does not need any memory barrier instructions during an MPU setup, because it accesses the MPU through the PPB, which is a Strongly-Ordered memory region.

For example, if the user wants all of the memory access behavior to take effect immediately after the programming sequence, a DSB instruction and an ISB instruction must be used. A DSB is required after changing MPU settings, such as at the end of a context switch. An ISB is required if the code that programs the MPU region or regions is entered using a branch or call. If the programming sequence is entered using a return from exception, or by taking an exception, then an ISB is not required.

#### 12.11.1.4 Updating an MPU Region Using Multi-word Writes

The user can program directly using multi-word writes, depending on how the information is divided. Consider the following reprogramming:

```
; R1 = region number
; R2 = address
; R3 = size, attributes in one
LDR R0, =MPU_RNR      ; 0xE000ED98, MPU region number register
STR R1, [R0, #0x0]    ; Region Number
STR R2, [R0, #0x4]    ; Region Base Address
STR R3, [R0, #0x8]    ; Region Attribute, Size and Enable
```

Use an STM instruction to optimize this:

```
; R1 = region number
; R2 = address
; R3 = size, attributes in one
LDR R0, =MPU_RNR      ; 0xE000ED98, MPU region number register
STM R0, {R1-R3}       ; Region Number, address, attribute, size and enable
```

This can be done in two words for pre-packed information. This means that the MPU\_RBAR contains the required region number and had the VALID bit set to 1. See [“MPU Region Base Address Register”](#) . Use this when the data is statically packed, for example in a boot loader:

```
; R1 = address and region number in one
; R2 = size and attributes in one
LDR R0, =MPU_RBAR     ; 0xE000ED9C, MPU Region Base register
STR R1, [R0, #0x0]    ; Region base address and
                        ; region number combined with VALID (bit 4) set to 1
STR R2, [R0, #0x4]    ; Region Attribute, Size and Enable
```

Use an STM instruction to optimize this:

```
; R1 = address and region number in one
; R2 = size and attributes in one
LDR R0, =MPU_RBAR     ; 0xE000ED9C, MPU Region Base register
STM R0, {R1-R2}       ; Region base address, region number and VALID bit,
                        ; and Region Attribute, Size and Enable
```

### 12.11.1.5 Subregions

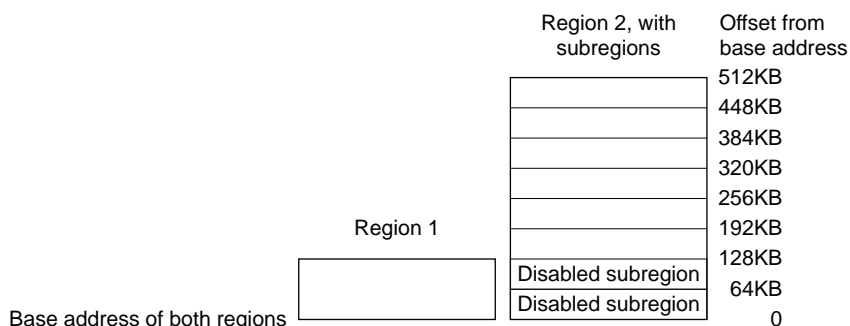
Regions of 256 bytes or more are divided into eight equal-sized subregions. Set the corresponding bit in the SRD field of the MPU\_RASR field to disable a subregion. See “MPU Region Attribute and Size Register”. The least significant bit of SRD controls the first subregion, and the most significant bit controls the last subregion. Disabling a subregion means another region overlapping the disabled range matches instead. If no other enabled region overlaps the disabled subregion, the MPU issues a fault.

Regions of 32, 64, and 128 bytes do not support subregions. With regions of these sizes, the SRD field must be set to 0x00, otherwise the MPU behavior is unpredictable.

### 12.11.1.6 Example of SRD Use

Two regions with the same base address overlap. Region 1 is 128 KB, and region 2 is 512 KB. To ensure the attributes from region 1 apply to the first 128 KB region, set the SRD field for region 2 to b00000011 to disable the first two subregions, as in Figure 12-13 below:

**Figure 12-13. SRD Use**



### 12.11.1.7 MPU Design Hints And Tips

To avoid unexpected behavior, disable the interrupts before updating the attributes of a region that the interrupt handlers might access.

Ensure the software uses aligned accesses of the correct size to access MPU registers:

- Except for the MPU\_RASR register, it must use aligned word accesses
- For the MPU\_RASR register, it can use byte or aligned halfword or word accesses.

The processor does not support unaligned accesses to MPU registers.

When setting up the MPU, and if the MPU has previously been programmed, disable unused regions to prevent any previous region settings from affecting the new MPU setup.

#### *MPU Configuration for a Microcontroller*

Usually, a microcontroller system has only a single processor and no caches. In such a system, program the MPU as follows:

**Table 12-40. Memory Region Attributes for a Microcontroller**

Memory Region	TEX	C	B	S	Memory Type and Attributes
Flash memory	b000	1	0	0	Normal memory, non-shareable, write-through
Internal SRAM	b000	1	0	1	Normal memory, shareable, write-through
External SRAM	b000	1	1	1	Normal memory, shareable, write-back, write-allocate
Peripherals	b000	0	1	1	Device memory, shareable

In most microcontroller implementations, the shareability and cache policy attributes do not affect the system behavior. However, using these settings for the MPU regions can make the application code more portable. The values given are for typical situations. In special systems, such as multiprocessor designs or designs with a separate DMA engine, the shareability attribute might be important. In these cases, refer to the recommendations of the memory device manufacturer.

## 12.11.2 Memory Protection Unit (MPU) User Interface

Table 12-41. Memory Protection Unit (MPU) Register Mapping

Offset	Register	Name	Access	Reset
0xE000ED90	MPU Type Register	MPU_TYPE	Read-only	0x00000800
0xE000ED94	MPU Control Register	MPU_CTRL	Read-write	0x00000000
0xE000ED98	MPU Region Number Register	MPU_RNR	Read-write	0x00000000
0xE000ED9C	MPU Region Base Address Register	MPU_RBAR	Read-write	0x00000000
0xE000EDA0	MPU Region Attribute and Size Register	MPU_RASR	Read-write	0x00000000
0xE000EDA4	Alias of RBAR, see MPU Region Base Address Register	MPU_RBAR_A1	Read-write	0x00000000
0xE000EDA8	Alias of RASR, see MPU Region Attribute and Size Register	MPU_RASR_A1	Read-write	0x00000000
0xE000EDAC	Alias of RBAR, see MPU Region Base Address Register	MPU_RBAR_A2	Read-write	0x00000000
0xE000EDB0	Alias of RASR, see MPU Region Attribute and Size Register	MPU_RASR_A2	Read-write	0x00000000
0xE000EDB4	Alias of RBAR, see MPU Region Base Address Register	MPU_RBAR_A3	Read-write	0x00000000
0xE000EDB8	Alias of RASR, see MPU Region Attribute and Size Register	MPU_RASR_A3	Read-write	0x00000000



### 12.11.2.1 MPU Type Register

**Name:** MPU\_TYPE

**Access:** Read-write

**Reset:** 0x00000800

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
IREGION							
15	14	13	12	11	10	9	8
DREGION							
7	6	5	4	3	2	1	0
—							SEPARATE

The MPU\_TYPE register indicates whether the MPU is present, and if so, how many regions it supports.

- **IREGION: Instruction Region**

Indicates the number of supported MPU instruction regions.

Always contains 0x00. The MPU memory map is unified and is described by the DREGION field.

- **DREGION: Data Region**

Indicates the number of supported MPU data regions:

0x08 = Eight MPU regions.

- **SEPARATE: Separate Instruction**

Indicates support for unified or separate instruction and data memory maps:

0: Unified.

### 12.11.2.2 MPU Control Register

**Name:** MPU\_CTRL

**Access:** Read-write

**Reset:** 0x00000800

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
—					PRIVDEFENA	HFNMENA	ENABLE

The MPU CTRL register enables the MPU, enables the default memory map background region, and enables the use of the MPU when in the hard fault, Non-maskable Interrupt (NMI), and FAULTMASK escalated handlers.

- **PRIVDEFENA: Privileged Default Memory Map Enabled**

Enables privileged software access to the default memory map:

0: If the MPU is enabled, disables the use of the default memory map. Any memory access to a location not covered by any enabled region causes a fault.

1: If the MPU is enabled, enables the use of the default memory map as a background region for privileged software accesses. When enabled, the background region acts as a region number -1. Any region that is defined and enabled has priority over this default map.

If the MPU is disabled, the processor ignores this bit.

- **HFNMENA: Hard Fault and NMI Enabled**

Enables the operation of MPU during hard fault, NMI, and FAULTMASK handlers.

When the MPU is enabled:

0: MPU is disabled during hard fault, NMI, and FAULTMASK handlers, regardless of the value of the ENABLE bit.

1: The MPU is enabled during hard fault, NMI, and FAULTMASK handlers.

When the MPU is disabled, if this bit is set to 1, the behavior is unpredictable.

- **ENABLE**

Enables the MPU:

0: MPU disabled.

1: MPU enabled.

When ENABLE and PRIVDEFENA are both set to 1:

- For privileged accesses, the *default memory map* is as described in “[Memory Model](#)”. Any access by privileged software that does not address an enabled memory region behaves as defined by the default memory map.
- Any access by unprivileged software that does not address an enabled memory region causes a memory management fault.

XN and Strongly-ordered rules always apply to the System Control Space regardless of the value of the ENABLE bit.

When the ENABLE bit is set to 1, at least one region of the memory map must be enabled for the system to function unless the PRIVDEFENA bit is set to 1. If the PRIVDEFENA bit is set to 1 and no regions are enabled, then only privileged software can operate.

When the ENABLE bit is set to 0, the system uses the default memory map. This has the same memory attributes as if the MPU is not implemented. The default memory map applies to accesses from both privileged and unprivileged software.

When the MPU is enabled, accesses to the System Control Space and vector table are always permitted. Other areas are accessible based on regions and whether PRIVDEFENA is set to 1.

Unless HFNMIENA is set to 1, the MPU is not enabled when the processor is executing the handler for an exception with priority –1 or –2. These priorities are only possible when handling a hard fault or NMI exception, or when FAULTMASK is enabled. Setting the HFNMIENA bit to 1 enables the MPU when operating with these two priorities.

### 12.11.2.3 MPU Region Number Register

**Name:** MPU\_RNR  
**Access:** Read-write  
**Reset:** 0x00000800

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
—							
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
REGION							

The MPU\_RNR selects which memory region is referenced by the MPU\_RBAR and MPU\_RASR registers.

- **REGION**

Indicates the MPU region referenced by the MPU\_RBAR and MPU\_RASR registers.

The MPU supports 8 memory regions, so the permitted values of this field are 0-7.

Normally, the required region number is written to this register before accessing the MPU\_RBAR or MPU\_RASR. However, the region number can be changed by writing to the MPU\_RBAR with the VALID bit set to 1; see [“MPU Region Base Address Register”](#). This write updates the value of the REGION field.

### 12.11.2.4 MPU Region Base Address Register

**Name:** MPU\_RBAR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
ADDR							
23	22	21	20	19	18	17	16
ADDR							
15	14	13	12	11	10	9	N
ADDR							
N-1	6	5	4	3	2	1	0
-		VALID		REGION			

**Note:** If the region size is 32B, the ADDR field is bits [31:5] and there is no Reserved field.

The MPU\_RBAR defines the base address of the MPU region selected by the MPU\_RNR, and can update the value of the MPU\_RNR.

Write MPU\_RBAR with the VALID bit set to 1 to change the current region number and update the MPU\_RNR.

- **ADDR: Region Base Address**

The value of N depends on the region size. The ADDR field is bits [31:N] of the MPU\_RBAR. The region size, as specified by the SIZE field in the MPU\_RASR, defines the value of N:

$N = \text{Log}_2(\text{Region size in bytes})$ ,

If the region size is configured to 4 GB, in the MPU\_RASR, there is no valid ADDR field. In this case, the region occupies the complete memory map, and the base address is 0x00000000.

The base address is aligned to the size of the region. For example, a 64 KB region must be aligned on a multiple of 64 KB, for example, at 0x00010000 or 0x00020000.

- **VALID: MPU Region Number Valid**

Write:

0: MPU\_RNR not changed, and the processor updates the base address for the region specified in the MPU\_RNR, and ignores the value of the REGION field.

1: The processor updates the value of the MPU\_RNR to the value of the REGION field, and updates the base address for the region specified in the REGION field.

Always reads as zero.

- **REGION: MPU Region**

For the behavior on writes, see the description of the VALID field.

On reads, returns the current region number, as specified by the MPU\_RNR.

### 12.11.2.5 MPU Region Attribute and Size Register

**Name:** MPU\_RASR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–			XN	–	AP		
23	22	21	20	19	18	17	16
–		TEX			S	C	B
15	14	13	12	11	10	9	8
SRD							
7	6	5	4	3	2	1	0
–		SIZE					ENABLE

The MPU\_RASR defines the region size and memory attributes of the MPU region specified by the MPU\_RNR, and enables that region and any subregions.

MPU\_RASR is accessible using word or halfword accesses:

- The most significant halfword holds the region attributes.
- The least significant halfword holds the region size, and the region and subregion enable bits.

- **XN: Instruction Access Disable**

0: Instruction fetches enabled.

1: Instruction fetches disabled.

- **AP: Access Permission**

See [Table 12-39](#).

- **TEX, C, B: Memory Access Attributes**

See [Table 12-37](#).

- **S: Shareable**

See [Table 12-37](#).

- **SRD: Subregion Disable**

For each bit in this field:

0: Corresponding sub-region is enabled.

1: Corresponding sub-region is disabled.

See [“Subregions”](#) for more information.

Region sizes of 128 bytes and less do not support subregions. When writing the attributes for such a region, write the SRD field as 0x00.

- **SIZE: Size of the MPU Protection Region**

The minimum permitted value is 3 (b00010).

The SIZE field defines the size of the MPU memory region specified by the MPU\_RNR. as follows:

$$(\text{Region size in bytes}) = 2^{(\text{SIZE}+1)}$$

The smallest permitted region size is 32B, corresponding to a SIZE value of 4. The table below gives an example of SIZE values, with the corresponding region size and value of N in the MPU\_RBAR.

SIZE Value	Region Size	Value of N <sup>(1)</sup>	Note
b00100 (4)	32 B	5	Minimum permitted size
b01001 (9)	1 KB	10	-
b10011 (19)	1 MB	20	-
b11101 (29)	1 GB	30	-
b11111 (31)	4 GB	b01100	Maximum possible size

Note: 1. In the MPU\_RBAR, see [“MPU Region Base Address Register”](#)

- **ENABLE: Region Enable**

Note: For information about access permission, see [“MPU Access Permission Attributes”](#) .

## 12.12 Floating Point Unit (FPU)

The Cortex-M4F FPU implements the FPv4-SP floating-point extension.

The FPU fully supports single-precision add, subtract, multiply, divide, multiply and accumulate, and square root operations. It also provides conversions between fixed-point and floating-point data formats, and floating-point constant instructions.

The FPU provides floating-point computation functionality that is compliant with the ANSI/IEEE Std 754-2008, IEEE Standard for Binary Floating-Point Arithmetic, referred to as the IEEE 754 standard.

The FPU contains 32 single-precision extension registers, which can also be accessed as 16 doubleword registers for load, store, and move operations.

### 12.12.1 Enabling the FPU

The FPU is disabled from reset. It must be enabled before any floating-point instructions can be used. Example 4-1 shows an example code sequence for enabling the FPU in both privileged and user modes. The processor must be in privileged mode to read from and write to the CPACR.

Example of Enabling the FPU:

```
; CPACR is located at address 0xE000ED88
LDR.W R0, =0xE000ED88
; Read CPACR
LDR R1, [R0]
; Set bits 20-23 to enable CP10 and CP11 coprocessors
ORR R1, R1, #(0xF << 20)
; Write back the modified value to the CPACR
STR R1, [R0]; wait for store to complete
DSB
;reset pipeline now the FPU is enabled
ISB
```



## 12.12.2 Floating Point Unit (FPU) User Interface

Table 12-42. Floating Point Unit (FPU) Register Mapping

Offset	Register	Name	Access	Reset
0xE000ED88	Coprocessor Access Control Register	CPACR	Read-write	0x00000000
0xE000EF34	Floating-point Context Control Register	FPCCR	Read-write	0xC0000000
0xE000EF38	Floating-point Context Address Register	FPCAR	Read-write	–
–	Floating-point Status Control Register	FPSCR	Read-write	–
0xE000E01C	Floating-point Default Status Control Register	FPDSCR	Read-write	0x00000000

### 12.12.2.1 Coprocessor Access Control Register

**Name:** CPACR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
—							
23	22	21	20	19	18	17	16
CP11		CP10		—			
15	14	13	12	11	10	9	8
—							
7	6	5	4	3	2	1	0
—							

The CPACR register specifies the access privileges for coprocessors.

- **CPn: Access Privileges for Coprocessor n [2n+1:2n], for n Values 10 and 11.**

The possible values of each field are:

0b00 = Access denied. Any attempted access generates a NOCP UsageFault.

0b01 = Privileged access only. An unprivileged access generates a NOCP fault.

0b10 = Reserved. The result of any access is unpredictable.

0b11 = Full access.

### 12.12.2.2 Floating-point Context Control Register

**Name:** FPCCR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
ASPEN	LSPEN	–					
23	22	21	20	19	18	17	16
–							
15	14	13	12	11	10	9	8
–							MONRDY
7	6	5	4	3	2	1	0
–	BFRDY	MMRDY	HFRDY	THREAD	–	USER	LSPACT

The FPCCR register sets or returns FPU control data.

- **ASPEN: Automatic Hardware State Preservation And Restoration**

Enables CONTROL bit [2] setting on execution of a floating-point instruction. This results in an automatic hardware state preservation and restoration, for floating-point context, on exception entry and exit.

0: Disable CONTROL bit [2] setting on execution of a floating-point instruction.

1: Enable CONTROL bit [2] setting on execution of a floating-point instruction.

- **LSPEN: Automatic Lazy State Preservation**

0: Disable automatic lazy state preservation for floating-point context.

1: Enable automatic lazy state preservation for floating-point context.

- **MONRDY: Debug Monitor Ready**

0: DebugMonitor is disabled or the priority did not permit to set MON\_PEND when the floating-point stack frame was allocated.

1: DebugMonitor is enabled and the priority permitted to set MON\_PEND when the floating-point stack frame was allocated.

- **BFRDY: Bus Fault Ready**

0: BusFault is disabled or the priority did not permit to set the BusFault handler to the pending state when the floating-point stack frame was allocated.

1: BusFault is enabled and the priority permitted to set the BusFault handler to the pending state when the floating-point stack frame was allocated.

- **MMRDY: Memory Management Ready**

0: MemManage is disabled or the priority did not permit to set the MemManage handler to the pending state when the floating-point stack frame was allocated.

1: MemManage is enabled and the priority permitted to set the MemManage handler to the pending state when the floating-point stack frame was allocated.

- **HFRDY: Hard Fault Ready**

0: The priority did not permit to set the HardFault handler to the pending state when the floating-point stack frame was allocated.

1: The priority permitted to set the HardFault handler to the pending state when the floating-point stack frame was allocated.

- **THREAD: Thread Mode**

0: The mode was not the Thread Mode when the floating-point stack frame was allocated.

1: The mode was the Thread Mode when the floating-point stack frame was allocated.

- **USER: User Privilege Level**

0: The privilege level was not User when the floating-point stack frame was allocated.

1: The privilege level was User when the floating-point stack frame was allocated.

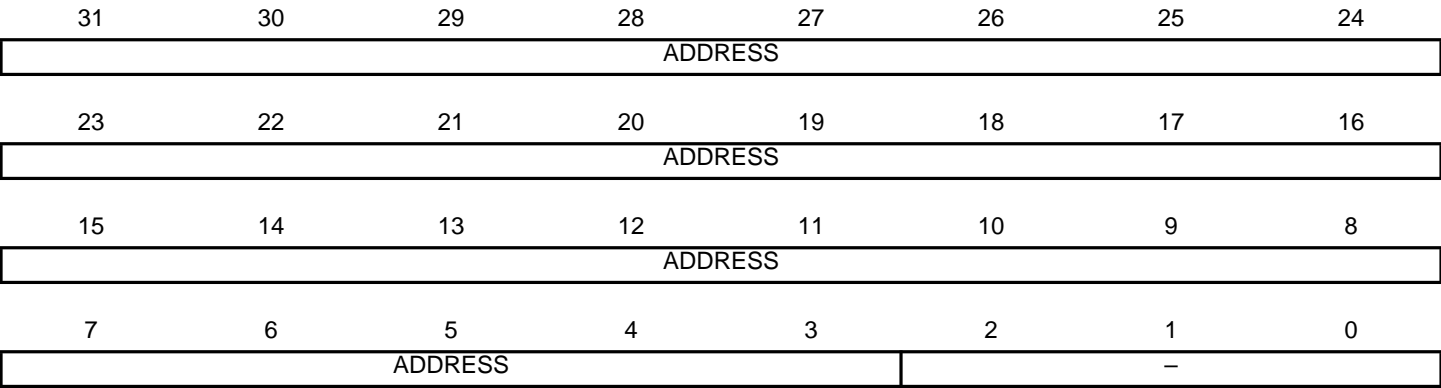
- **LSPACT: Lazy State Preservation Active**

0: The lazy state preservation is not active.

1: The lazy state preservation is active. The floating-point stack frame has been allocated but saving the state to it has been deferred.

12.12.2.3 Floating-point Context Address Register

**Name:** FPCAR  
**Access:** Read-write  
**Reset:** 0x00000000



The FPCAR register holds the location of the unpopulated floating-point register space allocated on an exception stack frame.

• ADDRESS

The location of the unpopulated floating-point register space allocated on an exception stack frame.

### 12.12.2.4 Floating-point Status Control Register

**Name:** FPSCR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
N	Z	C	V	–	AHP	DN	FZ
23	22	21	20	19	18	17	16
RMode		–					
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
IDC	–		IXC	UFC	OFC	DZC	IOC

The FPSCR register provides all necessary User level control of the floating-point system.

- **N: Negative Condition Code Flag**

Floating-point comparison operations update this flag.

- **Z: Zero Condition Code Flag**

Floating-point comparison operations update this flag.

- **C: Carry Condition Code Flag**

Floating-point comparison operations update this flag.

- **V: Overflow Condition Code Flag**

Floating-point comparison operations update this flag.

- **AHP: Alternative Half-precision Control**

0: IEEE half-precision format selected.

1: Alternative half-precision format selected.

- **DN: Default NaN Mode Control**

0: NaN operands propagate through to the output of a floating-point operation.

1: Any operation involving one or more NaNs returns the Default NaN.

- **FZ: Flush-to-zero Mode Control**

0: Flush-to-zero mode disabled. The behavior of the floating-point system is fully compliant with the IEEE 754 standard.

1: Flush-to-zero mode enabled.

- **RMode: Rounding Mode Control**

The encoding of this field is:

0b00 Round to Nearest (RN) mode

0b01 Round towards Plus Infinity (RP) mode.

0b10 Round towards Minus Infinity (RM) mode.

0b11 Round towards Zero (RZ) mode.

The specified rounding mode is used by almost all floating-point instructions.

- **IDC: Input Denormal Cumulative Exception**

IDC is a cumulative exception bit for floating-point exception; see also bits [4:0].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

- **IXC: Inexact Cumulative Exception**

IXC is a cumulative exception bit for floating-point exception; see also bit [7].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

- **UFC: Underflow Cumulative Exception**

UFC is a cumulative exception bit for floating-point exception; see also bit [7].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

- **OFC: Overflow Cumulative Exception**

OFC is a cumulative exception bit for floating-point exception; see also bit [7].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

- **DZC: Division by Zero Cumulative Exception**

DZC is a cumulative exception bit for floating-point exception; see also bit [7].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

- **IOC: Invalid Operation Cumulative Exception**

IOC is a cumulative exception bit for floating-point exception; see also bit [7].

This bit is set to 1 to indicate that the corresponding exception has occurred since 0 was last written to it.

### 12.12.2.5 Floating-point Default Status Control Register

**Name:** FPDSCR

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–					AHP	DN	FZ
23	22	21	20	19	18	17	16
RMode		–					
15	14	13	12	11	10	9	8
–							
7	6	5	4	3	2	1	0
–							

The FPDSCR register holds the default values for the floating-point status control data.

- **AHP**

Default value for FPSCR.AHP.

- **DN**

Default value for FPSCR.DN.

- **FZ**

Default value for FPSCR.FZ.

- **RMode**

Default value for FPSCR.RMode.



## 12.13 Glossary

This glossary describes some of the terms used in technical documents from ARM.

Abort	<p>A mechanism that indicates to a processor that the value associated with a memory access is invalid. An abort can be caused by the external or internal memory system as a result of attempting to access invalid instruction or data memory.</p>
Aligned	<p>A data item stored at an address that is divisible by the number of bytes that defines the data size is said to be aligned. Aligned words and halfwords have addresses that are divisible by four and two respectively. The terms word-aligned and halfword-aligned therefore stipulate addresses that are divisible by four and two respectively.</p>
Banked register	<p>A register that has multiple physical copies, where the state of the processor determines which copy is used. The Stack Pointer, SP (R13) is a banked register.</p>
Base register	<p>In instruction descriptions, a register specified by a load or store instruction that is used to hold the base value for the instruction's address calculation. Depending on the instruction and its addressing mode, an offset can be added to or subtracted from the base register value to form the address that is sent to memory.</p> <p><i>See also</i> <a href="#">"Index register"</a></p>
Big-endian (BE)	<p>Byte ordering scheme in which bytes of decreasing significance in a data word are stored at increasing addresses in memory.</p> <p><i>See also</i> <a href="#">"Byte-invariant"</a> , <a href="#">"Endianness"</a> , <a href="#">"Little-endian (LE)"</a> .</p>
Big-endian memory	<p>Memory in which:</p> <ul style="list-style-type: none"><li>a byte or halfword at a word-aligned address is the most significant byte or halfword within the word at that address,</li><li>a byte at a halfword-aligned address is the most significant byte within the halfword at that address.</li></ul> <p><i>See also</i> <a href="#">"Little-endian memory"</a> .</p>
Breakpoint	<p>A breakpoint is a mechanism provided by debuggers to identify an instruction at which program execution is to be halted. Breakpoints are inserted by the programmer to enable inspection of register contents, memory locations, variable values at fixed points in the program execution to test that the program is operating correctly. Breakpoints are removed after the program is successfully tested.</p>

Byte-invariant	<p>In a byte-invariant system, the address of each byte of memory remains unchanged when switching between little-endian and big-endian operation. When a data item larger than a byte is loaded from or stored to memory, the bytes making up that data item are arranged into the correct order depending on the endianness of the memory access.</p> <p>An ARM byte-invariant implementation also supports unaligned halfword and word memory accesses. It expects multi-word accesses to be word-aligned.</p>
Condition field	A four-bit field in an instruction that specifies a condition under which the instruction can execute.
Conditional execution	If the condition code flags indicate that the corresponding condition is true when the instruction starts executing, it executes normally. Otherwise, the instruction does nothing.
Context	The environment that each process operates in for a multitasking operating system. In ARM processors, this is limited to mean the physical address range that it can access in memory and the associated memory access permissions.
Coprocessor	A processor that supplements the main processor. Cortex-M4 does not support any coprocessors.
Debugger	A debugging system that includes a program, used to detect, locate, and correct software faults, together with custom hardware that supports software debugging.
Direct Memory Access (DMA)	An operation that accesses main memory directly, without the processor performing any accesses to the data concerned.
Doubleword	A 64-bit data item. The contents are taken as being an unsigned integer unless otherwise stated.
Doubleword-aligned	A data item having a memory address that is divisible by eight.
Endianness	<p>Byte ordering. The scheme that determines the order that successive bytes of a data word are stored in memory. An aspect of the system's memory mapping.</p> <p>See also <a href="#">“Little-endian (LE)”</a> and <a href="#">“Big-endian (BE)”</a></p>
Exception	<p>An event that interrupts program execution. When an exception occurs, the processor suspends the normal program flow and starts execution at the address indicated by the corresponding exception vector. The indicated address contains the first instruction of the handler for the exception.</p> <p>An exception can be an interrupt request, a fault, or a software-generated system exception. Faults include attempting an invalid memory access, attempting to execute an instruction in an invalid processor state, and attempting to execute an undefined instruction.</p>

Exception service routine	See <a href="#">“Interrupt handler”</a> .
Exception vector	See <a href="#">“Interrupt vector”</a> .
Flat address mapping	A system of organizing memory in which each physical address in the memory space is the same as the corresponding virtual address.
Halfword	A 16-bit data item.
Illegal instruction	An instruction that is architecturally Undefined.
Implementation-defined	The behavior is not architecturally defined, but is defined and documented by individual implementations.
Implementation-specific	The behavior is not architecturally defined, and does not have to be documented by individual implementations. Used when there are a number of implementation options available and the option chosen does not affect software compatibility.
Index register	In some load and store instruction descriptions, the value of this register is used as an offset to be added to or subtracted from the base register value to form the address that is sent to memory. Some addressing modes optionally enable the index register value to be shifted prior to the addition or subtraction. See also <a href="#">“Base register”</a> .
Instruction cycle count	The number of cycles that an instruction occupies the Execute stage of the pipeline.
Interrupt handler	A program that control of the processor is passed to when an interrupt occurs.
Interrupt vector	One of a number of fixed addresses in low memory, or in high memory if high vectors are configured, that contains the first instruction of the corresponding interrupt handler.
Little-endian (LE)	Byte ordering scheme in which bytes of increasing significance in a data word are stored at increasing addresses in memory. See also <a href="#">“Big-endian (BE)”</a> , <a href="#">“Byte-invariant”</a> , <a href="#">“Endianness”</a> .

Little-endian memory	<p>Memory in which:</p> <ul style="list-style-type: none"> <li>a byte or halfword at a word-aligned address is the least significant byte or halfword within the word at that address,</li> <li>a byte at a halfword-aligned address is the least significant byte within the halfword at that address.</li> </ul> <p>See also <a href="#">“Big-endian memory”</a> .</p>
Load/store architecture	A processor architecture where data-processing operations only operate on register contents, not directly on memory contents.
Memory Protection Unit (MPU)	Hardware that controls access permissions to blocks of memory. An MPU does not perform any address translation.
Prefetching	In pipelined processors, the process of fetching instructions from memory to fill up the pipeline before the preceding instructions have finished executing. Prefetching an instruction does not mean that the instruction has to be executed.
Preserved	Preserved by writing the same value back that has been previously read from the same field on the same processor.
Read	Reads are defined as memory operations that have the semantics of a load. Reads include the Thumb instructions LDM, LDR, LDRSH, LDRH, LDRSB, LDRB, and POP.
Region	A partition of memory space.
Reserved	<p>A field in a control register or instruction format is reserved if the field is to be defined by the implementation, or produces Unpredictable results if the contents of the field are not zero. These fields are reserved for use in future extensions of the architecture or are implementation-specific. All reserved bits not used by the implementation must be written as 0 and read as 0.</p>
Thread-safe	In a multi-tasking environment, thread-safe functions use safeguard mechanisms when accessing shared resources, to ensure correct operation without the risk of shared access conflicts.
Thumb instruction	One or two halfwords that specify an operation for a processor to perform. Thumb instructions must be halfword-aligned.
Unaligned	A data item stored at an address that is not divisible by the number of bytes that defines the data size is said to be unaligned. For example, a word stored at an address that is not divisible by four.

Undefined	Indicates an instruction that generates an Undefined instruction exception.
Unpredictable	One cannot rely on the behavior. Unpredictable behavior must not represent security holes. Unpredictable behavior must not halt or hang the processor, or any parts of the system.
Warm reset	Also known as a core reset. Initializes the majority of the processor excluding the debug controller and debug logic. This type of reset is useful if debugging features of a processor.
Word	A 32-bit data item.
Write	Writes are defined as operations that have the semantics of a store. Writes include the Thumb instructions STM, STR, STRH, STRB, and PUSH.

## 13. Debug and Test Features

### 13.1 Description

The SAM4 Series microcontrollers feature a number of complementary debug and test capabilities. The Serial Wire/JTAG Debug Port (SWJ-DP) combining a Serial Wire Debug Port (SW-DP) and JTAG Debug (JTAG-DP) port is used for standard debugging functions, such as downloading code and single-stepping through programs. It also embeds a serial wire trace.

### 13.2 Associated Documentation

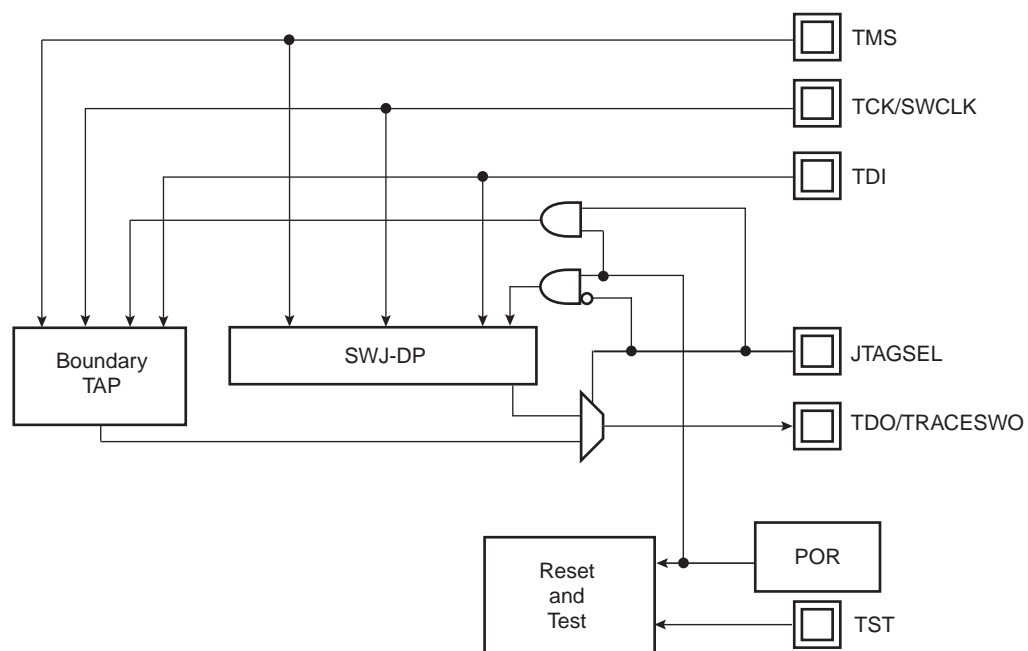
The SAM4C implements the standard ARM CoreSight™ Macrocell. For further detailed CoreSight information, the following reference documents are available from the ARM website:

- Cortex-M4/M4F Technical Reference Manual (ARM DDI 0439C)
- CoreSight Technology System Design Guide (ARM DGI 0012D)
- CoreSight Components Technical Reference Manual (ARM DDI 0314H)
- ARM Debug Interface v5 Architecture Specification (Doc. ARM IHI 0031A)
- ARMv7-M Architecture Reference Manual (ARM DDI 0403D)

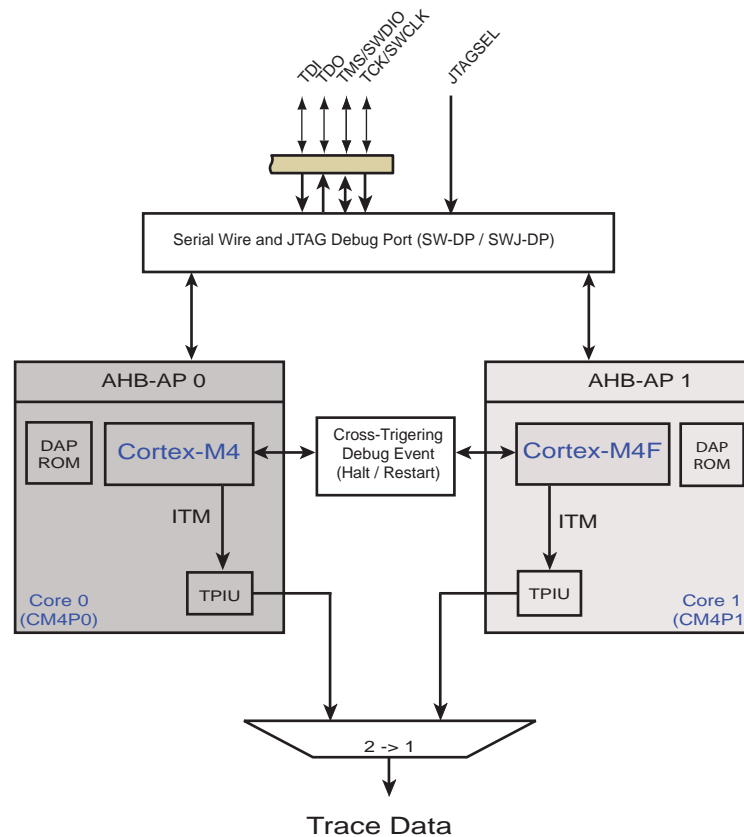
### 13.3 Embedded Characteristics

- Dual Core Debugging with common Serial Wire Debug Port (SW-DP) and Serial Wire JTAG Debug Port (SWJ-DP) debug access port connected to both cores.
- Star Topology AHB-AP Debug Access Port Implementation with common SW-DP / SWJ-DP providing higher performance than daisy-chain topology.
- Possibility to halt each core on debug event on the other core (hardware)
- Possibility to restart each core when the other core has restarted (hardware)
- Synchronization and software cross-triggering with Debugger
- Intrumentation Trace Macrocell (ITM) on both core for support of printf style debugging
- Mux 2-1 to trace chosen core (limit the number of out put pin)
- Single wire Viewer or clock mode (4-bit parallel output ports)
- Debug access to all memory and registers in the system, including Cortex-M4 register bank when the core is running, halted, or held in reset.
- Flash Patch and Breakpoint (FPB) unit for implementing breakpoints and code patches
- Data Watchpoint and Trace (DWT) unit for implementing watch points, data tracing, and system profiling
- IEEE 1149.1 JTAG Boundary scan on All Digital Pins

Figure 13-1. Debug and Test Block Diagram



**Figure 13-2. Dual Core Debug Architecture**



The above figure depicts the dual core debug implementation using only one SW-JTAG/SW-DP Debug Access Port. Star topology has been used to connect the AHB-AP 0 (Core 0) and AHB-AP 1 (Core) rather than legacy daisy chaining method. It is providing higher performance than daisy-chain topology. This core debug architecture is fully supported by debug tools vendors.



## 13.4 Cross Triggering Debut Events

The Cross Triggerring (CT) as shown in the above figure is an Atmel Module allowing the two cores to send and receive debug events to and from each other. It might be needed to debug two applications at the same time (one application running on each core).

For that case, the CT allows core 0 (or 1) to trig a debug event (halt) to core 1 (or 0) to enter debug mode. The debug event can be sent when the core 0 (or 1) is entering debug mode (such as brealpoint) or at run-time. It means that an user application running on core 0 (or 1) can put core 1 (or 0) without entering debug mode.

Once core 0 (or 1) gets out of debug mode, it releases core 1 (0) from debug mode as well.

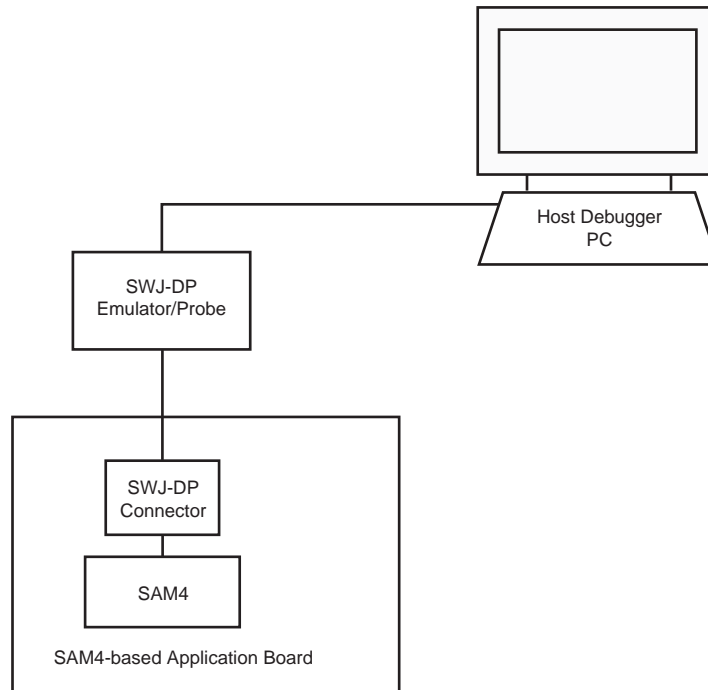
The Cross Triggerring configuration is located in the Special Function Register in the Matrix user Interface.

## 13.5 Application Examples

### 13.5.1 Debug Environment

Figure 13-3 shows a complete debug environment example. The SWJ-DP interface is used for standard debugging functions, such as downloading code and single-stepping through the program and viewing core and peripheral registers.

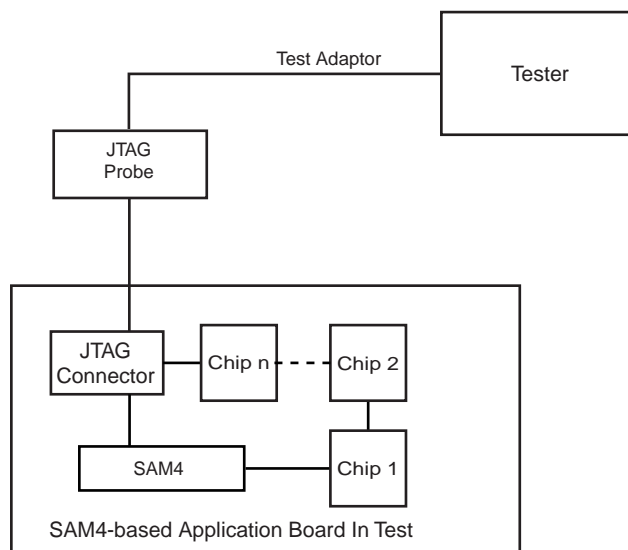
**Figure 13-3. Application Debug Environment Example**



### 13.5.2 Test Environment

Figure 13-4 shows a test environment example (JTAG Boundary scan). Test vectors are sent and interpreted by the tester. In this example, the “board in test” is designed using a number of JTAG-compliant devices. These devices can be connected to form a single scan chain.

Figure 13-4. Application Test Environment Example



### 13.6 Debug and Test Pin Description

Table 13-1. Debug and Test Signal List

Signal Name	Function	Type	Active Level
<b>Reset/Test</b>			
NRST	Microcontroller Reset	Input/Output	Low
TST	Test Select	Input	
<b>SWD/JTAG</b>			
TCK/SWCLK	Test Clock/Serial Wire Clock	Input	
TDI	Test Data In	Input	
TDO/TRACESWO	Test Data Out/Trace Asynchronous Data Out	Output	
TMS/SWDIO	Test Mode Select/Serial Wire Input/Output	Input	
JTAGSEL	JTAG Selection	Input	High

## 13.7 Functional Description

### 13.7.1 Test Pin

One dedicated pin, TST, is used to define the device operating mode. When this pin is at low level during power-up, the device is in normal operating mode. When at high level, the device is in test mode or FFPI mode. The TST pin integrates a permanent pull-down resistor of about 15 k $\Omega$ , so that it can be left unconnected for normal operation. Note that when setting the TST pin to low or high level at power up, it must remain in the same state during the duration of the whole operation.

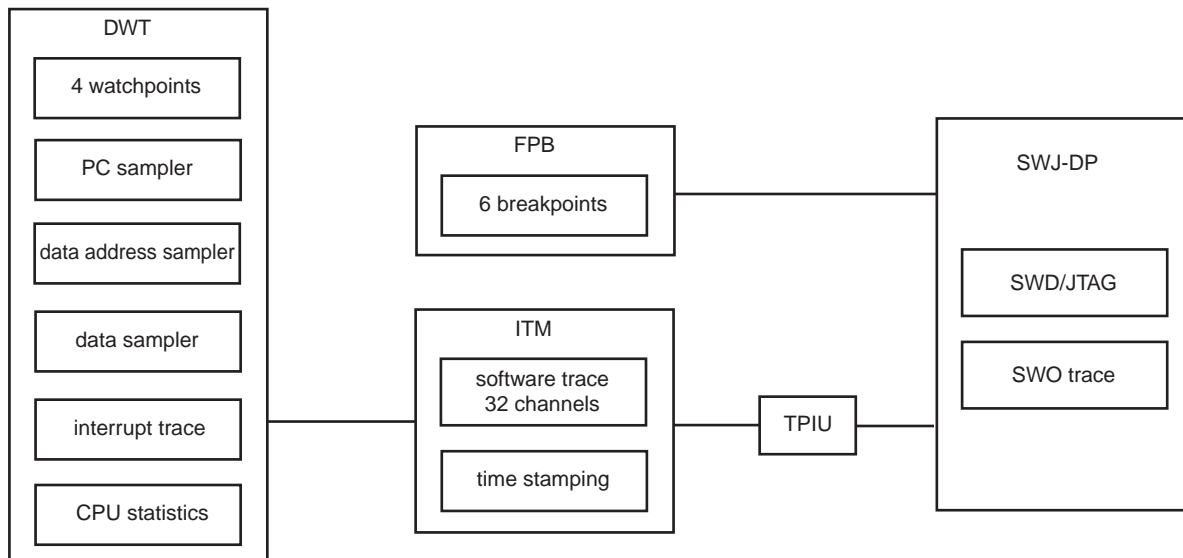
### 13.7.2 Debug Architecture

Figure 13-5 shows the Debug Architecture used in the SAM4. The Cortex-M4 embeds four functional units for debug:

- SWJ-DP (Serial Wire/JTAG Debug Port)
- FPB (Flash Patch Breakpoint)
- DWT (Data Watchpoint and Trace)
- ITM (Instrumentation Trace Macrocell)
- TPIU (Trace Port Interface Unit)

The debug architecture information that follows is mainly dedicated to developers of SWJ-DP Emulators/Probes and debugging tool vendors for Cortex-M4 based microcontrollers. For further details on SWJ-DP see the Cortex-M4 technical reference manual.

Figure 13-5. Debug Architecture



### 13.7.3 Serial Wire/JTAG Debug Port (SWJ-DP)

The Cortex-M4 embeds a SWJ-DP Debug port which is the standard CoreSight debug port. It combines Serial Wire Debug Port (SW-DP), from 2 to 3 pins and JTAG debug Port (JTAG-DP), 5 pins.

By default, the JTAG Debug Port is active. If the host debugger wants to switch to the Serial Wire Debug Port, it must provide a dedicated JTAG sequence on TMS/SWDIO and TCK/SWCLK which disables JTAG-DP and enables SW-DP.

When the Serial Wire Debug Port is active, TDO/TRACESWO can be used for trace. The asynchronous TRACE output (TRACESWO) is multiplexed with TDO. So the asynchronous trace can only be used with SW-DP, not JTAG-DP.

**Table 13-2. SWJ-DP Pin List**

Pin Name	JTAG Port	Serial Wire Debug Port
TMS/SWDIO	TMS	SWDIO
TCK/SWCLK	TCK	SWCLK
TDI	TDI	-
TDO/TRACESWO	TDO	TRACESWO (optional: trace)

SW-DP or JTAG-DP mode is selected when JTAGSEL is low. It is not possible to switch directly between SWJ-DP and JTAG boundary scan operations. A chip reset must be performed after JTAGSEL is changed.

#### 13.7.3.1 SW-DP and JTAG-DP Selection Mechanism

Debug port selection mechanism is done by sending specific **SWDIOTMS** sequence. The JTAG-DP is selected by default after reset.

- Switch from JTAG-DP to SW-DP. The sequence is:
  - Send more than 50 **SWCLKTCK** cycles with **SWDIOTMS** = 1
  - Send the 16-bit sequence on **SWDIOTMS** = 0111100111100111 (0x79E7 MSB first)
  - Send more than 50 **SWCLKTCK** cycles with **SWDIOTMS** = 1
- Switch from SWD to JTAG. The sequence is:
  - Send more than 50 **SWCLKTCK** cycles with **SWDIOTMS** = 1
  - Send the 16-bit sequence on **SWDIOTMS** = 0011110011100111 (0x3CE7 MSB first)
  - Send more than 50 **SWCLKTCK** cycles with **SWDIOTMS** = 1

### 13.7.4 FPB (Flash Patch Breakpoint)

The FPB:

- Implements hardware breakpoints
- Patches code and data from code space to system space.

The FPB unit contains:

- Two literal comparators for matching against literal loads from Code space, and remapping to a corresponding area in System space.
- Six instruction comparators for matching against instruction fetches from Code space and remapping to a corresponding area in System space.
- Alternatively, comparators can also be configured to generate a Breakpoint instruction to the processor core on a match.

### 13.7.5 DWT (Data Watchpoint and Trace)

The DWT contains four comparators which can be configured to generate the following:

- PC sampling packets at set intervals
- PC or Data watchpoint packets
- Watchpoint event to halt core

The DWT contains counters for the items that follow:

- Clock cycle (CYCCNT)
- Folded instructions
- Load Store Unit (LSU) operations
- Sleep Cycles
- CPI (all instruction cycles except for the first cycle)
- Interrupt overhead

### 13.7.6 ITM (Instrumentation Trace Macrocell)

The ITM is an application driven trace source that supports printf style debugging to trace Operating System (OS) and application events, and emits diagnostic system information. The ITM emits trace information as packets which can be generated by three different sources with several priority levels:

- **Software trace:** Software can write directly to ITM stimulus registers. This can be done thanks to the “printf” function. For more information, refer to [Section 13.7.6.1 “How to Configure the ITM”](#).
- **Hardware trace:** The ITM emits packets generated by the DWT.
- **Time stamping:** Timestamps are emitted relative to packets. The ITM contains a 21-bit counter to generate the timestamp.

#### 13.7.6.1 How to Configure the ITM

The following example describes how to output trace data in asynchronous trace mode.

- Configure the TPIU for asynchronous trace mode (refer to [Section 13.7.6.3 “How to Configure the TPIU”](#))
- Enable the write accesses into the ITM registers by writing “0xC5ACCE55” into the Lock Access Register (Address: 0xE0000FB0)
- Write 0x00010015 into the Trace Control Register:
  - Enable ITM
  - Enable Synchronization packets
  - Enable SWO behavior
  - Fix the ATB ID to 1
- Write 0x1 into the Trace Enable Register:
  - Enable the Stimulus port 0
- Write 0x1 into the Trace Privilege Register:
  - Stimulus port 0 only accessed in privileged mode (Clearing a bit in this register will result in the corresponding stimulus port being accessible in user mode.)
- Write into the Stimulus port 0 register: TPIU (Trace Port Interface Unit)

The TPIU acts as a bridge between the on-chip trace data and the Instruction Trace Macrocell (ITM).

The TPIU formats and transmits trace data off-chip at frequencies asynchronous to the core.

### 13.7.6.2 Asynchronous Mode

The TPIU is configured in asynchronous mode, trace data are output using the single TRACESWO pin. The TRACESWO signal is multiplexed with the TDO signal of the JTAG Debug Port. As a consequence, asynchronous trace mode is only available when the Serial Wire Debug mode is selected since TDO signal is used in JTAG debug mode.

Two encoding formats are available for the single pin output:

- Manchester encoded stream. This is the reset value.
- NRZ\_based UART byte structure

### 13.7.6.3 How to Configure the TPIU

This example only concerns the asynchronous trace mode.

- Set the TRCENA bit to 1 into the Debug Exception and Monitor Register (0xE00EDFC) to enable the use of trace and debug blocks.
- Write 0x2 into the Selected Pin Protocol Register
  - Select the Serial Wire Output – NRZ
- Write 0x100 into the Formatter and Flush Control Register
- Set the suitable clock prescaler value into the Async Clock Prescaler Register to scale the baud rate of the asynchronous output (this can be done automatically by the debugging tool).

### 13.7.7 IEEE 1149.1 JTAG Boundary Scan

IEEE 1149.1 JTAG Boundary Scan allows pin-level access independent of the device packaging technology.

IEEE 1149.1 JTAG Boundary Scan is enabled when TST is tied low, while JTAGSEL is high and PA7 is tied low during the power-up, and must be kept in this state during the whole boundary scan operation. The SAMPLE, EXTEST and BYPASS functions are implemented. In SWD/JTAG debug mode, the ARM processor responds with a non-JTAG chip ID that identifies the processor. This is not IEEE 1149.1 JTAG-compliant.

It is not possible to switch directly between JTAG Boundary Scan and SWJ Debug Port operations. A chip reset must be performed after JTAGSEL is changed.

A Boundary-scan Descriptor Language (BSDL) file is provided on [Atmel's web site](#) to set up the test.

#### 13.7.7.1 JTAG Boundary-scan Register

The Boundary-scan Register (BSR) contains a number of bits which correspond to active pins and associated control signals.

Each SAM4 input/output pin corresponds to a 3-bit register in the BSR. The OUTPUT bit contains data that can be forced on the pad. The INPUT bit facilitates the observability of data applied to the pad. The CONTROL bit selects the direction of the pad.

For more information, refer to BSDL files available for the SAM4 Series.

### 13.7.8 ID Code Register

Access: Read-only

31	30	29	28	27	26	25	24
VERSION				PART NUMBER			
23	22	21	20	19	18	17	16
PART NUMBER							
15	14	13	12	11	10	9	8
PART NUMBER				MANUFACTURER IDENTITY			
7	6	5	4	3	2	1	0
MANUFACTURER IDENTITY							1

- **VERSION[31:28]: Product Version Number**

Set to 0x0.

- **PART NUMBER[27:12]: Product Part Number**

Chip Name	Chip ID
SAM4C	0x05B34

- **MANUFACTURER IDENTITY[11:1]**

Set to 0x01F.

- **Bit[0] Required by IEEE Std. 1149.1.**

Set to 0x1.

Chip Name	JTAG ID Code
SAM4C	0x05B3_403F

## 14. SAM4C Boot Program

### 14.1 Description

The SAM-BA Boot Program integrates an array of programs permitting download and/or upload into the different memories of the product.

### 14.2 Hardware and Software Constraints

- SAM-BA Boot uses the first 4096 bytes of the SRAM for variables and stacks. The remaining available size can be used for user's code.
- UART0 requirements: None

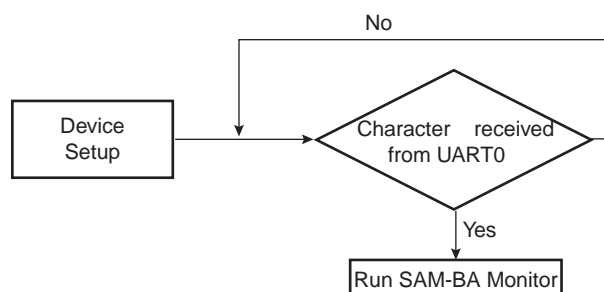
Table 14-1. Pins Driven during Boot Program Execution

Peripheral	Pin	PIO Line
UART0	URXD0	PB4
UART0	UTXD0	PB5

### 14.3 Flow Diagram

The Boot Program implements the algorithm in [Figure 14-1](#).

Figure 14-1. Boot Program Algorithm Flow Diagram



The SAM-BA Boot program uses the internal 12 MHz RC oscillator as source clock for PLL. The MCK runs from PLL divided by 2. The core runs at 48 MHz.

### 14.4 Device Initialization

Initialization follows the steps described below:

1. Stack setup
2. Setup the Embedded Flash Controller
3. Switch on internal 12 MHz RC oscillator
4. Configure PLLB to run at 48 MHz
5. Configure UART0
6. Disable Watchdog
7. Wait for a character on UART0
8. Jump to SAM-BA monitor (see [Section 14.5 "SAM-BA Monitor"](#))



## 14.5 SAM-BA Monitor

The SAM-BA boot principle:

Once the communication interface is identified, to run in an infinite loop waiting for different commands as shown in [Table 14-2](#).

**Table 14-2. Commands Available through the SAM-BA Boot**

Command	Action	Argument(s)	Example
<b>N</b>	Set Normal mode	No argument	<b>N#</b>
<b>T</b>	Set Terminal mode	No argument	<b>T#</b>
<b>O</b>	Write a byte	Address, Value#	<b>O200001,CA#</b>
<b>o</b>	Read a byte	Address,#	<b>o200001,#</b>
<b>H</b>	Write a half word	Address, Value#	<b>H200002,CAFE#</b>
<b>h</b>	Read a half word	Address,#	<b>h200002,#</b>
<b>W</b>	Write a word	Address, Value#	<b>W200000,CAFEDCA#</b>
<b>w</b>	Read a word	Address,#	<b>w200000,#</b>
<b>S</b>	Send a file	Address,#	<b>S200000,#</b>
<b>R</b>	Receive a file	Address, NbOfBytes#	<b>R200000,1234#</b>
<b>G</b>	Go	Address#	<b>G200200#</b>
<b>V</b>	Display version	No argument	<b>V#</b>

- Mode commands:
  - Normal mode configures SAM-BA Monitor to send/receive data in binary format
  - Terminal mode configures SAM-BA Monitor to send/receive data in ASCII format
- Write commands: Write a byte (**O**), a halfword (**H**) or a word (**W**) to the target
  - *Address*: Address in hexadecimal
  - *Value*: Byte, halfword or word to write in hexadecimal
- Read commands: Read a byte (**o**), a halfword (**h**) or a word (**w**) from the target
  - *Address*: Address in hexadecimal
  - *Output*: The byte, halfword or word read in hexadecimal
- Send a file (**S**): Send a file to a specified address
  - *Address*: Address in hexadecimal

**Note:** There is a time-out on this command which is reached when the prompt '>' appears before the end of the command execution.

- Receive a file (**R**): Receive data into a file from a specified address
  - *Address*: Address in hexadecimal
  - *NbOfBytes*: Number of bytes in hexadecimal to receive
- Go (**G**): Jump to a specified address and execute the code
  - *Address*: Address to jump in hexadecimal
- Get Version (**V**): Return the SAM-BA boot version

**Note:** In Terminal mode, when the requested command is performed, SAM-BA Monitor adds the following prompt sequence to its answer: <LF>+<CR>+>'.

### 14.5.1 UART0 Serial Port

Communication is performed through the UART0 initialized to 115200 Baud, 8, n, 1.

The Send and Receive File commands use the Xmodem protocol to communicate. Any terminal performing this protocol can be used to send the application file to the target. The size of the binary file to send depends on the SRAM size embedded in the product. In all cases, the size of the binary file must be lower than the SRAM size because the Xmodem protocol requires some SRAM memory to work. See [Section 14.2 "Hardware and Software Constraints"](#)

### 14.5.2 Xmodem Protocol

The supported Xmodem protocol is the 128-byte length block. This protocol uses a two-character CRC-16 to guarantee detection of a maximum bit error.

Xmodem protocol with CRC is accurate provided both sender and receiver report a successful transmission. Each block of the transfer looks like:

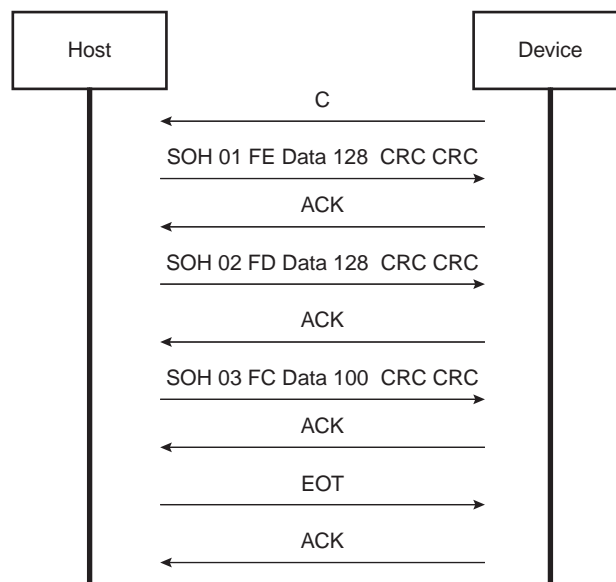
<SOH><blk #><255-blk #><--128 data bytes--><checksum>

where:

- <SOH> = 01 hex
- <blk #> = binary number, starts at 01, increments by 1, and wraps 0FFH to 00H (not to 01)
- <255-blk #> = 1's complement of the blk#.
- <checksum> = 2 bytes CRC16

[Figure 14-2](#) shows a transmission using this protocol.

**Figure 14-2. Xmodem Transfer Example**



### 14.5.3 In Application Programming (IAP) Feature

The IAP feature is a function located in ROM that can be called by any software application.

When called, this function sends the desired FLASH command to the EEFC and waits for the Flash to be ready (looping while the FRDY bit is not set in the MC\_FSR register).

Since this function is executed from ROM, this allows Flash programming (such as sector write) to be done by code running in Flash.

The IAP function entry point is retrieved by reading the NMI vector in ROM (0x02000008).

This function takes one argument in parameter: the command to be sent to the EEFC.

This function returns the value of the MC\_FSR register.

IAP software code example:

```
(unsigned int) (*IAP_Function)(unsigned long);
void main (void){

    unsigned long FlashSectorNum = 200; //
    unsigned long flash_cmd = 0;
    unsigned long flash_status = 0;
    unsigned long EFCIndex = 0; // 0:EEFC0, 1: EEFC1

    /* Initialize the function pointer (retrieve function address from NMI vector)
    */

    IAP_Function = ((unsigned long) (*)(unsigned long))
0x02000008;

    /* Send your data to the sector here */

    /* build the command to send to EEFC */

    flash_cmd = (0x5A << 24) | (FlashSectorNum << 8) |
AT91C_MC_FCMD_EWP;

    /* Call the IAP function with appropriate command */

    flash_status = IAP_Function (EFCIndex, flash_cmd);

}
```

## 15. Reset Controller (RSTC)

### 15.1 Description

The Reset Controller (RSTC), based on power-on reset cells, handles all the resets of the system without any external components. It reports which reset occurred last.

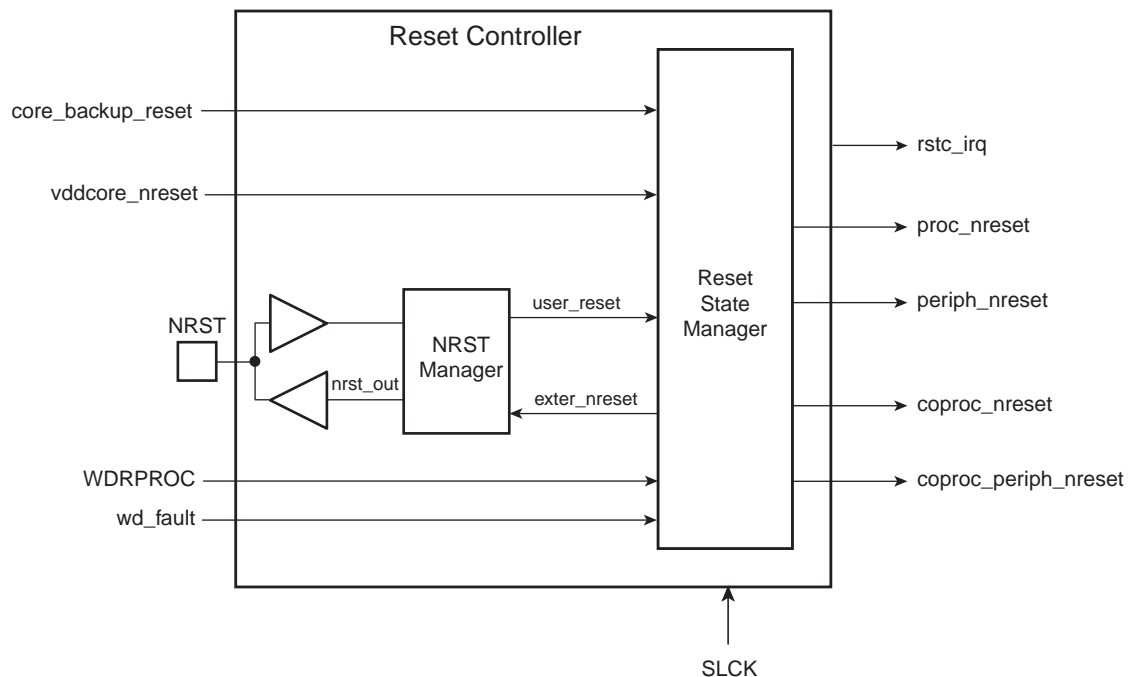
The Reset Controller also drives independently or simultaneously the external reset and the peripheral and processor resets.

### 15.2 Embedded Characteristics

- Manages all Resets of the System, Including
  - External Devices through the NRST Pin
  - Processor Reset and Coprocessor (second processor) Reset
  - Processor Peripheral Set Reset and Coprocessor Peripheral Set Reset
- Based on Embedded Power-on Cell
- Reset Source Status
  - Status of the Last Reset
  - Either Software Reset, User Reset, Watchdog Reset
- External Reset Signal Shaping

### 15.3 Block Diagram

Figure 15-1. Reset Controller Block Diagram



## 15.4 Functional Description

### 15.4.1 Reset Controller Overview

The Reset Controller is made up of an NRST Manager and a Reset State Manager. It runs at Slow Clock and generates the following reset signals:

- `proc_nreset`: Processor reset line. It also resets the Watchdog Timer
- `coproc_nreset`: Coprocessor (second processor) reset line
- `periph_nreset`: Affects the whole set of embedded peripherals
- `coproc_periph_nreset`: Affects the whole set of embedded peripherals driven by the Co-processor
- `nrst_out`: Drives the NRST pin

These reset signals are asserted by the Reset Controller, either on external events or on software action. The Reset State Manager controls the generation of reset signals and provides a signal to the NRST Manager when an assertion of the NRST pin is required.

The NRST Manager shapes the NRST assertion during a programmable time, thus controlling external device resets.

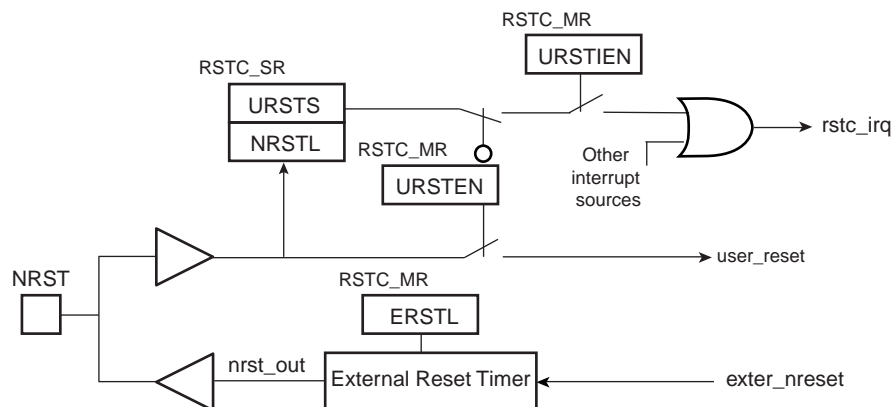
The Reset Controller Mode Register (`RSTC_MR`), allowing the configuration of the Reset Controller, is powered with `VDD_REG_BUP`, so that its configuration is saved as long as `VDD_REG_BUP` is on.

### 15.4.2 NRST Manager

After power-up, NRST is an output during the `ERSTL` time period defined in the `RSTC_MR`. When `ERSTL` has elapsed, the pin behaves as an input and all the system is held in reset if NRST is tied to GND by an external signal.

The NRST Manager samples the NRST input pin and drives this pin low when required by the Reset State Manager. [Figure 15-2](#) shows the block diagram of the NRST Manager.

Figure 15-2. NRST Manager



#### 15.4.2.1 NRST Signal or Interrupt

The NRST Manager samples the NRST pin at Slow Clock speed. When the line is detected low, a User Reset is reported to the Reset State Manager.

However, the NRST Manager can be programmed to not trigger a reset when an assertion of NRST occurs. Writing the bit `URSTEN` at 0 in `RSTC_MR` disables the User Reset trigger.

The level of the pin NRST can be read at any time in the bit `NRSTL` (NRST level) in `RSTC_SR`. As soon as the pin NRST is asserted, the bit `URSTS` in `RSTC_SR` is set. This bit clears only when `RSTC_SR` is read.

The Reset Controller can also be programmed to generate an interrupt instead of generating a reset. To do so, the bit `URSTIEN` in `RSTC_MR` must be written at 1.

### 15.4.2.2 NRST External Reset Control

The Reset State Manager asserts the signal `ext_nreset` to assert the NRST pin. When this occurs, the “`nrst_out`” signal is driven low by the NRST Manager for a time programmed by the field `ERSTL` in `RSTC_MR`. This assertion duration, named `EXTERNAL_RESET_LENGTH`, lasts  $2^{(ERSTL+1)}$  Slow Clock cycles. This gives the approximate duration of an assertion between 60  $\mu$ s and 2 seconds. Note that `ERSTL` at 0 defines a two-cycle duration for the NRST pulse.

This feature allows the Reset Controller to shape the NRST pin level, and thus to guarantee that the NRST line is driven low for a time compliant with potential external devices connected on the system reset.

As the `ERSTL` field is within `RSTC_MR` register, which is backed-up, it can be used to shape the system power-up reset for devices requiring a longer startup time than the Slow Clock Oscillator.

### 15.4.3 Brownout Manager

The Brownout manager is embedded within the Supply Controller, refer to the product Supply Controller section for a detailed description.

### 15.4.4 Reset States

The Reset State Manager handles the different reset sources and generates the internal reset signals. It reports the reset status in the field `RSTTYP` of the Status Register (`RSTC_SR`). The update of the field `RSTTYP` is performed when the processor reset is released.

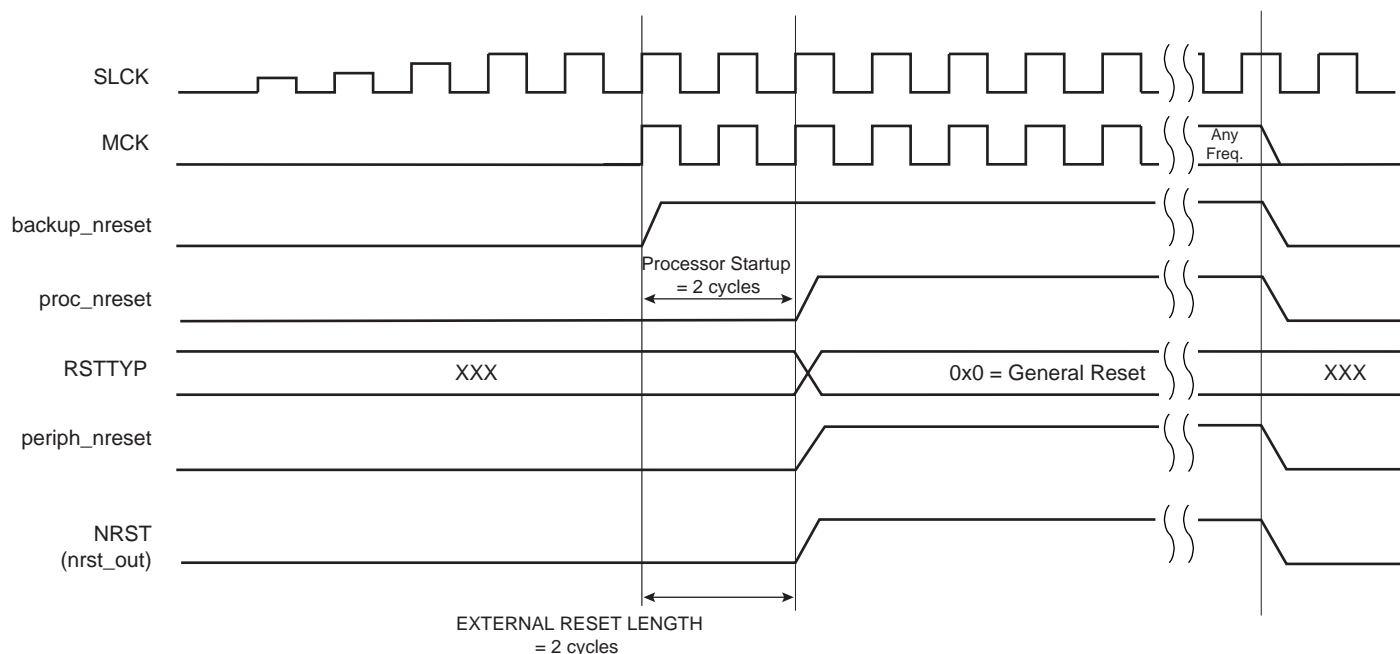
#### 15.4.4.1 General Reset

A general reset occurs when a Power-on-reset is detected, a Brownout or a Voltage regulation loss is detected by the Supply controller. The `vddcore_nreset` signal is asserted by the Supply Controller when a general reset occurs.

All the reset signals are released and the field `RSTTYP` in `RSTC_SR` reports a General Reset. As the `RSTC_MR` is reset, the NRST line rises 2 cycles after the `vddcore_nreset`, as `ERSTL` defaults at value 0x0.

Figure 15-3 shows how the General Reset affects the reset signals.

Figure 15-3. General Reset State



#### 15.4.4.2 Backup Reset

A Backup reset occurs when the chip returns from Backup Mode. The `core_backup_reset` signal is asserted by the Supply Controller when a Backup reset occurs.

The field `RSTTYP` in `RSTC_SR` is updated to report a Backup Reset.

#### 15.4.4.3 User Reset

The User Reset is entered when a low level is detected on the `NRST` pin and the bit `URSTEN` in `RSTC_MR` is at 1. The `NRST` input signal is resynchronized with `SLCK` to insure proper behavior of the system.

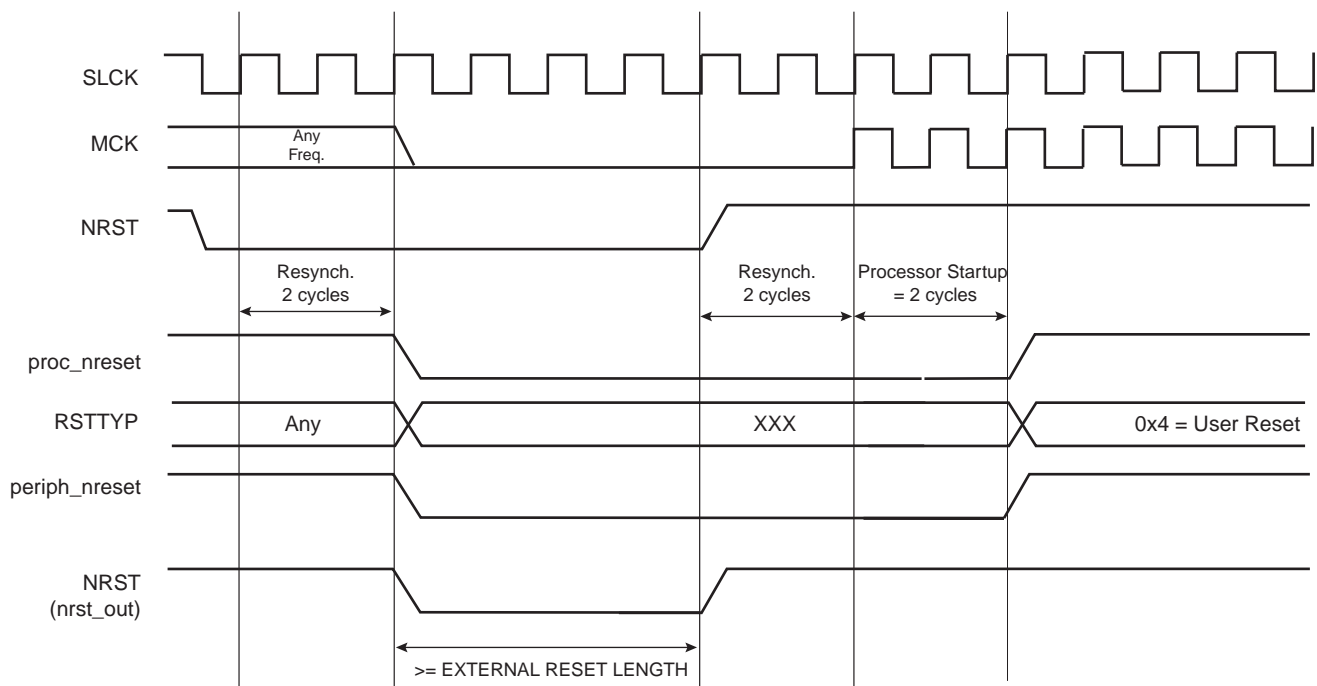
The User Reset is entered as soon as a low level is detected on `NRST`. The Processor and Coprocessor Reset and the Peripheral Resets are asserted.

The User Reset is left when `NRST` rises, after a two-cycle resynchronization time and a 3-cycle processor startup. The processor clock is re-enabled as soon as `NRST` is confirmed high.

When the processor reset signal is released, the `RSTTYP` field of the Status Register (`RSTC_SR`) is loaded with the value `0x4`, indicating a User Reset.

The `NRST` Manager guarantees that the `NRST` line is asserted for `EXTERNAL_RESET_LENGTH` Slow Clock cycles, as programmed in the field `ERSTL`. However, if `NRST` does not rise after `EXTERNAL_RESET_LENGTH` because it is driven low externally, the internal reset lines remain asserted until `NRST` actually rises.

**Figure 15-4. User Reset State**



#### 15.4.4.4 Software Reset

The Reset Controller offers several commands used to assert the different reset signals. These commands are performed by writing the Control Register (RSTC\_CR) or Coprocessor Mode Register with the following bits at 1:

- PROCRST: Writing PROCRST at 1 resets the processor and the watchdog timer
- PERRST: Writing PERRST at 1 resets all the embedded peripherals associated to processor whereas the coprocessor peripherals are not reset, including the memory system, and, in particular, the Remap Command. The Peripheral Reset is generally used for debug purposes. Except for debug purposes, PERRST must always be used in conjunction with PROCRST (PERRST and PROCRST set both at 1 simultaneously).
- CPROCEN: Writing CPROCEN at 0 resets the coprocessor only.
- CPEREN: Writing CPEREN at 0 resets all the embedded peripherals associated to coprocessor whereas the processor peripherals are not reset.
- EXTRST: Writing EXTRST at 1 asserts low the NRST pin during a time defined by the field ERSTL in the Mode Register (RSTC\_MR).

The software reset is entered if at least one of these bits is set by the software. All these commands can be performed independently or simultaneously. The software reset lasts 3 Slow Clock cycles.

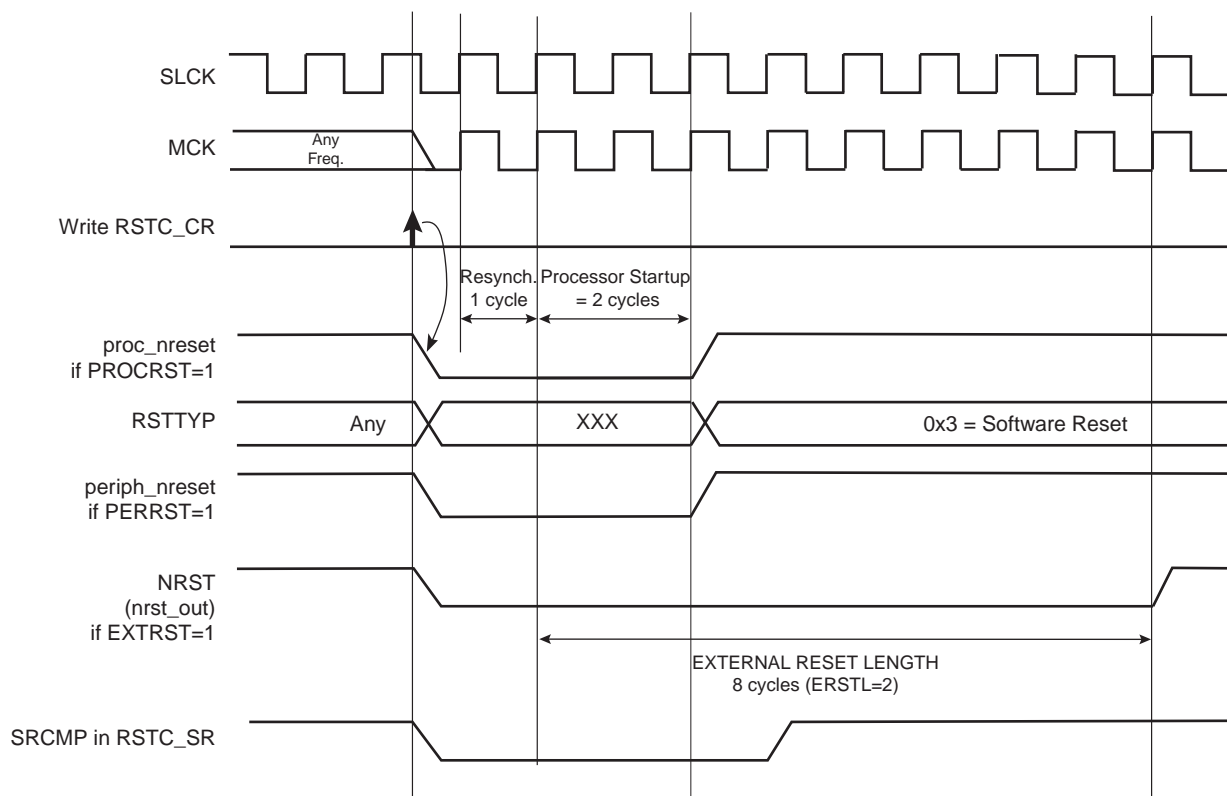
The internal reset signals are asserted as soon as the register write is performed. This is detected on the Master Clock (MCK). They are released when the software reset is left, i.e.; synchronously to SLCK.

If EXTRST is set, the nrst\_out signal is asserted depending on the programming of the field ERSTL. However, the resulting falling edge on NRST does not lead to a User Reset.

If and only if the PROCRST bit is set, the Reset Controller reports the software status in the field RSTTYP of the Status Register (RSTC\_SR). Other Software Resets are not reported in RSTTYP.

As soon as a software operation is detected, the bit SRCMP (Software Reset Command in Progress) is set in the Status Register (RSTC\_SR). It is cleared as soon as the software reset is left. No other software reset can be performed while the SRCMP bit is set, and writing any value in RSTC\_CR has no effect.

Figure 15-5. Software Reset





#### 15.4.4.5 Watchdog Reset

The Watchdog Reset is entered when a watchdog fault occurs. This state lasts 3 Slow Clock cycles.

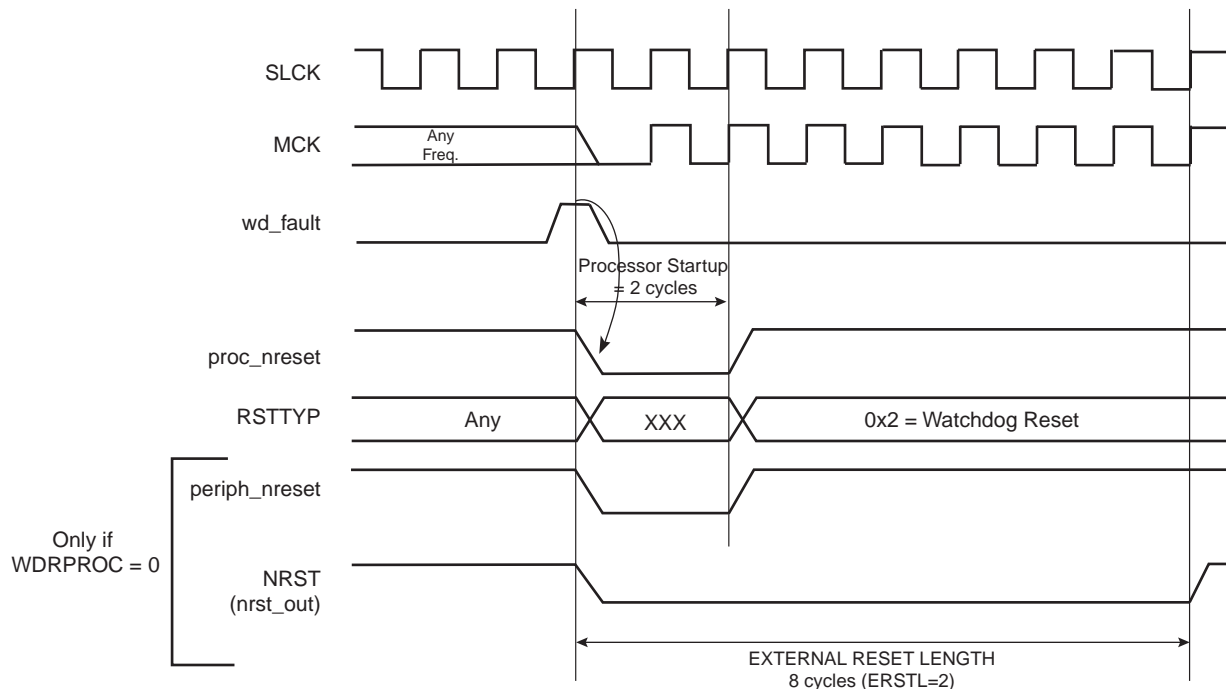
When in Watchdog Reset, assertion of the reset signals depends on the WDRPROC bit in WDT\_MR:

- If WDRPROC is 0, the Processor Reset and the Peripheral Reset are asserted. The NRST line is also asserted, depending on the programming of the field ERSTL. However, the resulting low level on NRST does not result in a User Reset state.
- If WDRPROC = 1, only the processor reset is asserted.

The Watchdog Timer is reset by the proc\_nreset signal. As the watchdog fault always causes a processor reset if WDRSTEN is set, the Watchdog Timer is always reset after a Watchdog Reset, and the Watchdog is enabled by default and with a period set to a maximum.

When the WDRSTEN in WDT\_MR bit is reset, the watchdog fault has no impact on the reset controller.

**Figure 15-6. Watchdog Reset**



### 15.4.5 Reset State Priorities

The Reset State Manager manages the following priorities between the different reset sources, given in descending order:

- General Reset
- Backup Reset
- Watchdog Reset
- Software Reset
- User Reset

Particular cases are listed below:

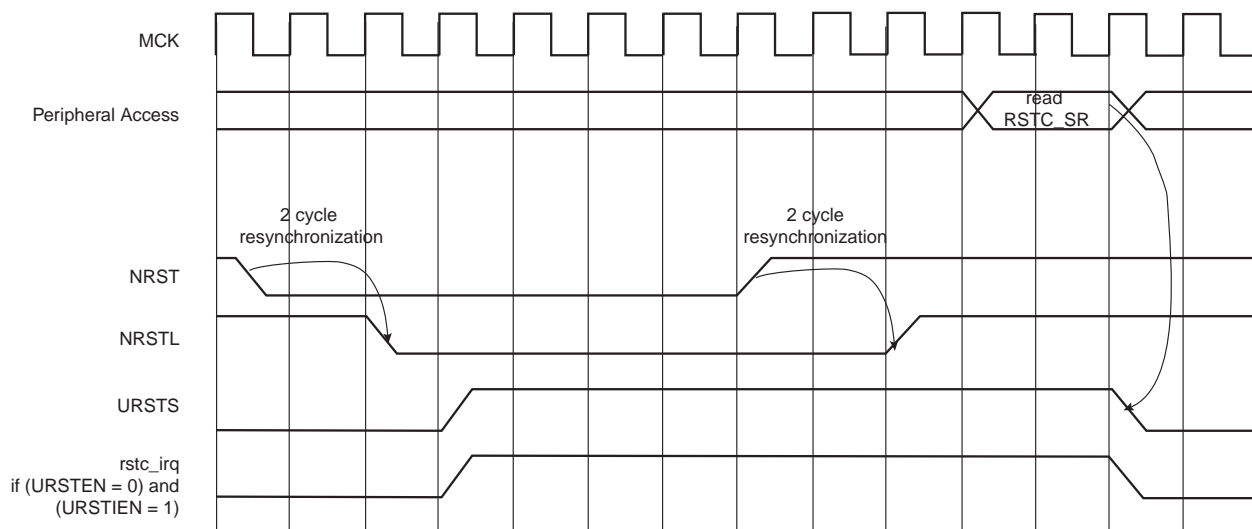
- When in User Reset:
  - A watchdog event is impossible because the Watchdog Timer is being reset by the `proc_nreset` signal.
  - A software reset is impossible, since the processor reset is being activated.
- When in Software Reset:
  - A watchdog event has priority over the current state.
  - The `NRST` has no effect.
- When in Watchdog Reset:
  - The processor reset is active and so a Software Reset cannot be programmed.
  - A User Reset cannot be entered.

### 15.4.6 Reset Controller Status Register

The Reset Controller status register (`RSTC_SR`) provides several status fields:

- `RSTTYP` field: This field gives the type of the last reset, as explained in previous sections.
- `SRCMP` bit: This field indicates that a Software Reset Command is in progress and that no further software reset should be performed until the end of the current one. This bit is automatically cleared at the end of the current software reset.
- `NRSTL` bit: The `NRSTL` bit of the Status Register gives the level of the `NRST` pin sampled on each `MCK` rising edge.
- `URSTS` bit: A high-to-low transition of the `NRST` pin sets the `URSTS` bit of the `RSTC_SR` register. This transition is also detected on the Master Clock (`MCK`) rising edge (see Figure 15-7). If the User Reset is disabled (`URSTEN` = 0) and if the interruption is enabled by the `URSTIEN` bit in the `RSTC_MR` register, the `URSTS` bit triggers an interrupt. Reading the `RSTC_SR` status register resets the `URSTS` bit and clears the interrupt.

Figure 15-7. Reset Controller Status and Interrupt



## 15.5 Reset Controller (RSTC) User Interface

Table 15-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	RSTC_CR	Write-only	—
0x04	Status Register	RSTC_SR	Read-only	0x0000_0000
0x08	Mode Register	RSTC_MR	Read-write	0x0000 0001
0x0C	Coprocessor Mode Register	RSTC_CPMR	Read-write	0x0000_0000

### 15.5.1 Reset Controller Control Register

**Name:** RSTC\_CR  
**Address:** 0x400E1400  
**Access:** Write-only

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	EXTRST	PERRST	–	PROCRST

- **PROCRST: Processor Reset**

0 = No effect.

1 = If KEY is correct, resets the processor.

- **PERRST: Peripheral Reset**

0 = No effect.

1 = If KEY is correct, resets the processor peripherals.

- **EXTRST: External Reset**

0 = No effect.

1 = If KEY is correct, asserts the NRST pin and resets the processor and the peripherals.

- **KEY: System Reset Key**

Value	Name	Description
0xA5	PASSWD	Writing any other value in this field aborts the write operation.

## 15.5.2 Reset Controller Status Register

**Name:** RSTC\_SR

**Address:** 0x400E1404

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	SRCMP	NRSTL
15	14	13	12	11	10	9	8
–	–	–	–	–	RSTTYP		
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	URSTS

- **URSTS: User Reset Status**

0 = No high-to-low edge on NRST happened since the last read of RSTC\_SR.

1 = At least one high-to-low transition of NRST has been detected since the last read of RSTC\_SR.

- **RSTTYP: Reset Type**

Value	Name	Description
0	General Reset	First power-up Reset
1	Backup Reset	Return from Backup Mode
2	Watchdog Reset	Watchdog fault occurred
3	Software Reset	Processor reset required by the software
4	User Reset	NRST pin detected low

Reports the cause of the last processor reset. Reading this RSTC\_SR does not reset this field.

- **NRSTL: NRST Pin Level**

Registers the NRST Pin Level at Master Clock (MCK).

- **SRCMP: Software Reset Command in Progress**

0 = No software command is being performed by the reset controller. The reset controller is ready for a software command.

1 = A software reset command is being performed by the reset controller. The reset controller is busy.

### 15.5.3 Reset Controller Mode Register

**Name:** RSTC\_MR  
**Address:** 0x400E1408  
**Access:** Read-write

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	ERSTL			
7	6	5	4	3	2	1	0
–	–		URSTIEN	–	–	–	URSTEN

- **URSTEN: User Reset Enable**

0 = The detection of a low level on the pin NRST does not generate a User Reset.

1 = The detection of a low level on the pin NRST triggers a User Reset.

- **URSTIEN: User Reset Interrupt Enable**

0 = USRTS bit in RSTC\_SR at 1 has no effect on rstc\_irq.

1 = USRTS bit in RSTC\_SR at 1 asserts rstc\_irq if URSTEN = 0.

- **ERSTL: External Reset Length**

This field defines the external reset length. The external reset is asserted during a time of  $2^{(ERSTL+1)}$  Slow Clock cycles. This allows assertion duration to be programmed between 60  $\mu$ s and 2 seconds.

- **KEY: Write Access Password**

Value	Name	Description
0xA5	PASSWD	Writing any other value in this field aborts the write operation. Always reads as 0.

### 15.5.4 Reset Controller Coprocessor Mode Register

**Name:** RSTC\_CPMR

**Address:** 0x400E140C

**Access:** Read-Write

31	30	29	28	27	26	25	24
CPKEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	CPEREN	–			CPROCEN

- **CPROCEN: Coprocessor (second processor) Enable**

0 = If CPKEY is correct, resets the coprocessor (power-on default value).

1 = If CPKEY is correct, deasserts the reset of the coprocessor.

- **CPEREN: Coprocessor Peripheral Enable**

0 = If CPKEY is correct, resets the coprocessor peripherals.

1 = If CPKEY is correct, deasserts the reset of the coprocessor peripherals.

- **CPKEY: Coprocessor System Enable Key**

Should be written at value 0x5A. Writing any other value in this field aborts the write operation.

## 16. Real-time Timer (RTT)

### 16.1 Description

The Real-time Timer is built around a 32-bit counter used to count roll-over events of the programmable 16-bit prescaler which enables counting elapsed seconds from a 32 kHz slow clock source. It generates a periodic interrupt and/or triggers an alarm on a programmed value.

It can be configured to be driven by the 1 Hz signal generated by the RTC, thus taking advantage of a calibrated 1 Hz clock.

The slow clock source can be fully disabled to reduce power consumption when RTT is not required.

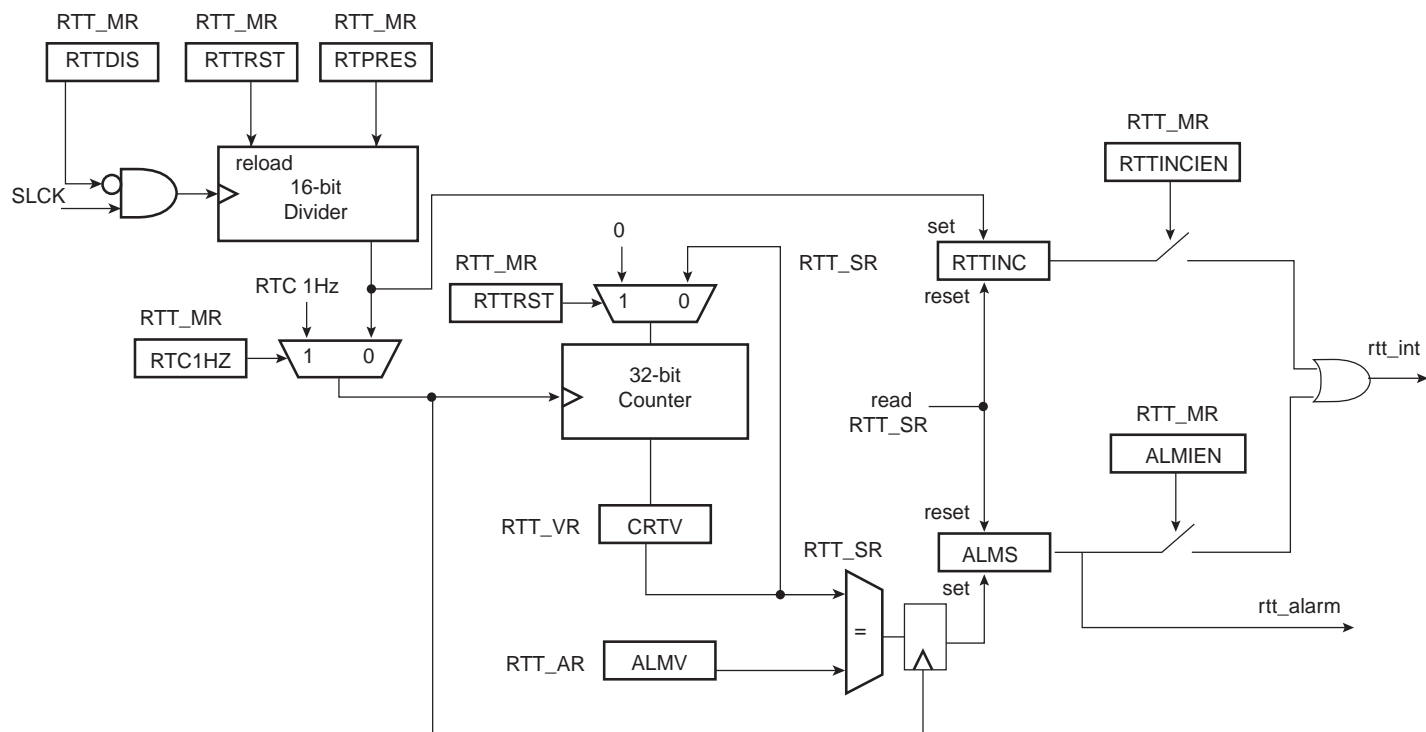
### 16.2 Embedded Characteristics

- 32-bit Free-running Counter on prescaled slow clock or RTC calibrated 1 Hz clock
- 16-bit Configurable Prescaler
- Interrupt on Alarm



## 16.3 Block Diagram

Figure 16-1. Real-time Timer



## 16.4 Functional Description

The Real-time Timer can be used to count elapsed seconds. It is built around a 32-bit counter fed by Slow Clock divided by a programmable 16-bit value. The value can be programmed in the field RTPRES of the Real-time Mode Register (RTT\_MR).

Programming RTPRES at 0x00008000 corresponds to feeding the real-time counter with a 1 Hz signal (if the Slow Clock is 32.768 kHz). The 32-bit counter can count up to  $2^{32}$  seconds, corresponding to more than 136 years, then roll over to 0.

The real-time 32-bit counter can also be supplied by the RTC 1 Hz clock. This mode is interesting when the RTC 1Hz is calibrated (CORRECTION field of RTC\_MR register differs from 0) in order to guaranty the synchronism between RTC and RTT counters.

Setting the RTC 1HZ clock to 1 in RTT\_MR register allows to drive the 32-bit RTT counter with the RTC 1Hz clock. In this mode, RTPRES field has no effect on 32-bit counter but RTTINC is still triggered by RTPRES.

The Real-time Timer can also be used as a free-running timer with a lower time-base. The best accuracy is achieved by writing RTPRES to 3. Programming RTPRES to 1 or 2 is possible, but may result in losing status events because the status register is cleared two Slow Clock cycles after read. Thus if the RTT is configured to trigger an interrupt, the interrupt occurs during 2 Slow Clock cycles after reading RTT\_SR. To prevent several executions of the interrupt handler, the interrupt must be disabled in the interrupt handler and re-enabled when the status register is clear.

The Real-time Timer value (CRTV) can be read at any time in the register RTT\_VR (Real-time Value Register). As this value can be updated asynchronously from the Master Clock, it is advisable to read this register twice at the same value to improve accuracy of the returned value.

The current value of the counter is compared with the value written in the alarm register RTT\_AR (Real-time Alarm Register). If the counter value matches the alarm, the bit ALMS in RTT\_SR is set. The alarm register is set to its maximum value, corresponding to 0xFFFF\_FFFF, after a reset.

The alarm interrupt must be disabled (ALMIEN must be cleared in RTT\_MR register) when writing a new ALMV value in Real-time Alarm Register.

The bit RTTINC in RTT\_SR is set each time there is a prescaler roll-over, so each time the Real-time Timer counter is incremented if RTC1HZ=0 else if RTC1HZ=1 the RTTINC bit can be triggered according to RTPRES value, in a fully independent way from the 32-bit counter increment. This bit can be used to start a periodic interrupt, the period being one second when the RTPRES is programmed with 0x8000 and Slow Clock equal to 32.768 Hz.

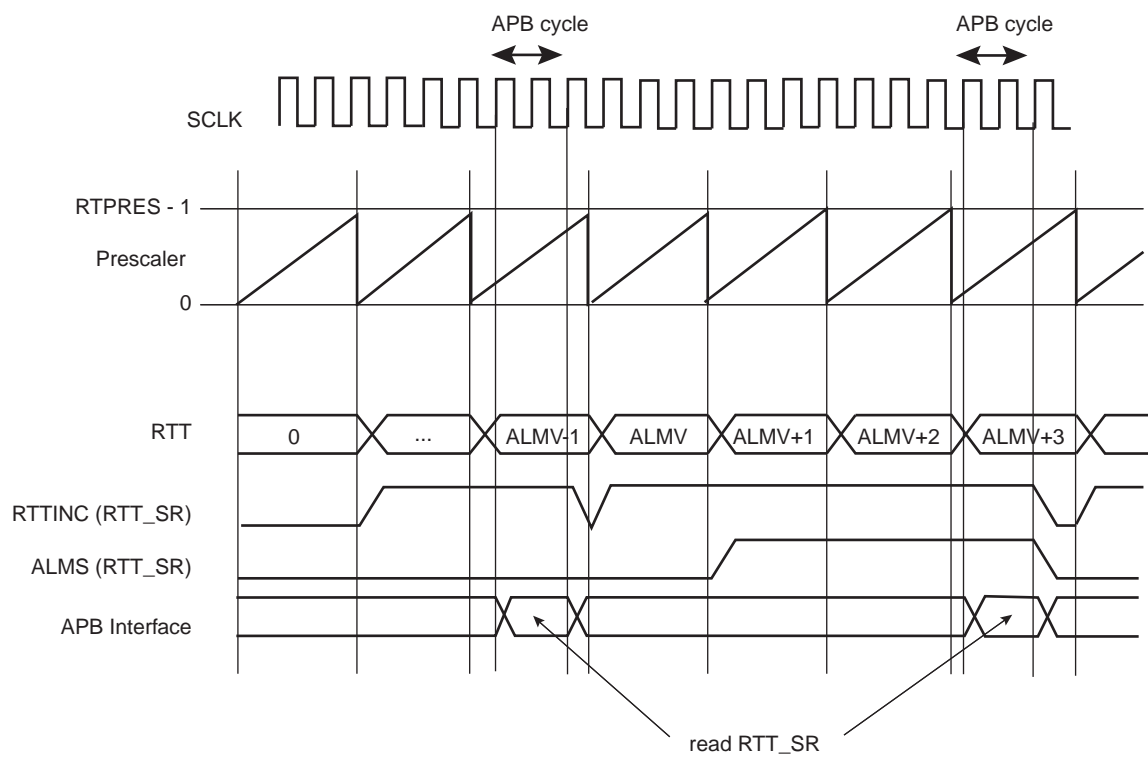
The RTTINCIEN field must be cleared prior to write a new RTPRES value in RTT\_MR register.

Reading the RTT\_SR status register resets the RTTINC and ALMS fields.

Writing the bit RTTRST in RTT\_MR immediately reloads and restarts the clock divider with the new programmed value. This also resets the 32-bit counter.

When not used, the Real-time Timer can be disabled in order to suppress dynamic power consumption in this module. This can be achieved by setting the RTTDIS field to 1 in RTT\_MR register.

Figure 16-2. RTT Counting



## 16.5 Real-time Timer (RTT) User Interface

Table 16-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Mode Register	RTT_MR	Read-write	0x0000_8000
0x04	Alarm Register	RTT_AR	Read-write	0xFFFF_FFFF
0x08	Value Register	RTT_VR	Read-only	0x0000_0000
0x0C	Status Register	RTT_SR	Read-only	0x0000_0000

### 16.5.1 Real-time Timer Mode Register

**Name:** RTT\_MR

**Address:** 0x400E1430

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	RTC1HZ
23	22	21	20	19	18	17	16
–	–	–	RTTDIS	–	RTTRST	RTTINCIEN	ALMIEN
15	14	13	12	11	10	9	8
RTPRES							
7	6	5	4	3	2	1	0
RTPRES							

- **RTPRES: Real-time Timer Prescaler Value**

Defines the number of SCLK periods required to increment the Real-time timer. RTPRES is defined as follows:

RTPRES = 0: The prescaler period is equal to  $2^{16} \times \text{SCLK period}$ .

RTPRES  $\neq$  0: The prescaler period is equal to RTPRES \* SCLK period.

Note: The RTTINCIEN field must be cleared prior to write a new RTPRES value.

- **ALMIEN: Alarm Interrupt Enable**

0 = The bit ALMS in RTT\_SR has no effect on interrupt.

1 = The bit ALMS in RTT\_SR asserts interrupt.

- **RTTINCIEN: Real-time Timer Increment Interrupt Enable**

0 = The bit RTTINC in RTT\_SR has no effect on interrupt.

1 = The bit RTTINC in RTT\_SR asserts interrupt.

- **RTTRST: Real-time Timer Restart**

0 = No effect.

1 = Reloads and restarts the clock divider with the new programmed value. This also resets the 32-bit counter.

- **RTTDIS: Real-time Timer Disable**

0 = The real-time timer is enabled.

1 = The real-time timer is disabled (no dynamic power consumption).

Note: RTTDIS is write only.

- **RTC1HZ: Real-Time Clock 1Hz Clock Selection**

0 = The RTT 32-bit counter is driven by the 16-bit prescaler roll-over events.

1 = The RTT 32-bit counter is driven by the RTC 1 Hz clock.

Note: RTC1HZ is write only.

### 16.5.2 Real-time Timer Alarm Register

**Name:** RTT\_AR  
**Address:** 0x400E1434  
**Access:** Read-write

31	30	29	28	27	26	25	24
ALMV							
23	22	21	20	19	18	17	16
ALMV							
15	14	13	12	11	10	9	8
ALMV							
7	6	5	4	3	2	1	0
ALMV							

- **ALMV: Alarm Value**

Defines the alarm value (ALMV+1) compared with the Real-time Timer.

**Note:** The alarm interrupt must be disabled (ALMIEN must be cleared in RTT\_MR register) when writing a new ALMV value.

### 16.5.3 Real-time Timer Value Register

**Name:** RTT\_VR  
**Address:** 0x400E1438  
**Access:** Read-only

31	30	29	28	27	26	25	24
CRTV							
23	22	21	20	19	18	17	16
CRTV							
15	14	13	12	11	10	9	8
CRTV							
7	6	5	4	3	2	1	0
CRTV							

- **CRTV: Current Real-time Value**

Returns the current value of the Real-time Timer.

#### 16.5.4 Real-time Timer Status Register

**Name:** RTT\_SR

**Address:** 0x400E143C

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	RTTINC	ALMS

- **ALMS: Real-time Alarm Status**

0 = The Real-time Alarm has not occurred since the last read of RTT\_SR.

1 = The Real-time Alarm occurred since the last read of RTT\_SR.

- **RTTINC: Real-time Timer Increment**

0 = The Real-time Timer has not been incremented since the last read of the RTT\_SR.

1 = The Real-time Timer has been incremented since the last read of the RTT\_SR.

## 17. Real-time Clock (RTC)

### 17.1 Description

The Real-time Clock (RTC) peripheral is designed for very low power consumption.

It combines a complete time-of-day clock with alarm and a two-hundred-year Gregorian or Persian calendar, complemented by a programmable periodic interrupt. The alarm and calendar registers are accessed by a 32-bit data bus.

The time and calendar values are coded in binary-coded decimal (BCD) format. The time format can be 24-hour mode or 12-hour mode with an AM/PM indicator.

Updating time and calendar fields and configuring the alarm fields are performed by a parallel capture on the 32-bit data bus. An entry control is performed to avoid loading registers with incompatible BCD format data or with an incompatible date according to the current month/year/century.

A clock divider calibration circuitry enables to compensate crystal oscillator frequency inaccuracy.

An RTC output can be programmed to generate several waveforms, including a prescaled clock derived from 32.768 kHz

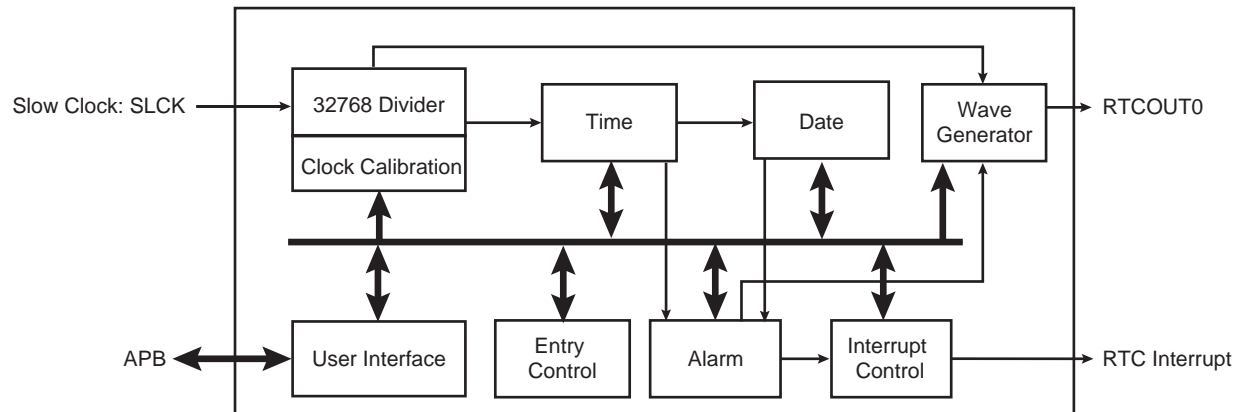
### 17.2 Embedded Characteristics

- Ultra Low Power Consumption
- Full Asynchronous Design
- Gregorian Calendar up to 2099 or Persian Calendar
- Programmable Periodic Interrupt
- Safety/security features:
  - Valid Time and Date Programmation Check
  - On-The-Fly Time and Date Validity Check
- Crystal Oscillator Clock Calibration
- Waveform Generation
- Tamper TimeStamping Registers



## 17.3 Block Diagram

Figure 17-1. RTC Block Diagram



## 17.4 Product Dependencies

### 17.4.1 Power Management

The Real-time Clock is continuously clocked at 32768 Hz. The Power Management Controller has no effect on RTC behavior.

### 17.4.2 Interrupt

RTC interrupt line is connected on one of the internal sources of the interrupt controller. RTC interrupt requires the interrupt controller to be programmed first.

## 17.5 Functional Description

The RTC provides a full binary-coded decimal (BCD) clock that includes century (19/20), year (with leap years), month, date, day, hours, minutes and seconds.

The valid year range is 1900 to 2099 in Gregorian mode, a two-hundred-year calendar (or 1300 to 1499 in Persian mode).

The RTC can operate in 24-hour mode or in 12-hour mode with an AM/PM indicator.

Corrections for leap years are included (all years divisible by 4 being leap years). This is correct up to the year 2099.

The RTC can generate configurable waveforms on RTCOUT0 output.

### 17.5.1 Reference Clock

The reference clock is Slow Clock (SLCK). It can be driven internally or by an external 32.768 kHz crystal.

During low power modes of the processor, the oscillator runs and power consumption is critical. The crystal selection has to take into account the current consumption for power saving and the frequency drift due to temperature effect on the circuit for time accuracy.

### 17.5.2 Timing

The RTC is updated in real time at one-second intervals in normal mode for the counters of seconds, at one-minute intervals for the counter of minutes and so on.

Due to the asynchronous operation of the RTC with respect to the rest of the chip, to be certain that the value read in the RTC registers (century, year, month, date, day, hours, minutes, seconds) are valid and stable, it is necessary to read these registers twice. If the data is the same both times, then it is valid. Therefore, a minimum of two and a maximum of three accesses are required.

### 17.5.3 Alarm

The RTC has five programmable fields: month, date, hours, minutes and seconds.

Each of these fields can be enabled or disabled to match the alarm condition:

- If all the fields are enabled, an alarm flag is generated (the corresponding flag is asserted and an interrupt generated if enabled) at a given month, date, hour/minute/second.
- If only the “seconds” field is enabled, then an alarm is generated every minute.

Depending on the combination of fields enabled, a large number of possibilities are available to the user ranging from minutes to 365/366 days.

Hour, minute and second matching alarm (SECEN, MINEN, HOUREN) can be enabled independently of SEC, MIN, HOUR fields.

**Note:** To change one of the SEC, MIN, HOUR, DATE, MONTH fields, it is recommended to disable the field before changing the value and then re-enable it after the change has been made. This requires up to 3 accesses to the RTC\_TIMALR or RTC\_CALALR registers. The first access clears the enable corresponding to the field to change (SECEN, MINEN, HOUREN, DATEEN, MTHEN). If the field is already cleared, this access is not required. The second access performs the change of the value (SEC, MIN, HOUR, DATE, MONTH). The third access is required to re-enable the field by writing 1 in SECEN, MINEN, HOUREN, DATEEN, MTHEN fields.

#### 17.5.4 Error Checking when Programming

Verification on user interface data is performed when accessing the century, year, month, date, day, hours, minutes, seconds and alarms. A check is performed on illegal BCD entries such as illegal date of the month with regard to the year and century configured.

If one of the time fields is not correct, the data is not loaded into the register/counter and a flag is set in the validity register. The user can not reset this flag. It is reset as soon as an acceptable value is programmed. This avoids any further side effects in the hardware. The same procedure is done for the alarm.

The following checks are performed:

1. Century (check if it is in range 19 - 20 or 13-14 in Persian mode)
2. Year (BCD entry check)
3. Date (check range 01 - 31)
4. Month (check if it is in BCD range 01 - 12, check validity regarding "date")
5. Day (check range 1 - 7)
6. Hour (BCD checks: in 24-hour mode, check range 00 - 23 and check that AM/PM flag is not set if RTC is set in 24-hour mode; in 12-hour mode check range 01 - 12)
7. Minute (check BCD and range 00 - 59)
8. Second (check BCD and range 00 - 59)

**Note:** If the 12-hour mode is selected by means of the RTC\_MR register, a 12-hour value can be programmed and the returned value on RTC\_TIMR will be the corresponding 24-hour value. The entry control checks the value of the AM/PM indicator (bit 22 of RTC\_TIMR register) to determine the range to be checked.

#### 17.5.5 RTC Internal Free Running Counter Error Checking

To improve the reliability and security of the RTC, a permanent check is performed on the internal free running counters to report non-BCD or invalid date/time values.

An error is reported by TDERR bit in the status register (RTC\_SR) if an incorrect value has been detected. The flag can be cleared by programming the TDERRCLR in the RTC status clear control register (RTC\_SCCR).

Anyway the TDERR error flag will be set again if the source of the error has not been cleared before clearing the TDERR flag. The clearing of the source of such error can be done either by reprogramming a correct value on RTC\_CALR and/or RTC\_TIMR registers.

The RTC internal free running counters may automatically clear the source of TDERR due to their roll-over (i.e. every 10 seconds for SECONDS[3:0] bitfield in RTC\_TIMR register). In this case the TDERR is held high until a clear command is asserted by TDERRCLR bit in RTC\_SCCR register.

#### 17.5.6 Updating Time/Calendar

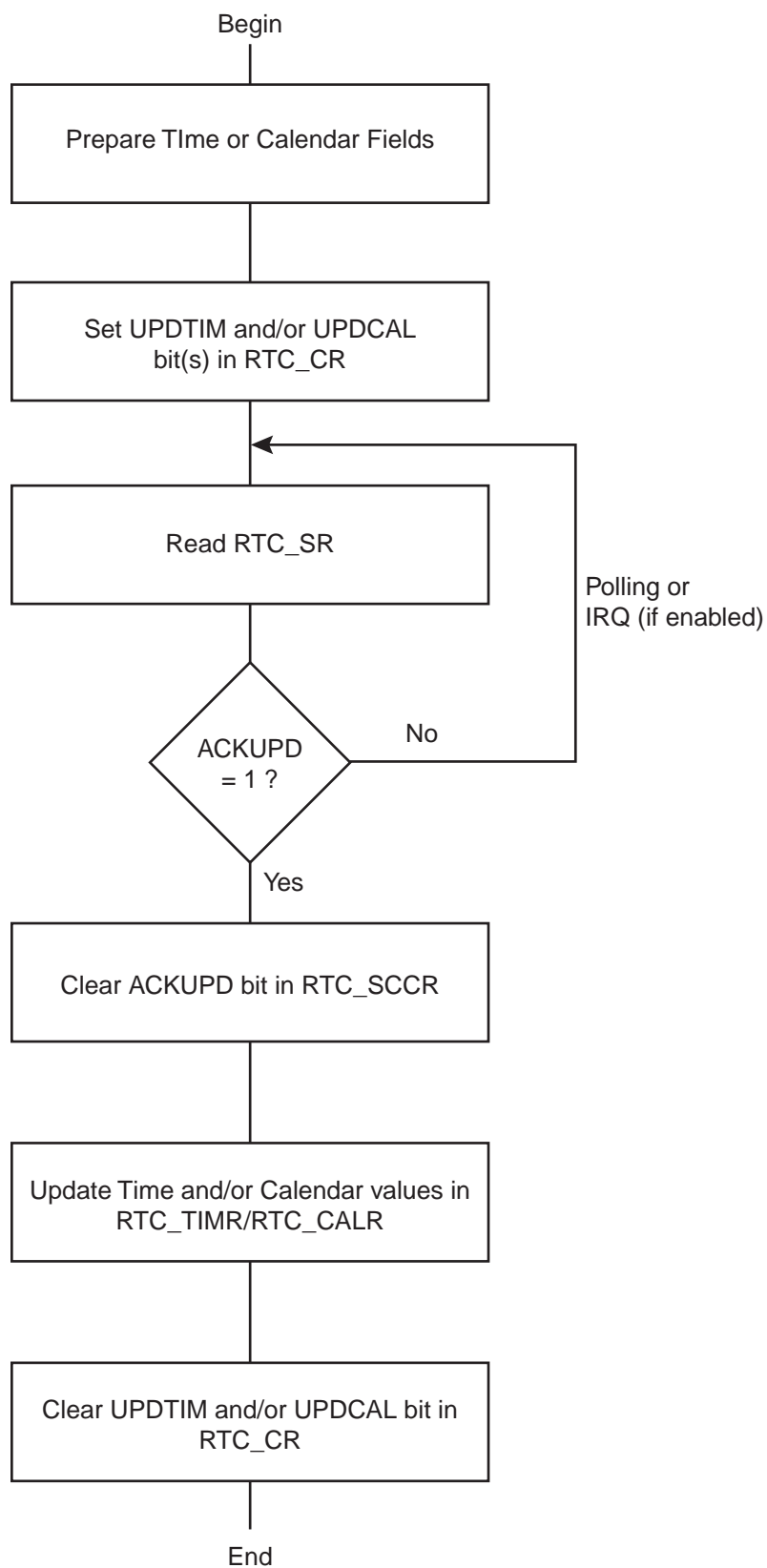
To update any of the time/calendar fields, the user must first stop the RTC by setting the corresponding field in the Control Register. Bit UPDTIM must be set to update time fields (hour, minute, second) and bit UPDCAL must be set to update calendar fields (century, year, month, date, day).

Then the user must poll or wait for the interrupt (if enabled) of bit ACKUPD in the Status Register. Once the bit reads 1, it is mandatory to clear this flag by writing the corresponding bit in RTC\_SCCR. The user can now write to the appropriate Time and Calendar register.

Once the update is finished, the user must reset (0) UPDTIM and/or UPDCAL in the Control

When entering programming mode of the calendar fields, the time fields remain enabled. When entering the programming mode of the time fields, both time and calendar fields are stopped. This is due to the location of the calendar logic circuitry (downstream for low-power considerations). It is highly recommended to prepare all the fields to be updated before entering programming mode. In successive update operations, the user must wait at least one second after resetting the UPDTIM/UPDCAL bit in the RTC\_CR (Control Register) before setting these bits again. This is done by waiting for the SEC flag in the Status Register before setting UPDTIM/UPDCAL bit. After resetting UPDTIM/UPDCAL, the SEC flag must also be cleared.

Figure 17-2. Update Sequence



### 17.5.7 RTC Accurate Clock Calibration

The crystal oscillator that drives the RTC may not be as accurate as expected mainly due to temperature variation. The RTC is equipped with circuitry able to correct slow clock crystal drift.

To compensate for possible temperature variations over time, this accurate clock calibration circuitry can be programmed on-the-fly and also programmed during application manufacturing, in order to correct the crystal frequency accuracy at room temperature (20-25°C). The typical clock drift range at room temperature is  $\pm 20$  ppm.

In the device operating temperature range, the 32.768 kHz crystal oscillator clock inaccuracy can be up to -200 ppm.

The RTC clock calibration circuitry allows positive or negative correction in a range of 1.5 ppm to 1950 ppm. After correction, the remaining crystal drift is as follows:

- Below 1 ppm, for an initial crystal drift between 1.5 ppm up to 90 ppm
- Below 2 ppm, for an initial crystal drift between 90 ppm up to 130 ppm
- Below 5 ppm, for an initial crystal drift between 130 ppm up to 200 ppm

The calibration circuitry acts by slightly modifying the 1 Hz clock period from time to time. When the period is modified, depending on the sign of the correction, the 1 Hz clock period increases or reduces by around 4 ms. The period interval between 2 correction events is programmable in order to cover the possible crystal oscillator clock variations.

The inaccuracy of a crystal oscillator at typical room temperature ( $\pm 20$  ppm at 20-25 degrees Celsius) can be compensated if a reference clock/signal is used to measure such inaccuracy. This kind of calibration operation can be set up during the final product manufacturing by means of measurement equipment embedding such a reference clock. The correction of value must be programmed into the RTC Mode Register (RTC\_MR), and this value is kept as long as the circuitry is powered (backup area). Removing the backup power supply cancels this calibration. This room temperature calibration can be further processed by means of the networking capability of the target application.

To ease the comparison of the inherent crystal accuracy with the reference clock/signal during manufacturing, an internal prescaled 32.768 kHz clock derivative signal can be assigned to drive RTC output. To accommodate the measure, several clock frequencies can be selected among 1 Hz, 32 Hz, 64 Hz, 512 Hz.

In any event, this adjustment does not take into account the temperature variation.

The frequency drift (up to -200 ppm) due to temperature variation can be compensated using a reference time if the application can access such a reference. If a reference time cannot be used, a temperature sensor can be placed close to the crystal oscillator in order to get the operating temperature of the crystal oscillator. Once obtained, the temperature may be converted using a lookup table (describing the accuracy/temperature curve of the crystal oscillator used) and RTC\_MR configured accordingly. The calibration can be performed on-the-fly. This adjustment method is not based on a measurement of the crystal frequency/drift and therefore can be improved by means of the networking capability of the target application.

If no crystal frequency adjustment has been done during manufacturing, it is still possible to do it. In the case where a reference time of the day can be obtained through LAN/WAN network, it is possible to calculate the drift of the application crystal oscillator by comparing the values read on RTC Time Register (RTC\_TIMR) and programming the HIGHPPM and CORRECTION bitfields on RTC\_MR according to the difference measured between the reference time and those of RTC\_TIMR.

### 17.5.8 Waveform Generation

Waveforms can be generated by the RTC in order to take advantage of the RTC inherent prescalers while the RTC is the only powered circuitry (low power mode of operation, backup mode) or in any active modes. Going into backup or low power operating modes does not affect the waveform generation outputs.

The RTC output (RTCOUT0) has a source driver selected among 7 possibilities.

The first selection choice sticks the associated output at 0 (This is the reset value and it can be used at any time to disable the waveform generation).

Selection choices 1 to 4 respectively select 1 Hz, 32 Hz, 64 Hz and 512 Hz.

32 Hz or 64 Hz can drive, for example, a TN LCD backplane signal while 1 Hz can be used to drive a blinking character like “:” for basic time display (hour, minute) on TN LCDs.

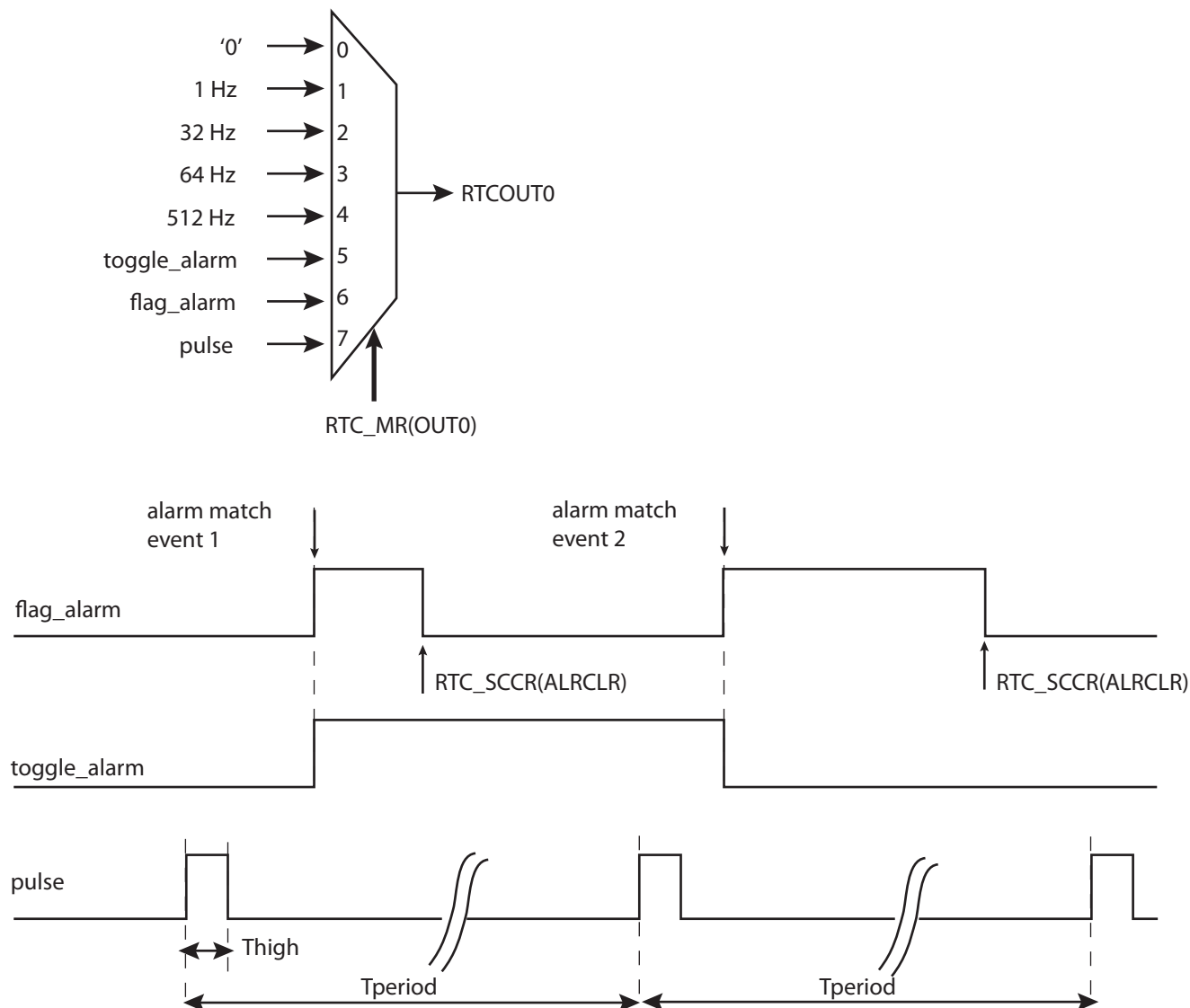
Selection choice 5 provides a toggling signal when the RTC alarm is reached.

Selection choice 6 provides a copy of the alarm flag, so the associated output is set high (logical 1) when an alarm occurs and immediately cleared when software clears the alarm interrupt source.

Selection choice 7 provides a 1 Hz periodic high pulse of 15  $\mu$ s duration that can be used to drive external devices for power consumption reduction or any other purpose.

PIO line associated to RTC output is automatically selecting these waveforms as soon as RTC\_MR register corresponding fields OUT0 differ from 0.

**Figure 17-3. Waveform Generation**



### 17.5.9 Tamper Timestamping

As soon as a tamper is detected, the tamper counter is incremented and the RTC stores the time of the day, the date and the source of the tamper event in registers located in the backup area. Up to 2 tamper events can be stored.

The tamper counter saturates at 15. Once this limit is reached, the exact number of tamper occurrence since the last read of stamping registers cannot be known.

The first set of timestamping registers (RTC\_TSTR0, RTC\_TSDR0, RTC\_TSSR0) cannot be overwritten, so once they have been written all data are stored until the registers are reset. Therefore these registers are storing the first tamper occurrence after a read.

The second set of timestamping registers (RTC\_TSTR1, RTC\_TSDR1, RTC\_TSSR1) are overwritten each time a tamper event is detected. This implies that the date and the time data of the first and the second stamping registers may be equal. This occurs when the tamper counter value carried on bitfield TEVCNT in RTC\_TSTR0 equals to 1. Thus this second set of registers allows to store the last occurrence of tamper before a read.

Reading a set of timestamping register requires three accesses, one for the time of the day, one for the date and one for the tamper source.

Reading the third part (RTC\_TSSR0/1) of a timestamping registers set clears the whole content of the registers (time, date and tamper source) and makes it available to store a new event.

## 17.6 Real-time Clock (RTC) User Interface

Table 17-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	RTC_CR	Read-write	0x0
0x04	Mode Register	RTC_MR	Read-write	0x0
0x08	Time Register	RTC_TIMR	Read-write	0x0
0x0C	Calendar Register	RTC_CALR	Read-write	0x01E111220
0x10	Time Alarm Register	RTC_TIMALR	Read-write	0x0
0x14	Calendar Alarm Register	RTC_CALALR	Read-write	0x01010000
0x18	Status Register	RTC_SR	Read-only	0x0
0x1C	Status Clear Command Register	RTC_SCCR	Write-only	–
0x20	Interrupt Enable Register	RTC_IER	Write-only	–
0x24	Interrupt Disable Register	RTC_IDR	Write-only	–
0x28	Interrupt Mask Register	RTC_IMR	Read-only	0x0
0x2C	Valid Entry Register	RTC_VER	Read-only	0x0
0xB0	TimeStamp Time Register 0	RTC_TSTR0	Read-only	0x0
0xB4	TimeStamp Date Register 0	RTC_TSDR0	Read-only	0x0
0xB8	TimeStamp Source Register 0	RTC_TSSR0	Read-only	0x0
0xBC	TimeStamp Time Register 1	RTC_TSTR1	Read-only	0x0
0xC0	TimeStamp Date Register 1	RTC_TSDR1	Read-only	0x0
0xC4	TimeStamp Source Register 1	RTC_TSSR1	Read-only	0x0
0xC8–0xF8	Reserved Register	–	–	–
0xFC	Reserved Register	–	–	–

Note: If an offset is not listed in the table it must be considered as reserved.



## 17.6.1 RTC Control Register

**Name:** RTC\_CR

**Address:** 0x400E1460

**Access:** Read-write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	CALEVSEL	
15	14	13	12	11	10	9	8
—	—	—	—	—	—	TIMEVSEL	
7	6	5	4	3	2	1	0
—	—	—	—	—	—	UPDCAL	UPDTIM

### • UPDTIM: Update Request Time Register

0 = No effect.

1 = Stops the RTC time counting.

Time counting consists of second, minute and hour counters. Time counters can be programmed once this bit is set and acknowledged by the bit ACKUPD of the Status Register.

### • UPDCAL: Update Request Calendar Register

0 = No effect.

1 = Stops the RTC calendar counting.

Calendar counting consists of day, date, month, year and century counters. Calendar counters can be programmed once this bit is set.

### • TIMEVSEL: Time Event Selection

The event that generates the flag TIMEV in RTC\_SR (Status Register) depends on the value of TIMEVSEL.

Value	Name	Description
0	MINUTE	Minute change
1	HOUR	Hour change
2	MIDNIGHT	Every day at midnight
3	NOON	Every day at noon

### • CALEVSEL: Calendar Event Selection

The event that generates the flag CALEV in RTC\_SR depends on the value of CALEVSEL

Value	Name	Description
0	WEEK	Week change (every Monday at time 00:00:00)
1	MONTH	Month change (every 01 of each month at time 00:00:00)
2	YEAR	Year change (every January 1 at time 00:00:00)

## 17.6.2 RTC Mode Register

**Name:** RTC\_MR  
**Address:** 0x400E1464  
**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	TPERIOD		–	THIGH		
23	22	21	20	19	18	17	16
–	–	–	–	–	OUT0		
15	14	13	12	11	10	9	8
HIGHPPM	CORRECTION						
7	6	5	4	3	2	1	0
–	–	–	NEGPPM	–	–	PERSIAN	HRMOD

- **HRMOD: 12-/24-hour Mode**

0 = 24-hour mode is selected.

1 = 12-hour mode is selected.

- **PERSIAN: PERSIAN Calendar**

0 = Gregorian Calendar.

1 = Persian Calendar.

- **NEGPPM: NEGative PPM Correction**

0 = positive correction (the divider will be slightly lower than 32768).

1 = negative correction (the divider will be slightly higher than 32768).

Refer to CORRECTION and HIGHPPM field descriptions.

- **CORRECTION: Slow Clock Correction**

0 = No correction

1..127 = The slow clock will be corrected according to the formula given below in HIGHPPM description.

- **HIGHPPM: HIGH PPM Correction**

0 = lower range ppm correction with accurate correction.

1 = higher range ppm correction with accurate correction.

If the absolute value of the correction to be applied is lower than 30ppm, it is recommended to clear HIGHPPM. HIGHPPM set to 1 is recommended for 30 ppm correction and above.

Formula:

If HIGHPPM = 0, then the clock frequency correction range is from 1.5 ppm up to 98 ppm. The RTC accuracy is less than 1 ppm for a range correction from 1.5 ppm up to 30 ppm..

The correction field must be programmed according to the required correction in ppm, the formula is as follows:

$$CORRECTION = \frac{3906}{20 \times ppm} - 1$$

The value obtained must be rounded to the nearest integer prior to being programmed into CORRECTION field.

If HIGHPPM = 1, then the clock frequency correction range is from 30.5 ppm up to 1950 ppm. The RTC accuracy is less than 1 ppm for a range correction from 30.5 ppm up to 90 ppm.

The correction field must be programmed according to the required correction in ppm, the formula is as follows:

$$CORRECTION = \frac{3906}{ppm} - 1$$

The value obtained must be rounded to the nearest integer prior to be programmed into CORRECTION field.

If NEGPPM is set to 1, the ppm correction is negative.

- **OUT0: RTCOUT0 OutputSource Selection**

Value	Name	Description
0	NO_WAVE	no waveform, stuck at '0'
1	FREQ1HZ	1 Hz square wave
2	FREQ32HZ	32 Hz square wave
3	FREQ64HZ	64 Hz square wave
4	FREQ512HZ	512 Hz square wave
5	ALARM_TOGGLE	output toggles when alarm flag rises
6	ALARM_FLAG	output is a copy of the alarm flag
7	PROG_PULSE	duty cycle programmable pulse

- **THIGH: High Duration of the Output Pulse**

Value	Name	Description
0	H_31MS	31.2 ms
1	H_16MS	15.6 ms
2	H_4MS	3.91 6.00ms
3	H_976US	976 μs
4	H_488US	488 μs
5	H_122US	122 μs
6	H_30US	30.5 μs
7	H_15US	15.2 μs

- **TPERIOD: Period of the Output Pulse**

Value	Name	Description
0	P_1S	1 second
1	P_500MS	500 ms
2	P_250MS	250 ms
3	P_125MS	125 ms

### 17.6.3 RTC Time Register

**Name:** RTC\_TIMR

**Address:** 0x400E1468

**Access:** Read-write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	AMPM	HOUR					
15	14	13	12	11	10	9	8
—	MIN						
7	6	5	4	3	2	1	0
—	SEC						

- **SEC: Current Second**

The range that can be set is 0 - 59 (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

- **MIN: Current Minute**

The range that can be set is 0 - 59 (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

- **HOUR: Current Hour**

The range that can be set is 1 - 12 (BCD) in 12-hour mode or 0 - 23 (BCD) in 24-hour mode.

- **AMPM: Ante Meridiem Post Meridiem Indicator**

This bit is the AM/PM indicator in 12-hour mode.

0 = AM.

1 = PM.

All non-significant bits read zero.

## 17.6.4 RTC Calendar Register

**Name:** RTC\_CALR  
**Address:** 0x400E146C  
**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	DATE					
23	22	21	20	19	18	17	16
DAY				MONTH			
15	14	13	12	11	10	9	8
YEAR							
7	6	5	4	3	2	1	0
–	CENT						

- **CENT: Current Century**

The range that can be set is 19 - 20 (gregorian) or 13-14 (persian) (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

- **YEAR: Current Year**

The range that can be set is 00 - 99 (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

- **MONTH: Current Month**

The range that can be set is 01 - 12 (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

- **DAY: Current Day in Current Week**

The range that can be set is 1 - 7 (BCD).

The coding of the number (which number represents which day) is user-defined as it has no effect on the date counter.

- **DATE: Current Day in Current Month**

The range that can be set is 01 - 31 (BCD).

The lowest four bits encode the units. The higher bits encode the tens.

All non-significant bits read zero.

### 17.6.5 RTC Time Alarm Register

**Name:** RTC\_TIMALR

**Address:** 0x400E1470

**Access:** Read-write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
HOUREN	AMPM	HOUR					
15	14	13	12	11	10	9	8
MINEN	MIN						
7	6	5	4	3	2	1	0
SECEN	SEC						

**Note:** To change one of the SEC, MIN, HOUR fields, it is recommended to disable the field before changing the value and then re-enable it after the change has been made. This requires up to 3 accesses to the RTC\_TIMALR register. The first access clears the enable corresponding to the field to change (SECEN, MINEN, HOUREN). If the field is already cleared, this access is not required. The second access performs the change of the value (SEC, MIN, HOUR). The third access is required to re-enable the field by writing 1 in SECEN, MINEN, HOUREN fields.

- **SEC: Second Alarm**

This field is the alarm field corresponding to the BCD-coded second counter.

- **SECEN: Second Alarm Enable**

0 = The second-matching alarm is disabled.

1 = The second-matching alarm is enabled.

- **MIN: Minute Alarm**

This field is the alarm field corresponding to the BCD-coded minute counter.

- **MINEN: Minute Alarm Enable**

0 = The minute-matching alarm is disabled.

1 = The minute-matching alarm is enabled.

- **HOUR: Hour Alarm**

This field is the alarm field corresponding to the BCD-coded hour counter.

- **AMPM: AM/PM Indicator**

This field is the alarm field corresponding to the BCD-coded hour counter.

- **HOUREN: Hour Alarm Enable**

0 = The hour-matching alarm is disabled.

1 = The hour-matching alarm is enabled.

### 17.6.6 RTC Calendar Alarm Register

**Name:** RTC\_CALALR

**Address:** 0x400E1474

**Access:** Read-write

31	30	29	28	27	26	25	24
DATEEN	–	DATE					
23	22	21	20	19	18	17	16
MTHEN	–	–	MONTH				
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

**Note:** To change one of the DATE, MONTH fields, it is recommended to disable the field before changing the value and then re-enable it after the change has been made. This requires up to 3 accesses to the RTC\_CALALR register. The first access clears the enable corresponding to the field to change (DATEEN,MTHEN). If the field is already cleared, this access is not required. The second access performs the change of the value (DATE,MONTH). The third access is required to re-enable the field by writing 1 in DATEEN, MTHEN fields.

- **MONTH: Month Alarm**

This field is the alarm field corresponding to the BCD-coded month counter.

- **MTHEN: Month Alarm Enable**

0 = The month-matching alarm is disabled.

1 = The month-matching alarm is enabled.

- **DATE: Date Alarm**

This field is the alarm field corresponding to the BCD-coded date counter.

- **DATEEN: Date Alarm Enable**

0 = The date-matching alarm is disabled.

1 = The date-matching alarm is enabled.

### 17.6.7 RTC Status Register

**Name:** RTC\_SR

**Address:** 0x400E1478

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
—	—	TDERR	CALEV	TIMEV	SEC	ALARM	ACKUPD

- **ACKUPD: Acknowledge for Update**

0 (FREERUN) = Time and calendar registers cannot be updated.

1 (UPDATE) = Time and calendar registers can be updated.

- **ALARM: Alarm Flag**

0 (NO\_ALARM\_EVENT) = No alarm matching condition occurred.

1 (ALARMEVENT) = An alarm matching condition has occurred.

- **SEC: Second Event**

0 (NO\_SECEVENT) = No second event has occurred since the last clear.

1 (SECEVENT) = At least one second event has occurred since the last clear.

- **TIMEV: Time Event**

0 (NO\_TIMEEVENT) = No time event has occurred since the last clear.

1 (TIMEEVENT) = At least one time event has occurred since the last clear.

The time event is selected in the TIMEVSEL field in RTC\_CR (Control Register) and can be any one of the following events: minute change, hour change, noon, midnight (day change).

- **CALEV: Calendar Event**

0 (NO\_CALEVENT) = No calendar event has occurred since the last clear.

1 (CALEVENT) = At least one calendar event has occurred since the last clear.

The calendar event is selected in the CALEVSEL field in RTC\_CR and can be any one of the following events: week change, month change and year change.

- **TDERR: Time and/or Date Free Running Error**

0 (CORRECT) = The internal free running counters are carrying valid values since the last read of RTC\_SR.

1 (ERR\_TIMEDATE) = The internal free running counters have been corrupted (invalid date or time, non-BCD values) since the last read and/or they are still invalid.



## 17.6.8 RTC Status Clear Command Register

**Name:** RTC\_SCCR  
**Address:** 0x400E147C  
**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	TDERRCLR	CALCLR	TIMCLR	SECCLR	ALRCLR	ACKCLR

- **ACKCLR: Acknowledge Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

- **ALRCLR: Alarm Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

- **SECCLR: Second Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

- **TIMCLR: Time Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

- **CALCLR: Calendar Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

- **TDERRCLR: Time and/or Date Free Running Error Clear**

0 = No effect.

1 = Clears corresponding status flag in the Status Register (RTC\_SR).

### 17.6.9 RTC Interrupt Enable Register

**Name:** RTC\_IER  
**Address:** 0x400E1480  
**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	TDERREN	CALEN	TIMEN	SECEN	ALREN	ACKEN

- **ACKEN: Acknowledge Update Interrupt Enable**

0 = No effect.

1 = The acknowledge for update interrupt is enabled.

- **ALREN: Alarm Interrupt Enable**

0 = No effect.

1 = The alarm interrupt is enabled.

- **SECEN: Second Event Interrupt Enable**

0 = No effect.

1 = The second periodic interrupt is enabled.

- **TIMEN: Time Event Interrupt Enable**

0 = No effect.

1 = The selected time event interrupt is enabled.

- **CALEN: Calendar Event Interrupt Enable**

0 = No effect.

1 = The selected calendar event interrupt is enabled.

- **TDERREN: Time and/or Date Error Interrupt Enable**

0 = No effect.

1 = The time and date error interrupt is enabled.

### 17.6.10 RTC Interrupt Disable Register

**Name:** RTC\_IDR

**Address:** 0x400E1484

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	TDERRDIS	CALDIS	TIMDIS	SECDIS	ALRDIS	ACKDIS

- **ACKDIS: Acknowledge Update Interrupt Disable**

0 = No effect.

1 = The acknowledge for update interrupt is disabled.

- **ALRDIS: Alarm Interrupt Disable**

0 = No effect.

1 = The alarm interrupt is disabled.

- **SECDIS: Second Event Interrupt Disable**

0 = No effect.

1 = The second periodic interrupt is disabled.

- **TIMDIS: Time Event Interrupt Disable**

0 = No effect.

1 = The selected time event interrupt is disabled.

- **CALDIS: Calendar Event Interrupt Disable**

0 = No effect.

1 = The selected calendar event interrupt is disabled.

- **TDERRDIS: Time and/or Date Error Interrupt Disable**

0 = No effect.

- 1 = The time and date error interrupt is disabled.

### 17.6.11 RTC Interrupt Mask Register

**Name:** RTC\_IMR

**Address:** 0x400E1488

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	CAL	TIM	SEC	ALR	ACK

- **ACK: Acknowledge Update Interrupt Mask**

0 = The acknowledge for update interrupt is disabled.

1 = The acknowledge for update interrupt is enabled.

- **ALR: Alarm Interrupt Mask**

0 = The alarm interrupt is disabled.

1 = The alarm interrupt is enabled.

- **SEC: Second Event Interrupt Mask**

0 = The second periodic interrupt is disabled.

1 = The second periodic interrupt is enabled.

- **TIM: Time Event Interrupt Mask**

0 = The selected time event interrupt is disabled.

1 = The selected time event interrupt is enabled.

- **CAL: Calendar Event Interrupt Mask**

0 = The selected calendar event interrupt is disabled.

1 = The selected calendar event interrupt is enabled.

### 17.6.12 RTC Valid Entry Register

**Name:** RTC\_VER  
**Address:** 0x400E148C  
**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
—	—	—	—	NVCALALR	NVTIMALR	NVCAL	NVTIM

- **NVTIM: Non-valid Time**

0 = No invalid data has been detected in RTC\_TIMR (Time Register).

1 = RTC\_TIMR has contained invalid data since it was last programmed.

- **NVCAL: Non-valid Calendar**

0 = No invalid data has been detected in RTC\_CALR (Calendar Register).

1 = RTC\_CALR has contained invalid data since it was last programmed.

- **NVTIMALR: Non-valid Time Alarm**

0 = No invalid data has been detected in RTC\_TIMALR (Time Alarm Register).

1 = RTC\_TIMALR has contained invalid data since it was last programmed.

- **NVCALALR: Non-valid Calendar Alarm**

0 = No invalid data has been detected in RTC\_CALALR (Calendar Alarm Register).

1 = RTC\_CALALR has contained invalid data since it was last programmed.

### 17.6.13 RTC TimeStamp Time Register 0

**Name:** RTC\_TSTR0

**Address:** 0x400E1510

**Access:** Read-only

31	30	29	28	27	26	25	24
BACKUP	–	–	–	TEVCNT			
23	22	21	20	19	18	17	16
–	AMPM	HOUR					
15	14	13	12	11	10	9	8
–	MIN						
7	6	5	4	3	2	1	0
–	SEC						

- **SEC: SECONDS of the Tamper**
- **MIN: MINutes of the Tamper**
- **HOUR: HOURs of the Tamper**
- **AMPM: AMPM Indicator of the Tamper**
- **TEVCNT: Tamper Events Counter**

Each time a tamper event occurs, this counter is incremented. This counter saturates at 15. Once this value is reached, it is no more possible to know the exact number of tamper events.

If this field is not null, this implies that at least one tamper event occurs since last register reset and that the values stored in time-stamping registers are valid.

- **BACKUP: System Mode of the Tamper**

0 = The state of the system is different from backup mode when the tamper event occurs.

1 = The system is in backup mode when the tamper event occurs.

This register is cleared by reading RTC\_TSSR0 register.

All non-significant bits read zero.

#### 17.6.14 RTC TimeStamp Time Register 1

**Name:** RTC\_TSTR1

**Address:** 0x400E151C

**Access:** Read-only

31	30	29	28	27	26	25	24
BACKUP	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	AMPM	HOUR					
15	14	13	12	11	10	9	8
–	MIN						
7	6	5	4	3	2	1	0
–	SEC						

- **SEC: SEConds of the Tamper**
- **MIN: MINutes of the Tamper**
- **HOUR: HOURs of the Tamper**
- **AMPM: AMPM Indicator of the Tamper**

This register is cleared by reading RTC\_TSSR1 register.

- **BACKUP: System Mode of the Tamper**

0 = The state of the system is different from backup mode when the tamper event occurs.

1 = The system is in backup mode when the tamper event occurs.

All non-significant bits read zero.

### 17.6.15 RTC TimeStamp Date Register

**Name:** RTC\_TSDRx

**Address:** 0x400E1514 [0], 0x400E1520 [1]

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	DATE					
23	22	21	20	19	18	17	16
DAY				MONTH			
15	14	13	12	11	10	9	8
YEAR							
7	6	5	4	3	2	1	0
–	CENT						

- **CENT:** Century of the Tamper
- **YEAR:** Year of the Tamper
- **MONTH:** Month of the Tamper
- **DAY:** Day of the Tamper
- **DATE:** Date of the Tamper

The fields contains the date and the source of a tamper occurrence if the TEVCNT is not null.

This register is cleared by reading RTC\_TSSR register.

All non-significant bits read zero.



### 17.6.16 RTC TimeStamp Source Register

**Name:** RTC\_TSSRx

**Address:** 0x400E1518 [0], 0x400E1524 [1]

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
—	—	—	—	—	—	TSRC	

- **TSRC: Tamper Source**

This field contains the tamper source. It is valid only if the TEVCNT is not null.

This register is cleared after read and the read access also performs a clear on RTC\_TSTR and RTC\_TSDR registers.

All non-significant bits read zero.



## 18.4 Functional Description

The Watchdog Timer can be used to prevent system lock-up if the software becomes trapped in a deadlock. It is supplied with VDDCORE. It restarts with initial values on processor reset.

The Watchdog is built around a 12-bit down counter, which is loaded with the value defined in the field WDV of the Mode Register (WDT\_MR). The Watchdog Timer uses the Slow Clock divided by 128 to establish the maximum Watchdog period to be 16 seconds (with a typical Slow Clock of 32.768 kHz).

After a Processor Reset, the value of WDV is 0xFFFF, corresponding to the maximum value of the counter with the external reset generation enabled (field WDRSTEN at 1 after a Backup Reset). This means that a default Watchdog is running at reset, i.e., at power-up. The user must either disable it (by setting the WDDIS bit in WDT\_MR) if he does not expect to use it or must reprogram it to meet the maximum Watchdog period the application requires.

If the watchdog is restarted by writing into the WDT\_CR register, the WDT\_MR register must not be programmed during a period of time of 3 slow clock periods following the WDT\_CR write access. In any case, programming a new value in the WDT\_MR register automatically initiates a restart instruction.

The Watchdog Mode Register (WDT\_MR) can be written only once. Only a processor reset resets it. Writing the WDT\_MR register reloads the timer with the newly programmed mode parameters.

In normal operation, the user reloads the Watchdog at regular intervals before the timer underflow occurs, by writing the Control Register (WDT\_CR) with the bit WDRSTT to 1. The Watchdog counter is then immediately reloaded from WDT\_MR and restarted, and the Slow Clock 128 divider is reset and restarted. The WDT\_CR register is write-protected. As a result, writing WDT\_CR without the correct hard-coded key has no effect. If an underflow does occur, the “wdt\_fault” signal to the Reset Controller is asserted if the bit WDRSTEN is set in the Mode Register (WDT\_MR). Moreover, the bit WDUNF is set in the Watchdog Status Register (WDT\_SR).

To prevent a software deadlock that continuously triggers the Watchdog, the reload of the Watchdog must occur while the Watchdog counter is within a window between 0 and WDD, WDD is defined in the WatchDog Mode Register WDT\_MR.

Any attempt to restart the Watchdog while the Watchdog counter is between WDV and WDD results in a Watchdog error, even if the Watchdog is disabled. The bit WDERR is updated in the WDT\_SR and the “wdt\_fault” signal to the Reset Controller is asserted.

Note that this feature can be disabled by programming a WDD value greater than or equal to the WDV value. In such a configuration, restarting the Watchdog Timer is permitted in the whole range [0; WDV] and does not generate an error. This is the default configuration on reset (the WDD and WDV values are equal).

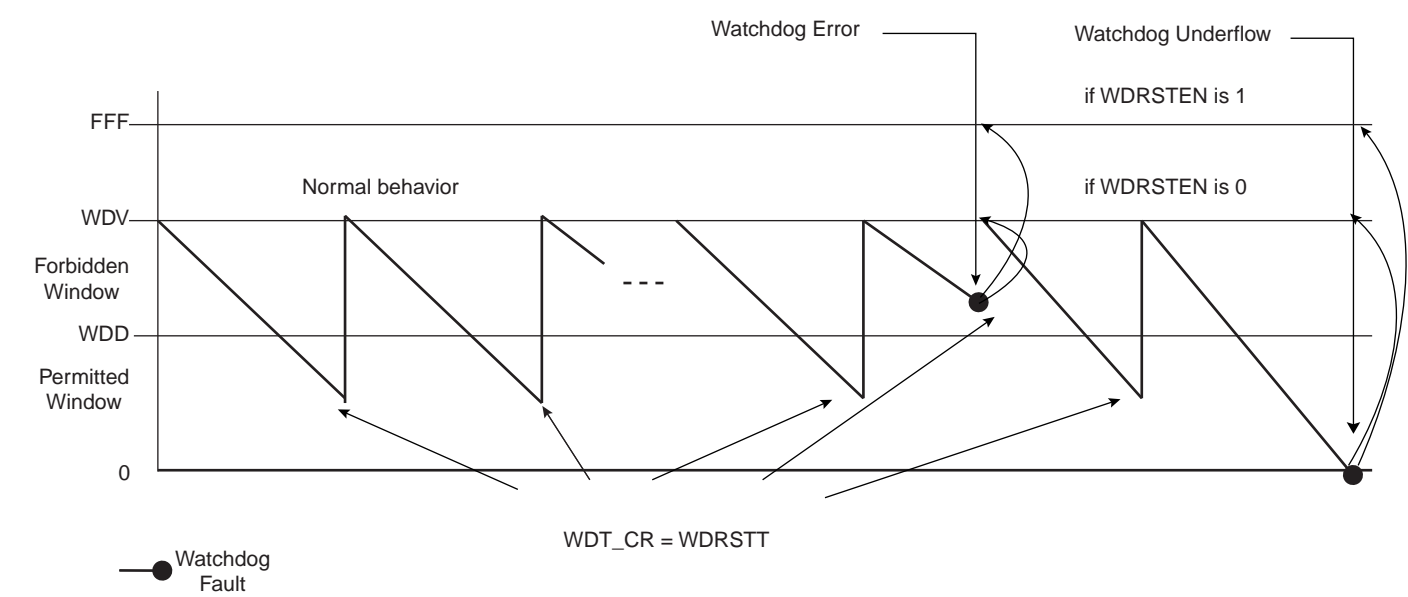
The status bits WDUNF (Watchdog Underflow) and WDERR (Watchdog Error) trigger an interrupt, provided the bit WDFIEN is set in the mode register. The signal “wdt\_fault” to the reset controller causes a Watchdog reset if the WDRSTEN bit is set as already explained in the reset controller programmer Datasheet. In that case, the processor and the Watchdog Timer are reset, and the WDERR and WDUNF flags are reset.

If a reset is generated or if WDT\_SR is read, the status bits are reset, the interrupt is cleared, and the “wdt\_fault” signal to the reset controller is deasserted.

Writing the WDT\_MR reloads and restarts the down counter.

While the processor is in debug state or in idle mode, the counter may be stopped depending on the value programmed for the bits WDIDLEHLT and WDDBGHLT in the WDT\_MR.

Figure 18-2. Watchdog Behavior



## 18.5 Watchdog Timer (WDT) User Interface

Table 18-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	WDT_CR	Write-only	–
0x04	Mode Register	WDT_MR	Read-write Once	0x3FFF_2FFF
0x08	Status Register	WDT_SR	Read-only	0x0000_0000

### 18.5.1 Watchdog Timer Control Register

**Name:** WDT\_CR  
**Address:** 0x400E1450  
**Access:** Write-only

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WDRSTT

- **WDRSTT: Watchdog Restart**

0: No effect.

1: Restarts the Watchdog if KEY is written to 0xA5.

- **KEY: Password.**

Value	Name	Description
0xA5	PASSWD	Writing any other value in this field aborts the write operation.

## 18.5.2 Watchdog Timer Mode Register

**Name:** WDT\_MR

**Address:** 0x400E1454

**Access:** Read-write Once

31	30	29	28	27	26	25	24
		WDIDLEHLT	WDDBGHLT	WDD			
23	22	21	20	19	18	17	16
WDD							
15	14	13	12	11	10	9	8
WDDIS	WDRPROC	WDRSTEN	WDFIEN	WDV			
7	6	5	4	3	2	1	0
WDV							

**Note:** The first write access prevents any further modification of the value of this register, read accesses remain possible.

**Note:** The WDD and WDV values must not be modified within a period of time of 3 slow clock periods following a restart of the watchdog performed by means of a write access in the WDT\_CR register, else the watchdog may trigger an end of period earlier than expected.

- **WDV: Watchdog Counter Value**

Defines the value loaded in the 12-bit Watchdog Counter.

- **WDFIEN: Watchdog Fault Interrupt Enable**

0: A Watchdog fault (underflow or error) has no effect on interrupt.

1: A Watchdog fault (underflow or error) asserts interrupt.

- **WDRSTEN: Watchdog Reset Enable**

0: A Watchdog fault (underflow or error) has no effect on the resets.

1: A Watchdog fault (underflow or error) triggers a Watchdog reset.

- **WDRPROC: Watchdog Reset Processor**

0: If WDRSTEN is 1, a Watchdog fault (underflow or error) activates all resets.

1: If WDRSTEN is 1, a Watchdog fault (underflow or error) activates the processor reset.

- **WDD: Watchdog Delta Value**

Defines the permitted range for reloading the Watchdog Timer.

If the Watchdog Timer value is less than or equal to WDD, writing WDT\_CR with WDRSTT = 1 restarts the timer.

If the Watchdog Timer value is greater than WDD, writing WDT\_CR with WDRSTT = 1 causes a Watchdog error.

- **WDDBGHLT: Watchdog Debug Halt**

0: The Watchdog runs when the processor is in debug state.

1: The Watchdog stops when the processor is in debug state.

- **WDIDLEHLT: Watchdog Idle Halt**

0: The Watchdog runs when the system is in idle mode.

1: The Watchdog stops when the system is in idle state.

- **WDDIS: Watchdog Disable**

0: Enables the Watchdog Timer.

1: Disables the Watchdog Timer.



### 18.5.3 Watchdog Timer Status Register

**Name:** WDT\_SR

**Address:** 0x400E1458

**Access** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	WDERR	WDUNF

- **WDUNF: Watchdog Underflow**

0: No Watchdog underflow occurred since the last read of WDT\_SR.

1: At least one Watchdog underflow occurred since the last read of WDT\_SR.

- **WDERR: Watchdog Error**

0: No Watchdog error occurred since the last read of WDT\_SR.

1: At least one Watchdog error occurred since the last read of WDT\_SR.

## 19. Reinforced Safety Watchdog Timer (RSWDT)

### 19.1 Description

When two watchdog timers are implemented in a device, the second one, the Reinforced Safety Watchdog Timer (RSWDT) works in parallel with the Watchdog Timer (WDT) to reinforce safe watchdog operations.

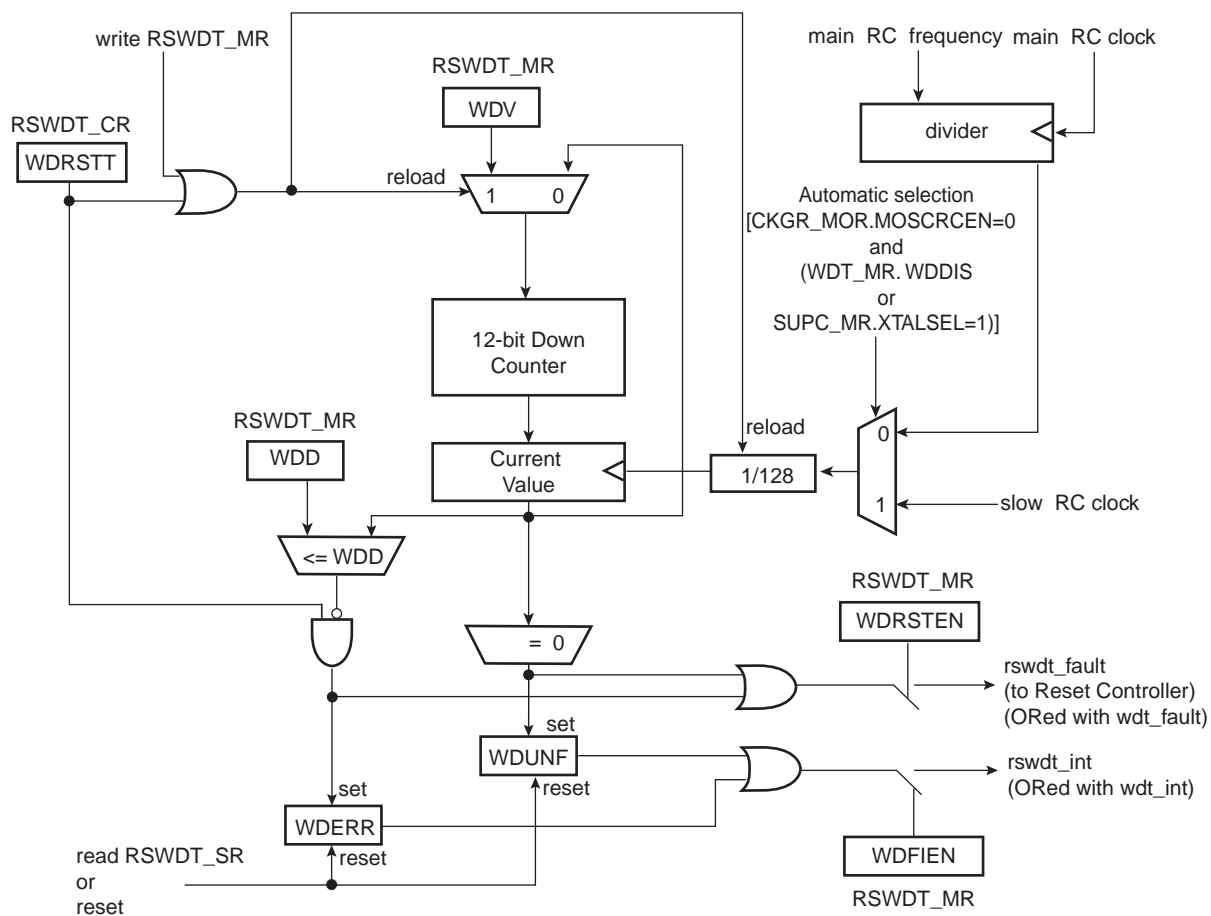
The Reinforced Safety Watchdog Timer (RSWDT) can be used to reinforce the safety level provided by the Watchdog Timer (WDT) in order to prevent system lock-up if the software becomes trapped in a deadlock. The RSWDT works in a fully operable mode, independent of the Watchdog Timer. Its clock source is automatically selected from either the slow RC oscillator clock or main RC oscillator divided clock to get an equivalent slow RC oscillator clock. If the Watchdog Timer clock source (for example the 32 kHz crystal oscillator) fails, the system lock-up is no longer monitored by the Watchdog Timer as the second watchdog timer, RSWDT, will perform the monitoring. Thus, there is no lack of safety irrespective of the external operating conditions. This Reinforced Safety Watchdog Timer shares the same features as the Watchdog Timer (i.e. a 12-bit down counter that allows a watchdog period of up to 16 seconds with slow clock at 32.768 kHz). It can generate a general reset or a processor reset only. In addition, it can be stopped while the processor is in debug mode or idle mode.

### 19.2 Embedded Characteristics

- System safety level reinforced by means of an independent second watchdog timer
- Automatically selected reliable independent clock source other than that of the first watchdog timer
- 12-bit Key-protected programmable counter
- Provides reset or interrupt signals to the system
- Counter may be stopped while the processor is in debug state or in idle mode

## 19.3 Block Diagram

Figure 19-1. Reinforced Safety Watchdog Timer Block Diagram



## 19.4 Functional Description

The Reinforced Safety Watchdog Timer (RSWDT) can be used to prevent system lock-up if the software becomes trapped in a deadlock. It is supplied with VDDCORE. RSWDT is initialized with default values on processor reset, or power-on sequence and is disabled (it's default mode) under such conditions.

The Reinforced Safety Watchdog Timer works in a fully independent mode distinct from the Watchdog Timer (WDT). Its clock source is automatically selected from either slow RC oscillator clock or main RC oscillator divided clock to get an equivalent slow RC oscillator clock. If the Watchdog Timer (WDT), clock source (for example the 32 kHz crystal oscillator) fails, the system lock-up is no longer monitored by the WDT, but the second watchdog timer, the RSWDT will perform the monitoring. Therefore, continuous safety is assured regardless of the external operating conditions.

The selection of the Reinforced Safety Watchdog Timer clock source consists of a combination of the state of the main RC oscillator (field MOSCRGEN in CKGR\_MOR register), Watchdog Timer (field WDDIS of WDT\_MR register) and slow clock selection (field XTALSEL in the SUPC\_MR register). The Reinforced Safety Watchdog Timer is driven by the slow RC oscillator if the main RC oscillator is not already in use, and either the selected slow clock is the 32 kHz crystal oscillator, or the Watchdog Timer (WDT) is disabled. Accordingly, slow or main RC oscillators are automatically enabled.

The RSWDT is built around a 12-bit down counter, which is loaded with a slow clock value other than that of the slow clock in the Watchdog Timer, defined in the WDV field of the Mode Register (RSWDT\_MR). The Reinforced Safety Watchdog Timer uses the Slow Clock divided by 128 to establish the maximum watchdog period to be 16 seconds (with a typical Slow Clock of 32.768 kHz).

After a processor reset, the value of WDV is 0xFFFF, corresponding to the maximum value of the counter with the external reset generation enabled (WDRSTEN field at 1 after a backup reset). This means that a default watchdog is running at reset, i.e., at power-up.

The Mode Register (RSWDT\_MR) can be written only once. Only a processor reset resets it. Writing the RSWDT\_MR register reloads the timer with the newly programmed mode parameters.

In normal operation, the user reloads the watchdog at regular intervals before the timer underflow occurs, by writing the Control Register (RSWDT\_CR) with the bit WDRSTT to 1. The watchdog counter is then immediately reloaded from RSWDT\_MR and restarted, and the Slow Clock 128 divider is reset and restarted. The RSWDT\_CR register is write-protected. As a result, writing RSWDT\_CR without the correct hard-coded key has no effect. If an underflow does occur, the “wdt\_fault” signal to the reset controller is asserted if the bit WDRSTEN is set in the Mode Register (RSWDT\_MR). Moreover, the bit WDUNF is set in the Status Register (RSWDT\_SR).

To prevent a software deadlock that continuously triggers the RSWDT, the reload of the RSWDT must occur while the watchdog counter is within a window between 0 and WDD, WDD is defined in the Mode Register, RSWDT\_MR.

Any attempt to restart the watchdog while the watchdog counter is between WDV and WDD results in a watchdog error, even if the RSWDT is disabled. The WDERR bit is updated in the RSWDT\_SR and the “wdt\_fault” signal to the reset controller is asserted.

Note that this feature can be disabled by programming a WDD value greater than or equal to the WDV value. In such a configuration, restarting the Reinforced Safety Watchdog Timer is permitted in the whole range [0; WDV] and does not generate an error. This is the default configuration on reset (the WDD and WDV values are equal).

The status bits, WDUNF (Watchdog Underflow) and WDERR (Watchdog Error) trigger an interrupt, provided the WDFIEN bit is set in the mode register. The signal “wdt\_fault” to the reset controller causes a Watchdog reset if the WDRSTEN bit is set as explained in the reset controller programmer's documentation. In that case, the processor and the watchdog timer are reset, and the WDERR and WDUNF flags are reset.

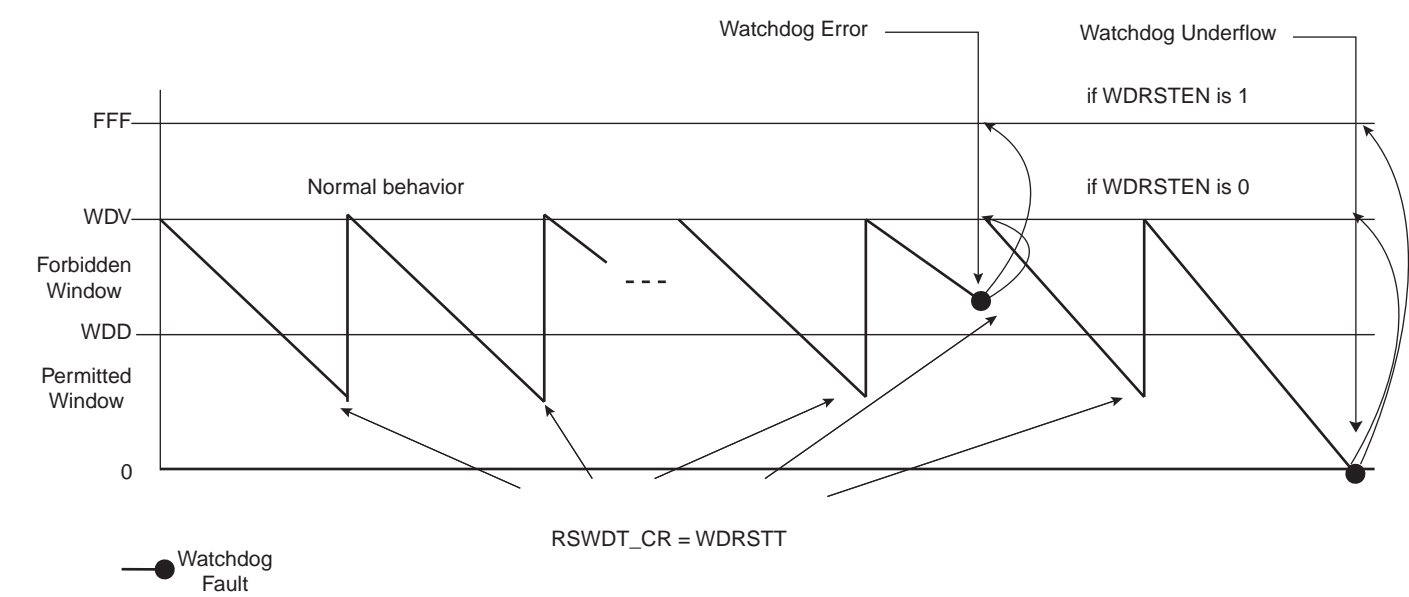
If a reset is generated, or if RSWDT\_SR is read, the status bits are reset, the interrupt is cleared, and the “wdt\_fault” signal to the reset controller is deasserted.

Writing the RSWDT\_MR reloads and restarts the down counter.

The RSWDT is disabled after any power-on sequence.

While the processor is in debug state or in idle mode, the counter may be stopped depending on the value programmed for the WDIDLEHLT and WDDBGHLT bits in the RSWDT\_MR.

Figure 19-2. Watchdog Behavior



## 19.5 Reinforced Safety Watchdog Timer (RSWDT) User Interface

Table 19-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	RSWDT_CR	Write-only	–
0x04	Mode Register	RSWDT_MR	Read-write Once	0x3FFF_AFFF
0x08	Status Register	RSWDT_SR	Read-only	0x0000_0000

### 19.5.1 Reinforced Safety Watchdog Timer Control Register

**Name:** RSWDT\_CR

**Address:** 0x400E1500

**Access:** Write-only

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WDRSTT

- **WDRSTT: Watchdog Restart**

0: No effect.

1: Restarts the watchdog.

- **KEY: Password**

Should be written at value 0xC4. Writing any other value in this field aborts the write operation.

### 19.5.2 Reinforced Safety Watchdog Timer Mode Register

**Name:** RSWDT\_MR

**Address:** 0x400E1504

**Access:** Read-write Once

31	30	29	28	27	26	25	24
		WDIDLEHLT	WDDBGHLT	WDD			
23	22	21	20	19	18	17	16
WDD							
15	14	13	12	11	10	9	8
WDDIS	WDRPROC	WDRSTEN	WDFIEN	WDV			
7	6	5	4	3	2	1	0
WDV							

- **WDV: Watchdog Counter Value**

Defines the value loaded in the 12-bit watchdog counter.

- **WDFIEN: Watchdog Fault Interrupt Enable**

0: A Watchdog fault (underflow or error) has no effect on interrupt.

1: A Watchdog fault (underflow or error) asserts interrupt.

- **WDRSTEN: Watchdog Reset Enable**

0: A Watchdog fault (underflow or error) has no effect on the resets.

1: A Watchdog fault (underflow or error) triggers a watchdog reset.

- **WDRPROC: Watchdog Reset Processor**

0: If WDRSTEN is 1, a watchdog fault (underflow or error) activates all resets.

1: If WDRSTEN is 1, a watchdog fault (underflow or error) activates the processor reset.

- **WDD: Watchdog Delta Value**

Defines the permitted range for reloading the watchdog timer.

If the watchdog timer value is less than or equal to WDD, writing RSWDT\_CR with WDRSTT = 1 restarts the timer.

If the watchdog timer value is greater than WDD, writing RSWDT\_CR with WDRSTT = 1 causes a Watchdog error.

- **WDDBGHLT: Watchdog Debug Halt**

0: The watchdog runs when the processor is in debug state.

1: The watchdog stops when the processor is in debug state.

- **WDIDLEHLT: Watchdog Idle Halt**

0: The watchdog runs when the system is in idle mode.

1: The watchdog stops when the system is in idle state.

- **WDDIS: Watchdog Disable**

0: Enables the watchdog timer.

1: Disables the watchdog timer.



### 19.5.3 Reinforced Safety Watchdog Timer Status Register

**Name:** RSWDT\_SR

**Address:** 0x400E1508

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	WDERR	WDUNF

- **WDUNF: Watchdog Underflow**

0: No watchdog underflow occurred since the last read of RSWDT\_SR.

1: At least one watchdog underflow occurred since the last read of RSWDT\_SR.

- **WDERR: Watchdog Error**

0: No watchdog error occurred since the last read of RSWDT\_SR.

1: At least one watchdog error occurred since the last read of RSWDT\_SR.

## 20. Supply Controller (SUPC)

### 20.1 Description

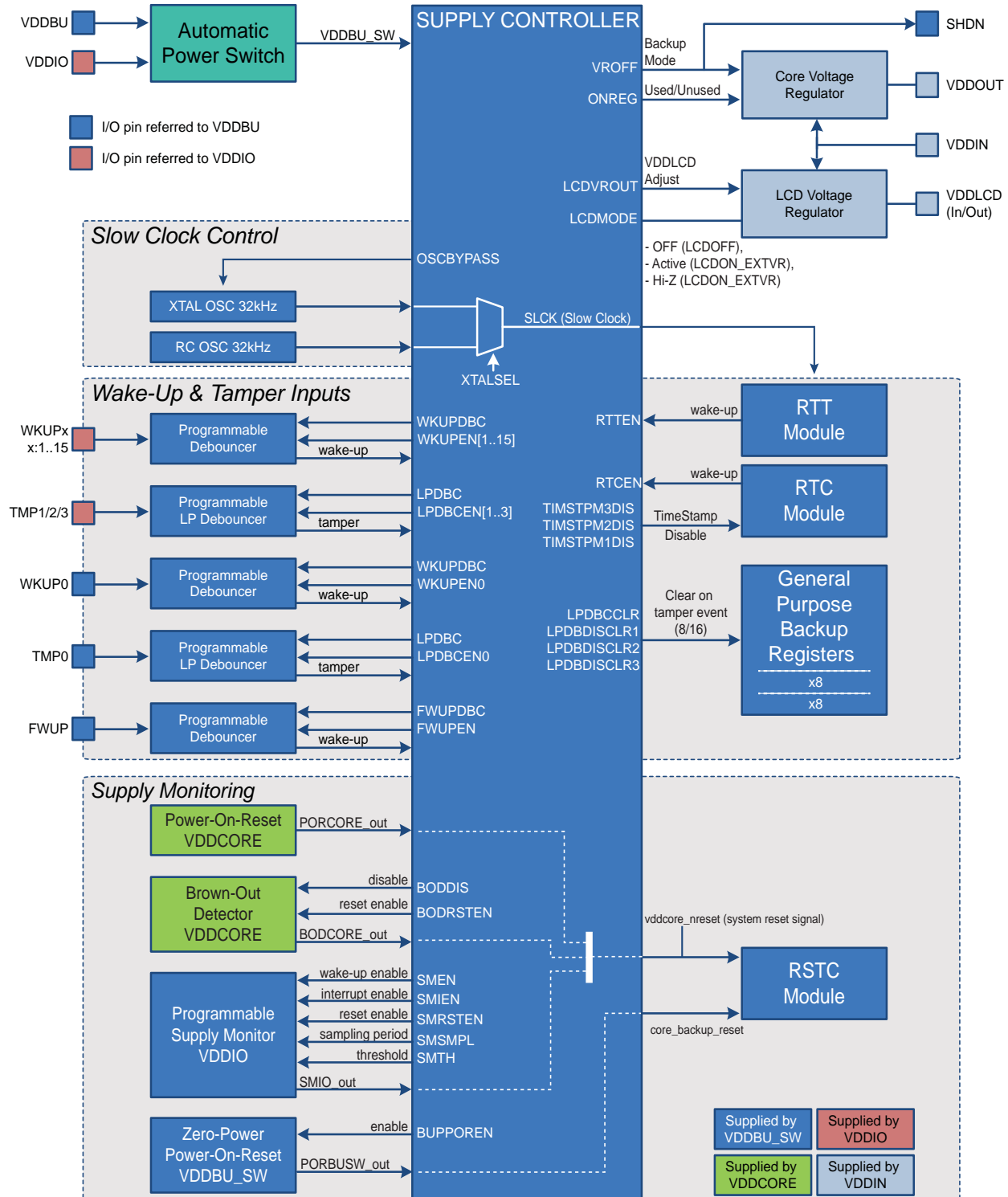
The Supply Controller (SUPC) controls the supply voltages of the system and manages the backup low-power mode. In this mode, current consumption is reduced to less than 1 microamp (typ) for backup power retention. Exit from this mode is possible on multiple wake-up sources. The SUPC also generates the slow clock by selecting either the low-power RC oscillator or the low-power crystal oscillator.

### 20.2 Embedded Characteristics

- Manages VDDCORE and the Backup Low-Power Mode by Controlling the Embedded Voltage Regulator
- Manages the LCD Power Supply VDDLCD and the Backup Low-Power Mode by Controlling the Embedded LCD Voltage Regulator
- A Supply Monitor Detection on VDDIO or a Brownout Detection on VDDCORE Triggers a System Reset
- A Supply Monitor Detection on VDDBU\_SW Triggers a System Reset
- Generates the Slow Clock SLCK by Selecting Either the 32 kHz Low-Power RC Oscillator or the 32 kHz Low-Power Crystal Oscillator
- Supports Multiple Wake-up Sources for Exit from Backup Low Power Mode
  - Force Wake-up Pin, with Programmable Debouncing
  - Up to 16 Wake-up Inputs (including Tamper Inputs), with Programmable Debouncing
  - Real-Time Clock Alarm
  - Real-Time Timer Alarm
  - Supply Monitor Detection on VDDIO, with Programmable Scan Period and Voltage Threshold

## 20.3 Block Diagram

Figure 20-1. Supply Controller Block Diagram



**Note:** TMPx signals and WKUPx signals are multiplexed on the same pins (ex. TMP0/WKUP0, TMP1/WKUP10, etc.). This generates a wake-up event only, a tamper event only or a wake-up and a tamper event.

## 20.4 Supply Controller Functional Description

### 20.4.1 Supply Controller Overview

The device can be divided into two power supply areas:

- The backup VDDBU\_SW power supply that includes the Supply Controller, a part of the Reset Controller, the slow clock switch, the general-purpose backup registers, the supply monitor and the clock which includes the Real-time Timer and the Real-time Clock.
- The core power supply that includes the other part of the Reset Controller, the Brownout Detector, the processor, the SRAM memory, the Flash memory and the peripherals.

The Supply Controller (SUPC) controls the core power supply. The SUPC intervenes when the VDDBU\_SW power supply rises (when the system is starting) or when the backup low-power mode is entered.

The SUPC also integrates the slow clock generator which is based on a 32 kHz crystal oscillator and an embedded 32 kHz RC oscillator. The slow clock defaults to the RC oscillator, but the software can enable the crystal oscillator and select it as the slow clock source.

The Supply Controller and the VDDBU\_SW power supply have a reset circuitry based on a zero-power power-on reset cell. The zero-power power-on reset allows the SUPC to start properly as soon as the VDDBU\_SW voltage becomes valid.

At start-up of the system, once the backup voltage VDDBU\_SW is valid and the embedded 32 kHz RC oscillator is stabilized, the SUPC starts up the core voltage regulator and ties the SHDN pin to VDDBU. Once the VDDCORE voltage is valid, it releases the system reset signal (`vddcore_nreset`) to the RSTC. The RSTC module then releases the sub-system 0 reset signals (`proc_nreset` and `periph_nreset`). Note that the sub-system 1 remains in reset after power-up.

Once the system has started, the user can program a supply monitor and/or a brownout detector. If a powerfail condition occurs on either VDDIO or on VDDCORE power supplies, the SUPC asserts the system reset signal (`vddcore_nreset`). This signal is released when the powerfail condition is cleared.

When the backup low-power mode is entered, the SUPC sequentially asserts the system reset signal and disables the voltage regulator, in order to maintain only the VDDBU\_SW power supply. Current consumption is reduced to less than one microamp for the backup part retention. Exit from this mode is possible on multiple wake-up sources including an event on the FWUP pin or WKUPx pins, or a clock alarm. To exit this mode, the SUPC operates in the same way as system start-up, in particular, the SUPC starts by enabling the core voltage regulator and the SHDN pin.

## 20.4.2 Slow Clock Generator

The Supply Controller embeds a slow clock generator that is supplied with the VDDBU\_SW power supply. As soon as the VDDBU\_SW is supplied, both the crystal oscillator and the embedded RC oscillator are powered up, but only the embedded RC oscillator is enabled. This allows the slow clock to be valid in a short time (about 100 µs).

The user can select the crystal oscillator to be the source of the slow clock, as it provides a more accurate frequency. The command is executed by writing the Supply Controller Control register (SUPC\_CR) with the XTALSEL bit at 1, resulting in the following sequence:

1. The crystal oscillator is enabled.
2. A number of slow RC oscillator clock periods is counted to cover the start-up time of the crystal oscillator (refer to the electrical characteristics for information on 32 kHz crystal oscillator start-up time).
3. The slow clock is switched to the output of the crystal oscillator.
4. The RC oscillator is disabled to save power.

The switching time may vary depending on the slow RC oscillator clock frequency range. The switch of the slow clock source is glitch-free. The OSCSEL bit of the Supply Controller Status register (SUPC\_SR) indicates when the switch sequence is finished.

Coming back on the RC oscillator is only possible by shutting down the VDDBU\_SW power supply.

If the user does not need the crystal oscillator, the XIN32 and XOUT32 pins should be left unconnected.

The user can also set the crystal oscillator in bypass mode instead of connecting a crystal. In this case, the user has to provide the external clock signal on XIN32. The input characteristics of the XIN32 pin are given in the electrical characteristics section. In order to set the bypass mode, the OSCBYPASS bit of the Supply Controller Mode register (SUPC\_MR) must be set at 1.

## 20.4.3 Core Voltage Regulator Control/Backup Low Power Mode

The Supply Controller can be used to control the embedded voltage regulator.

The voltage regulator automatically adapts its quiescent current depending on the required load current. More information can be found in the Electrical Characteristics section.

The user can switch off the voltage regulator, and thus put the device in backup mode, by writing SUPC\_CR with the VROFF bit at 1.

This asserts the system reset signal after the write resynchronization time which lasts two slow clock cycles (worst case). Once the system reset signal is asserted, the processor and the peripherals are stopped one slow clock cycle before shutting down the core voltage regulator and pulling the SHDN pin to ground.

When the user does not use the internal voltage regulator and wants to supply VDDCORE by an external supply, it is possible to disable the voltage regulator. This is done through ONREG bit in SUPC\_MR.

## 20.4.4 LCD Voltage Regulator Control

The Supply Controller can be used to select the power supply source of the LCD voltage regulator.

This selection is done by the LCDMODE field in SUPC\_MR. After a backup reset, the LCDMODE field is at 0x0. No power supply source is selected and the SLCD Controller reset signal is asserted.

The status of the LCD Controller reset is given by the LCDS field in SUPC\_SR.

- If LCDMODE is written to 0x2 while it is at 0x0, after the write resynchronization time (about 2 slow clock cycles), the external power supply source is selected, then after one slow clock cycle, the SLCDC reset signal is released.
- If LCDMODE is written to 0x0 while it is at 0x2, after the write resynchronization time (about 2 slow clock cycles), the SLCDC reset signal is asserted, then after one slow clock cycle, the external power supply source is deselected.
- If LCDMODE is written to 0x3 while it is at 0x0, after the write resynchronization time (about 2 slow clock cycles), the internal power supply source is selected and the embedded regulator is turned on, then after 15 slow clock cycles, the SLCDC reset signal is released.

- If LCDMODE is written to 0x0 while it is at 0x3, after the write resynchronization time (about 2 slow clock cycles), the SLDC reset signal, then after one slow clock cycle, the internal power supply source is deselected.

There are several restrictions concerning the write of the LCDMODE field:

- The user must check that the previous power supply selection is done before writing LCDMODE again. To do that, the user must check that the LCDS flag has the correct value. If LCDMODE is written to 0x0, the LCDS flag is reset at 0. If LCDMODE is written to 0x0, the LCDS flag is set at 1.
- Writing LCDMODE to 0x2 while it is at 0x3 or writing LCDMODE to 0x3 while it is at 0x2 is forbidden and has no effect.
- Before writing LCDMODE to 0x2, the user must ensure that the external power supply is ready and supplies the VDDLCD pin.
- Before writing LCDMODE to 0x3, the user must ensure that the external power supply does not supply the VDDLCD pin.

To use the LCD controller to maintain display in backup mode, all configuration registers must be kept when backup mode entered.

#### 20.4.5 Using Backup Battery/Automatic Power Switch

The power switch is able to automatically select one power source among VDDBU and VDDIO.

As soon as VDDIO is present (higher than 1.9V), it supplies the backup area of the device (VDDBU\_SW = VDDIO) even if the voltage of VDDBU is higher than VDDIO. If not, the backup area is supplied by the VDDBU voltage source (VDDBU\_SW = VDDBU). For more information about power supply schematics, refer to the section on Power Considerations.

#### 20.4.6 Supply Monitor

The Supply Controller embeds a supply monitor located in the VDDBU\_SW power domain and which monitors VDDIO power supply.

The supply monitor can be used to prevent the processor from falling into an unpredictable state if the main power supply drops below a certain level.

The threshold of the supply monitor is programmable. It can be selected from 1.9V to 3.4V by steps of 100 mV. This threshold is programmed in the SMTH field of the Supply Controller Supply Monitor Mode register (SUPC\_SMMR).

The supply monitor can also be enabled during one slow clock period on every one of **either** 32, 256 or 2048 slow clock periods, depending on what the user selects. This can be configured by programming the SMSMPL field in SUPC\_SMMR.

Enabling the supply monitor for such reduced times divides the typical supply monitor power consumption by factors of 32, 256 or 2048, respectively, if the user does not need a continuous monitoring of the VDDIO power supply.

A supply monitor detection can either generate a system reset (vddcore\_nreset signal is asserted) or a system wake-up. Generating a system reset when a supply monitor detection occurs is enabled by writing the SMRSTEN bit to 1 in SUPC\_SMMR.

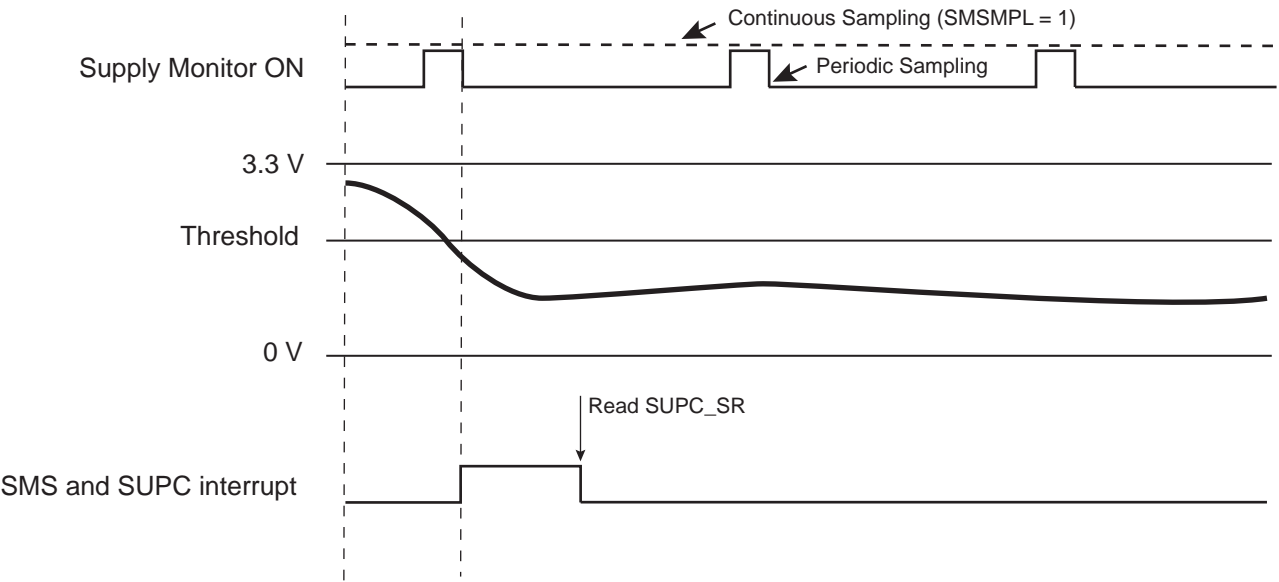
Waking up the system when a supply monitor detection occurs can be enabled by writing the SMEN bit to 1 in the Supply Controller Wake-up Mode register (SUPC\_WUMR).

The Supply Controller provides two status bits in the Supply Controller Status register for the supply monitor which determines whether the last wake-up was due to the supply monitor:

- The SMOS bit provides real-time information, updated at each measurement cycle or updated at each Slow Clock cycle, if the measurement is continuous.
- The SMS bit provides saved information and shows a supply monitor detection has occurred since the last read of SUPC\_SR.

The SMS bit can generate an interrupt if the SMIEN bit is set to 1 in SUPC\_SMMR.

Figure 20-2. Supply Monitor Status Bit and Associated Interrupt



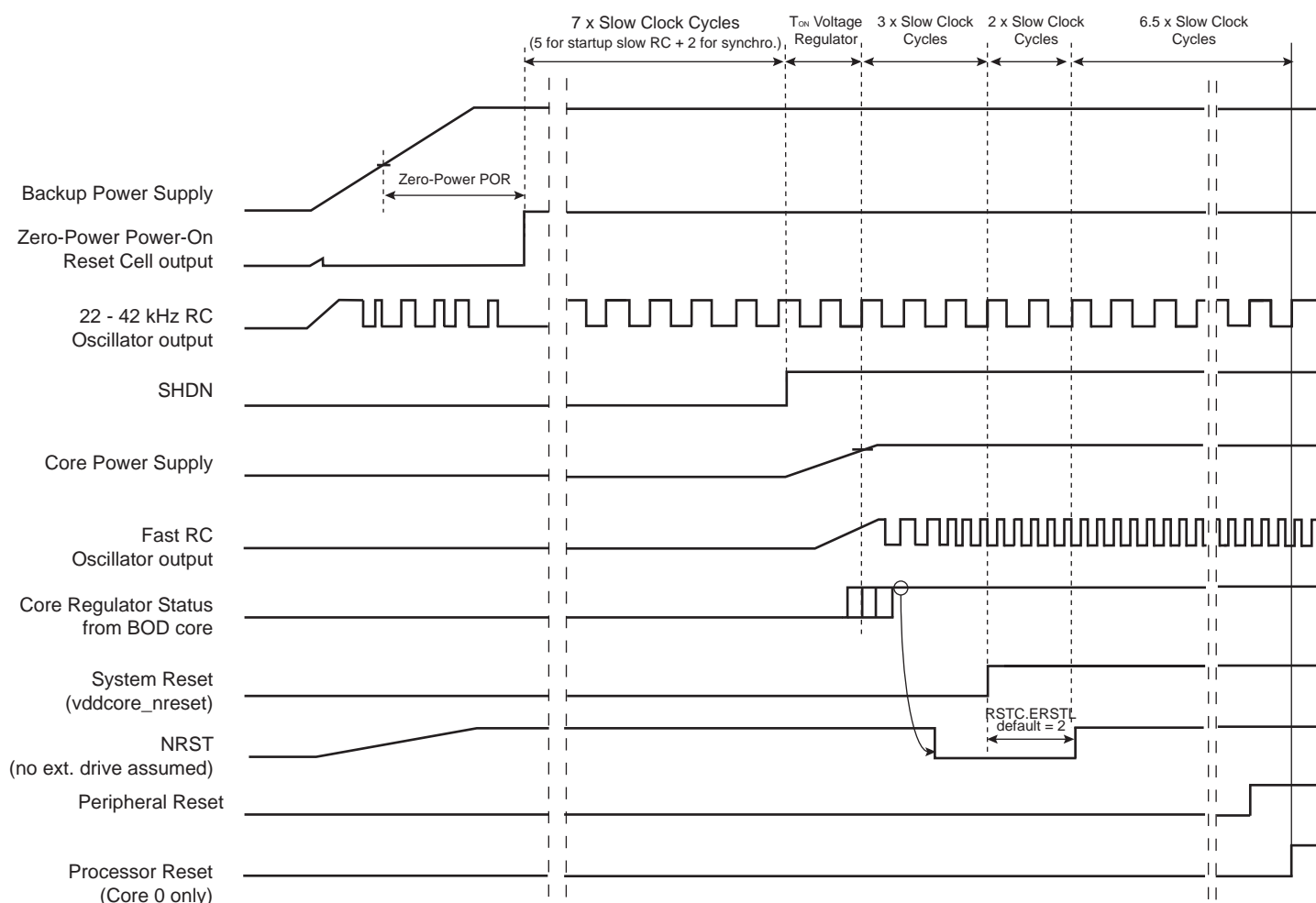
## 20.4.7 Backup Power Supply Reset

### 20.4.7.1 Raising the Backup Power Supply

As soon as the backup voltage VDDBU\_SW rises, the 32 kHz RC oscillator is powered up and the zero-power power-on reset cell maintains its output low as long as VDDBU\_SW has not reached its target voltage. During this time, the Supply Controller is entirely reset. When the VDDBU\_SW voltage becomes valid and zero-power power-on reset signal is released, a counter is started for five slow clock cycles, which is the time required for the 32 kHz RC oscillator to stabilize.

After this time, the SHDN pin is asserted high and the core voltage regulator is enabled. The core power supply rises and the brownout detector provides the core regulator status as soon as the core voltage VDDCORE is valid. The system reset signal is then released to the Reset Controller after the core voltage status has been confirmed as being valid for at least one slow clock cycle.

Figure 20-3. Raising the VDDBU\_SW Power Supply



Note: After processor reset rising, the core starts fetching instructions from Flash at 4 MHz.

### 20.4.7.2 SHDN Output Pin

The SHDN pin is designed to drive the enable pin of an external voltage regulator. This pin is controlled by the VROFF bit in SUPC\_CR. When the device goes into backup mode (VROFF written to 1), the SHDN pin is asserted low. Upon a wake-up event, the SHDN pin is released (VDDBU level).



## 20.4.8 System Reset

The Supply Controller manages the system reset signal (`vddcore_nreset`) to the Reset Controller, as described in [Section 20.4.7 "Backup Power Supply Reset"](#). The system reset signal is normally asserted before shutting down the core power supply and released as soon as the core power supply is correctly regulated.

There are two additional sources which can be programmed to activate the system reset signal:

- a supply monitor detection
- a brownout detection

### 20.4.8.1 Supply Monitor Reset

The supply monitor is capable of generating a reset of the system. This can be enabled by setting the `SMRSTEN` bit in `SUPC_SMMR`.

If `SMRSTEN` is set and if a supply monitor detection occurs, the system reset signal is immediately activated for a minimum of 1 slow clock cycles.

### 20.4.8.2 Brownout Detector Reset

The brownout detector provides the core voltage status signal (`BODCORE_out`) to the SUPC which indicates that the voltage regulation is operating as programmed. If this signal is lost for longer than 1 slow clock period while the voltage regulator is enabled, the Supply Controller can assert system reset signal. This feature is enabled by writing `BODRSTEN` to 1 in `SUPC_MR`.

If `BODRSTEN` is set and the voltage regulation is lost (output voltage of the regulator too low), the system reset signal is asserted for a minimum of 1 slow clock cycle and then released if the core voltage status has been reactivated. The `BODRSTS` bit is set in `SUPC_SR` so that the user knows the source of the last reset.

The system reset signal remains active as long as the core voltage status signal (`BODCORE_out`) indicates a powerfail condition.

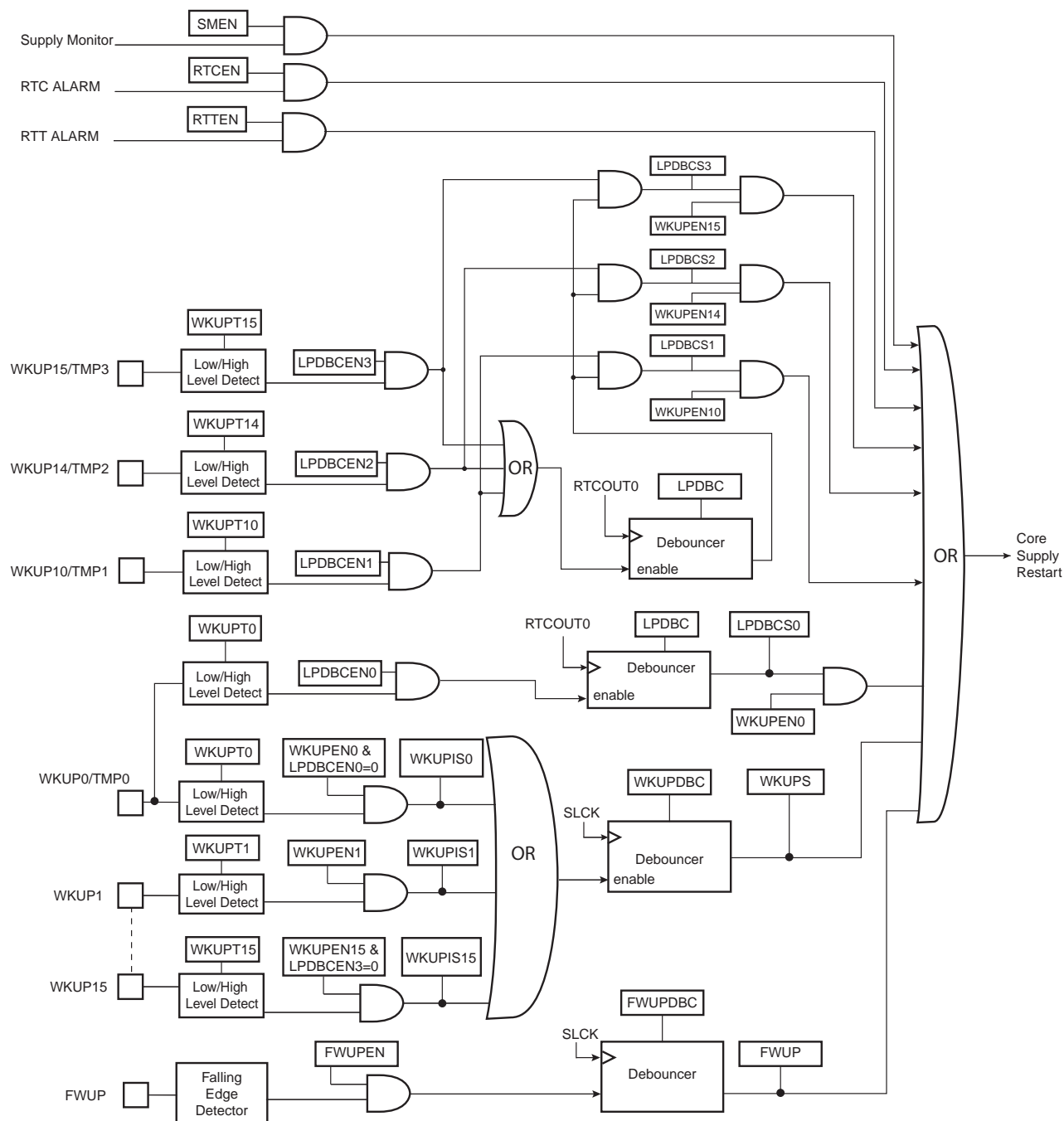
### 20.4.8.3 Power-on-Reset on `VDDBU_SW`

The power-on-reset monitors `VDDBU_SW`. It is active by default and monitors voltage at start up but also during power down. It can be deactivated by software through `SUPC_MR`. If `VDDBU_SW` goes below the threshold voltage, the entire chip is reset. Note that due to the automatic power switch, `VDDBU_SW` can be either `VDDIO` or `VDDBU`.

## 20.4.9 Wake-up Sources

The wake-up events allow the device to exit backup mode. When a wake-up event is detected, the Supply Controller performs a sequence which automatically reenables the core power supply.

**Figure 20-4. SAM4C16/8 Wake-up Sources**



#### 20.4.9.1 Force Wake-up

The FWUP pin is enabled as a wake-up source by writing the FWUPEN bit to 1 in SUPC\_WUMR. The FWUPDBC field in the same register then selects the debouncing period, which can be selected between 3, 32, 512, 4,096 or 32,768 slow clock cycles, corresponding to about 100  $\mu$ s, about 1 ms, about 16 ms, about 128 ms and about 1 second, respectively (for a typical slow clock frequency of 32 kHz). Programming FWUPDBC to 0x0 selects an immediate wake-up, i.e., the FWUP must be low during at least one slow clock period to wake up the system.

If the FWUP pin is asserted for a time longer than the debouncing period, a system wake-up is started and the FWUP bit in SUPC\_SR is set and remains high until the register is read.

### 20.4.9.2 Wake-up Inputs

The wake-up inputs WKUPx can be programmed to perform a system wake-up. Each input can be enabled by writing to 1 the corresponding bit, WKUPENx, in the Wake-up Inputs register (SUPC\_WUIR). The wake-up level can be selected with the corresponding polarity bit, WKUPPLx, also located in SUPC\_WUIR.

All the resulting signals are wired-ORed to trigger a debounce counter, which can be programmed with the WKUPDBC field in SUPC\_WUMR. The WKUPDBC field can select a debouncing period of 3, 32, 512, 4,096 or 32,768 slow clock cycles. This corresponds respectively to about 100  $\mu$ s, about 1 ms, about 16 ms, about 128 ms and about 1 second (for a typical slow clock frequency of 32 kHz). Programming WKUPDBC to 0x0 selects an immediate wake-up, i.e., an enabled WKUP pin must be active according to its polarity during a minimum of one slow clock period to wake up the core power supply.

If an enabled WKUPx pin is asserted for a time longer than the debouncing period, a system wake-up is started and the signals, WKUPx as shown in [Figure 20-4](#), are latched in SUPC\_SR. This allows the user to identify the source of the wake-up, however, if a new wake-up condition occurs, the primary information is lost. No new wake-up can be detected since the primary wake-up condition has disappeared.

### 20.4.9.3 Low-power Debouncer Inputs (Tamper Detection Pins)

Low-power debouncer inputs are dedicated to tamper detection. If the tamper sensor is biased through a resistor and constantly driven by the power supply, this leads to power consumption as long as the tamper detection switch is in its active state. To prevent power consumption when the switch is in active state, the tamper sensor circuitry can be intermittently powered, thus, a specific waveform must be generated.

It is possible to generate a waveform (on pin RTCOUT0) in all modes (including backup mode). Refer to the RTC section for waveform generation.

For SAM4C16/8 devices, separate debouncers are embedded, one for WKUP0/TMP0 input and a shared one for WKUP10/TMP1, WKUP14/TMP2, WKUP15/TMP3 inputs. See [Figure 20-4](#).

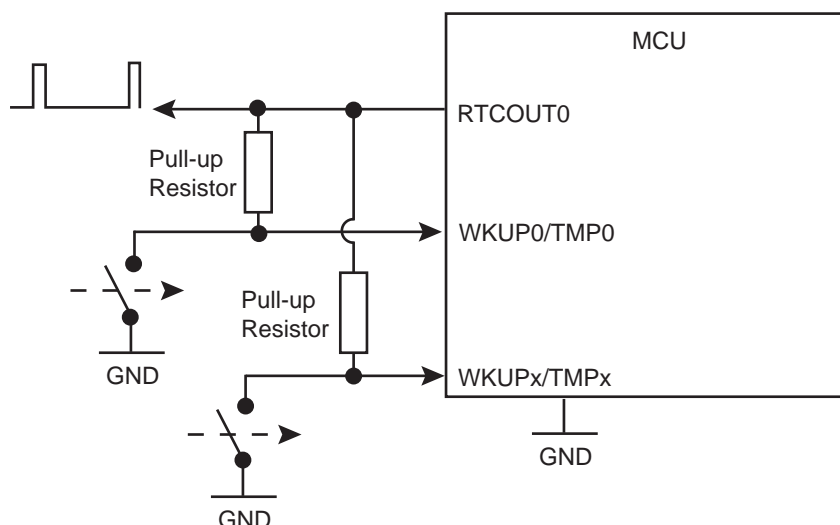
The WKUP0/TMP0 and/or WKUP10/TMP1, WKUP14/TMP2, WKUP15/TMP3 inputs can be programmed to perform a system wake-up with a debouncing done by RTCOUT0. This can be enabled by setting LPDBCEN0/1/2/3 bit in the [Supply Controller Wake-up Mode Register](#) (SUPC\_WUMR).

These inputs can be also used when VDDCORE is powered to get tamper detection function with a low power debounce function and to raise an interrupt.

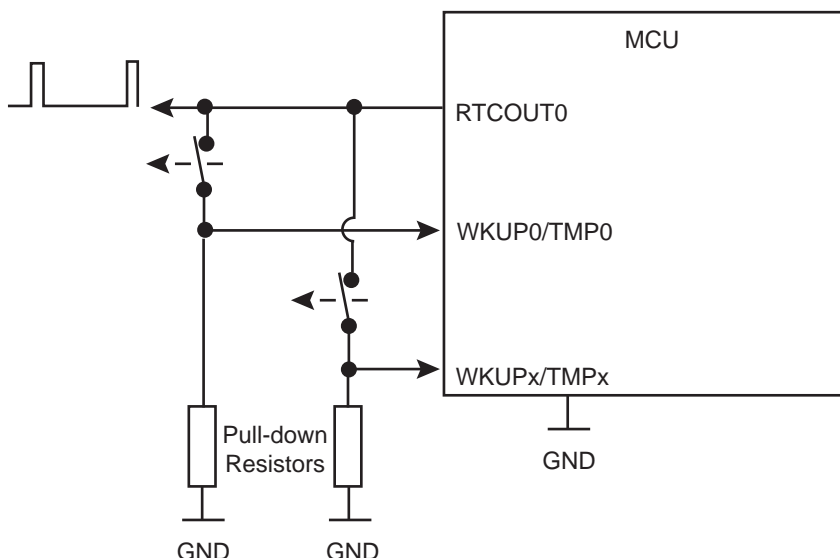
This mode of operation requires the RTC output (RTCOUT0) to be configured to generate a duty cycle programmable pulse (i.e., OUT0 = 0x7 in RTC\_MR) in order to create the sampling points of both debouncers. The sampling point is the falling edge of the RTCOUT0 waveform.

[Figure 20-5](#) shows an example of an application where two tamper switches are used. RTCOUT0 powers the external pull-up used by the tamperers.

**Figure 20-5. Low Power Debouncer (Push-to-Make Switch, Pull-up Resistors)**



**Figure 20-6. Low Power Debouncer (Push-to-Break Switch, Pull-down Resistors)**



The debouncing period duration is configurable. The period is identical for all debouncers (i.e., the duration cannot be adjusted separately for each debouncer). The number of successive identical samples to wake up the system can be configured from 2 up to 8 in the LPDBC field of SUPC\_WUMR. The period of time between two samples can be configured by programming the TPERIOD field in the RTC\_MR. Power parameters can be adjusted by modifying the period of time in the THIGH field in RTC\_MR.

The wake-up polarity of the inputs can be independently configured by writing WKUPT0 /WKUPT10/WKUPT14/WKUPT15 fields in SUPC\_WUMR.

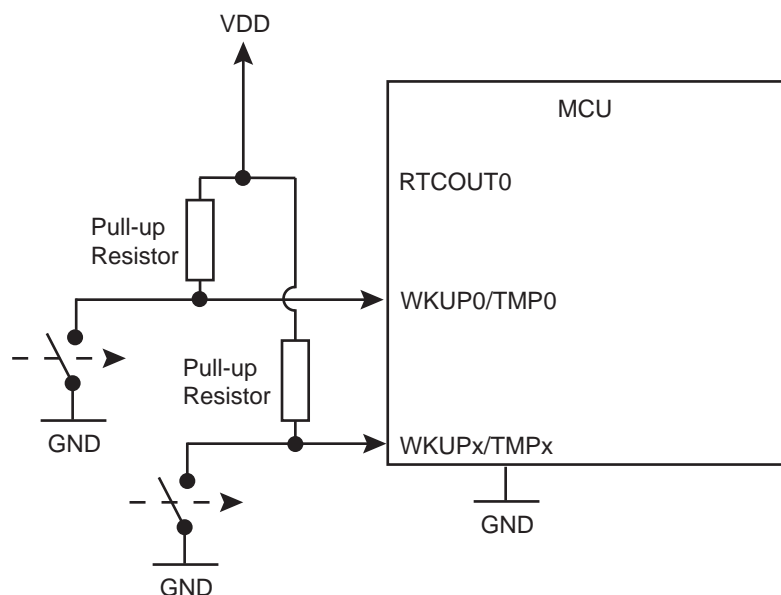
In order to determine which wake-up/tamper pin triggers the system wake-up, a status flag is associated for each low-power debouncer. These flags can be read in the SUPC\_SR.

A debounce event (tamper detection) can perform an immediate clear (0 delay) on first half the general-purpose backup registers (GPBR). The LPDBCCLR bit must be set in [Section 20.6.5 "Supply Controller Mode Register"](#) and it is possible to individually disable the clear capability for TMP1/TMP2/TMP3 by writing a 1 in the corresponding bit DISTMPCLR1/2/3.

Note that it is not mandatory to use the RTCOUT0 pin when using the WKUP0/WKUP10/WKUP14/WKUP15 pins as tampering inputs (TMP0/TMP1/TMP2/TMP3) in any mode. Using RTCOUT0 pin provides a “sampling mode” to further reduce the power consumption of the tamper detection circuitry. If RTCOUT0 is not used, the RTC must be configured to create an internal sampling point for the debouncer logic. The period of time between two samples can be configured by programming the TPERIOD field in the RTC\_MR.

Figure 20-7 shows how to use WKUPx/TMPx without RTCOUT0 pin.

**Figure 20-7. Using WKUP/TMP Pins Without RTCOUT Pins**



#### 20.4.9.4 Clock Alarms

The RTC and the RTT alarms generates a system wake-up. This can be enabled by writing respectively, the bits RTCEN and RTTEN to 1 in SUPC\_WUMR.

The Supply Controller does not provide any status as the information is available in the user interface of either the Real-Time Timer or the Real-Time Clock.

#### 20.4.9.5 Supply Monitor Detection

The supply monitor generates a system wake-up when configured as a wake-up source. See [Section 20.4.6 "Supply Monitor"](#).

## 20.5 Register Write Protection

To prevent any single software error from corrupting SUPC behavior, certain registers in the address space can be write-protected by setting the WPEN bit in the [System Controller Write Protection Mode Register](#) (SYSC\_WPMR).

The following registers can be write-protected:

- RSTC Mode Register
- RTT Mode Register
- RTT Alarm Register
- RTC Control Register
- RTC Mode Register
- RTC Time Alarm Register
- RTC Calendar Alarm Register
- General Purpose Backup Registers
- [Supply Controller Control Register](#)
- [Supply Controller Supply Monitor Mode Register](#)
- [Supply Controller Mode Register](#)
- [Supply Controller Wake-up Mode Register](#)
- [Supply Controller Wake-up Inputs Register](#)

## 20.6 Supply Controller (SUPC) User Interface

The user interface of the Supply Controller is part of the System Controller user interface.

### 20.6.1 System Controller (SYSC) User Interface

**Table 20-1. System Controller Peripheral Offsets**

Offset	System Controller Peripheral	Name
0x00-0x0c	Reset Controller	RSTC
0x10-0x2C	Supply Controller	SUPC
0x30-0x3C	Real Time Timer	RTT
0x50-0x5C	Watchdog Timer	WDT
0x60-0x8C	Real Time Clock	RTC
0x90-0xDC	General Purpose Backup Register	GPBR
0xE0	Reserved	
0xE4	Write Protection Mode Register	SYSC_WPMR
0xE8-0xF8	Reserved	
0xFC	Reserved	
0x100-0x10C	Reinforced Safety Watchdog Timer	RSWDT
0x110-0x124	TimeStamping Registers	RTC

### 20.6.2 Supply Controller (SUPC) User Interface

**Table 20-2. Register Mapping**

Offset	Register	Name	Access	Reset
0x00	Supply Controller Control Register	SUPC_CR	Write-only	N/A
0x04	Supply Controller Supply Monitor Mode Register	SUPC_SMMR	Read/Write	0x0000_0000
0x08	Supply Controller Mode Register	SUPC_MR	Read/Write	0x0000_DA00
0x0C	Supply Controller Wake-up Mode Register	SUPC_WUMR	Read/Write	0x0000_0000
0x10	Supply Controller Wake-up Inputs Register	SUPC_WUIR	Read/Write	0x0000_0000
0x14	Supply Controller Status Register	SUPC_SR	Read-only	0x0000_0000
0x18	Reserved			
0xFC	Reserved			

### 20.6.3 Supply Controller Control Register

**Name:** SUPC\_CR  
**Address:** 0x400E1410  
**Access:** Write-only

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	XTALSEL	VROFF	–	–

- **VROFF: Voltage Regulator Off**

0 (NO\_EFFECT): No effect.

1 (STOP\_VREG): If KEY is correct, asserts the system reset signal and stops the voltage regulator.

- **XTALSEL: Crystal Oscillator Select**

0 (NO\_EFFECT): No effect.

1 (CRYSTAL\_SEL): If KEY is correct, switches the slow clock on the crystal oscillator output.

- **KEY: Password**

Value	Name	Description
0xA5	PASSWD	Writing any other value in this field aborts the write operation.



## 20.6.4 Supply Controller Supply Monitor Mode Register

**Name:** SUPC\_SMMR

**Address:** 0x400E1414

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	SMIEN	SMRSTEN	–	SMSMPL		
7	6	5	4	3	2	1	0
–	–	–	–	SMTH			

- **SMTH: Supply Monitor Threshold**

Selects the threshold voltage of the supply monitor. Refer to the Electrical Characteristics for voltage values.

- **SMSMPL: Supply Monitor Sampling Period**

Value	Name	Description
0x0	SMD	Supply Monitor disabled
0x1	CSM	Continuous Supply Monitor
0x2	32SLCK	Supply Monitor enabled one SLCK period every 32 SLCK periods
0x3	256SLCK	Supply Monitor enabled one SLCK period every 256 SLCK periods
0x4	2048SLCK	Supply Monitor enabled one SLCK period every 2,048 SLCK periods

- **SMRSTEN: Supply Monitor Reset Enable**

0 (NOT\_ENABLE): The system reset signal is not affected when a supply monitor detection occurs.

1 (ENABLE): The system reset signal is asserted when a supply monitor detection occurs.

- **SMIEN: Supply Monitor Interrupt Enable**

0 (NOT\_ENABLE): The SUPC interrupt signal is not affected when a supply monitor detection occurs.

1 (ENABLE): The SUPC interrupt signal is asserted when a supply monitor detection occurs.

## 20.6.5 Supply Controller Mode Register

**Name:** SUPC\_MR  
**Address:** 0x400E1418  
**Access:** Read/Write

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
–	–	–	OSCBYPASS	–	–	–	–
15	14	13	12	11	10	9	8
BUPPOREN	ONREG	BODDIS	BODRSTEN	–	–	–	–
7	6	5	4	3	2	1	0
–	–	LCDMODE		LCDVROUT			

- **LCDVROUT: LCD Voltage Regulator Output**

Adjusts the output voltage of the LCD Voltage Regulator. Refer to the electrical characteristics for voltage values.

- **LCDMODE: LCD Controller Mode of Operation**

Value	Name	Description
0x0	LCDOFF	The internal supply source and the external supply source are both deselected (OFF Mode).
0x2	LCDON_EXTVR	The external supply source for LCD (VDDLCD) is selected (the LCD voltage regulator is in Hi-Z Mode).
0x3	LCDON_INVR	The internal supply source for LCD (the LCD Voltage Regulator) is selected (Active Mode).

- **BODRSTEN: Brownout Detector Reset Enable**

0 (NOT\_ENABLE): The system reset signal is not affected when a brownout detection occurs.

1 (ENABLE): The system reset signal is asserted when a brownout detection occurs.

- **BODDIS: Brownout Detector Disable**

0 (ENABLE): The core brownout detector is enabled.

1 (DISABLE): The core brownout detector is disabled.

- **ONREG: Voltage Regulator enable**

0 (ONREG\_UNUSED): Internal voltage regulator is not used (external power supply is used).

1 (ONREG\_USED): Internal voltage regulator is used.

- **BUPPOREN: Backup Area Power-On Reset Enable**

0 (BUPPOR\_DISABLE): Disables the backup POR.

1 (BUPPOR\_ENABLE): Enables the backup POR.

**Note:** The value written in BUPPOREN is effective when BUPPORS has the same value in [Supply Controller Status Register](#).

- **OSCBYPASS: Oscillator Bypass**

0 (NO\_EFFECT): No effect. Clock selection depends on XTALSEL value.

1 (BYPASS): The 32 kHz crystal oscillator is selected and put in bypass mode.

- **KEY: Password Key**

Value	Name	Description
0xA5	PASSWD	Writing any other value in this field aborts the write operation.

## 20.6.6 Supply Controller Wake-up Mode Register

**Name:** SUPC\_WUMR

**Address:** 0x400E141C

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	DISTSTMP3	DISTSTMP2	DISTSTMP1	–	DISTMPCLR 3	DISTMPCLR 2	DISTMPCLR 1
23	22	21	20	19	18	17	16
–	–	LPDBCEN3	LPDBCEN2	–	LPDBC		
15	14	13	12	11	10	9	8
–	WKUPDBC			–	FWUPDBC		
7	6	5	4	3	2	1	0
LPDBCCLR	LPDBCEN1	LPDBCEN0	–	RTCEN	RTTEN	SMEN	FWUPEN

- **FWUPEN: Force Wake-up Enable**

0 (NOT\_ENABLE): The force wake-up pin has no wake-up effect.

1 (ENABLE): The force wake-up pin low forces a system wake-up.

- **SMEN: Supply Monitor Wake-up Enable**

0 (NOT\_ENABLE): The supply monitor detection has no wake-up effect.

1 (ENABLE): The supply monitor detection forces a system wake-up.

- **RTTEN: Real-time Timer Wake-up Enable**

0 (NOT\_ENABLE): The RTT alarm signal has no wake-up effect.

1 (ENABLE): The RTT alarm signal forces a system wake-up.

- **RTCEN: Real-time Clock Wake-up Enable**

0 (NOT\_ENABLE): The RTC alarm signal has no wake-up effect.

1 (ENABLE): The RTC alarm signal forces a system wake-up.

- **LPDBCEN0: Low-Power Debouncer Enable WKUP0/TMP0**

0 (NOT\_ENABLE): The WKUP0/TMP0 input pin is not connected to the low-power debouncer.

1 (ENABLE): The WKUP0/TMP0 input pin is connected to the low-power debouncer and can force a system wake-up.

- **LPDBCEN1: Low-Power Debouncer Enable WKUP10/TMP1**

0 (NOT\_ENABLE): The WKUP10/TMP1 input pin is not connected to the low-power debouncer.

1 (ENABLE): The WKUP10/TMP1 input pin is connected to the low-power debouncer and can force a system wake-up.

- **LPDBCCLR: Low-Power Debouncer Clear**

0 (NOT\_ENABLE): A low-power debounce event does not create an immediate clear on the first half of GPBR registers.

1 (ENABLE): A low-power debounce event on WKUP0/TMP0 or WKUP10/14/15/TMP1/2/3 (if DISTMPCLR1/2/3 is cleared) generates an immediate clear on the first half of GPBR registers.

- **FWUPDBC: Force Wake-up Debouncer Period**

Value	Name	Description
0	IMMEDIATE	Immediate, no debouncing, detected active at least on one Slow Clock edge.
1	3_SCLK	FWUP shall be low for at least 3 SLCK periods
2	32_SCLK	FWUP shall be low for at least 32 SLCK periods
3	512_SCLK	FWUP shall be low for at least 512 SLCK periods
4	4096_SCLK	FWUP shall be low for at least 4,096 SLCK periods
5	32768_SCLK	FWUP shall be low for at least 32,768 SLCK periods

- **WKUPDBC: Wake-up Inputs Debouncer Period**

Value	Name	Description
0	IMMEDIATE	Immediate, no debouncing, detected active at least on one Slow Clock edge.
1	3_SCLK	WKUPx shall be in its active state for at least 3 SLCK periods
2	32_SCLK	WKUPx shall be in its active state for at least 32 SLCK periods
3	512_SCLK	WKUPx shall be in its active state for at least 512 SLCK periods
4	4096_SCLK	WKUPx shall be in its active state for at least 4,096 SLCK periods
5	32768_SCLK	WKUPx shall be in its active state for at least 32,768 SLCK periods

- **LPDBC: Low Power Debouncer Period**

Value	Name	Description
0	DISABLE	Disable the low-power debouncers.
1	2_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 2 RTCOUT0 periods
2	3_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 3 RTCOUT0 periods
3	4_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 4 RTCOUT0 periods
4	5_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 5 RTCOUT0 periods
5	6_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 6 RTCOUT0 periods
6	7_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 7 RTCOUT0 periods
7	8_RTCOUT0	WKUP0/10/14/15/TMP0/1/2/3 in active state for at least 8 RTCOUT0 periods

- **LPDBCEN2: Low Power Debouncer Enable WKUP14/TMP2**

0 (NOT\_ENABLE): The WKUP14/TMP2 input pin is not connected to the low-power debouncer.

1 (ENABLE): The WKUP14/TMP2 input pin is connected to the low-power debouncer and can force a system wake-up.

- **LPDBCEN3: Low Power Debouncer Enable WKUP15/TMP3**

0 (NOT\_ENABLE): The WKUP15/TMP3 input pin is not connected to the low-power debouncer.

1 (ENABLE): The WKUP15/TMP3 input pin is connected to the low-power debouncer and can force a system wake-up.

- **DISTMPCLR1: Disable GPBR Clear Command from WKUP10/TMP1 pin**

0 (ENABLE): The WKUP10/TMP1 input pin can clear the GPBR (if LPDBCCLR is enabled) when tamper is detected.

1 (DISABLE): The WKUP10/TMP1 input pin has no effect on the GPBR value (no clear on tamper detection).

- **DISTMPCLR2: Disable GPBR Clear Command from WKUP14/TMP2 Pin**

0 (ENABLE): The WKUP14/TMP2 input pin can clear the GPBR (if LPDBCCLR is enabled) when tamper is detected.

1 (DISABLE): The WKUP14/TMP2 input pin has no effect on the GPBR value (no clear on tamper detection).

- **DISTMPCLR3: Disable GPBR Clear Command from WKUP15/TMP3 Pin**

0 (ENABLE): The WKUP15/TMP3 input pin can clear the GPBR (if LPDBCCLR is enabled) when tamper is detected.

1 (DISABLE): The WKUP15/TMP3 input pin has no effect on the GPBR value (no clear on tamper detection).

- **DISTSTMP1: Disable Timestamp from WKUP10/TMP1 Pin**

0 (ENABLE): A tamper detection on WKUP10/TMP1 pin generates a timestamp.

1 (DISABLE): A tamper detection on WKUP10/TMP1 does NOT generate a report in timestamp register.

- **DISTSTMP2: Disable Timestamp from WKUP14/TMP2 Pin**

0 (ENABLE): A tamper detection on WKUP14/TMP2 pin generates a timestamp.

1 (DISABLE): A tamper detection on WKUP14/TMP2 does NOT generate a report in timestamp register.

- **DISTSTMP3: Disable Timestamp from WKUP15/TMP3 Pin**

0 (ENABLE): A tamper detection on WKUP15/TMP3 pin generates a timestamp.

1 (DISABLE): A tamper detection on WKUP15/TMP3 does NOT generate a report in timestamp register.

## 20.6.7 Supply Controller Wake-up Inputs Register

**Name:** SUPC\_WUIR

**Address:** 0x400E1420

**Access:** Read/Write

31	30	29	28	27	26	25	24
WKUPT15	WKUPT14	WKUPT13	WKUPT12	WKUPT11	WKUPT10	WKUPT9	WKUPT8
23	22	21	20	19	18	17	16
WKUPT7	WKUPT6	WKUPT5	WKUPT4	WKUPT3	WKUPT2	WKUPT1	WKUPT0
15	14	13	12	11	10	9	8
WKUPEN15	WKUPEN14	WKUPEN13	WKUPEN12	WKUPEN11	WKUPEN10	WKUPEN9	WKUPEN8
7	6	5	4	3	2	1	0
WKUPEN7	WKUPEN6	WKUPEN5	WKUPEN4	WKUPEN3	WKUPEN2	WKUPEN1	WKUPEN0

- **WKUPENx: WKUPx Input Enable**

0 (DISABLE): The corresponding wake-up input has no wake-up effect.

1 (ENABLE): The corresponding wake-up input forces a system wake-up.

- **WKUPTx: WKUPx Input Type**

0 (LOW): A low level for a period defined by WKUPDBC on the corresponding wake-up input forces a system wake-up.

1 (HIGH): A high level for a period defined by WKUPDBC on the corresponding wake-up input forces a system wake-up.

## 20.6.8 Supply Controller Status Register

**Name:** SUPC\_SR

**Address:** 0x400E1424

**Access:** Read-only

31	30	29	28	27	26	25	24
WKUPIS15	WKUPIS14	WKUPIS13	WKUPIS12	WKUPIS11	WKUPIS10	WKUPIS9	WKUPIS8
23	22	21	20	19	18	17	16
WKUPIS7	WKUPIS6	WKUPIS5	WKUPIS4	WKUPIS3	WKUPIS2	WKUPIS1	WKUPIS0
15	14	13	12	11	10	9	8
BUPPORS	LPDBCS1	LPDBCS0	FWUPIS	–	LPDBCS3	LPDBCS2	LCDS
7	6	5	4	3	2	1	0
OSCSEL	SMOS	SMS	SMRSTS	BODRSTS	SMWS	WKUPS	FWUPS

**Note:** Because of the asynchronism between the Slow Clock (SCLK) and the System Clock (MCK), the status register flag reset is taken into account only 2 slow clock cycles after the read of the SUPC\_SR.

### • FWUPS: FWUP Wake-up Status

0 (NO): No wake-up due to the assertion of the FWUP pin has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one wake-up due to the assertion of the FWUP pin has occurred since the last read of SUPC\_SR.

### • WKUPS: WKUP Wake-up Status

0 (NO): No wake-up due to the assertion of the WKUP pins has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one wake-up due to the assertion of the WKUP pins has occurred since the last read of SUPC\_SR.

### • SMWS: Supply Monitor Detection Wake-up Status

0 (NO): No wake-up due to a supply monitor detection has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one wake-up due to a supply monitor detection has occurred since the last read of SUPC\_SR.

### • BODRSTS: Brownout Detector Reset Status

0 (NO): No core brownout rising edge event has been detected since the last read of SUPC\_SR.

1 (PRESENT): At least one brownout output rising edge event has been detected since the last read of SUPC\_SR.

When the voltage remains below the defined threshold, there is no rising edge event at the output of the brownout detection cell. The rising edge event occurs only when there is a voltage transition below the threshold.

### • SMRSTS: Supply Monitor Reset Status

0 (NO): No supply monitor detection has generated a system reset since the last read of the SUPC\_SR.

1 (PRESENT): At least one supply monitor detection has generated a system reset since the last read of the SUPC\_SR.

### • SMS: Supply Monitor Status

0 (NO): No supply monitor detection since the last read of SUPC\_SR.

1 (PRESENT): At least one supply monitor detection since the last read of SUPC\_SR.

### • SMOS: Supply Monitor Output Status

0 (HIGH): The supply monitor detected **VDDIO** higher than its threshold at its last measurement.



1 (LOW): The supply monitor detected VDDIO lower than its threshold at its last measurement.

- **OSCSEL: 32 kHz Oscillator Selection Status**

0 (RC): The slow clock, SLCK, is generated by the embedded 32 kHz RC oscillator.

1 (CRYST): The slow clock, SLCK, is generated by the 32 kHz crystal oscillator.

- **LCDS: LCD Status**

0 (DISABLED): LCD controller is disabled.

1 (ENABLED): LCD controller is enabled.

- **LPDBCS2: Low Power Debouncer Tamper Status on WKUP14/TMP2**

0 (NO): No tamper detection or wake-up due to the assertion of the WKUP14/TMP2 pin has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one tamper detection and wake-up (if enabled by WKUPEN14) due to the assertion of the WKUP14/TMP2 pin has occurred since the last read of SUPC\_SR. The SUPC interrupt line is asserted while LPDBCS2 is 1.

- **LPDBCS3: Low Power Debouncer Tamper Status on WKUP15/TMP3**

0 (NO): No tamper detection or wake-up due to the assertion of the WKUP15/TMP3 pin has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one tamper detection and wake-up (if enabled by WKUPEN15) due to the assertion of the WKUP15/TMP3 pin has occurred since the last read of SUPC\_SR. The SUPC interrupt line is asserted while LPDBCS2 is 1.

- **FWUPIS: FWUP Input Status**

0 (LOW): FWUP input is tied low.

1 (HIGH): FWUP input is tied high.

- **LPDBCS0: Low Power Debouncer Wake-up Status on WKUP0/TMP0**

0 (NO): No tamper detection or wake-up due to the assertion of the WKUP0/TMP0 pin has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one tamper detection and wake-up (if enabled by WKUPEN0) due to the assertion of the WKUP0/TMP0 pin has occurred since the last read of SUPC\_SR. The SUPC interrupt line is asserted while LPDBCS0 is 1.

- **LPDBCS1: Low Power Debouncer Wake-up Status on WKUP10/TMP1**

0 (NO): No tamper detection or wake-up due to the assertion of the WKUP10 pin has occurred since the last read of SUPC\_SR.

1 (PRESENT): At least one tamper detection and wake-up (if enabled by WKUPEN10) due to the assertion of the WKUP10/TMP1 pin has occurred since the last read of SUPC\_SR. The SUPC interrupt line is asserted while LPDBCS1 is 1.

- **BUPPORS: Backup Area Power-On Reset Status**

0 (BUPPOR\_DISABLED): Backup POR is disabled.

1 (BUPPOR\_ENABLED): Backup POR is enabled.

Note: The value written in BUPPORN is effective when BUPPOREN is carrying the same value in [Supply Controller Status Register](#).

- **WKUPISx: WKUPx Input Status**

0 (DIS): The corresponding wake-up input is disabled, or was inactive at the time the debouncer triggered a wake-up event.

1 (EN): The corresponding wake-up input was active at the time the debouncer triggered a wake-up event.

### 20.6.9 System Controller Write Protection Mode Register

**Name:** SYSC\_WPMR

**Access:** Read/Write

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

For more information on Write Protection registers, refer to [Section 20.5 "Register Write Protection"](#).

- **WPEN:**

0: Disables the write protection if WPKEY corresponds to 0x525443 (SYSC in ASCII).

1: Enables the write protection if WPKEY corresponds to 0x525443 (SYSC in ASCII).

See [Section 20.5 "Register Write Protection"](#) for the list of registers that can be protected.

- **WPKEY: Write Protection Key**

Value	Name	Description
0x525443	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

## 21. General-Purpose Backup Registers (GPBR)

### 21.1 Description

The System Controller embeds 16 General-purpose Backup registers.

It is possible to generate an immediate clear of the content of General-purpose Backup registers 0 to 7 (first half), if a Tamper event is detected on one of the tamper pins, TMP0 to TMP3. The content of the other General-purpose Backup registers (second half) remains unchanged. The Tamper events on pins TMP1 to TMP3 to perform a GPBR clear are configurable in the Supply Controller. The TMP0 Tamper event always performs an immediate clear.

The Supply Controller module must be programmed accordingly. In the register SUPC\_WUMR in the Supply Controller module, LPDBCCLR, LPDBCEN0, LPDBCEN1, LPDBCEN2 and LPDBCEN3 bit must be configured to 1 and LPDBC must be other than 0.

If a Tamper event has been detected, it is not possible to write to the General-purpose Backup registers while the LPBCSx flags are not cleared in the Supply Controller Status register SUPC\_SR.

### 21.2 Embedded Characteristics

- 16 32-bit General Purpose Backup Registers

## 21.3 General Purpose Backup Registers (GPBR) User Interface

Table 21-1. Register Mapping

Offset	Register	Name	Access	Reset
0x0	General Purpose Backup Register 0	SYS_GPBR0	Read-write	—
...	...	...	...	...
0xCC	General Purpose Backup Register 15	SYS_GPBR15	Read-write	—

### 21.3.1 General Purpose Backup Register x

**Name:** SYS\_GPBRx

**Address:** 0x400E1490

**Access:** Read-write

31	30	29	28	27	26	25	24
GPBR_VALUE							
23	22	21	20	19	18	17	16
GPBR_VALUE							
15	14	13	12	11	10	9	8
GPBR_VALUE							
7	6	5	4	3	2	1	0
GPBR_VALUE							

- **GPBR\_VALUE:** Value of GPBR x

If a Tamper event has been detected, it is not possible to write GPBR\_VALUE as long as the LPDBCS0 or LPDBCS3 flags have not been cleared in Supply Controller Status register SUPC\_SR.

## 22. Enhanced Embedded Flash Controller (EEFC)

### 22.1 Description

The Enhanced Embedded Flash Controller (EEFC) ensures the interface of the Flash block with the 32-bit internal bus. Its 128-bit or 64-bit wide memory interface increases performance. It also manages the programming, erasing, locking and unlocking sequences of the Flash using a full set of commands. One of the commands returns the embedded Flash descriptor definition that informs the system about the Flash organization, thus making the software generic.

### 22.2 Embedded Characteristics

- Interface of the Flash Block with the 32-bit Internal Bus
- Increases Performance in Thumb-2 Mode with 128-bit or 64-bit-wide Memory Interface up to 100 MHz
- Code Loop Optimization
- 128 Lock Bits, Each Protecting a Lock Region
- 2 General-purpose GPNVM Bits
- One-by-one Lock Bit Programming
- Commands Protected by a Keyword
- Erase the Entire Flash
- Erase by Plane
- Erase by Sector
- Erase by Pages
- Possibility of Erasing before Programming
- Locking and Unlocking Operations
- ECC Single and Multiple Error Flags Report
- Possibility to Read the Calibration Bits

### 22.3 Product Dependencies

#### 22.3.1 Power Management

The Enhanced Embedded Flash Controller (EEFC) is continuously clocked. The Power Management Controller has no effect on its behavior.

#### 22.3.2 Interrupt Sources

The EEFC interrupt line is connected to the interrupt controller. Using the EEFC interrupt requires the interrupt controller to be programmed first. The EEFC interrupt is generated only on FRDY bit rising.

**Table 22-1. Peripheral IDs**

Instance	ID
EFC	6

## 22.4 Functional Description

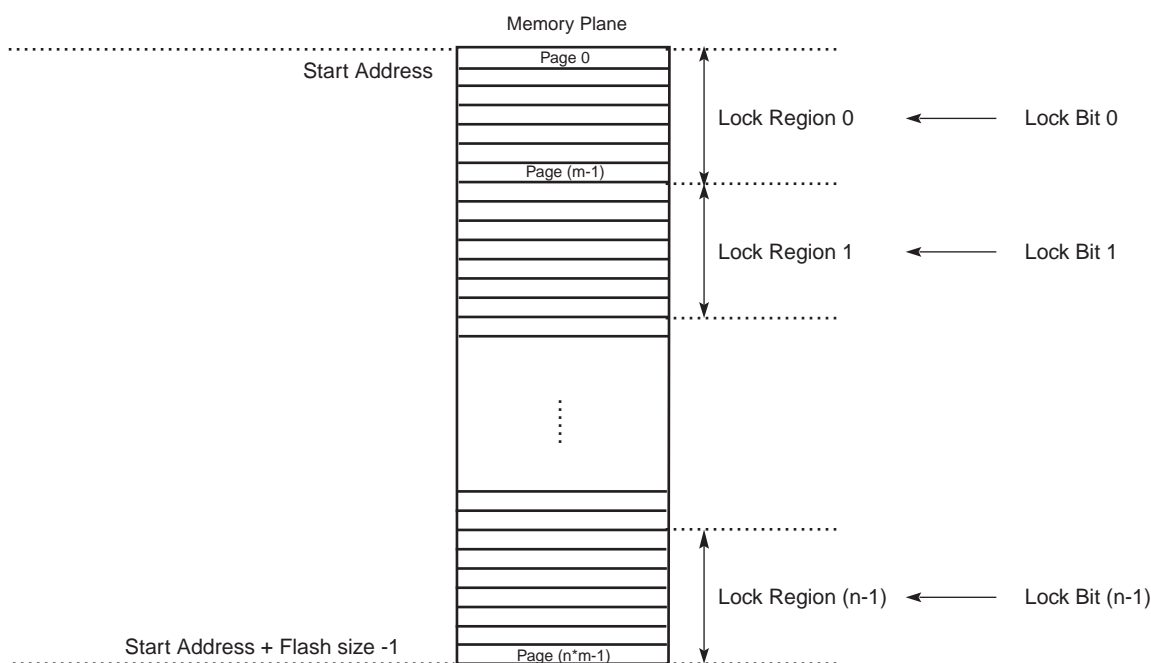
### 22.4.1 Embedded Flash Organization

The embedded Flash interfaces directly with the 32-bit internal bus. The embedded Flash is composed of:

- One memory plane organized in several pages of the same size.
- Two 128-bit or 64-bit read buffers used for code read optimization.
- One 128-bit or 64-bit read buffer used for data read optimization.
- One write buffer that manages page programming. The write buffer size is equal to the page size. This buffer is write-only and accessible all along the 1 MByte address space, so that each word can be written to its final address.
- Several lock bits used to protect write/erase operation on several pages (lock region). A lock bit is associated with a lock region composed of several pages in the memory plane.
- Several bits that may be set and cleared through the EEFC interface, called general-purpose non-volatile memory bits (GPNVM bits).

The embedded Flash size, the page size, the organization of lock regions and the definition of GPNVM bits are specific to the product. The EEFC returns a descriptor of the Flash controlled after a Get descriptor command issued by the application (see [“Get Flash Descriptor Command” on page 379](#)).

**Figure 22-1. Embedded Flash Organization**



## 22.4.2 Read Operations

An optimized controller manages embedded Flash reads, thus increasing performance when the processor is running in Thumb-2 mode by means of the 128- or 64-bit-wide memory interface.

The Flash memory is accessible through 8-, 16- and 32-bit reads.

As the Flash block size is smaller than the address space reserved for the internal memory area, the embedded Flash wraps around the address space and appears to be repeated within it.

The read operations can be performed with or without wait states. Wait states must be programmed in the field FWS (Flash Read Wait State) in the Flash Mode register (EEFC\_FMR). Defining FWS as 0 enables the single-cycle access of the embedded Flash. Refer to the Electrical Characteristics section for more details.

### 22.4.2.1 128-bit or 64-bit Access Mode

By default, the read accesses of the Flash are performed through a 128-bit wide memory interface. It improves system performance especially when 2 or 3 wait states are needed.

For systems requiring only 1 wait state, or to focus on current consumption rather than performance, the user can select a 64-bit wide memory access via the FAM bit in EEFC\_FMR.

Refer to the Electrical Characteristics section for more details.

### 22.4.2.2 Code Read Optimization

Code read optimization is enabled if the SCOD bit in EEFC\_FMR is cleared.

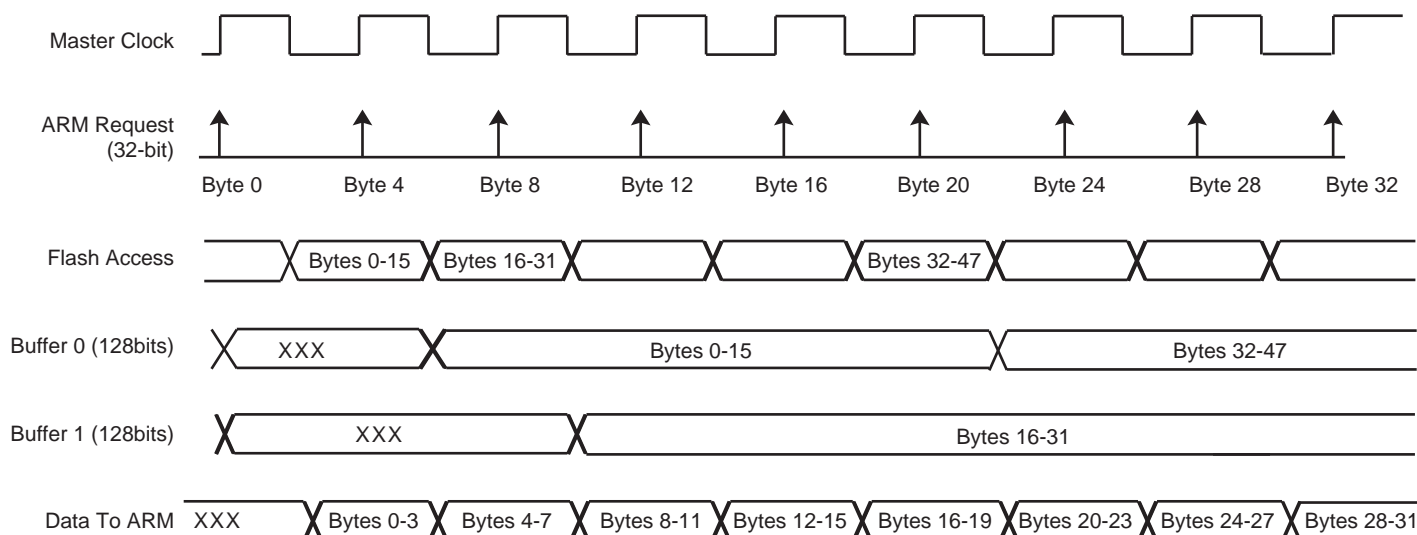
A system of 2 x 128-bit or 2 x 64-bit buffers is added in order to optimize sequential code fetch.

Note: Immediate consecutive code read accesses are not mandatory to benefit from this optimization.

The sequential code read optimization is enabled by default. If the SCOD bit in EEFC\_FMR is set to 1, these buffers are disabled and the sequential code read is no longer optimized.

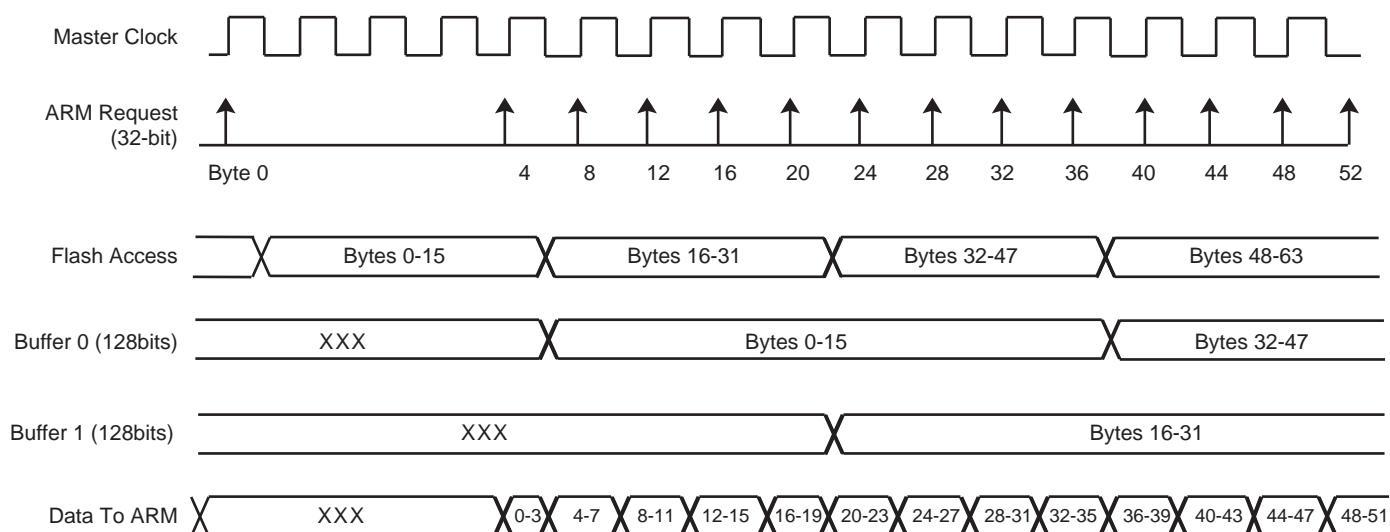
Another system of 2 x 128-bit or 2 x 64-bit buffers is added in order to optimize loop code fetch. Refer to [“Code Loop Optimization” on page 376](#) for more details.

**Figure 22-2. Code Read Optimization for FWS = 0**



Note: When FWS is equal to 0, all the accesses are performed in a single-cycle access.

**Figure 22-3. Code Read Optimization for FWS = 3**



**Note:** When FWS is included between 1 and 3, in case of sequential reads, the first access takes (FWS+1) cycles, the other ones only 1 cycle.

#### 22.4.2.3 Code Loop Optimization

Code loop optimization is enabled when the CLOE bit of EEFC\_FMR is set to 1.

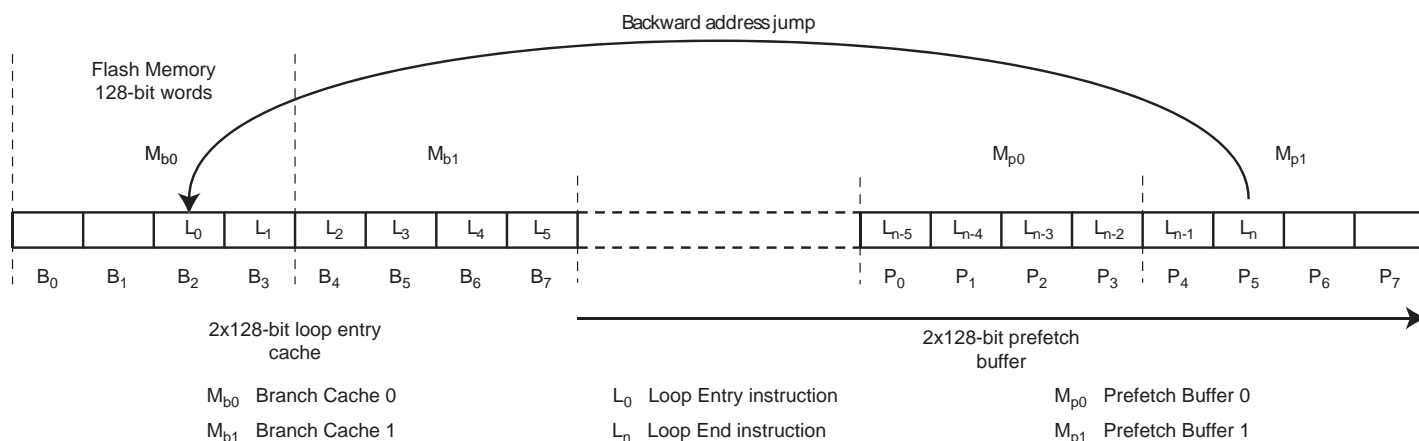
When a backward jump is inserted in the code, the pipeline of the sequential optimization is broken and becomes inefficient. In this case, the loop code read optimization takes over from the sequential code read optimization to prevent the insertion of wait states. The loop code read optimization is enabled by default. In EEFC\_FMR, if the bit CLOE is reset to 0 or the bit SCOD is set to 1, these buffers are disabled and the loop code read is not optimized.

When code loop optimization is enabled, if inner loop body instructions  $L_0$  to  $L_n$  are positioned from the 128-bit Flash memory cell  $M_{b0}$  to the memory cell  $M_{p1}$ , after recognition of a first backward branch, the first two Flash memory cells  $M_{b0}$  and  $M_{b1}$  targeted by this branch are cached for fast access from the processor at the next loop iteration.

Then by combining the sequential prefetch (described in [Section 22.4.2.2 "Code Read Optimization"](#)) through the loop body with the fast read access to the loop entry cache, the entire loop can be iterated with no wait state.

[Figure 22-4](#) illustrates code loop optimization.

**Figure 22-4. Code Loop Optimization**



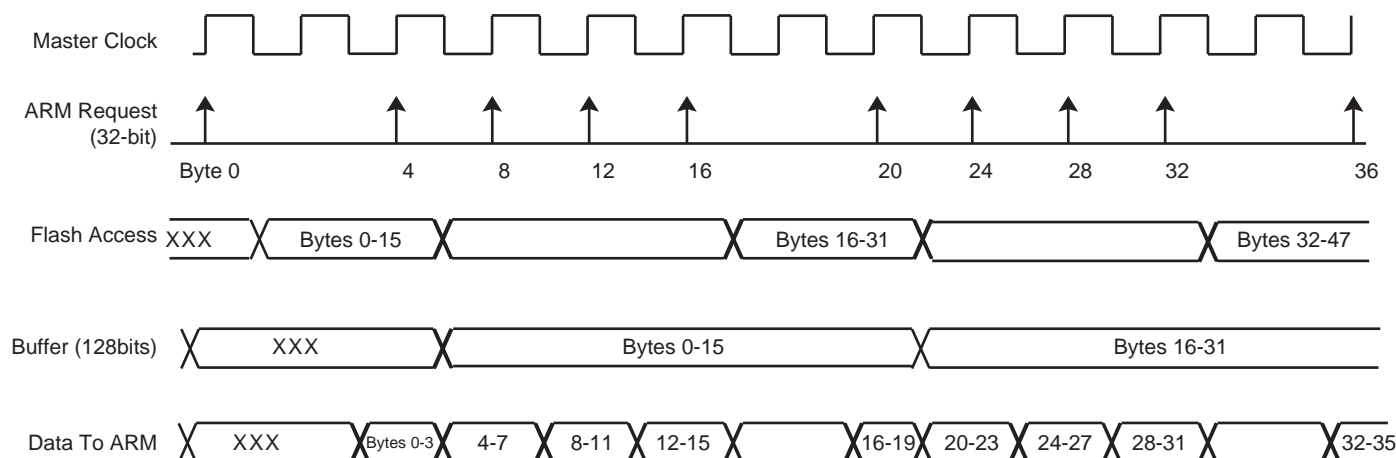


### 22.4.2.4 Data Read Optimization

The organization of the Flash in 128 bits (or 64 bits) is associated with two 128-bit (or 64-bit) prefetch buffers and one 128-bit (or 64-bit) data read buffer, thus providing maximum system performance. This buffer is added in order to store the requested data plus all the data contained in the 128-bit (64-bit) aligned data. This speeds up sequential data reads if, for example, FWS is equal to 1 (see Figure 22-5). The data read optimization is enabled by default. If the SCOD bit in EEFC\_FMR is set to 1, this buffer is disabled and the data read is no longer optimized.

Note: No consecutive data read accesses are mandatory to benefit from this optimization.

**Figure 22-5. Data Read Optimization for FWS = 1**



### 22.4.3 Flash Commands

The EEFC offers a set of commands to manage programming the Flash memory, locking and unlocking lock regions, consecutive programming, locking and full Flash erasing, etc.

**Table 22-2. Set of Commands**

Command	Value	Mnemonic
Get Flash descriptor	0x00	GETD
Write page	0x01	WP
Write page and lock	0x02	WPL
Erase page and write page	0x03	EWP
Erase page and write page then lock	0x04	EWPL
Erase all	0x05	EA
Erase pages	0x07	EPA
Set lock bit	0x08	SLB
Clear lock bit	0x09	CLB
Get lock bit	0x0A	GLB
Set GPNVM bit	0x0B	SGPB
Clear GPNVM bit	0x0C	CGPB
Get GPNVM bit	0x0D	GGPB
Start read unique identifier	0x0E	STUI

**Table 22-2. Set of Commands**

Command	Value	Mnemonic
Stop read unique identifier	0x0F	SPUI
Get CALIB bit	0x10	GCALB
Erase sector	0x11	ES
Write user signature	0x12	WUS
Erase user signature	0x13	EUS
Start read user signature	0x14	STUS
Stop read user signature	0x15	SPUS

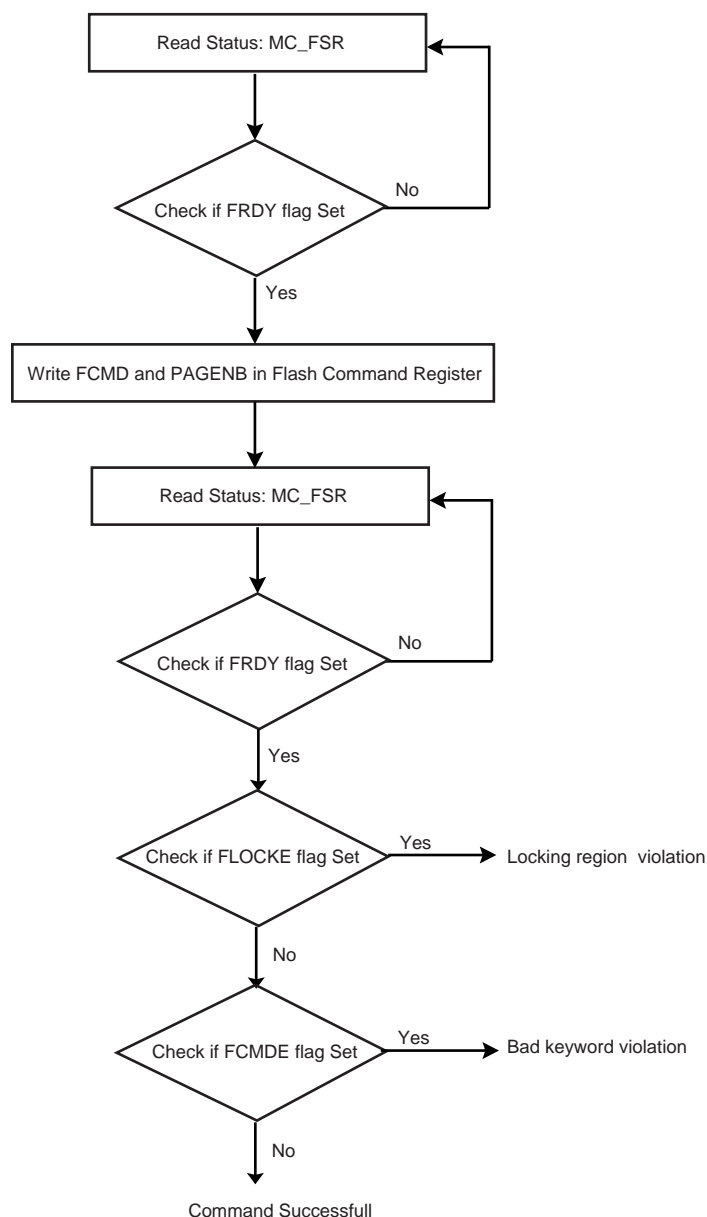
In order to perform one of these commands, the Flash Command register (EEFC\_FCR) must be written with the correct command using the FCMD field. As soon as EEFC\_FCR is written, the FRDY flag and the FVALUE field in the Flash Result register (EEFC\_FRR) are automatically cleared. Once the current command is achieved, then the FRDY flag is automatically set. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the corresponding interrupt line of the interrupt controller is activated. (Note that this is true for all commands except for the STUI Command. The FRDY flag is not set when the STUI command is achieved.)

All the commands are protected by the same keyword, which must be written in the eight highest bits of EEFC\_FCR.

Writing EEFC\_FCR with data that does not contain the correct key and/or with an invalid command has no effect on the whole memory plane, but the FCMDE flag is set in the Flash Status register (EEFC\_FSR). This flag is automatically cleared by a read access to EEFC\_FSR.

When the current command writes or erases a page in a locked region, the command has no effect on the whole memory plane, but the FLOCKE flag is set in EEFC\_FSR. This flag is automatically cleared by a read access to EEFC\_FSR.

**Figure 22-6. Command State Chart**



#### 22.4.3.1 Get Flash Descriptor Command

This command provides the system with information on the Flash organization. The system can take full advantage of this information. For instance, a device could be replaced by one with more Flash capacity, and so the software is able to adapt itself to the new configuration.

To get the embedded Flash descriptor, the application writes the GETD command in EEFC\_FCR. The first word of the descriptor can be read by the software application in EEFC\_FRR as soon as the FRDY flag in EEFC\_FSR rises. The next reads of EEFC\_FRR provide the following word of the descriptor. If extra read operations to EEFC\_FRR are done after the last word of the descriptor has been returned, then EEFC\_FRR value is 0 until the next valid command.

**Table 22-3. Flash Descriptor Definition**

Symbol	Word Index	Description
FL_ID	0	Flash interface description
FL_SIZE	1	Flash size in bytes
FL_PAGE_SIZE	2	Page size in bytes
FL_NB_PLANE	3	Number of planes.
FL_PLANE[0]	4	Number of bytes in the plane.
FL_NB_LOCK	4 + FL_NB_PLANE	Number of lock bits. A bit is associated with a lock region. A lock bit is used to prevent write or erase operations in the lock region.
FL_LOCK[0]	4 + FL_NB_PLANE + 1	Number of bytes in the first lock region.

#### 22.4.3.2 Write Commands

Several commands are used to program the Flash.

Only 0 values can be programmed using Flash technology; 1 is the erased value. In order to program words in a page, the page must first be erased. Commands are available to erase the full memory plane or a given number of pages. With the EWP and EWPL commands, a page erase is done automatically before a page programming.

After programming, the page (the entire lock region) can be locked to prevent miscellaneous write or erase sequences. The lock bit can be automatically set after page programming using WPL or EWPL commands.

Data to be programmed in the Flash must be written in an internal latch buffer before writing the programming command in EEFC\_FCR. Data can be written at their final destination address, as the latch buffer is mapped into the Flash memory address space and wraps around within this Flash address space.

Byte and half-word AHB accesses to the latch buffer are not allowed. Only 32-bit word accesses are supported.

32-bit words must be written continuously, in either ascending or descending order. Writing the latch buffer in a random order is not permitted. This prevents mapping a C-code structure to the latch buffer and accessing the data of the structure in any order. It is instead recommended to fill in a C-code structure in SRAM and copy it in the latch buffer in a continuous order.

Write operations in the latch buffer are performed with the number of wait states programmed for reading the Flash.

The latch buffer is automatically re-initialized, i.e., written with logical 1, after execution of each programming command.

The programming sequence is as follows:

- Write the data to be programmed in the latch buffer.
- Write the programming command in EEFC\_FCR. This will automatically clear the FRDY bit in EEFC\_TSR.
- When Flash programming is completed, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the bit FRDY in EEFC\_FMR, the interrupt line of the EEFC is activated.

Three errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Lock Error: the page to be programmed belongs to a locked region. A command must be run previously to unlock the corresponding region.
- Flash Error: when programming is completed, the WriteVerify test of the Flash memory has failed.

Only one page can be programmed at a time. It is possible to program all the bits of a page (full page programming) or only some of the bits of the page (partial page programming).

Depending on the number of bits to be programmed within the page, the EEFC adapts the write operations required to program the Flash.

When the Programming Page command is given, the EEFC starts the programming sequence and all the bits written at 0 in the latch buffer are cleared in the Flash memory array.

During programming, i.e. until FDRY rises, access to the Flash is not allowed.

#### *Full Page Programming*

To program a full page, all the bits of the page must be erased before writing the latch buffer and launching the WP command. The latch buffer must be written in ascending order, starting from the first address of the page. See [Figure 22-7, "Full Page Programming"](#).

#### *Partial Page Programming*

To program only part of a page using the WP command, the following constraints must be respected:

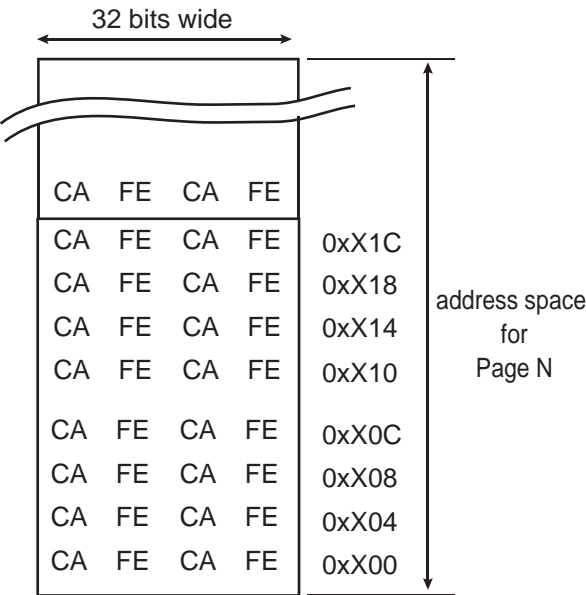
- Data to be programmed must be contained in integer multiples of 64-bit address-aligned words.
- 64-bit words can be programmed only if all the corresponding bits in the Flash array are erased (at logical value 1).

See [Figure 22-8, "Partial Page Programming"](#)

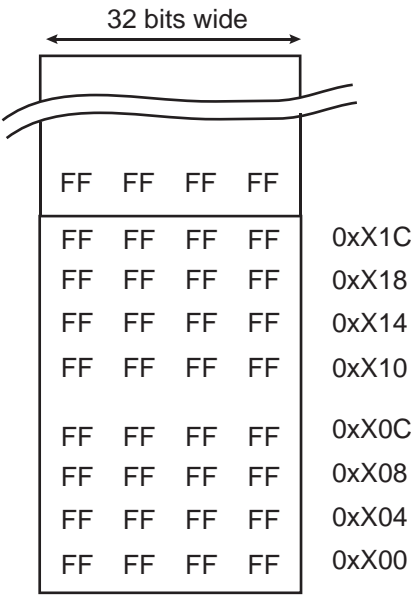
#### *Programming Bytes*

Individual bytes can be programmed using the partial page programming mode. In this case, an area of 64 bits must be reserved for each byte, as shown in [Figure 22-9, "Programming Bytes in the Flash"](#).

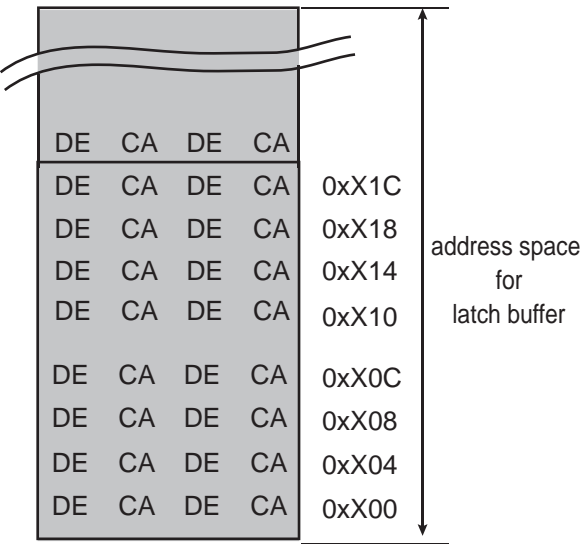
Figure 22-7. Full Page Programming



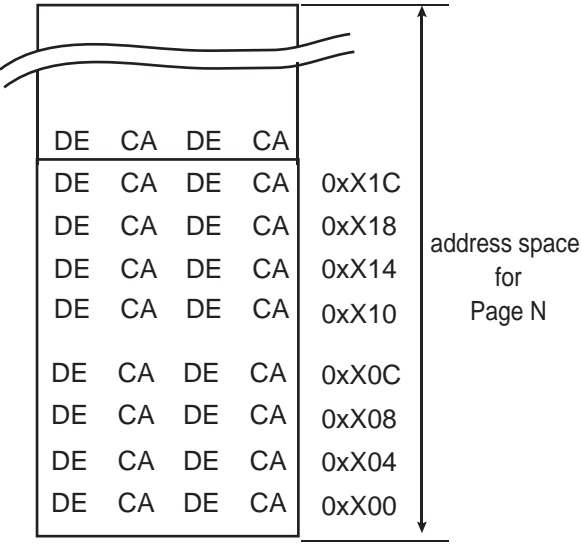
Before programming: Unerased page in Flash array



Step 1: Flash array after page erase

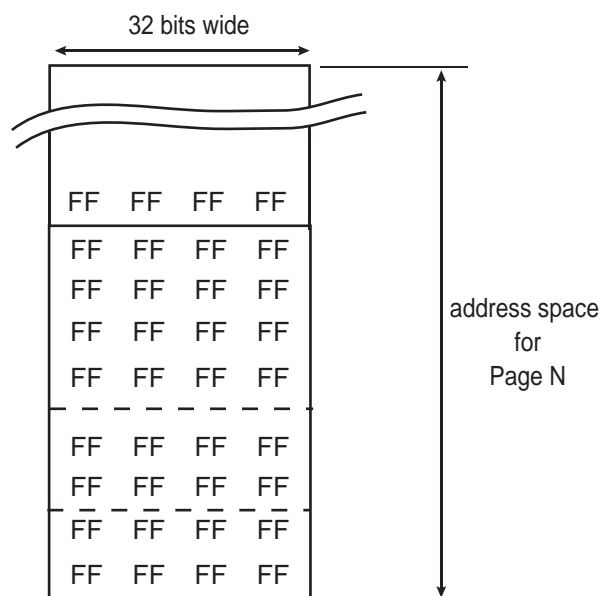


Step 2: Writing a page in the latch buffer

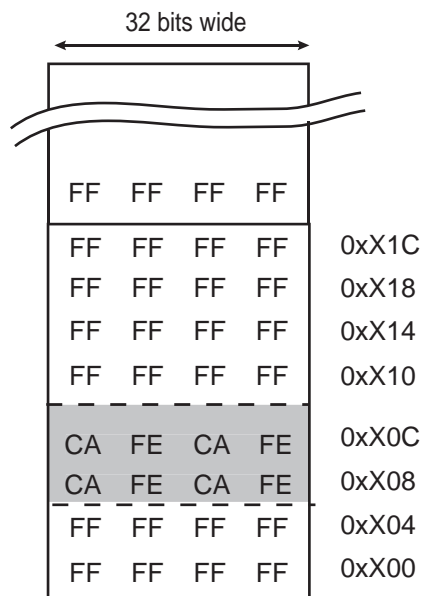


Step 3: Page in Flash array after issuing WP command and FRDY=1

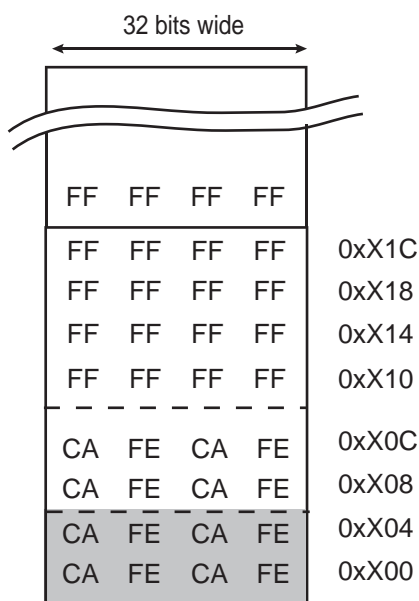
### Figure 22-8. Partial Page Programming



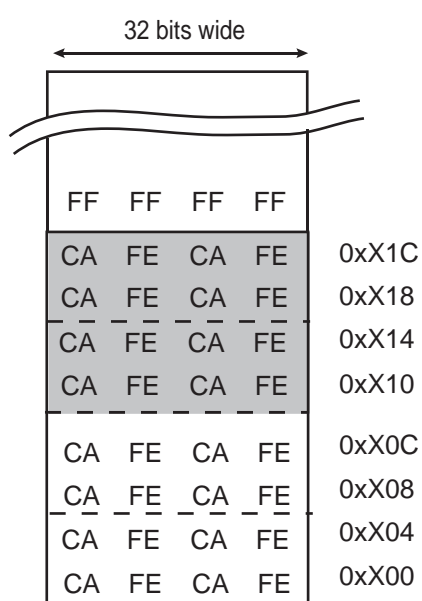
### Step 1: Flash array after page erase



Step 2: Flash array after programming  
64-bit at address 0xX08 (write latch buffer + WP)

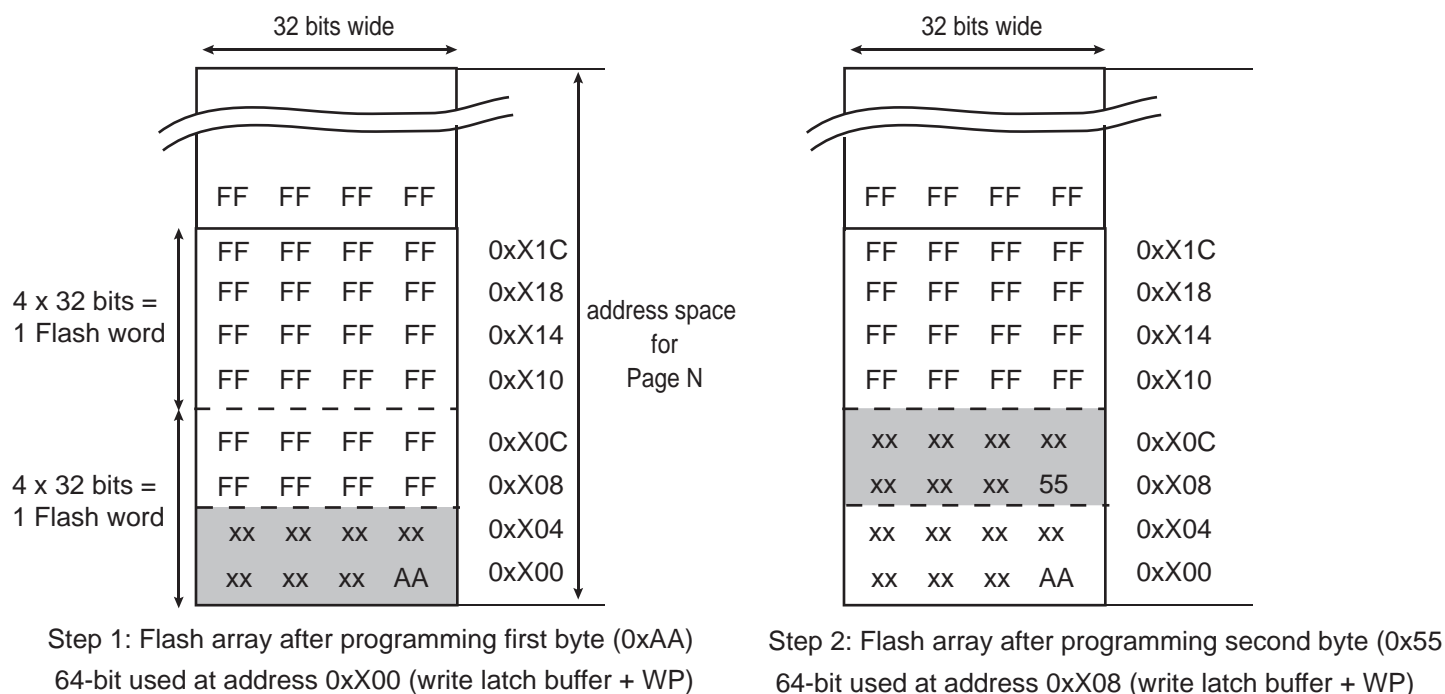


Step 3: Flash array after programming a second 64-bit data at address 0xX00 (write latch buffer + WP)



Step 4: Flash array after programming a 128-bit data word at address 0xX10 (write latch buffer + WP)

**Figure 22-9. Programming Bytes in the Flash**



#### 22.4.3.3 Erase Commands

Erase commands are allowed only on unlocked regions. Depending on the Flash memory, several commands can be used to erase the Flash:

- Erase all memory (EA): all memory is erased. The processor must not fetch code from the Flash memory.
- Erase pages (EPA): 8 or 16 pages are erased in the Flash sector selected. The first page to be erased is specified in the FARG[15:2] field of the MC\_FCR. The first page number must be modulo 8, 16 or 32 depending on the number of pages to erase at the same time.
- Erase sector (ES): a full memory sector is erased. Sector size depends on the Flash memory. FARG must be set with a page number that is in the sector to be erased.

If the processor is fetching code from the Flash memory while the EPA or ES command is being performed, the processor accesses will be stalled until the EPA command is completed. To avoid stalling the processor, the code can be run out of internal SRAM.

The erase sequence is:

- Erase starts as soon as one of the erase commands and the FARG field are written in EEFC\_FCR.
  - For the EPA command, the two lowest bits of the FARG field define the number of pages to be erased (FARG[1:0]):

**Table 22-4. FARG Field for EPA Command**

FARG[1:0]	Number of pages to be erased with EPA command
0	4 pages (only valid for small 8 KB sectors)
1	8 pages
2	16 pages
3	32 pages (not valid for small 8 KB sectors)



- When programming is completed, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.

Three errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Lock Error: at least one page to be erased belongs to a locked region. The erase command has been refused, no page has been erased. A command must be run previously to unlock the corresponding region.
- Flash Error: at the end of the programming, the EraseVerify test of the Flash memory has failed.

#### 22.4.3.4 Lock Bit Protection

Lock bits are associated with several pages in the embedded Flash memory plane. This defines lock regions in the embedded Flash memory plane. They prevent writing/erasing protected pages.

The lock sequence is:

- The Set lock bit command (SLB) and a page number to be protected are written in EEFC\_FCR.
- When the locking completes, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.
- The result of the SLB command can be checked running a Get Lock Bit (GLB) command.

Note: The value of the FARG argument passed together with SLB command must not exceed the higher lock bit index available in the product.

Two errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the EraseVerify or WriteVerify test of the Flash memory has failed.

It is possible to clear lock bits previously set. Then the locked region can be erased or programmed. The unlock sequence is:

- The Clear lock bit command (CLB) and a page number to be unprotected are written in EEFC\_FCR.
- When the unlock completes, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.

Note: The value of the FARG argument passed together with CLB command must not exceed the higher lock bit index available in the product.

Two errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the EraseVerify or WriteVerify test of the Flash memory has failed.

The status of lock bits can be returned by the EEFC. The Get lock bit status sequence is:

- The Get lock bit command (GLB) is written in EEFC\_FCR, FARG field is meaningless.
- Lock bits can be read by the software application in EEFC\_FRR. The first word read corresponds to the 32 first lock bits, next reads providing the next 32 lock bits as long as it is meaningful. Extra reads to EEFC\_FRR return 0.

For example, if the third bit of the first word read in EEFC\_FRR is set, then the third lock region is locked.

Two errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR
- Flash Error: at the end of the programming, the EraseVerify or WriteVerify test of the Flash memory has failed.

Note: Access to the Flash in read is permitted when a set, clear or get lock bit command is performed.

#### 22.4.3.5 GPNVM Bit

GPNVM bits do not interfere with the embedded Flash memory plane. Refer to specific product details for information on GPNVM bit action.

The Set GPNVM bit sequence is:

- Start the Set GPNVM bit command (SGPB) by writing EEFC\_FCR with the SGPB command and the number of the GPNVM bits to be set.

- When the GPNVM bit is set, the bit FRDY in EEFC\_FSR rises. If an interrupt was enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.
- The result of the SGPB command can be checked by running a Get GPNVM bit (GGPB) command.

Note: The value of the FARG argument passed together with SGPB command must not exceed the higher GPNVM index available in the product. Flash data content is not altered if FARG exceeds the limit. Command Error is detected only if FARG is greater than 8.

Two errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the EraseVerify or WriteVerify test of the Flash memory has failed.

It is possible to clear GPNVM bits previously set. The Clear GPNVM bit sequence is:

- Start the Clear GPNVM bit command (CGPB) by writing EEFC\_FCR with CGPB and the number of the GPNVM bits to be cleared.
- When the clear completes, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.

Note: The value of the FARG argument passed together with CGPB command must not exceed the higher GPNVM index available in the product. Flash data content is not altered if FARG exceeds the limit. Command Error is detected only if FARG is greater than 8.

Two errors can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the EraseVerify or WriteVerify test of the Flash memory has failed.

The status of GPNVM bits can be returned by the EEFC. The sequence is:

- Start the Get GPNVM bit command by writing EEFC\_FCR with GGPB. The FARG field is meaningless.
- GPNVM bits can be read by the software application in EEFC\_FRR. The first word read corresponds to the 32 first GPNVM bits, following reads provide the next 32 GPNVM bits as long as it is meaningful. Extra reads to EEFC\_FRR return 0.

For example, if the third bit of the first word read in EEFC\_FRR is set, then the third GPNVM bit is active.

One error can be detected in EEFC\_FSR after a programming sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.

Note: Access to the Flash in read is permitted when a set, clear or get GPNVM bit command is performed.

#### 22.4.3.6 Calibration Bit

Calibration bits do not interfere with the embedded Flash memory plane.

The calibration bits cannot be modified.

The status of calibration bits are returned by the EEFC. The sequence is:

- Issue the Get CALIB bit command by writing EEFC\_FCR with GCALB (see [Table 22-2](#)). The FARG field is meaningless.
- Calibration bits can be read by the software application in EEFC\_FRR. The first word read corresponds to the first 32 calibration bits. The following reads provide the next 32 calibration bits as long as it is meaningful. Extra reads to EEFC\_FRR return 0.

The 4/8/12 MHz Fast RC oscillator is calibrated in production. This calibration can be read through the Get CALIB bit command. The table below shows the bit implementation for each frequency:

**Table 22-5. Calibration Bit Indexes**

RC Calibration Frequency	EEFC_FRR Bits
8 MHz output	[28 - 22]
12 MHz output	[38 - 32]

The RC calibration for the 4 MHz is set to '1000000'.

#### 22.4.3.7 Security Bit Protection

When the security is enabled, access to the Flash, either through the JTAG/SWD interface or through the Fast Flash Programming interface, is forbidden. This ensures the confidentiality of the code programmed in the Flash.

The security bit is GPNVM0.

Disabling the security bit can only be achieved by asserting the ERASE pin at 1, and after a full Flash erase is performed. When the security bit is deactivated, all accesses to the Flash are permitted.

#### 22.4.3.8 Unique Identifier

Each part is programmed with a 2x512-bytes unique identifier. It can be used to generate keys for example. To read the unique identifier, the sequence is:

- Send the Start read unique identifier command (STUI) by writing EEFC\_FCR with the STUI command.
- When the unique identifier is ready to be read, the FRDY bit in EEFC\_FSR falls.
- The unique identifier is located in the first 128 bits of the Flash memory mapping, thus, at the address 0x1000000-0x10003FF.
- To stop the unique identifier mode, the user needs to send the Stop read unique identifier command (SPUI) by writing EEFC\_FCR with the SPUI command.
- When the SPUI command has been performed, the FRDY bit in EEFC\_FSR rises. If an interrupt was enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.

Note that during the sequence, the software cannot run out of Flash.

#### 22.4.3.9 User Signature

Each part contains a user signature of 512 bytes. It can be used for storage. Read, write and erase of this area is allowed.

To read the user signature, the sequence is as follows:

- Send the Start read user signature command (STUS) by writing EEFC\_FCR with the STUS command.
- When the user signature is ready to be read, the FRDY bit in EEFC\_FSR falls.
- The user signature is located in the first 512 bytes of the Flash memory mapping, thus, at the address 0x1000000-0x10001FF.
- To stop the user signature mode, the user needs to send the Stop read user signature command (SPUS) by writing EEFC\_FCR with the SPUS command.
- When the SPUI command has been performed, the FRDY bit in EEFC\_FSR rises. If an interrupt was enabled by setting the FRDY bit in EEFC\_FMR, the interrupt line of the interrupt controller is activated.

Note that during the sequence, the software cannot run out of Flash or the second plane in case of dual plane.

One error can be detected in EEFC\_FSR after this sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.

To write the user signature, the sequence is:

- Write the full page, at any page address, within the internal memory area address space.
- Send the Write user signature command (WUS) by writing EEFC\_FCR with the WUS command.
- When programming is completed, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the corresponding interrupt line of the interrupt controller is activated.

Two errors can be detected in EEFC\_FSR after this sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the WriteVerify test of the Flash memory has failed.

To erase the user signature, the sequence is:

- Send the Erase user signature command (EUS) by writing EEFC\_FCR with the EUS command.

- When programming is completed, the FRDY bit in EEFC\_FSR rises. If an interrupt has been enabled by setting the FRDY bit in EEFC\_FMR, the corresponding interrupt line of the interrupt controller is activated.

Two errors can be detected in EEFC\_FSR after this sequence:

- Command Error: a bad keyword has been written in EEFC\_FCR.
- Flash Error: at the end of the programming, the EraseVerify test of the Flash memory has failed.

#### 22.4.3.10 ECC Errors and Corrections

The Flash embeds an ECC module able to correct one unique error and able to detect two errors. The errors are detected while a read access is performed into memory array and stored in EEFC\_FSR (see [Section 22.5.3 “EEFC Flash Status Register” on page 393](#)). The error report is kept until EEFC\_FSR is read.

There is one flag for a unique error on lower half part of the Flash word (64 LSB) and one flag for the upper half part (MSB). The multiple errors are reported in the same way.

Due to the anticipation mechanism to improve bandwidth throughput on instruction fetch, a reported error can be located in the next sequential Flash word compared to the location of the instruction being executed, which is located in the previously fetched Flash word.

If a software routine processes the error detection independently from the main software routine, the entire Flash located software must be rewritten because there is no storage of the error location.

If only a software routine is running to program and check pages by reading EEFC\_FSR, the situation differs from previous case. Performing a check for ECC unique errors just after page programming completion involves a read of the newly programmed page. This read sequence is viewed as data accesses and is not optimized by the Flash controller. Thus, in case of unique error, only the current page must be reprogrammed.

## 22.5 Enhanced Embedded Flash Controller (EEFC) User Interface

The User Interface of the Embedded Flash Controller (EEFC) is integrated within the System Controller with base address 0x400E0A00.

**Table 22-6. Register Mapping**

Offset	Register	Name	Access	Reset State
0x00	EEFC Flash Mode Register	EEFC_FMR	Read/Write	0x0400_0000
0x04	EEFC Flash Command Register	EEFC_FCR	Write-only	–
0x08	EEFC Flash Status Register	EEFC_FSR	Read-only	0x00000001
0x0C	EEFC Flash Result Register	EEFC_FRR	Read-only	0x0
0x10	Reserved	–	–	–

## 22.5.1 EEFC Flash Mode Register

**Name:** EEFC\_FMR  
**Address:** 0x400E0A00  
**Access:** Read/Write  
**Offset:** 0x00

31	30	29	28	27	26	25	24
–	–	–	–	–	CLOE	–	FAM
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	SCOD
15	14	13	12	11	10	9	8
–	–	–	–	FWS			
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	FRDY

- **FRDY: Ready Interrupt Enable**

0: Flash Ready does not generate an interrupt.

1: Flash Ready (to accept a new command) generates an interrupt.

- **FWS: Flash Wait State**

This field defines the number of wait states for read and write operations:

$$\text{Number of cycles for Read/Write operations} = \text{FWS} + 1$$

- **SCOD: Sequential Code Optimization Disable**

0: The sequential code optimization is enabled.

1: The sequential code optimization is disabled.

No Flash read should be done during change of this register.

- **FAM: Flash Access Mode**

0: 128-bit access in read mode only, to enhance access speed.

1: 64-bit access in read mode only, to enhance power consumption.

No Flash read should be done during change of this register.

- **CLOE: Code Loop Optimization Enable**

0: The opcode loop optimization is disabled.

1: The opcode loop optimization is enabled.

No Flash read should be done during change of this register.

## 22.5.2 EEFC Flash Command Register

**Name:** EEFC\_FCR  
**Address:** 0x400E0A04  
**Access:** Write-only  
**Offset:** 0x04

31	30	29	28	27	26	25	24
FKEY							
23	22	21	20	19	18	17	16
FARG							
15	14	13	12	11	10	9	8
FARG							
7	6	5	4	3	2	1	0
FCMD							

### • FCMD: Flash Command

Value	Name	Description
0x00	GETD	Get Flash descriptor
0x01	WP	Write page
0x02	WPL	Write page and lock
0x03	EWP	Erase page and write page
0x04	EWPL	Erase page and write page then lock
0x05	EA	Erase all
0x07	EPA	Erase pages
0x08	SLB	Set lock bit
0x09	CLB	Clear lock bit
0x0A	GLB	Get lock bit
0x0B	SGPB	Set GPNVM bit
0x0C	CGPB	Clear GPNVM bit
0x0D	GGPB	Get GPNVM bit
0x0E	STUI	Start read unique identifier
0x0F	SPUI	Stop read unique identifier
0x10	GCALB	Get CALIB bit
0x11	ES	Erase sector
0x12	WUS	Write user signature
0x13	EUS	Erase user signature
0x14	STUS	Start read user signature
0x15	SPUS	Stop read user signature

- **FARG: Flash Command Argument**

GETD, GLB, GGPB, STUI, SPUI, GCALB, WUS, EUS, STUS, SPUS, EA	Commands requiring no argument, including Erase all command	FARG is meaningless, must be written with 0
ES	Erase sector command	FARG must be written with any page number within the sector to be erased
EPA	Erase pages command	<p>FARG[1:0] defines the number of pages to be erased The start page must be written in FARG[15:2].</p> <p>FARG[1:0] = 0: Four pages to be erased. FARG[15:2] = Page_Number Modulo 4 FARG[1:0] = 1: Eight pages to be erased. FARG[15:2] = Page_Number Modulo 8 FARG[1:0] = 2: Sixteen pages to be erased. FARG[15:2] = Page_Number Modulo 16 FARG[1:0] = 3: Thirty-two pages to be erased. FARG[15:2] = Page_Number Modulo 32</p> <p>Refer to <a href="#">Table 22-4 on page 384</a>.</p>
WP, WPL, EWP, EWPL	Programming commands	FARG must be written with the page number to be programmed
SLB, CLB	Lock bit commands	FARG defines the page number to be locked or unlocked
SGPB, CGPB	GPNVM commands	FARG defines the GPNVM number to be programmed

- **FKEY: Flash Writing Protection Key**

Value	Name	Description
0x5A	PASSWD	The 0x5A value enables the command defined by the bits of the register. If the field is written with a different value, the write is not performed and no action is started.



### 22.5.3 EEFC Flash Status Register

**Name:** EEFC\_FSR  
**Address:** 0x400E0A08  
**Access:** Read-only  
**Offset:** 0x08

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	MECCMSB	UECCMSB	MECCLSB	UECCLSB
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	FLERR	FLOCKE	FCMDE	FRDY

- **FRDY: Flash Ready Status**

0: The EEFC is busy.

1: The EEFC is ready to start a new command.

When set, this flag triggers an interrupt if the FRDY flag is set in EEFC\_FMR.

This flag is automatically cleared when the EEFC is busy.

- **FCMDE: Flash Command Error Status**

0: No invalid commands and no bad keywords were written in EEFC\_FMR.

1: An invalid command and/or a bad keyword was/were written in EEFC\_FMR.

This flag is automatically cleared when EEFC\_FSR is read or EEFC\_FCR is written.

- **FLOCKE: Flash Lock Error Status**

0: No programming/erase of at least one locked region has happened since the last read of EEFC\_FSR.

1: Programming/erase of at least one locked region has happened since the last read of EEFC\_FSR.

This flag is automatically cleared when EEFC\_FSR is read or EEFC\_FCR is written.

- **FLERR: Flash Error Status**

0: No Flash memory error occurred at the end of programming (EraseVerify or WriteVerify test has passed).

1: A Flash memory error occurred at the end of programming (EraseVerify or WriteVerify test has failed).

- **UECCLSB: Unique ECC Error on LSB Part of the Memory Flash Data Bus**

0: No unique error detected on 64 LSB data bus of the Flash memory since the last read of EEFC\_FSR.

1: One unique error detected but corrected on 64 LSB data bus of the Flash memory since the last read of EEFC\_FSR.

- **MECCLSB: Multiple ECC Error on LSB Part of the Memory Flash Data Bus**

0: No multiple error detected on 64 LSB part of the Flash memory data bus since the last read of EEFC\_FSR.

1: Multiple errors detected and NOT corrected on 64 LSB part of the Flash memory data bus since the last read of EEFC\_FSR.

- **UECCMSB: Unique ECC Error on MSB Part of the Memory Flash Data Bus**

0: No unique error detected on 64 MSB data bus of the Flash memory since the last read of EEFC\_FSR.

1: One unique error detected but corrected on 64 MSB data bus of the Flash memory since the last read of EEFC\_FSR.

- **MECCEMSB: Multiple ECC Error on MSB Part of the Memory Flash Data Bus**

0: No multiple error detected on 64 MSB part of the Flash memory data bus since the last read of EEFC\_FSR.

1: Multiple errors detected and NOT corrected on 64 MSB part of the Flash memory data bus since the last read of EEFC\_FSR.

## 22.5.4 EEFC Flash Result Register

**Name:** EEFC\_FRR  
**Address:** 0x400E0A0C  
**Access:** Read-only  
**Offset:** 0x0C

31	30	29	28	27	26	25	24
FVALUE							
23	22	21	20	19	18	17	16
FVALUE							
15	14	13	12	11	10	9	8
FVALUE							
7	6	5	4	3	2	1	0
FVALUE							

- **FVALUE: Flash Result Value**

The result of a Flash command is returned in this register. If the size of the result is greater than 32 bits, then the next resulting value is accessible at the next register read.

## 23. Fast Flash Programming Interface (FFPI)

### 23.1 Description

The Fast Flash Programming Interface (FFPI) provides parallel high-volume programming using a standard gang programmer. The parallel interface is fully handshaked and the device is considered to be a standard EEPROM. Additionally, the parallel protocol offers an optimized access to all the embedded Flash functionalities.

Although the Fast Flash Programming Mode is a dedicated mode for high volume programming, this mode is not designed for in-situ programming.

### 23.2 Embedded Characteristics

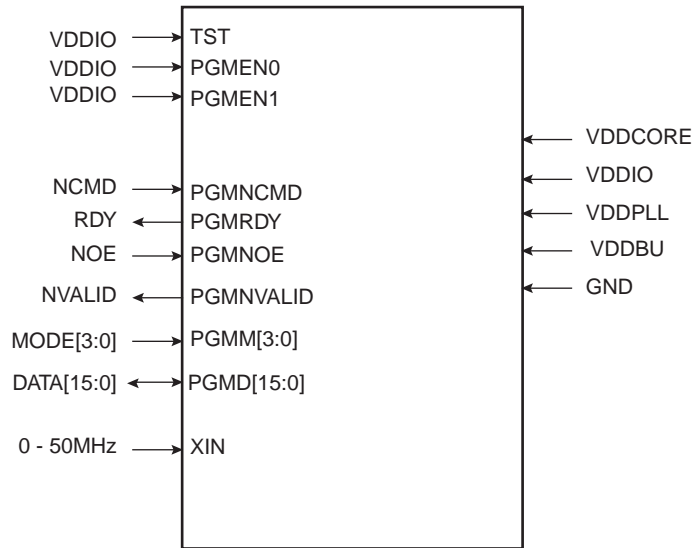
- Programming Mode for High-volume Flash Programming Using Gang Programmer
  - Offers Read and Write Access to the Flash Memory Plane
  - Enables Control of Lock Bits and General-purpose NVM Bits
  - Enables Security Bit Activation
  - Disabled Once Security Bit is Set
- Parallel Fast Flash Programming Interface
  - Provides an 16-bit Parallel Interface to Program the Embedded Flash
  - Full Handshake Protocol

## 23.3 Parallel Fast Flash Programming

### 23.3.1 Device Configuration

In Fast Flash Programming Mode, the device is in a specific test mode. Only a certain set of pins is significant. The rest of the PIOs are used as inputs with a pull-up. The crystal oscillator is in bypass mode. Other pins must be left unconnected.

Figure 23-1. Parallel Programming Interface



### 23.3.2 Signal Names

**Table 23-1. Signal Description List**

Signal Name	Function	Type	Active Level	Comments
<b>Power</b>				
VDDIO	I/O Lines Power Supply	Power		
VDDCORE	Core Power Supply	Power		
VDDPLL	PLL Power Supply	Power		
GND	Ground	Ground		
<b>Clocks</b>				
XIN	Main Clock Input. This input can be tied to GND. In this case, the device is clocked by the internal RC oscillator.	Input		32 kHz to 50 MHz
<b>Test</b>				
TST	Test Mode Select	Input	High	Must be connected to VDDIO
PGMEN0	Test Mode Select	Input	High	Must be connected to VDDIO
PGMEN1	Test Mode Select	Input	High	Must be connected to VDDIO
<b>PIO</b>				
PGMNCMD	Valid command available	Input	Low	Pulled-up input at reset
PGMRDY	0: Device is busy 1: Device is ready for a new command	Output	High	Pulled-up input at reset
PGMNOE	Output Enable (active high)	Input	Low	Pulled-up input at reset
PGMNVAILD	0: DATA[15:0] is in input mode 1: DATA[15:0] is in output mode	Output	Low	Pulled-up input at reset
PGMM[3:0]	Specifies DATA type (see <a href="#">Table 23-2</a> )	Input		Pulled-up input at reset
PGMD[15:0]	Bi-directional data bus	Input/Output		Pulled-up input at reset

Depending on the MODE settings, DATA is latched in different internal registers.

**Table 23-2. Mode Coding**

MODE[3:0]	Symbol	Data
0000	CMDE	Command Register
0001	ADDR0	Address Register LSBs
0010	ADDR1	
0101	DATA	Data Register
Default	IDLE	No register

When MODE is equal to CMDE, then a new command (strobed on DATA[15:0] signals) is stored in the command register.

**Table 23-3. Command Bit Coding**

DATA[15:0]	Symbol	Command Executed
0x0011	READ	Read Flash
0x0012	WP	Write Page Flash
0x0022	WPL	Write Page and Lock Flash
0x0032	EWP	Erase Page and Write Page
0x0042	EWPL	Erase Page and Write Page then Lock
0x0013	EA	Erase All
0x0014	SLB	Set Lock Bit
0x0024	CLB	Clear Lock Bit
0x0015	GLB	Get Lock Bit
0x0034	SGPB	Set General Purpose NVM bit
0x0044	CGPB	Clear General Purpose NVM bit
0x0025	GGPB	Get General Purpose NVM bit
0x0054	SSE	Set Security Bit
0x0035	GSE	Get Security Bit
0x001F	WRAM	Write Memory
0x001E	GVE	Get Version

### 23.3.3 Entering Programming Mode

The following algorithm puts the device in Parallel Programming Mode:

- Apply the supplies as described in [Table 23-1](#).
- Apply XIN clock within  $T_{POR\_RESET}$  if an external clock is available.
- Wait for  $T_{POR\_RESET}$
- Start a read or write handshaking.

**Note:** After reset, the device is clocked by the internal RC oscillator. Before clearing RDY signal, if an external clock (>32 kHz) is connected to XIN, then the device switches on the external clock. Else, XIN input is not considered. A higher frequency on XIN speeds up the programmer handshake.

23.3.4 Programmer Handshaking

An handshake is defined for read and write operations. When the device is ready to start a new operation (RDY signal set), the programmer starts the handshake by clearing the NCMD signal. The handshaking is achieved once NCMD signal is high and RDY is high.

23.3.4.1 Write Handshaking

For details on the write handshaking sequence, refer to [Figure 23-2](#) and [Table 23-4](#).

Figure 23-2. Parallel Programming Timing, Write Sequence

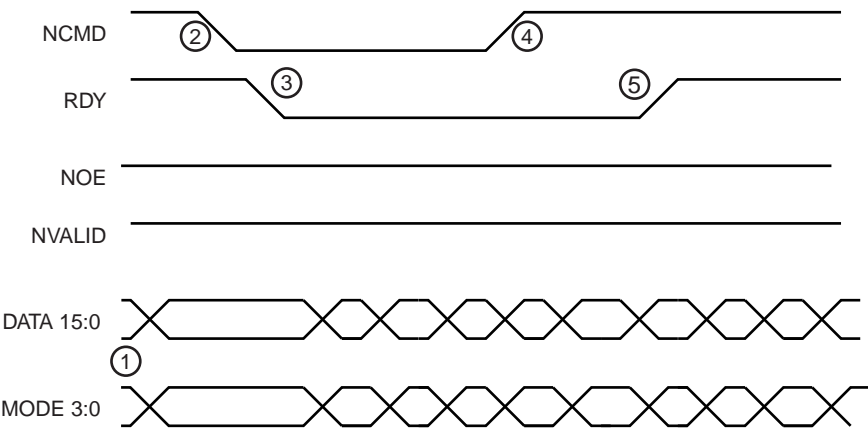


Table 23-4. Write Handshake

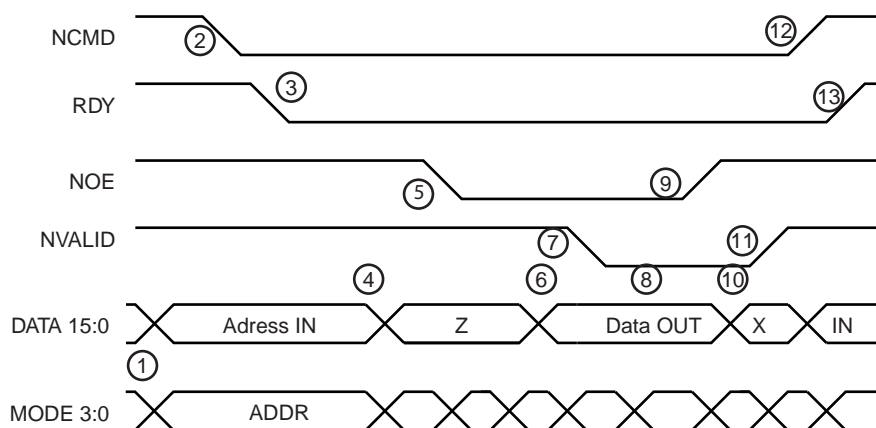
Step	Programmer Action	Device Action	Data I/O
1	Sets MODE and DATA signals	Waits for NCMD low	Input
2	Clears NCMD signal	Latches MODE and DATA	Input
3	Waits for RDY low	Clears RDY signal	Input
4	Releases MODE and DATA signals	Executes command and polls NCMD high	Input
5	Sets NCMD signal	Executes command and polls NCMD high	Input
6	Waits for RDY high	Sets RDY	Input



### 23.3.4.2 Read Handshaking

For details on the read handshaking sequence, refer to [Figure 23-3](#) and [Table 23-5](#).

**Figure 23-3. Parallel Programming Timing, Read Sequence**



**Table 23-5. Read Handshake**

Step	Programmer Action	Device Action	DATA I/O
1	Sets MODE and DATA signals	Waits for NCMD low	Input
2	Clears NCMD signal	Latch MODE and DATA	Input
3	Waits for RDY low	Clears RDY signal	Input
4	Sets DATA signal in tristate	Waits for NOE Low	Input
5	Clears NOE signal		Tristate
6	Waits for NVALID low	Sets DATA bus in output mode and outputs the flash contents.	Output
7		Clears NVALID signal	Output
8	Reads value on DATA Bus	Waits for NOE high	Output
9	Sets NOE signal		Output
10	Waits for NVALID high	Sets DATA bus in input mode	X
11	Sets DATA in output mode	Sets NVALID signal	Input
12	Sets NCMD signal	Waits for NCMD high	Input
13	Waits for RDY high	Sets RDY signal	Input

### 23.3.5 Device Operations

Several commands on the Flash memory are available. These commands are summarized in [Table 23-3 on page 399](#). Each command is driven by the programmer through the parallel interface running several read/write handshaking sequences.

When a new command is executed, the previous one is automatically achieved. Thus, chaining a read command after a write automatically flushes the load buffer in the Flash.

#### 23.3.5.1 Flash Read Command

This command is used to read the contents of the Flash memory. The read command can start at any valid address in the memory plane and is optimized for consecutive reads. Read handshaking can be chained; an internal address buffer is automatically increased.

**Table 23-6. Read Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	READ
2	Write handshaking	ADDR0	Memory Address LSB
3	Write handshaking	ADDR1	Memory Address
4	Read handshaking	DATA	*Memory Address++
5	Read handshaking	DATA	*Memory Address++
...	...	...	...
n	Write handshaking	ADDR0	Memory Address LSB
n+1	Write handshaking	ADDR1	Memory Address
n+2	Read handshaking	DATA	*Memory Address++
n+3	Read handshaking	DATA	*Memory Address++
...	...	...	...

#### 23.3.5.2 Flash Write Command

This command is used to write the Flash contents.

The Flash memory plane is organized into several pages. Data to be written are stored in a load buffer that corresponds to a Flash memory page. The load buffer is automatically flushed to the Flash:

- before access to any page other than the current one
- when a new command is validated (MODE = CMDE)

The Write Page command (WP) is optimized for consecutive writes. Write handshaking can be chained; an internal address buffer is automatically increased.

**Table 23-7. Write Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	WP or WPL or EWP or EWPL
2	Write handshaking	ADDR0	Memory Address LSB
3	Write handshaking	ADDR1	Memory Address
4	Write handshaking	DATA	*Memory Address++
5	Write handshaking	DATA	*Memory Address++
...	...	...	...
n	Write handshaking	ADDR0	Memory Address LSB

**Table 23-7. Write Command (Continued)**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
n+1	Write handshaking	ADDR1	Memory Address
n+2	Write handshaking	DATA	*Memory Address++
n+3	Write handshaking	DATA	*Memory Address++
...	...	...	...

The Flash command Write Page and Lock (WPL) is equivalent to the Flash Write Command. However, the lock bit is automatically set at the end of the Flash write operation. As a lock region is composed of several pages, the programmer writes to the first pages of the lock region using Flash write commands and writes to the last page of the lock region using a Flash write and lock command.

The Flash command Erase Page and Write (EWP) is equivalent to the Flash Write Command. However, before programming the load buffer, the page is erased.

The Flash command Erase Page and Write the Lock (EWPL) combines EWP and WPL commands.

### 23.3.5.3 Flash Full Erase Command

This command is used to erase the Flash memory planes.

All lock regions must be unlocked before the Full Erase command by using the CLB command. Otherwise, the erase command is aborted and no page is erased.

**Table 23-8. Full Erase Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	EA
2	Write handshaking	DATA	0

### 23.3.5.4 Flash Lock Commands

Lock bits can be set using WPL or EWPL commands. They can also be set by using the **Set Lock** command (**SLB**). With this command, several lock bits can be activated. A Bit Mask is provided as argument to the command. When bit 0 of the bit mask is set, then the first lock bit is activated.

In the same way, the **Clear Lock** command (**CLB**) is used to clear lock bits.

**Table 23-9. Set and Clear Lock Bit Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	SLB or CLB
2	Write handshaking	DATA	Bit Mask

Lock bits can be read using Get Lock Bit command (GLB). The  $n^{\text{th}}$  lock bit is active when the bit  $n$  of the bit mask is set..

**Table 23-10. Get Lock Bit Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	GLB
2	Read handshaking	DATA	Lock Bit Mask Status 0 = Lock bit is cleared 1 = Lock bit is set

### 23.3.5.5 Flash General-purpose NVM Commands

General-purpose NVM bits (GP NVM bits) can be set using the Set GPNVM command (SGPB). This command also activates GP NVM bits. A bit mask is provided as argument to the command. When bit 0 of the bit mask is set, then the first GP NVM bit is activated.

In the same way, the Clear GPNVM command (CGPB) is used to clear general-purpose NVM bits. The general-purpose NVM bit is deactivated when the corresponding bit in the pattern value is set to 1.

**Table 23-11. Set/Clear GP NVM Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	SGPB or CGPB
2	Write handshaking	DATA	GP NVM bit pattern value

General-purpose NVM bits can be read using the **Get GPNVM Bit** command (**GGPB**). The  $n^{\text{th}}$  GP NVM bit is active when bit  $n$  of the bit mask is set.

**Table 23-12. Get GP NVM Bit Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	GGPB
2	Read handshaking	DATA	GP NVM Bit Mask Status 0 = GP NVM bit is cleared 1 = GP NVM bit is set

### 23.3.5.6 Flash Security Bit Command

A security bit can be set using the Set Security Bit command (SSE). Once the security bit is active, the Fast Flash programming is disabled. No other command can be run. An event on the Erase pin can erase the security bit once the contents of the Flash have been erased.

**Table 23-13. Set Security Bit Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	SSE
2	Write handshaking	DATA	0

Once the security bit is set, it is not possible to access FFPI. The only way to erase the security bit is to erase the Flash.

In order to erase the Flash, the user must perform the following:

- Power-off the chip
- Power-on the chip with TST = 0
- Assert Erase during a period of more than 220 ms
- Power-off the chip

Then it is possible to return to FFPI mode and check that Flash is erased.

### 23.3.5.7 Memory Write Command

This command is used to perform a write access to any memory location.

The Memory Write command (WRAM) is optimized for consecutive writes. Write handshaking can be chained; an internal address buffer is automatically increased.

**Table 23-14. Write Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	WRAM
2	Write handshaking	ADDR0	Memory Address LSB
3	Write handshaking	ADDR1	Memory Address
4	Write handshaking	DATA	*Memory Address++
5	Write handshaking	DATA	*Memory Address++
...	...	...	...
n	Write handshaking	ADDR0	Memory Address LSB
n+1	Write handshaking	ADDR1	Memory Address
n+2	Write handshaking	DATA	*Memory Address++
n+3	Write handshaking	DATA	*Memory Address++
...	...	...	...

### 23.3.5.8 Get Version Command

The Get Version (GVE) command retrieves the version of the FFPI interface.

**Table 23-15. Get Version Command**

Step	Handshake Sequence	MODE[3:0]	DATA[15:0]
1	Write handshaking	CMDE	GVE
2	Read handshaking	DATA	Version

## 24. Cortex M Cache Controller (CMCC)

### 24.1 Description

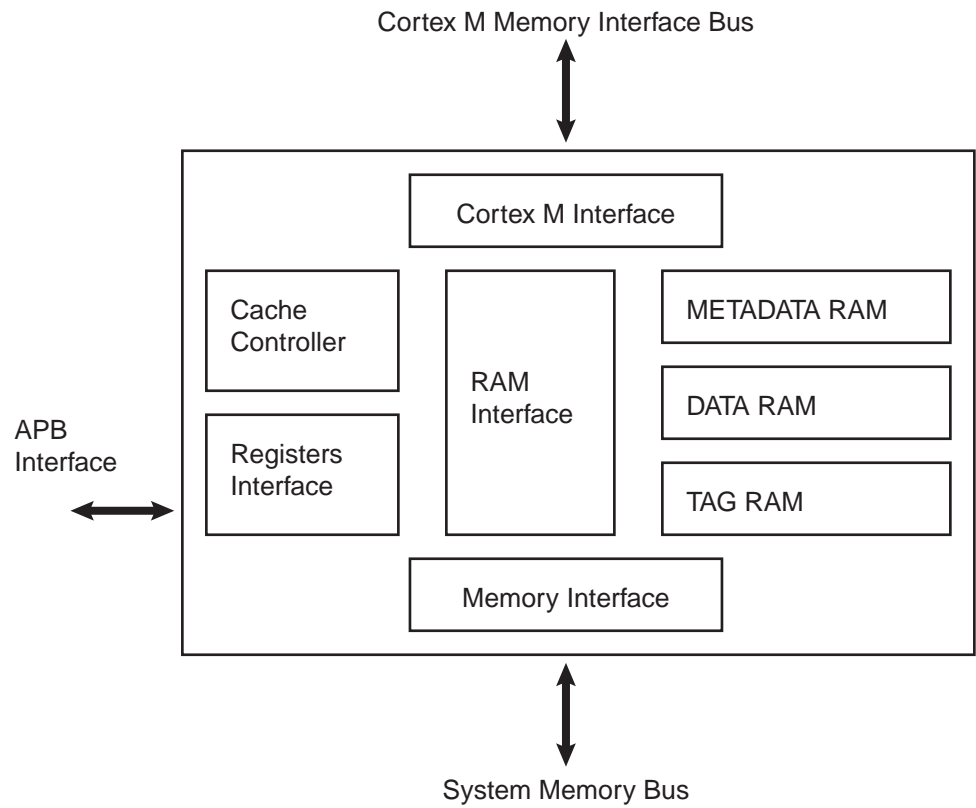
The Cortex M Cache Controller (CMCC) is a 4-Way set associative unified cache controller. It integrates a controller, a tag directory, data memory, metadata memory and a configuration interface.

### 24.2 Embedded Characteristics

- Physically addressed and physically tagged
- L1 data cache set to 2 Kbytes
- L1 cache line size set to 16 Bytes
- L1 cache integrates 32-bit bus master interface
- Unified Direct mapped cache architecture
- Unified 4-Way set associative cache architecture
- Write through cache operations, read allocate
- Round Robin victim selection policy
- Event Monitoring, with one programmable 32-bit counter
- Configuration registers accessible through Cortex M Private Peripheral Bus
- Cache Interface includes cache maintenance operations registers

### 24.3 Block Diagram

Figure 24-1. Block Diagram



## 24.4 Functional Description

### 24.4.1 Cache Operation

On reset, the cache controller data entries are all invalidated and the cache is enabled. The cache is transparent to processor operations. The cache controller is activated with its configuration registers. The configuration interface is memory mapped in the private peripheral bus.

The cache must always be enabled, even if the code is running out of a non-cached region.

When the cache is disabled, the accesses to the cache on its slave port are “forwarded” to the master port. In this case, there are two simultaneous accesses on the matrix: one on a non-cached region, and another “dummy” access on the cache master port. These two accesses can slow down the system due to the wait error introduction on the cache master port.

### 24.4.2 Cache Maintenance

If the contents seen by the cache has changed, the user needs to invalidate the cache entries. It can be done line by line or for all cache entries.

#### 24.4.2.1 Cache Invalidate by Line Operation

When an invalidate by line command is issued the cache controller resets the valid bit information of the decoded cache line. As the line is no longer valid the replacement counter points to that line.

Use the following sequence to invalidate one line of cache.

1. Disable the cache controller, writing 0 to the CEN field of the CMCC\_CTRL register.
2. Check CSTS field of the CMCC\_SR to verify that the cache is successfully disabled.
3. Perform an invalidate by line writing the bit set {index, way} in the CMCC\_MAINT1 register.
4. Enable the cache controller, writing 1 to the CEN field of the CMCC\_CTRL register.

#### 24.4.2.2 Cache Invalidate All Operation

To invalidate all cache entries:

Write 1 to the INVALL field of the CMCC\_MAINT0 register.

### 24.4.3 Cache Performance Monitoring

The Cortex M cache controller includes a programmable 32-bit monitor counter. The monitor can be configured to count the number of clock cycles, the number of data hits or the number of instruction hits.

Use the following sequence to activate the counter

1. Configure the monitor counter, writing the MODE field of the CMCC\_CFG register.
2. Enable the counter, writing one to the MENABLE field of the CMCC\_MEN register.
3. If required, reset the counter, writing one to the SWRST field of the CMCC\_MCTRL register.
4. Check the value of the monitor counter, reading EVENT\_CNT field of the CMCC\_SR.



## 24.5 Cortex M Cache Controller (CMCC) User Interface

Table 24-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Cache Type Register	CMCC_TYPE	Read-only	0x000011D7
0x04	Reserved	–	–	–
0x08	Cache Control Register	CMCC_CTRL	Write-only	0x00000000
0x0C	Cache Status Register	CMCC_SR	Read-only	0x00000001
0x10 - 0x1C	Reserved	–	–	–
0x20	Cache Maintenance Register 0	CMCC_MAINT0	Write-only	–
0x24	Cache Maintenance Register 1	CMCC_MAINT1	Write-only	–
0x28	Cache Monitor Configuration Register	CMCC_MCFG	Read-write	0x00000000
0x2C	Cache Monitor Enable Register	CMCC_MEN	Read-write	0x00000000
0x30	Cache Monitor Control Register	CMCC_MCTRL	Write-only	–
0x34	Cache Monitor Status Register	CMCC_MSR	Read-only	0x00000000
0x38 - 0xFC	Reserved	–	–	–

### 24.5.1 Cache Controller Type Register

**Name:** CMCC\_TYPE

**Address:** 0x4007C000 (0), 0x48018000 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–		CLSIZE			CSIZE		
7	6	5	4	3	2	1	0
LCKDOWN	WAYNUM		RRP	LRUP	RANDP	–	–

- **RANDP: Random Selection Policy Supported**

0: Random victim selection is not supported.

1: Random victim selection is supported.

- **LRUP: Least Recently Used Policy Supported**

0: Least Recently Used Policy is not supported.

1: Least Recently Used Policy is supported.

- **RRP: Random Selection Policy Supported**

0: Random Selection Policy is not supported.

1: Random Selection Policy is supported.

- **WAYNUM: Number of Way**

Value	Name	Description
0	DMAPPED	Direct Mapped Cache
1	ARCH2WAY	2-WAY set associative
2	ARCH4WAY	4-WAY set associative
3	ARCH8WAY	8-WAY set associative

- **LCKDOWN: Lock Down Supported**

0: Lock Down is not supported.

1: Lock Down is supported.

- **CSIZE: Cache Size**

Value	Name	Description
0	CSIZE_1KB	Cache Size 1 Kbytes
1	CSIZE_2KB	Cache Size 2 Kbytes
2	CSIZE_4KB	Cache Size 4 Kbytes
3	CSIZE_8KB	Cache Size 8 Kbytes

- **CLSIZE: Cache Size**

Value	Name	Description
0	CLSIZE_1KB	4 Bytes
1	CLSIZE_2KB	8 Bytes
2	CLSIZE_4KB	16 Bytes
3	CLSIZE_8KB	32 Bytes

## 24.5.2 Cache Controller Control Register

**Name:** CMCC\_CTRL

**Address:** 0x4007C008 (0), 0x48018008 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	CEN

- **CEN: Cache Controller Enable**

0: When set to 0, this field disables the cache controller.

1: When set to 1, this field enables the cache controller.

### 24.5.3 Cache Controller Status Register

**Name:** CMCC\_SR

**Address:** 0x4007C00C (0), 0x4801800C (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	CSTS

- **CSTS: Cache Controller Status**

0: When read as 0, this field indicates that the cache controller is disabled.

1: When read as 1, this field indicates that the cache controller is enabled.

#### 24.5.4 Cache Controller Maintenance Register 0

**Name:** CMCC\_MAINT0

**Address:** 0x4007C020 (0), 0x48018020 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	INVALL

- **INVALL: Cache Controller Invalidate All**

0: No effect.

1: When set to 1, this field invalidates all cache entries.

### 24.5.5 Cache Controller Maintenance Register 1

**Name:** CMCC\_MAINT1

**Address:** 0x4007C024 (0), 0x48018024 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
WAY		–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	INDEX
7	6	5	4	3	2	1	0
INDEX				–	–	–	–

- **INDEX: Invalidate Index**

This field indicates the cache line that is being invalidated.

The size of the INDEX field depends on the cache size:

- for 2 Kbytes: 5 bits
- for 4 Kbytes: 6 bits
- for 8 Kbytes: 7 bits, and so on

- **WAY: Invalidate Way**

Value	Name	Description
0	WAY0	Way 0 is selection for index invalidation
1	WAY1	Way 1 is selection for index invalidation
2	WAY2	Way 2 is selection for index invalidation
3	WAY3	Way 3 is selection for index invalidation

### 24.5.6 Cache Controller Monitor Configuration Register

**Name:** CMCC\_MCFG

**Address:** 0x4007C028 (0), 0x48018028 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	MODE	

- **MODE: Cache Controller Monitor Counter Mode**

Value	Name	Description
0	CYCLE_COUNT	Cycle counter
1	IHIT_COUNT	Instruction hit counter
2	DHIT_COUNT	Data hit counter



### 24.5.7 Cache Controller Monitor Enable Register

**Name:** CMCC\_MEN

**Address:** 0x4007C02C (0), 0x4801802C (1)

**Access:** Write-only

**Reset:** 0x00002000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	MENABLE

- **MENABLE: Cache Controller Monitor Enable**

0: When set to 0, the monitor counter is disabled.

1: When set to 1, the monitor counter is activated.

### 24.5.8 Cache Controller Monitor Control Register

**Name:** CMCC\_MCTRL

**Address:** 0x4007C030 (0), 0x48018030 (1)

**Access:** Write-only

**Reset:** 0x00002000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	SWRST

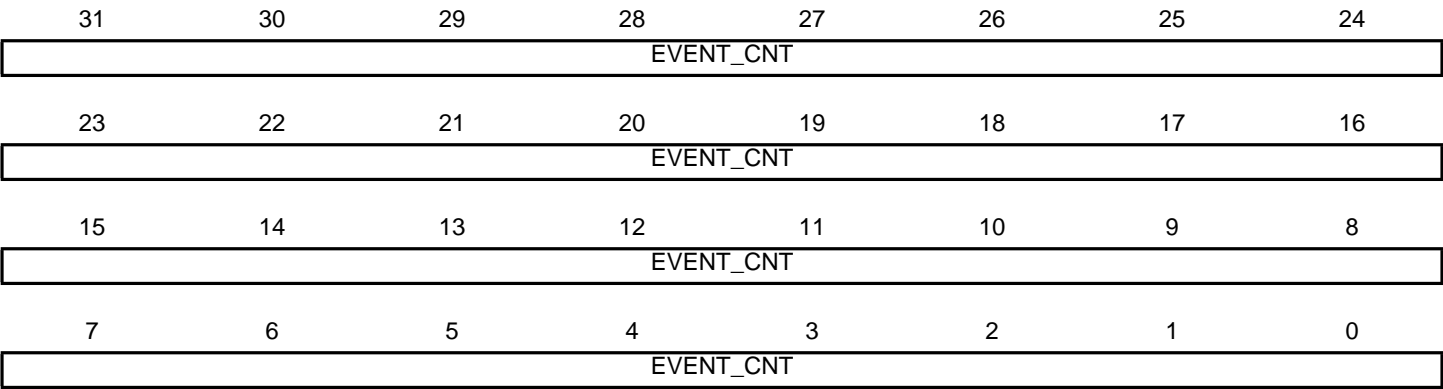
- **SWRST: Monitor**

0: No effect.

1: When set to 1, this field resets the event counter register.

24.5.9 Cache Controller Monitor Status Register

**Name:** CMCC\_MSR  
**Address:** 0x4007C034 (0), 0x48018034 (1)  
**Access:** Read-only  
**Reset:** 0x00002000



- **EVENT\_CNT:** Monitor Event Counter

## 25. Interprocessor Communication (IPC)

### 25.1 Description

The Interprocessor Communication (IPC) module has 32 interrupt sources. Each source has a set of enable, disable, clear, set, mask and status registers. The interrupt sources are ORed, and the IPC interrupt output line is connected to the Interrupt Controller input.

### 25.2 Block Diagram

Figure 25-1. IPC Block Diagram

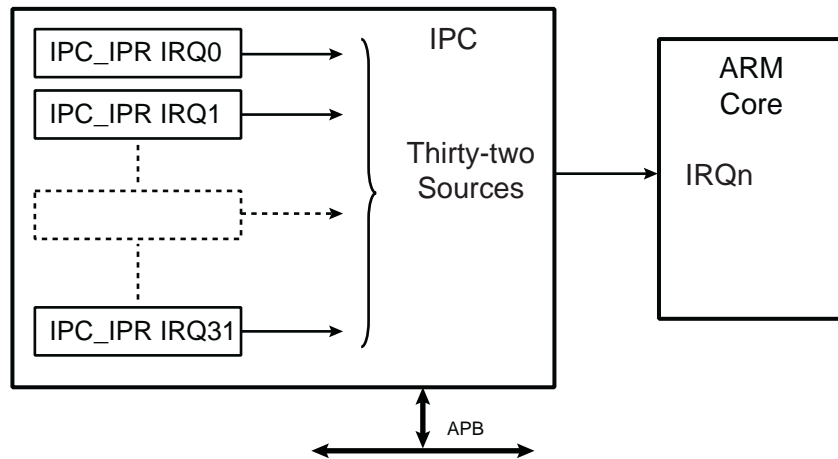
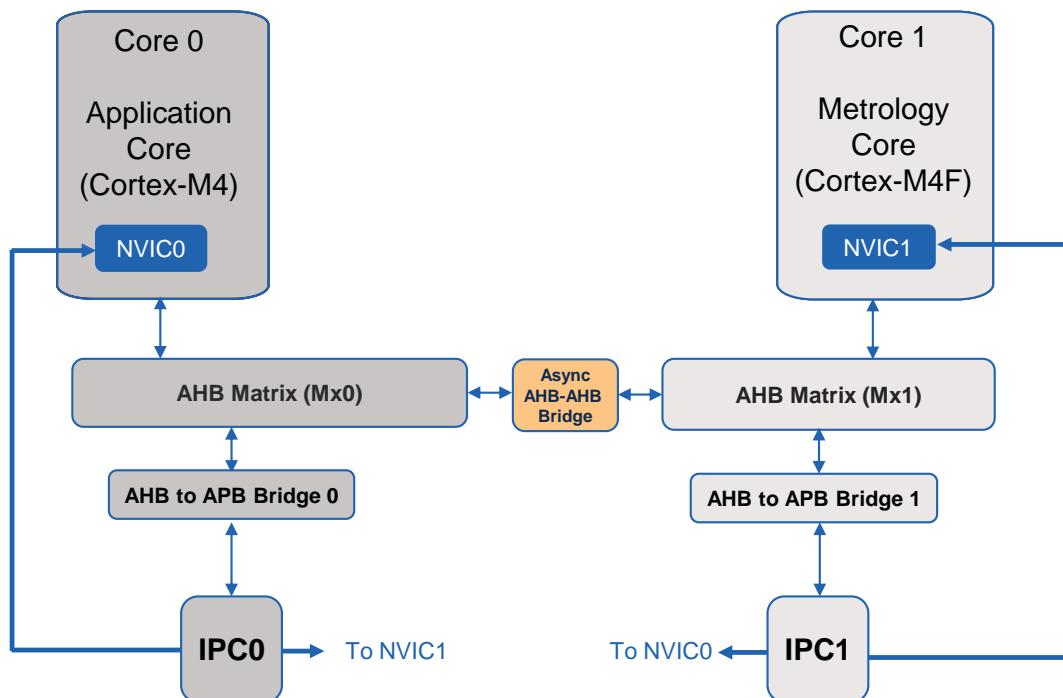


Figure 25-2. Dual Core IPC Implementation



## 25.3 Product Dependencies

### 25.3.1 Power Management

The Interprocessor Communication module is not continuously clocked. The IPC interface is clocked through the Power Management Controller (PMC), therefore the programmer must first configure the PMC to enable the IPC clock.

### 25.3.2 Interrupt Line

The IPC module has an interrupt line connected to the Interrupt Controller. Handling interrupts requires programming the Interrupt Controller before configuring the IPC.

**Table 25-1. Peripheral IDs**

Instance	ID
IPC0	31
IPC1	39

## 25.4 Functional Description

### 25.4.1 Interrupt Sources

#### 25.4.1.1 Interrupt Generation

Interrupt sources can be individually generated by writing respectively the IPC\_ISCR and IPC\_ICCR registers.

#### 25.4.1.2 Interrupt Source Control

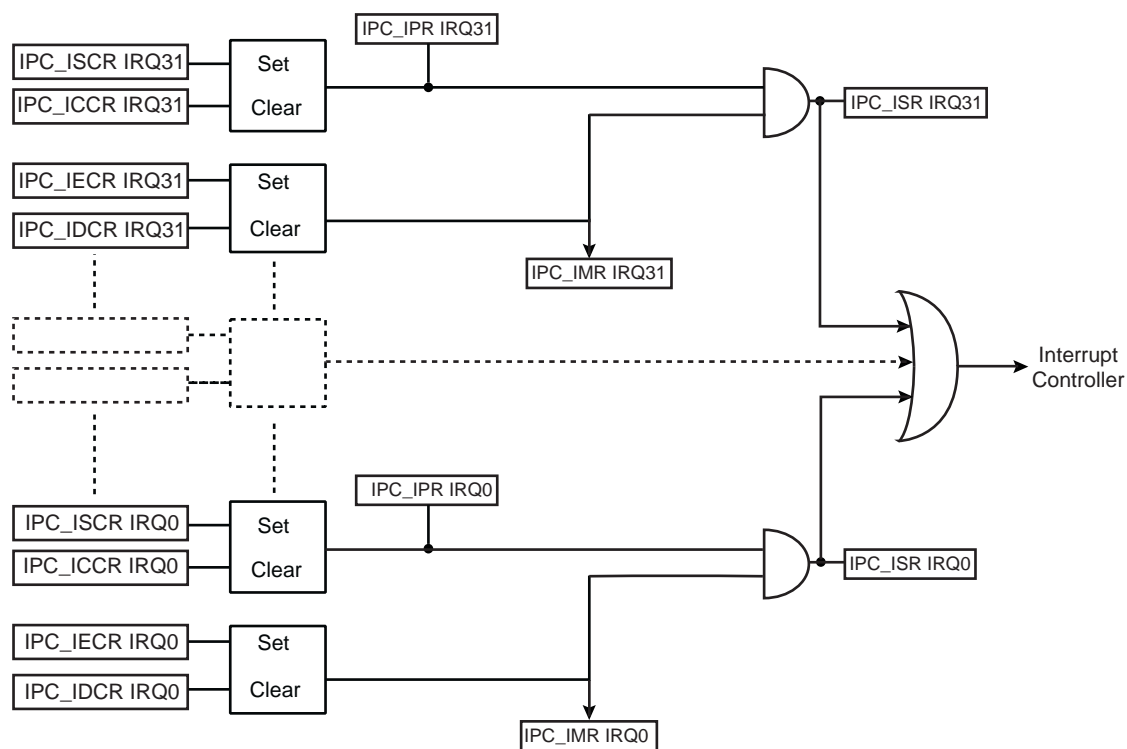
Each interrupt source (IRQ0 to IRQ31) can be enabled or disabled by using the command registers: IPC\_IECR (Interrupt Enable Command Register) and IPC\_IDCR (Interrupt Disable Command Register). This set of registers conducts enabling or disabling of an instruction. The interrupt mask can be read in the IPC\_IMR register. All IPC interrupts can be enabled/disabled, thus configuring the IPC Interrupt mask register. Each pending and unmasked IPC interrupt asserts the IPC output interrupt line. A disabled interrupt does not affect servicing of other interrupts.

#### 25.4.1.3 Interrupt Status

The IPC\_IECR and IPC\_IDCR registers are used to determine which interrupt sources are active/inhibited to generate an interrupt output. The IPC\_IMR register is a status of the interrupt source selection (a result from write into the IPC\_IECR and IPC\_IDCR registers). The IPC\_ISCR and IPC\_ICCR registers are used to activate/inhibit interrupt sources. The IPC\_IPR register is a status register giving active interrupt sources.

The IPC\_ISR register reports which interrupt source(s) is(are) currently asserting an interrupt output. IPC\_ISR is basically equivalent to an AND between the IPC\_IPR and IPC\_IMR registers.

Figure 25-3. Interrupt Input Stage



## 25.5 Inter-processor Communication (IPC) User Interface

Table 25-2. Register Mapping

Offset	Register	Name	Access	Reset
0x0000	Interrupt Set Command Register	IPC_ISCR	Write-only	–
0x0004	Interrupt Clear Command Register	IPC_ICCR	Write-only	–
0x0008	Interrupt Pending Register	IPC_IPR	Read-only	0x0
0x000C	Interrupt Enable Command Register	IPC_IECR	Write-only	–
0x0010	Interrupt Disable Command Register	IPC_IDCR	Write-only	–
0x0014	Interrupt Mask Register	IPC_IMR	Read-only	0x0
0x0018	Interrupt Status Register	IPC_ISR	Read-only	0x0

### 25.5.1 IPC Interrupt Set Command Register

**Name:** IPC\_ISCR

**Address:** 0x4004C000 (0), 0x48014000 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Set**

0: No effect.

1: Sets the corresponding interrupt.



## 25.5.2 IPC Interrupt Clear Command Register

**Name:** IPC\_ICCR

**Address:** 0x4004C004 (0), 0x48014004 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Clear**

0: No effect.

1: Clears the corresponding interrupt.

### 25.5.3 IPC Interrupt Pending Register

**Name:** IPC\_IPR

**Address:** 0x4004C008 (0), 0x48014008 (1)

**Access:** Read-only

**Reset:** 0x0

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Pending**

0: The corresponding interrupt is not pending.

1: The corresponding interrupt is pending.

## 25.5.4 IPC Interrupt Enable Command Register

**Name:** IPC\_IECR

**Address:** 0x4004C00C (0), 0x4801400C (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Enable**

0: No effect.

1: Enables the corresponding interrupt.

## 25.5.5 IPC Interrupt Disable Command Register

**Name:** IPC\_IDCR

**Address:** 0x4004C010 (0), 0x48014010 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Disable**

0: No effect.

1: Disables the corresponding interrupt.

## 25.5.6 IPC Interrupt Mask Register

**Name:** IPC\_IMR

**Address:** 0x4004C014 (0), 0x48014014 (1)

**Access:** Read-only

**Reset:** 0x0

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Interrupt Mask**

0: The corresponding interrupt is disabled.

1: The corresponding interrupt is enabled.

## 25.5.7 IPC Interrupt Status Register

**Name:** IPC\_ISR

**Access:** Read-only

**Reset:** 0x0

31	30	29	28	27	26	25	24
IRQ31	IRQ30	IRQ29	IRQ28	IRQ27	IRQ26	IRQ25	IRQ24
23	22	21	20	19	18	17	16
IRQ23	IRQ22	IRQ21	IRQ20	IRQ19	IRQ18	IRQ17	IRQ16
15	14	13	12	11	10	9	8
IRQ15	IRQ14	IRQ13	IRQ12	IRQ11	IRQ10	IRQ9	IRQ8
7	6	5	4	3	2	1	0
IRQ7	IRQ6	IRQ5	IRQ4	IRQ3	IRQ2	IRQ1	IRQ0

- **IRQ0-IRQ31: Current Interrupt Identifier**

0: The corresponding interrupt source is not currently asserting the interrupt output.

1: The corresponding interrupt source is currently asserting the interrupt output.

## 26. Bus Matrix (MATRIX)

### 26.1 Description

The Bus Matrix implements a multi-layer AHB, based on the AHB-Lite protocol, that enables parallel access paths between multiple AHB masters and slaves in a system, thus increasing the overall bandwidth. The Bus Matrix interconnects AHB masters to AHB slaves. The normal latency to connect a master to a slave is one cycle except for the default master of the accessed slave which is connected directly (zero cycle latency).

### 26.2 Embedded Characteristics

- One Decoder for Each Master
- Support for Long Bursts of 32, 64 and 128 Beats and Up to the 256-beat Word Burst AHB Limit
- Enhanced Programmable Mixed Arbitration for Each Slave
  - Round-robin
  - Fixed Priority
  - Latency Quality of Service
- Programmable Default Master for Each Slave
  - No Default Master
  - Last Accessed Default Master
  - Fixed Default Master
- Deterministic Maximum Access Latency for Masters
- Zero or One Cycle Arbitration Latency for the First Access of a Burst
- Bus Lock Forwarding to Slaves
- Master Number Forwarding to Slaves
- Write Protection of User Interface Registers

## 26.2.1 Matrix 0

### 26.2.1.1 Matrix 0 Masters

The Bus Matrix 0, which corresponds to the sub-system 0 (Core 0 - CM4P0), manages the masters listed in [Table 26-1](#). Each master can perform an access to an available slave concurrently with other masters.

Each master has its own specifically-defined decoder. In order to simplify the addressing, all the masters have the same decodings.

**Table 26-1. List of Bus Matrix Masters**

Master 0	Cortex-M4 Instruction/Data (CM4P0 I/D Bus)
Master 1	Cortex-M4 System (CM4P0 S Bus)
Master 2	Peripheral DMA Controller 0 (PDC0)
Master 3	Integrity Check Module (ICM)
Master 4	Matrix1
Master 5	EBI Matrix1
Master 6	CMCC0

### 26.2.1.2 Matrix 0 Slaves

The Bus Matrix manages the slaves listed in [Table 26-2](#). Each slave has its own arbiter providing a dedicated arbitration per slave.

**Table 26-2. List of Bus Matrix Slaves**

Slave 0	Internal SRAM0
Slave 1	Internal ROM
Slave 2	Internal Flash
Slave 3	External Bus Interface
Slave 4	Peripheral Bridge 0
Slave 5	CPKCC RAM and ROM
Slave 6	Matrix1
Slave 7	CMCC0

### 26.2.1.3 Master to Slave Access (Matrix 0)

[Table 26-3](#) gives valid paths for master to slave access on Matrix 0. The paths shown as “-” are forbidden or not wired, e.g. access from the Cortex-M4 S Bus to the Internal ROM.

**Table 26-3. Matrix 0 Master to Slave Access**

Slaves	Masters	0	1	2	3	4	5	6	Reserved
		Cortex-M4 I/D Bus	Cortex-M4 S Bus	PDC0	ICM	Matrix1	EBI Matrix 1	CMCC0	
0	Internal SRAM0	-	X	X	X	X	-	-	-
1	Internal ROM	X	-	X	X	-	-	-	-
2	Internal Flash	X	-	-	X	X	-	X	-
3	External Bus Interface	X	X	X	X	-	X	X	-
4	Peripheral Bridge 0	-	X	X	-	X	-	-	-

**Table 26-3. Matrix 0 Master to Slave Access**

5	CPKCC SRAM, ROM	-	X	-	X	-	-	-	-
6	Matrix1	-	X	-	X	-	-	-	-
7	CMCC0	X	-	-	-	-	-	-	-

**26.2.1.4 Accesses through Matrix 0**

- CM4P0 I/D Bus access to:
  - Flash, ROM
  - EBI (0x03000000 to 0x06FFFFFF)
  - Flash and EBI through Cache Controller CMCC0 (respectively through 0x11000000 to 0x11FFFFFF and 0x13000000 to 0x16FFFFFF)
- CMP4P0 S Bus access to:
  - SRAM0, SRAM1 through Matrix1, SRAM2 through Matrix1
  - PKCC
  - EBI (through 0x60000000 to 0x63FFFFFF, 0xA0000000 to A3FFFFFF)
- PDC0 access to:
  - SRAM0, ROM
  - EBI (through 0x60000000 to 0x63FFFFFF)
  - HBRIDGE0
- ICM access to:
  - Flash, ROM, SRAM0, SRAM1 through Matrix1, SRAM2 through Matrix1
  - PKCC
  - EBI through 0x60000000 to 0x63FFFFFF)
  - HBRIDGE1 through Matrix1
- Matrix1 access to
  - FLASH through 0x01000000 to 0x01FFFFFF and 0x11000000 to 0x11FFFFFF)
  - SRAM0
  - HBRIDGE0
- EBI Matrix1 access to:
  - EBI through 0x03000000 to 0x06FFFFFF, 0x13000000 to 0x16FFFFFF,
  - 0x60000000 to 0x63FFFFFF, 0xA0000000 to 0xA3FFFFFF)
- Cache Controller CMCC0 access to:
  - Flash (through 0x11000000 to 0x11FFFFFF)
  - EBI (through 0x13000000 to 0x16FFFFFF)

**26.2.2 Matrix 1****26.2.2.1 Matrix 1 Masters**

The Bus Matrix 1, which corresponds to the sub-system 1 (Core 1 - CM4P1), manages the masters listed in [Table 26-4](#). Each master can perform an access to an available slave concurrently with other masters.

Each master has its own specifically-defined decoder. In order to simplify the addressing, all the masters have the same decodings.

**Table 26-4. List of Bus Matrix Masters**

Master 0	Cortex-M4 Instruction/Data (CM4P1 I/D Bus)
Master 1	Cortex-M4 System (CM4P1 S Bus)
Master 2	Peripheral DMA Controller 1 (PDC1)



**Table 26-4. List of Bus Matrix Masters**

Master 3	Matrix0
Master 4	EBI Matrix0
Master 5	CMCC1

**26.2.2.2 Matrix 1 Slaves**

The Bus Matrix manages the slaves listed in [Table 26-2](#). Each slave has its own arbiter providing a dedicated arbitration per slave.

**Table 26-5. List of Bus Matrix Slaves**

Slave 0	Internal SRAM1
Slave 1	Internal SRAM2
Slave 2	External Bus Interface
Slave 3	Peripheral Bridge 1
Slave 4	Matrix0
Slave 5	CMCC1

**26.2.2.3 Master to Slave Access (Matrix 1)**

[Table 26-6](#) gives valid paths for master to slave access on Matrix 1. The paths shown as “-” are forbidden or not wired, e.g. access from the Cortex-M4 S Bus to the Internal ROM.

**Table 26-6. Matrix 1 Master to Slave Access**

Slaves	Masters	0	1	2	3	4	5
		Cortex-M4 I/D Bus	Cortex-M4 S Bus	PDC1	Matrix0	EBI Matrix 0	CMCC1
0	Internal SRAM1	X	X	X	X	-	-
1	Internal SRAM2	-	X	X	X	-	-
2	External Bus Interface	X	X	X	-	X	X
3	Peripheral Bridge 1	-	X	X	X	-	-
4	Matrix0	X	X	-	-	-	X
5	CMCC1	X	-	-	-	-	-

**26.2.2.4 Accesses through Matrix 1**

- CM4P1 I/D Bus access to:
  - Flash (through 0x01000000 to 0x01FFFFFF)
  - EBI (through 0x03000000 to 0x06FFFFFF)
  - FLASH and EBI through Cache CMCC1
- CM4P1 S-Bus access to:
  - SRAM1, SRAM2, SRAM0 through Matrix0 (0x20000000),
  - EBI (0x60000000 to 0x63FFFFFF and 0xA0000000 to 0xA3FFFFFF),
  - HBRIDGE1, HBRIDGE0 through Matrix0 (0x40000000)
- PDC1 access to:
  - SRAM1, SRAM2
  - EBI (0x60000000 to 0x63FFFFFF),

- HBRIDGE1
- Matrix0 access to:
  - SRAM1, SRAM2,
  - HBRIDGE1
- EBI from Matrix 0 access to:
  - EBI (through 0x03000000 to 0x06FFFFFFF, 0x60000000 to 0x63FFFFFFF, 0xA0000000 to A3FFFFFFF)
- Cache CMCC1 access to:
  - Flash through 0x11000000,
  - EBI through 0x13000000 to 0x16FFFFFFF

## 26.3 Special Bus Granting Mechanism

The Bus Matrix provides some speculative bus granting techniques in order to anticipate access requests from masters. This mechanism reduces latency at first access of a burst, or for a single transfer, as long as the slave is free from any other master access. However, the technique does not provide any benefits if the slave is continuously accessed by more than one master, since arbitration is pipelined and has no negative effect on the slave bandwidth or access latency.

This bus granting mechanism sets a different default master for every slave.

At the end of the current access, if no other request is pending, the slave remains connected to its associated default master. A slave can be associated with three kinds of default masters:

- No default master
- Last access master
- Fixed default master

To change from one type of default master to another, the Bus Matrix user interface provides Slave Configuration registers, one for every slave which set a default master for each slave. The Slave Configuration register contains two fields to manage master selection: DEFMSTR\_TYPE and FIXED\_DEFMSTR. The 2-bit DEFMSTR\_TYPE field selects the default master type (no default, last access master, fixed default master), whereas the 4-bit FIXED\_DEFMSTR field selects a fixed default master provided that DEFMSTR\_TYPE is set to fixed default master. Refer to [Section 26.9.2 “Bus Matrix Slave Configuration Registers” on page 441](#).

## 26.4 No Default Master

After the end of the current access, if no other request is pending, the slave is disconnected from all masters.

This configuration incurs one latency clock cycle for the first access of a burst after bus idle. Arbitration without the default master may be used for masters that perform significant bursts or several transfers with no idle in between, or if the slave bus bandwidth is widely used by one or more masters.

This configuration provides no benefit on access latency or bandwidth when reaching maximum slave bus throughput regardless of the number of requesting masters.

## 26.5 Last Access Master

After the end of the current access, if no other request is pending, the slave remains connected to the last master that performed an access request.

This allows the Bus Matrix to remove one latency cycle for the last master that accessed the slave. Other non-privileged masters still get one latency clock cycle if they need to access the same slave. This technique is used for masters that perform single accesses or short bursts with some idle cycles in between.

This configuration provides no benefit on access latency or bandwidth when reaching maximum slave bus throughput whatever is the number of requesting masters.

## 26.6 Fixed Default Master

After the end of the current access, if no other request is pending, the slave connects to its fixed default master. Unlike the last access master, the fixed default master does not change unless the user modifies it by software (FIXED\_DEFMSTR field of the related MATRIX\_SCFG).

This allows the Bus Matrix arbiters to remove the one latency clock cycle for the fixed default master of the slave. All requests attempted by the fixed default master do not cause any arbitration latency, whereas other non-privileged masters will get one latency cycle. This technique is used for a master that mainly performs single accesses or short bursts with idle cycles in between.

This configuration provides no benefit on access latency or bandwidth when reaching maximum slave bus throughput, regardless of the number of requesting masters.

## 26.7 Arbitration

The Bus Matrix provides an arbitration mechanism that reduces latency when a conflict occurs, i.e. when two or more masters try to access the same slave at the same time. One arbiter per AHB slave is provided, thus arbitrating each slave specifically.

The Bus Matrix provides the user with the possibility of choosing between two arbitration types or mixing them for each slave:

1. Round-robin arbitration (default)
2. Fixed priority arbitration

The resulting algorithm may be complemented by selecting a default master configuration for each slave.

When re-arbitration must be done, specific conditions apply. See [Section 26.7.1 “Arbitration Scheduling” on page 435](#).

### 26.7.1 Arbitration Scheduling

Each arbiter has the ability to arbitrate between two or more master requests. In order to avoid burst breaking and also to provide the maximum throughput for slave interfaces, arbitration may only take place during the following cycles:

1. Idle cycles: When a slave is not connected to any master or is connected to a master which is not currently accessing it
2. Single cycles: When a slave is currently doing a single access
3. End of Burst cycles: When the current cycle is the last cycle of a burst transfer. For defined burst length, predicted end of burst matches the size of the transfer but is managed differently for undefined burst length. See [“Undefined Length Burst Arbitration” on page 435](#)
4. Slot cycle limit: When the slot cycle counter has reached the limit value, indicating that the current master access is too long and must be broken. See [“Slot Cycle Limit Arbitration” on page 436](#)

#### 26.7.1.1 Undefined Length Burst Arbitration

In order to prevent long AHB burst lengths that can lock the access to the slave for an excessive period of time, the user can trigger the re-arbitration before the end of the incremental bursts. The re-arbitration period can be selected from the following Undefined Length Burst Type (ULBT) possibilities:

1. Unlimited: no predetermined end of burst is generated. This value enables 1-kbyte burst lengths.
2. 1-beat bursts: predetermined end of burst is generated at each single transfer during the INCR transfer.
3. 4-beat bursts: predetermined end of burst is generated at the end of each 4-beat boundary during INCR transfer.
4. 8-beat bursts: predetermined end of burst is generated at the end of each 8-beat boundary during INCR transfer.
5. 16-beat bursts: predetermined end of burst is generated at the end of each 16-beat boundary during INCR transfer.
6. 32-beat bursts: predetermined end of burst is generated at the end of each 32-beat boundary during INCR transfer.
7. 64-beat bursts: predetermined end of burst is generated at the end of each 64-beat boundary during INCR transfer.

8. 128-beat bursts: predetermined end of burst is generated at the end of each 128-beat boundary during INCR transfer.

The use of undefined length 8-beat bursts or less is discouraged since this may decrease the overall bus bandwidth due to arbitration and slave latencies at each first access of a burst.

However, if the usual length of undefined length bursts is known for a master, it is recommended to configure the ULBT according to this length.

This selection can be done through the ULBT field of the Master Configuration registers (MATRIX\_MCFG).

#### 26.7.1.2 Slot Cycle Limit Arbitration

The Bus Matrix contains specific logic to break long accesses, such as very long bursts on a very slow slave (e.g., an external low speed memory). At each arbitration time, a counter is loaded with the value previously written in the SLOT\_CYCLE field of the related Slave Configuration register (MATRIX\_SCFG) and decreased at each clock cycle. When the counter elapses, the arbiter has the ability to re-arbitrate at the end of the current AHB bus access cycle.

Unless a master has a very tight access latency constraint, which could lead to data overflow or underflow due to a badly undersized internal FIFO with respect to its throughput, the Slot Cycle Limit should be disabled (SLOT\_CYCLE = 0) or set to its default maximum value in order not to inefficiently break long bursts performed by some Atmel masters.

In most cases, this feature is not needed and should be disabled for power saving.

Warning: This feature cannot prevent any slave from locking its access indefinitely.

#### 26.7.2 Arbitration Priority Scheme

The Bus Matrix arbitration scheme is organized in priority pools. The corresponding access criticality class is assigned to each priority pool as shown in the “Latency Quality of Service” column in [Table 26-7](#). Latency Quality of Service is determined through the Bus Matrix user interface. See [Section 26.9.3 “Bus Matrix Priority Registers A For Slaves”](#) for details.

**Table 26-7. Arbitration Priority Pools**

Priority Pool	Latency Quality of Service
3	Latency Critical
2	Latency Sensitive
1	Bandwidth Sensitive
0	Background Transfers

Round-robin priority is used in the highest and lowest priority pools 3 and 0, whereas fixed level priority is used between priority pools and in the intermediate priority pools 2 and 1. See [Section 26.7.2.2 “Round-robin Arbitration”](#).

For each slave, each master is assigned to one of the slave priority pools through the Latency Quality of Service inputs or through the priority registers for slaves (MxPR fields of MATRIX\_PRAS and MATRIX\_PRBS). When evaluating master requests, this priority pool level always takes precedence.

After reset, most of the masters belong to the lowest priority pool (MxPR = 0, Background Transfer) and, therefore, are granted bus access in a true round-robin order.

The highest priority pool must be specifically reserved for masters requiring very low access latency. If more than one master belongs to this pool, they will be granted bus access in a biased round-robin manner which allows tight and deterministic maximum access latency from AHB bus requests. In the worst case, any currently occurring high-priority master request will be granted after the current bus master access has ended and other high priority pool master requests, if any, have been granted once each.

The lowest priority pool shares the remaining bus bandwidth between AHB Masters.

Intermediate priority pools allow fine priority tuning. Typically, a latency-sensitive master or a bandwidth-sensitive master will use such a priority level. The higher the priority level (MxPR value), the higher the master priority.

To ensure a good level of CPU performance, it is recommended to configure the CPU priority with the default reset value 2 (Latency Sensitive).

All combinations of MxPR values are allowed for all masters and slaves. For example, some masters might be assigned the highest priority pool (round-robin), and remaining masters the lowest priority pool (round-robin), with no master for intermediate fix priority levels.

#### 26.7.2.1 Fixed Priority Arbitration

Fixed priority arbitration algorithm is the first and only arbitration algorithm applied between masters from distinct priority pools. It is also used in priority pools other than the highest and lowest priority pools (intermediate priority pools).

Fixed priority arbitration allows the Bus Matrix arbiters to dispatch the requests from different masters to the same slave by using the fixed priority defined by the user in the MxPR field for each master in the Priority registers, MATRIX\_PRAS and MATRIX\_PRBS. If two or more master requests are active at the same time, the master with the highest priority MxPR number is serviced first.

In intermediate priority pools, if two or more master requests with the same priority are active at the same time, the master with the highest number is serviced first.

#### 26.7.2.2 Round-robin Arbitration

This algorithm is only used in the highest and lowest priority pools. It allows the Bus Matrix arbiters to properly dispatch requests from different masters to the same slave. If two or more master requests are active at the same time in the priority pool, they are serviced in a round-robin increasing master number order.

## 26.8 Register Write Protection

To prevent any single software error from corrupting the Bus Matrix behavior, certain registers in the address space can be write-protected by setting the WPEN bit in the [“Write Protection Mode Register”](#) (MATRIX\_WPMR).

If a write access to a write-protected register is detected, the WPVS flag in the [“Write Protection Status Register”](#) (MATRIX\_WPSR) is set and the field WPVSR indicates the register in which the write access has been attempted.

The WPVS bit is automatically cleared after reading the MATRIX\_WPSR.

The following registers can be write-protected:

- [“Bus Matrix Master Configuration Registers”](#)
- [“Bus Matrix Slave Configuration Registers”](#)
- [“Bus Matrix Priority Registers A For Slaves”](#)
- [“System I/O Configuration Register”](#)

## 26.9 AHB Bus Matrix (MATRIX) User Interface

Table 26-8. Register Mapping

Offset	Register	Name	Access	Reset
0x0000	Master Configuration Register 0	MATRIX_MCFG0	Read/Write	0x00000004
0x0004	Master Configuration Register 1	MATRIX_MCFG1	Read/Write	0x00000004
0x0008	Master Configuration Register 2	MATRIX_MCFG2	Read/Write	0x00000004
0x000C	Master Configuration Register 3	MATRIX_MCFG3	Read/Write	0x00000004
0x0010	Master Configuration Register 4	MATRIX_MCFG4	Read/Write	0x00000004
0x0014	Master Configuration Register 5	MATRIX_MCFG5	Read/Write	0x00000004
0x0018	Master Configuration Register 6	MATRIX_MCFG6	Read/Write	0x00000004
0x001C - 0x003C	Reserved	—	—	—
0x0040	Slave Configuration Register 0	MATRIX_SCFG0	Read/Write	0x000001FF
0x0044	Slave Configuration Register 1	MATRIX_SCFG1	Read/Write	0x000001FF
0x0048	Slave Configuration Register 2	MATRIX_SCFG2	Read/Write	0x000001FF
0x004C	Slave Configuration Register 3	MATRIX_SCFG3	Read/Write	0x000001FF
0x0050	Slave Configuration Register 4	MATRIX_SCFG4	Read/Write	0x000001FF
0x0054	Slave Configuration Register 5	MATRIX_SCFG5	Read/Write	0x000001FF
0x0058	Slave Configuration Register 6	MATRIX_SCFG6	Read/Write	0x000001FF
0x005C	Slave Configuration Register 7	MATRIX_SCFG7	Read/Write	0x000001FF
0x005C - 0x007C	Reserved	—	—	—
0x0080	Priority Register A for Slave 0	MATRIX_PRAS0	Read/Write	0x00000000 <sup>(1)</sup>
0x0084	Reserved	—	—	—
0x0088	Priority Register A for Slave 1	MATRIX_PRAS1	Read/Write	0x00000000 <sup>(1)</sup>
0x008C	Reserved	—	—	—
0x0090	Priority Register A for Slave 2	MATRIX_PRAS2	Read/Write	0x00000000 <sup>(1)</sup>
0x0094	Reserved	—	—	—
0x0098	Priority Register A for Slave 3	MATRIX_PRAS3	Read/Write	0x00000000 <sup>(1)</sup>
0x009C	Reserved	—	—	—
0x00A0	Priority Register A for Slave 4	MATRIX_PRAS4	Read/Write	0x00000000 <sup>(1)</sup>
0x00A4	Reserved	—	—	—
0x00A8	Priority Register A for Slave 5	MATRIX_PRAS5	Read/Write	0x00000000 <sup>(1)</sup>
0x00AC	Reserved	—	—	—
0x00B0	Priority Register A for Slave 6	MATRIX_PRAS6	Read/Write	0x00000000 <sup>(1)</sup>
0x00B4	Reserved	—	—	—
0x00B8	Priority Register A for Slave 7	MATRIX_PRAS7	Read/Write	0x00000000 <sup>(1)</sup>
0x00BC - 0x0110	Reserved	—	—	—
0x0114	System I/O Configuration Register	MATRIX_SYSIO	Read/Write	0x00000000
0x0118	Reserved	—	—	—

**Table 26-8. Register Mapping (Continued)**

Offset	Register	Name	Access	Reset
0x011C	SMC Nand Flash Chip Select Configuration Register	MATRIX_SMCNFCS	Read/Write	0x00000000
0x0120	Reserved	–	–	–
0x0124	Reserved	–	–	–
0x0128	Core Debug Configuration Register	MATRIX_CORE_DEBUG	Read/Write	0x00000000
0x012C - 0x01E0	Reserved	–	–	–
0x01E4	Write Protection Mode Register	MATRIX_WPMR	Read/Write	0x00000000
0x01E8	Write Protection Status Register	MATRIX_WPSR	Read-only	0x00000000

Note: 1. Values in the Bus Matrix Priority Registers are product dependent.

## 26.9.1 Bus Matrix Master Configuration Registers

**Name:** MATRIX\_MCFGx [x=0..6]

**Address:** 0x400E0200 (0), 0x48010000 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	ULBT		

This register can only be written if the WPEN bit is cleared in the [“Write Protection Mode Register”](#).

### • ULBT: Undefined Length Burst Type

0: Unlimited Length Burst

No predicted end of burst is generated, therefore INCR bursts coming from this master can only be broken if the Slave Slot Cycle Limit is reached. If the Slot Cycle Limit is not reached, the burst is normally completed by the master, at the latest, on the next AHB 1 KByte address boundary, allowing up to 256-beat word bursts or 128-beat double-word bursts.

This value should not be used in the very particular case of a master capable of performing back-to-back undefined length bursts on a single slave, since this could indefinitely freeze the slave arbitration and thus prevent another master from accessing this slave.

1: Single Access

The undefined length burst is treated as a succession of single accesses, allowing re-arbitration at each beat of the INCR burst or bursts sequence.

2: 4-beat Burst

The undefined length burst or bursts sequence is split into 4-beat bursts or less, allowing re-arbitration every 4 beats.

3: 8-beat Burst

The undefined length burst or bursts sequence is split into 8-beat bursts or less, allowing re-arbitration every 8 beats.

4: 16-beat Burst

The undefined length burst or bursts sequence is split into 16-beat bursts or less, allowing re-arbitration every 16 beats.

5: 32-beat Burst

The undefined length burst or bursts sequence is split into 32-beat bursts or less, allowing re-arbitration every 32 beats.

6: 64-beat Burst

The undefined length burst or bursts sequence is split into 64-beat bursts or less, allowing re-arbitration every 64 beats.

7: 128-beat Burst

The undefined length burst or bursts sequence is split into 128-beat bursts or less, allowing re-arbitration every 128 beats.

Unless duly needed, the ULBT should be left at its default 0 value for power saving.



## 26.9.2 Bus Matrix Slave Configuration Registers

**Name:** MATRIX\_SCFGx [x=0..7]

**Address:** 0x400E0240 (0), 0x48010040 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	FIXED_DEFMSTR				DEFMSTR_TYPE	
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	SLOT_CYCLE
7	6	5	4	3	2	1	0
SLOT_CYCLE							

This register can only be written if the WPEN bit is cleared in the [“Write Protection Mode Register”](#).

- **SLOT\_CYCLE: Maximum Bus Grant Duration for Masters**

When SLOT\_CYCLE AHB clock cycles have elapsed since the last arbitration, a new arbitration takes place to let another master access this slave. If another master is requesting the slave bus, then the current master burst is broken.

If SLOT\_CYCLE = 0, the Slot Cycle Limit feature is disabled and bursts always complete unless broken according to the ULBT.

This limit has been placed in order to enforce arbitration so as to meet potential latency constraints of masters waiting for slave access.

This limit must not be too small. Unreasonably small values break every burst and the Bus Matrix arbitrates without performing any data transfer. The default maximum value is usually an optimal conservative choice.

In most cases, this feature is not needed and should be disabled for power saving.

See [Section 26.7.1.2 “Slot Cycle Limit Arbitration”](#) for details.

- **DEFMSTR\_TYPE: Default Master Type**

0: No Default Master

At the end of the current slave access, if no other master request is pending, the slave is disconnected from all masters.

This results in a one clock cycle latency for the first access of a burst transfer or for a single access.

1: Last Default Master

At the end of the current slave access, if no other master request is pending, the slave stays connected to the last master having accessed it.

This results in not having one clock cycle latency when the last master tries to access the slave again.

2: Fixed Default Master

At the end of the current slave access, if no other master request is pending, the slave connects to the fixed master the number that has been written in the FIXED\_DEFMSTR field.

This results in not having one clock cycle latency when the fixed master tries to access the slave again.

- **FIXED\_DEFMSTR: Fixed Default Master**

This is the number of the Default Master for this slave. Only used if DEFMSTR\_TYPE is 2. Specifying the number of a master which is not connected to the selected slave is equivalent to setting DEFMSTR\_TYPE to 0.

### 26.9.3 Bus Matrix Priority Registers A For Slaves

**Name:** MATRIX\_PRASx [x=0..7]

**Address:** 0x400E0280 (0)[0], 0x400E0288 (0)[1], 0x400E0290 (0)[2], 0x400E0298 (0)[3], 0x400E02A0 (0)[4], 0x400E02A8 (0)[5], 0x400E02B0 (0)[6], 0x400E02B8 (0)[7], 0x48010080 (1)[0], 0x48010088 (1)[1], 0x48010090 (1)[2], 0x48010098 (1)[3], 0x480100A0 (1)[4], 0x480100A8 (1)[5], 0x480100B0 (1)[6], 0x480100B8 (1)[7]

**Access:** Read/Write

31	30	29	28	27	26	25	24
–		M7PR		–		M6PR	
23	22	21	20	19	18	17	16
–		M5PR		–		M4PR	
15	14	13	12	11	10	9	8
–		M3PR		–		M2PR	
7	6	5	4	3	2	1	0
–		M1PR		–		M0PR	

This register can only be written if the WPE bit is cleared in the [“Write Protection Mode Register”](#).

- **MxPR: Master x Priority**

Fixed priority of Master x for accessing the selected slave. The higher the number, the higher the priority.

All the masters programmed with the same MxPR value for the slave make up a priority pool.

Round-robin arbitration is used in the lowest (MxPR = 0) and highest (MxPR = 3) priority pools.

Fixed priority is used in intermediate priority pools (MxPR = 1) and (MxPR = 2).

See [“Arbitration Priority Scheme” on page 436](#) for details.

## 26.9.4 System I/O Configuration Register

**Name:** MATRIX\_SYSIO

**Address:** 0x400E0314 (0), 0x48010114 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	SYSIO9	–
7	6	5	4	3	2	1	0
–	–	–	–	SYSIO3	SYSIO2	SYSIO1	SYSIO0

This register can only be written if the WPEN bit is cleared in the [“Write Protection Mode Register”](#).

- **SYSIO0: PB0 or TDI Assignment**

0: TDI function selected.

1: PB0 function selected.

- **SYSIO1: PB1 or TDO/TRACESWO Assignment**

0: TDO/TRACESWO function selected.

1: PB1 function selected.

- **SYSIO2: PB2 or TMS/SWDIO Assignment**

0: TMS/SWDIO function selected.

1: PB2 function selected.

- **SYSIO3: PB3 or TCK/SWCLK Assignment**

0: TCK/SWCLK function selected.

1: PB3 function selected.

- **SYSIO9: PC9 or ERASE Assignment**

0: ERASE function selected.

1: PC9 function selected.

## 26.9.5 SMC NAND Flash Chip Select Configuration Register

**Name:** MATRIX\_SMCNFCS

**Address:** 0x400E031C (0), 0x4801011C (1)

**Access:** Read/Write

**Reset:** 0x0000\_0000

31	30	29	28	27	26	25	24
SMC_SEL	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	SMC_NFCS3	SMC_NFCS2	SMC_NFCS1	SMC_NFCS0

- **SMC\_NFCS0: SMC NAND Flash Chip Select 0 Assignment**

0: NCS0 is not assigned to a NAND Flash (NANDOE and NANWE not used for NCS0)

1: NCS0 is assigned to a NAND Flash (NANDOE and NANWE used for NCS0)

- **SMC\_NFCS1: SMC NAND Flash Chip Select 1 Assignment**

0: NCS1 is not assigned to a NAND Flash (NANDOE and NANWE not used for NCS1)

1: NCS1 is assigned to a NAND Flash (NANDOE and NANWE used for NCS1)

- **SMC\_NFCS2: SMC NAND Flash Chip Select 2 Assignment**

0: NCS2 is not assigned to a NAND Flash (NANDOE and NANWE not used for NCS2)

1: NCS2 is assigned to a NAND Flash (NANDOE and NANWE used for NCS2)

- **SMC\_NFCS3: SMC NAND Flash Chip Select 3 Assignment**

0: NCS3 is not assigned to a NAND Flash (NANDOE and NANWE not used for NCS3)

1: NCS3 is assigned to a NAND Flash (NANDOE and NANWE used for NCS3)

- **SMC\_SEL: SMC Selection for EBI pins**

0: EBI pins are used by SMC0

1: EBI pins are used by SMC1

## 26.9.6 Core Debug Configuration Register

**Name:** MATRIX\_CORE\_DEBUG

**Address:** 0x400E0328 (0), 0x48010128 (1)

**Access:** Read/Write

**Reset:** 0x0000\_0000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	CROSS_TRG0	CROSS_TRG1	–

- **CROSS\_TRIG1: Core 1 --> Core 0 Cross Triggering**

0: Core 1 is not able to trigger an event on core 0.

1: Core 1 is able to trigger an event on core 0.

- **CROSS\_TRIG0: Core 0 --> Core 1 Cross Triggering**

0: Core 0 is not able to trigger an event on core 1.

1: Core 0 is able to trigger an event on core 1.

## 26.9.7 Write Protection Mode Register

**Name:** MATRIX\_WPMR

**Address:** 0x400E03E4 (0), 0x480101E4 (1)

**Access:** Read/Write

**Reset:** See [Table 26-8](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

For more information on Write Protection registers, refer to [Section 26.8 “Register Write Protection”](#).

- **WPEN: Write Protect Enable**

0: Disables the write protection if WPKEY corresponds to 0x4D4154 (“MAT” in ASCII).

1: Enables the write protection if WPKEY corresponds to 0x4D4154 (“MAT” in ASCII).

See [Section 26.8 “Register Write Protection”](#) for the list of registers that can be protected.

- **WPKEY: Write Protect Key**

Value	Name	Description
0x4D4154	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

## 26.9.8 Write Protection Status Register

**Name:** MATRIX\_WPSR

**Address:** 0x400E03E8 (0), 0x480101E8 (1)

**Access:** Read-only

**Reset:** See [Table 26-8](#)

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

For more information on Write Protection registers, refer to [Section 26.8 “Register Write Protection”](#).

- **WPVS: Write Protection Violation Status**

0: No write protection violation has occurred since the last read of the MATRIX\_WPSR register.

1: A write protection violation has occurred since the last read of the MATRIX\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protection Violation Source**

When WPVS = 1, WPVSR indicates the register address offset at which a write access has been attempted.

## 27. Static Memory Controller (SMC)

### 27.1 Description

The External Bus Interface is designed to ensure the successful data transfer between several external devices and the Cortex-M4 based device. The External Bus Interface of the SAM4CM consists of a Static Memory Controller (SMC).

This SMC is capable of handling several types of external memory and peripheral devices, such as SRAM, PSRAM, PROM, EPROM, EEPROM, LCD Module, NOR Flash and NAND Flash.

The Static Memory Controller (SMC) generates the signals that control the access to the external memory devices or peripheral devices. It has 4 Chip Selects, a 24-bit address bus, and a configurable 8 or 16-bit data bus. Separate read and write control signals allow for direct memory and peripheral interfacing. Read and write signal waveforms are fully adjustable.

The SMC can manage wait requests from external devices to extend the current access. The SMC is provided with an automatic slow clock mode. In slow clock mode, it switches from user-programmed waveforms to slow-rate specific waveforms on read and write signals. The SMC supports asynchronous burst read in page mode access for page size up to 32 bytes.

The External Data Bus can be scrambled/unscrambled by means of user keys.

### 27.2 Embedded Characteristics

- 4 Chip Selects Available
- 16-Mbyte Address Space per Chip Select
- 8-bit or 16-bit Data Bus
- Zero Wait State Scrambling/Unscrambling function with User Key
- Word, Halfword, Byte Transfers
- Programmable Setup, Pulse And Hold Time for Read Signals per Chip Select
- Programmable Setup, Pulse And Hold Time for Write Signals per Chip Select
- Programmable Data Float Time per Chip Select
- External Data Bus Scrambling/Unscrambling Function
- External Wait Request
- Automatic Switch to Slow Clock Mode
- Asynchronous Read in Page Mode Supported: Page Size Ranges from 4 to 32 Bytes
- Write Protected Registers



## 27.3 I/O Lines Description

Table 27-1. I/O Line Description

Name	Description	Type	Active Level
NCS[3:0]	Static Memory Controller Chip Select Lines	Output	Low
NRD	Read Signal	Output	Low
NWR0/NWE	Write 0/Write Enable Signal	Output	Low
NWR1/NBS1	Write 1/Byte 1 Select Signal	Output	Low
A0/NBS0	Address Bit 0/Byte 0 Select Signal	Output	Low
A[23:1]	Address Bus	Output	
D[15:0]	Data Bus	I/O	
NWAIT	External Wait Signal	Input	Low
NANDCS	NAND Flash Chip Select Line	Output	Low
NANDOE	NAND Flash Output Enable	Output	Low
NANDWE	NAND Flash Write Enable	Output	Low

## 27.4 Product Dependencies

### 27.4.1 I/O Lines

The pins used for interfacing the Static Memory Controller are multiplexed with the PIO lines. The programmer must first program the PIO controller to assign the Static Memory Controller pins to their peripheral function. If I/O Lines of the SMC are not used by the application, they can be used for other purposes by the PIO Controller.

### 27.4.2 Power Management

The SMC is clocked through the Power Management Controller (PMC), thus the programmer must first configure the PMC to enable the SMC clock.

## 27.5 Multiplexed Signals

Table 27-2. Static Memory Controller (SMC) Multiplexed Signals

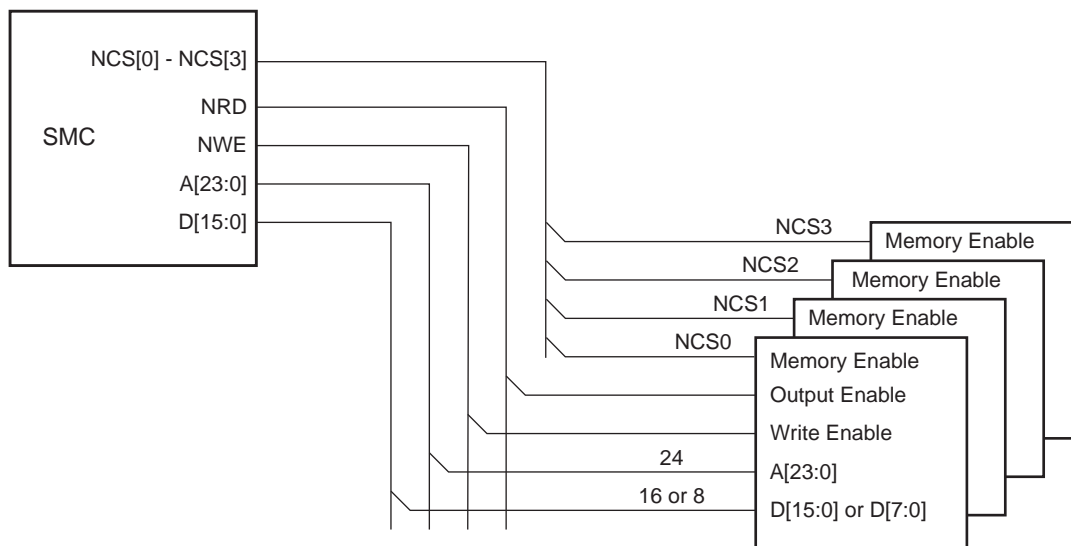
Multiplexed Signals		Related Function
NWR0	NWE	
A0	NBS0	
NWR1	NBS1	

## 27.6 External Memory Mapping

The SMC provides up to 24 address lines, A[23:0]. This allows each chip select line to address up to 16 Mbytes of memory.

If the physical memory device connected on one chip select is smaller than 16 Mbytes, it wraps around and appears to be repeated within this space. The SMC correctly handles any valid access to the memory device within the page (see [Figure 27-1](#)).

**Figure 27-1. Memory Connections for Four External Devices**



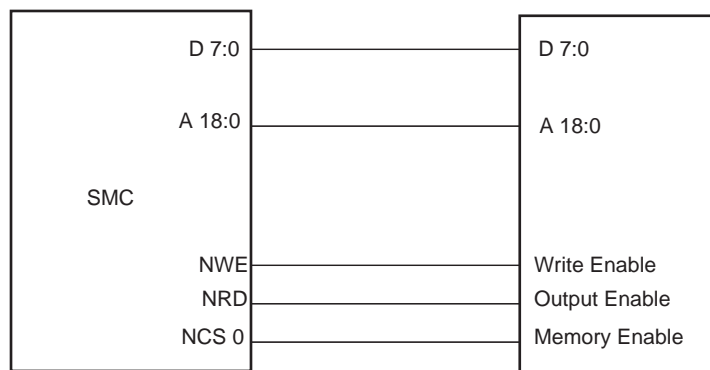
## 27.7 Connection to External Devices

### 27.7.1 Data Bus Width

The data bus width is 8 bits.

[Figure 27-2](#) shows how to connect a 512K x 8-bit memory on NCS0.

**Figure 27-2. Memory Connection for an 8-bit Data Bus**



### 27.7.1.1 NAND Flash Support

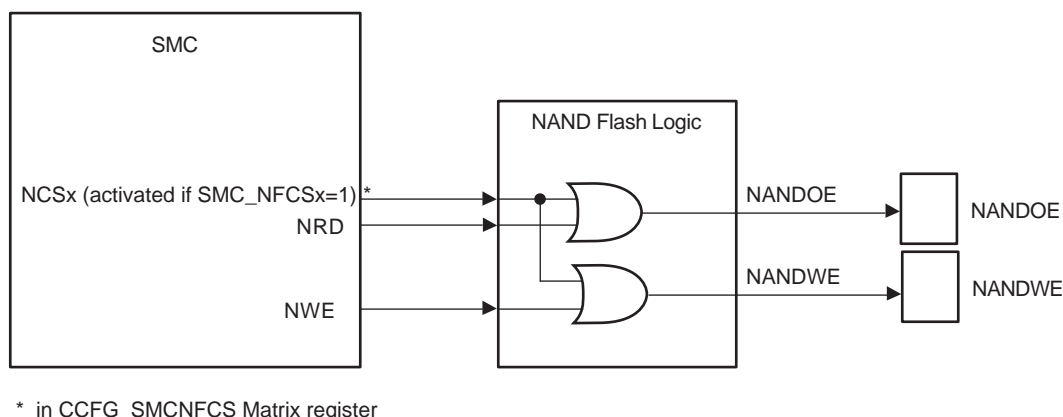
The SMC integrates circuitry that interfaces to NAND Flash devices.

The NAND Flash logic is driven by the Static Memory Controller. It depends on the programming of the SMC\_NFCSx field in the CCFG\_SMCNFCS Register on the Bus Matrix User Interface. For details on this register, refer to the Bus Matrix User Interface section. Access to an external NAND Flash device via the address space reserved to the chip select programmed.

The user can connect up to 4 NAND Flash devices with separated chip select.

The NAND Flash logic drives the read and write command signals of the SMC on the NANDOE and NANDWE signals when the NCSx programmed is active. NANDOE and NANDWE are disabled as soon as the transfer address fails to lie in the NCSx programmed address space.

Figure 27-3. NAND Flash Signal Multiplexing on SMC Pins



**Note:** When NAND Flash logic is activated, (SMCNFCSx=1), NWE pin cannot be used in PIO Mode but only in peripheral mode (NWE function). If NWE function is not used for other external memories (SRAM, LCD), it must be configured in one of the following modes.

- PIO Input with pull-up enabled (default state after reset)
- PIO Output set at level 1

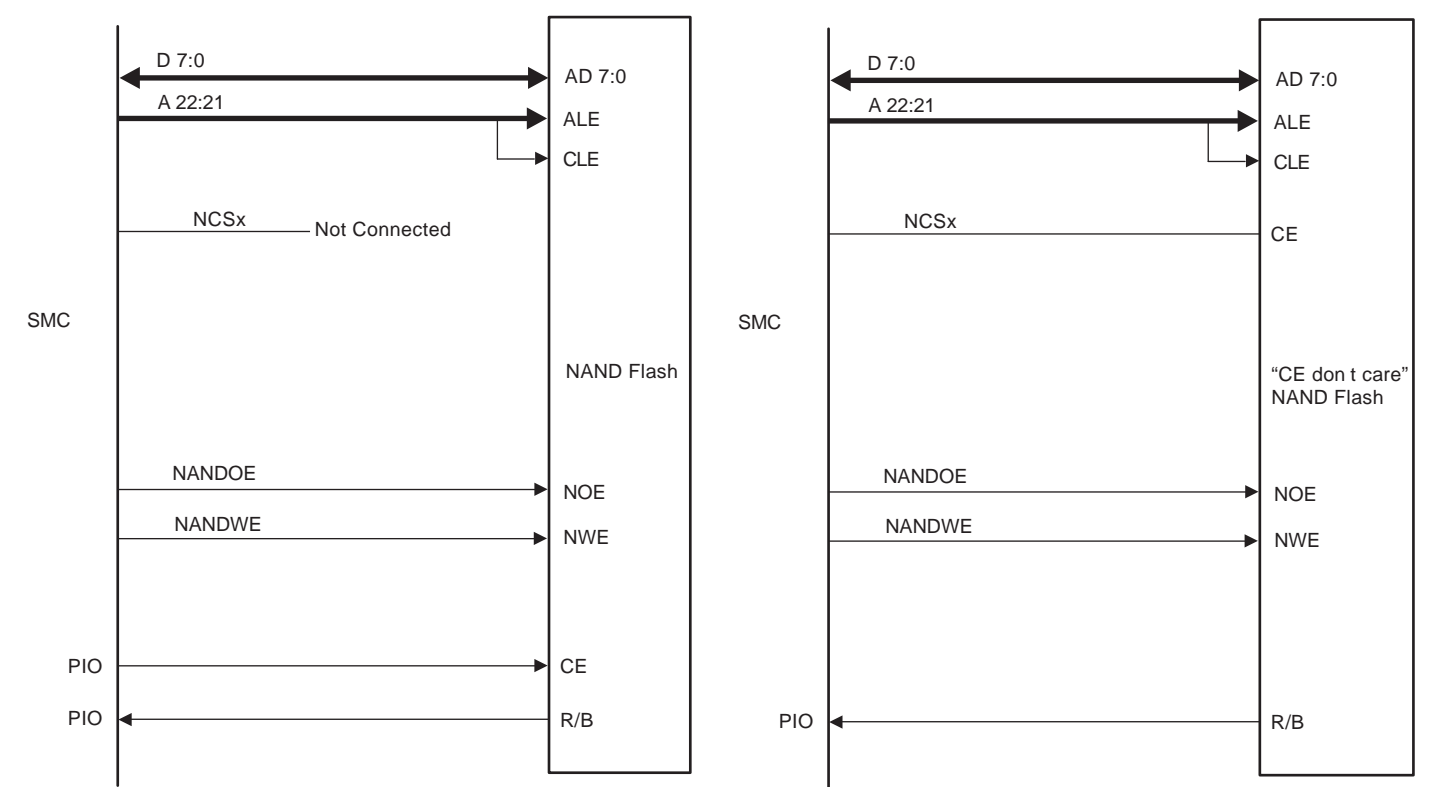
The address latch enable and command latch enable signals on the NAND Flash device are driven by address bits A22 and A21 of the address bus. Any bit of the address bus can also be used for this purpose. The command, address or data words on the data bus of the NAND Flash device use their own addresses within the NCSx address space (configured by CCFG\_SMCNFCS Register on the Bus Matrix User Interface). The chip enable (CE) signal of the device and the ready/busy (R/B) signals are connected to PIO lines. The CE signal then remains asserted even when NCS3 is not selected, preventing the device from returning to standby mode. The NANDCS output signal should be used in accordance with the external NAND Flash device type.

Two types of CE behavior exist depending on the NAND flash device:

- Standard NAND Flash devices require that the CE pin remains asserted Low continuously during the read busy period to prevent the device from returning to standby mode. Since the Static Memory Controller (SMC) asserts the NCSx signal High, it is necessary to connect the CE pin of the NAND Flash device to a GPIO line, in order to hold it low during the busy period preceding data read out.
- This restriction has been removed for “CE don’t care” NAND Flash devices. The NCSx signal can be directly connected to the CE pin of the NAND Flash device.

Figure 27-4 illustrates both topologies: Standard and “CE don’t care” NAND Flash.

Figure 27-4. Standard and “CE don’t care” NAND Flash Application Examples



## 27.8 Application Example

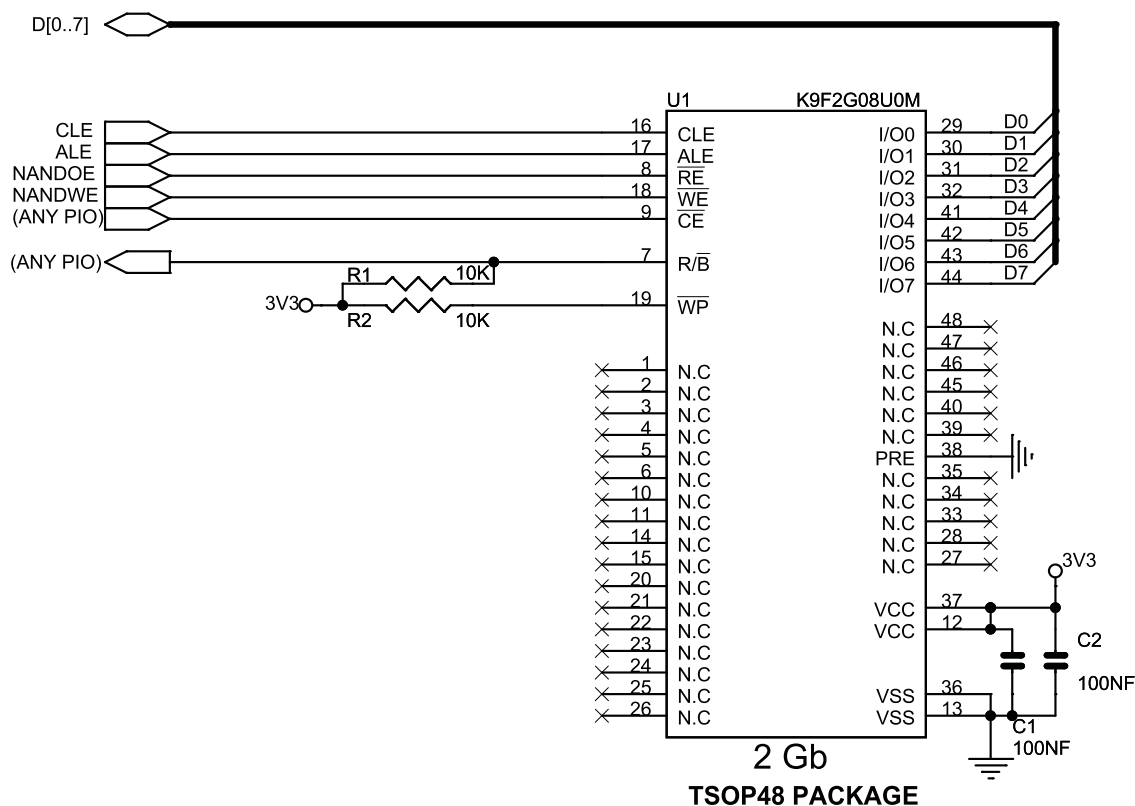
### 27.8.1 Implementation Examples

Hardware configurations are given for illustration only. The user should refer to the manufacturer web site to check for memory device availability.

For hardware implementation examples, refer to SAM4CM-EK schematics, which show examples of a connection to an LCD module and NAND Flash.

#### 27.8.1.1 8-bit NAND Flash

##### Hardware Configuration



##### Software Configuration

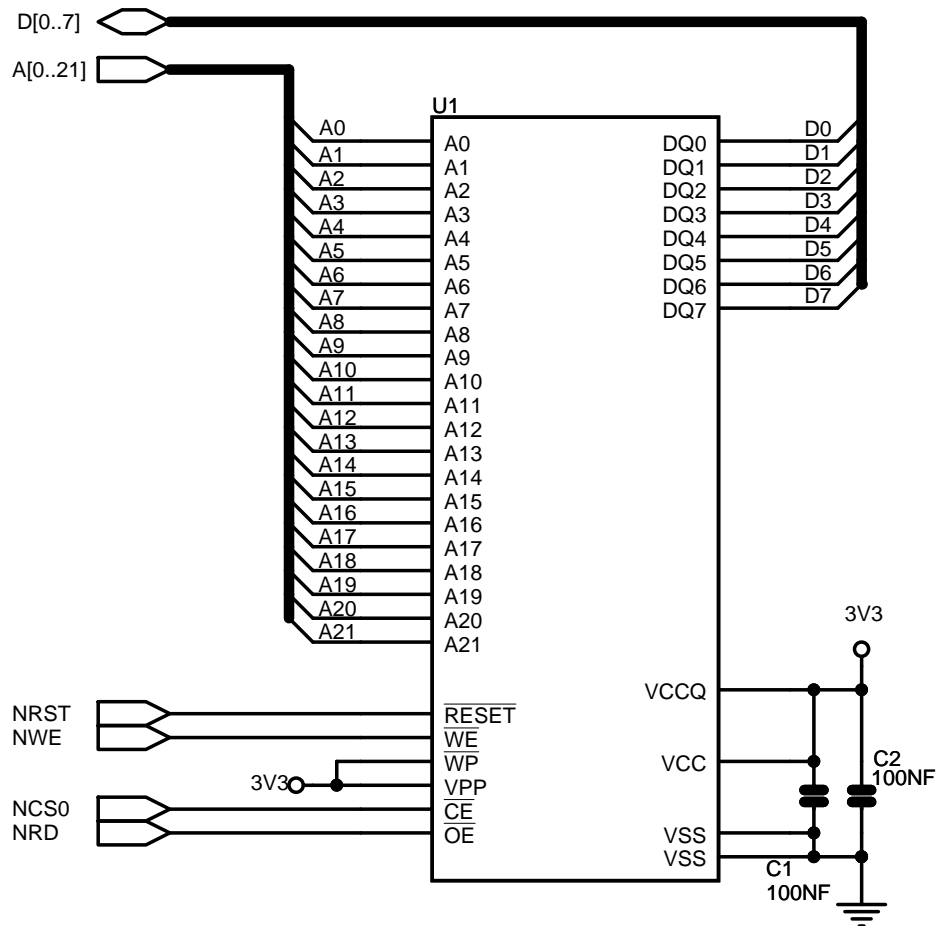
Perform the following configuration:

- Assign the SMC\_NFCSx (for example SMC\_NFCS3) field in the CCFG\_SMCNFCS Register on the Bus Matrix User Interface.
- Reserve A21 / A22 for ALE / CLE functions. Address and Command Latches are controlled respectively by setting to 1 the address bits A21 and A22 during accesses.
- NANDOE and NANDWE signals are multiplexed with PIO lines. Thus, the dedicated PIOs must be programmed in peripheral mode in the PIO controller.
- Configure a PIO line as an input to manage the Ready/Busy signal.
- Configure Static Memory Controller CS3 Setup, Pulse, Cycle and Mode according to NAND Flash timings, the data bus width and the system bus frequency.

In this example, the NAND Flash is not addressed as a “CE don’t care”. To address it as a “CE don’t care”, connect NCS3 (if SMC\_NFCS3 is set) to the NAND Flash CE.

### 27.8.1.2 NOR Flash

#### Hardware Configuration



#### Software Configuration

Configure the Static Memory Controller CS0 Setup, Pulse, Cycle and Mode depending on Flash timings and system bus frequency.

## 27.9 Standard Read and Write Protocols

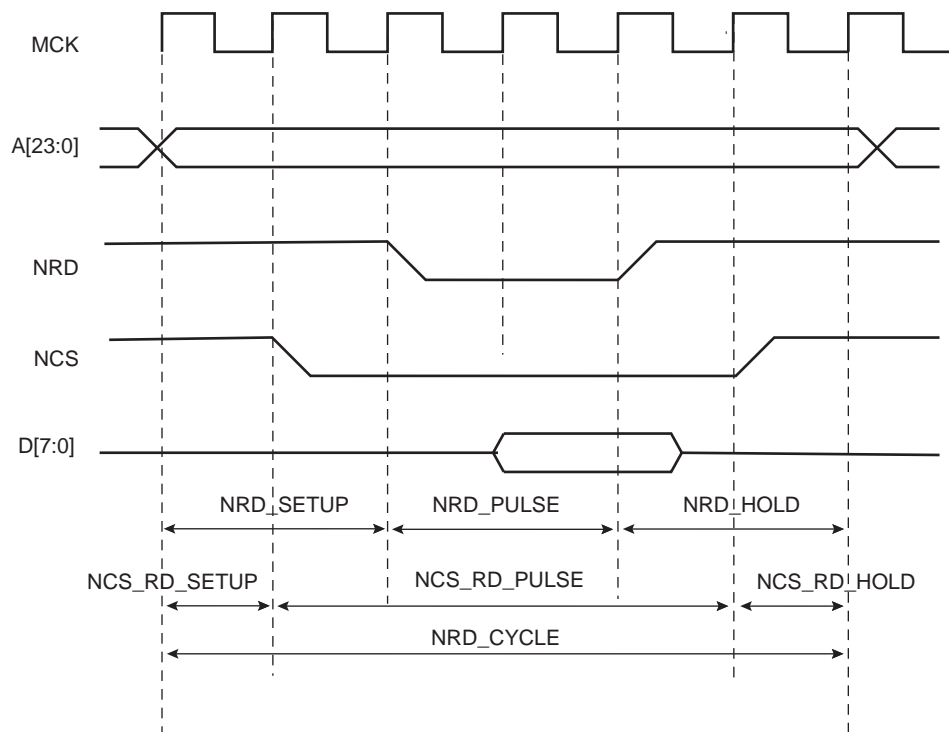
In the following sections, NCS represents one of the NCS[0..3] chip select lines.

### 27.9.1 Read Waveforms

The read cycle is shown on Figure 27-5.

The read cycle starts with the address setting on the memory address bus.

Figure 27-5. Standard Read Cycle



#### 27.9.1.1 NRD Waveform

The NRD signal is characterized by a setup timing, a pulse width and a hold timing.

1. NRD\_SETUP: the NRD setup time is defined as the setup of address before the NRD falling edge;
2. NRD\_PULSE: the NRD pulse length is the time between NRD falling edge and NRD rising edge;
3. NRD\_HOLD: the NRD hold time is defined as the hold time of address after the NRD rising edge.

#### 27.9.1.2 NCS Waveform

Similarly, the NCS signal can be divided into a setup time, pulse length and hold time:

1. NCS\_RD\_SETUP: the NCS setup time is defined as the setup time of address before the NCS falling edge.
2. NCS\_RD\_PULSE: the NCS pulse length is the time between NCS falling edge and NCS rising edge;
3. NCS\_RD\_HOLD: the NCS hold time is defined as the hold time of address after the NCS rising edge.

#### 27.9.1.3 Read Cycle

The NRD\_CYCLE time is defined as the total duration of the read cycle, i.e., from the time where address is set on the address bus to the point where address may change. The total read cycle time is equal to:

$$\begin{aligned} \text{NRD\_CYCLE} &= \text{NRD\_SETUP} + \text{NRD\_PULSE} + \text{NRD\_HOLD} \\ &= \text{NCS\_RD\_SETUP} + \text{NCS\_RD\_PULSE} + \text{NCS\_RD\_HOLD} \end{aligned}$$

All NRD and NCS timings are defined separately for each chip select as an integer number of Master Clock cycles. To ensure that the NRD and NCS timings are coherent, user must define the total read cycle instead of the hold timing. NRD\_CYCLE implicitly defines the NRD hold time and NCS hold time as:

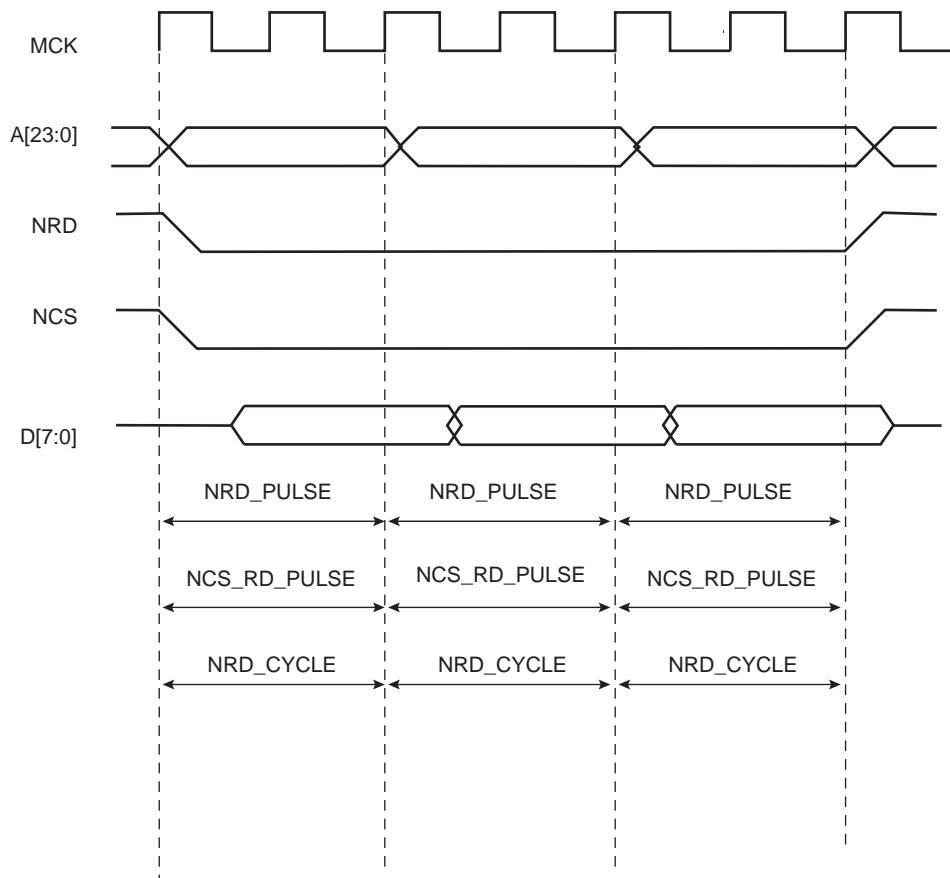
$$\text{NRD\_HOLD} = \text{NRD\_CYCLE} - \text{NRD\_SETUP} - \text{NRD\_PULSE}$$

$$\text{NCS\_RD\_HOLD} = \text{NRD\_CYCLE} - \text{NCS\_RD\_SETUP} - \text{NCS\_RD\_PULSE}$$

#### 27.9.1.4 Null Delay Setup and Hold

If null setup and hold parameters are programmed for NRD and/or NCS, NRD and NCS remain active continuously in case of consecutive read cycles in the same memory (see Figure 27-6).

**Figure 27-6. No Setup, No Hold on NRD and NCS Read Signals**



#### 27.9.1.5 Null Pulse

Programming null pulse is not permitted. Pulse must be at least set to 1. A null value leads to unpredictable behavior.

#### 27.9.2 Read Mode

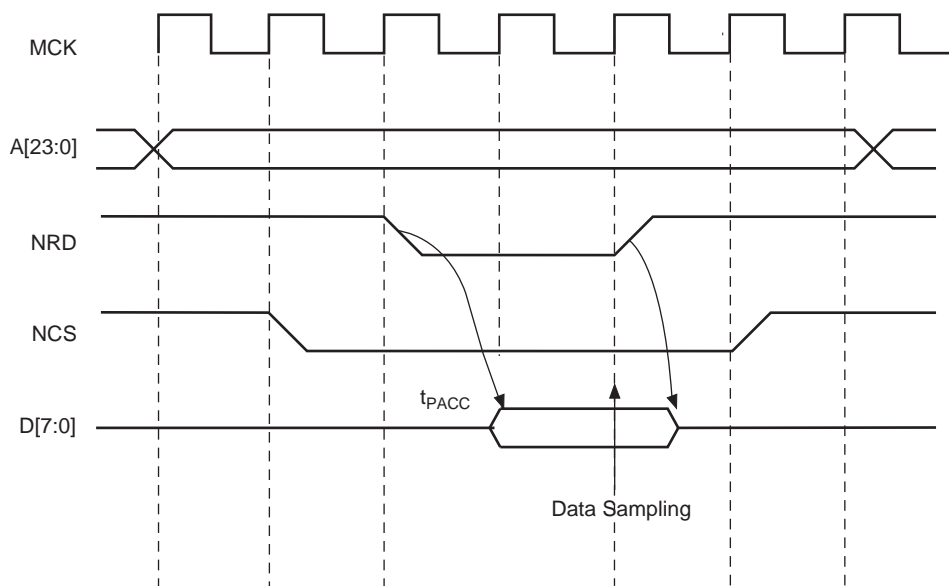
As NCS and NRD waveforms are defined independently of one other, the SMC needs to know when the read data is available on the data bus. The SMC does not compare NCS and NRD timings to know which signal rises first. The READ\_MODE parameter in the SMC\_MODE register of the corresponding chip select indicates which signal of NRD and NCS controls the read operation.



### 27.9.2.1 Read is Controlled by NRD (READ\_MODE = 1):

Figure 27-7 shows the waveforms of a read operation of a typical asynchronous RAM. The read data is available  $t_{PACC}$  after the falling edge of NRD, and turns to 'Z' after the rising edge of NRD. In this case, the READ\_MODE must be set to 1 (read is controlled by NRD), to indicate that data is available with the rising edge of NRD. The SMC samples the read data internally on the rising edge of Master Clock that generates the rising edge of NRD, whatever the programmed waveform of NCS may be.

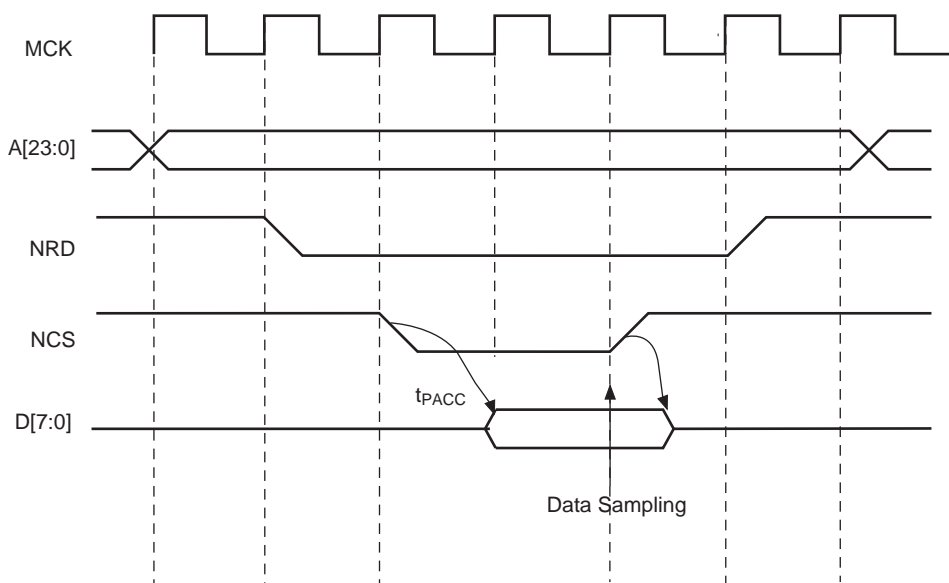
Figure 27-7. READ\_MODE = 1: Data is sampled by SMC before the rising edge of NRD



### 27.9.2.2 Read is Controlled by NCS (READ\_MODE = 0)

Figure 27-8 shows the typical read cycle of an LCD module. The read data is valid  $t_{PACC}$  after the falling edge of the NCS signal and remains valid until the rising edge of NCS. Data must be sampled when NCS is raised. In that case, the READ\_MODE must be set to 0 (read is controlled by NCS): the SMC internally samples the data on the rising edge of Master Clock that generates the rising edge of NCS, whatever the programmed waveform of NRD may be.

Figure 27-8. READ\_MODE = 0: Data is sampled by SMC before the rising edge of NCS



### 27.9.3 Write Waveforms

The write protocol is similar to the read protocol. It is depicted in Figure 27-9. The write cycle starts with the address setting on the memory address bus.

#### 27.9.3.1 NWE Waveforms

The NWE signal is characterized by a setup timing, a pulse width and a hold timing.

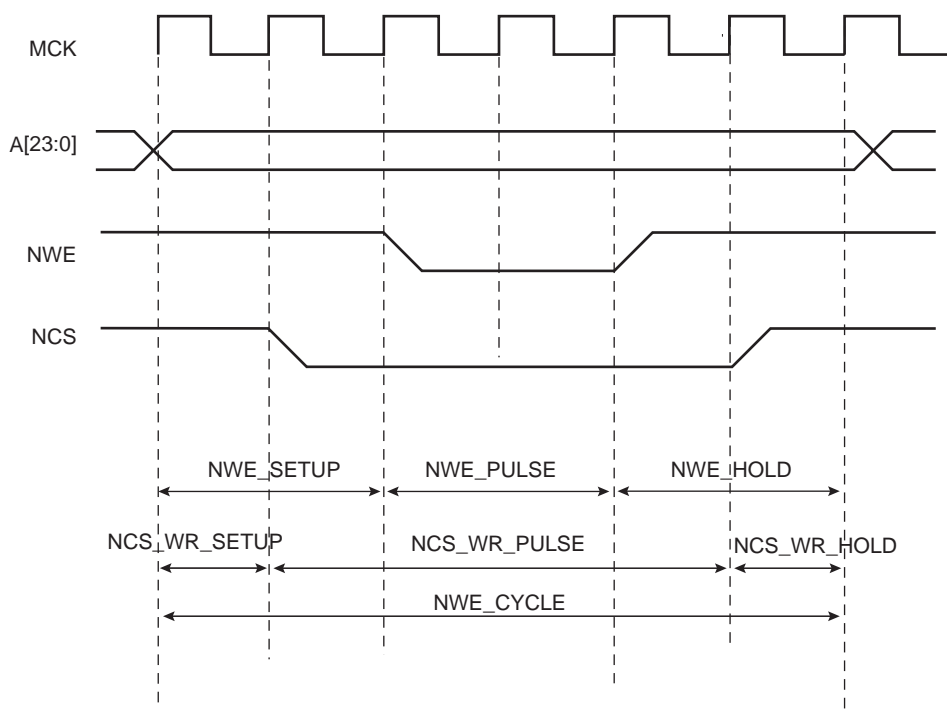
1. NWE\_SETUP: the NWE setup time is defined as the setup of address and data before the NWE falling edge;
2. NWE\_PULSE: The NWE pulse length is the time between NWE falling edge and NWE rising edge;
3. NWE\_HOLD: The NWE hold time is defined as the hold time of address and data after the NWE rising edge.

#### 27.9.3.2 NCS Waveforms

The NCS signal waveforms in write operation are not the same that those applied in read operations, but are separately defined:

1. NCS\_WR\_SETUP: the NCS setup time is defined as the setup time of address before the NCS falling edge.
2. NCS\_WR\_PULSE: the NCS pulse length is the time between NCS falling edge and NCS rising edge;
3. NCS\_WR\_HOLD: the NCS hold time is defined as the hold time of address after the NCS rising edge.

Figure 27-9. Write Cycle



#### 27.9.3.3 Write Cycle

The write\_cycle time is defined as the total duration of the write cycle, that is, from the time where address is set on the address bus to the point where address may change. The total write cycle time is equal to:

$$\begin{aligned} \text{NWE\_CYCLE} &= \text{NWE\_SETUP} + \text{NWE\_PULSE} + \text{NWE\_HOLD} \\ &= \text{NCS\_WR\_SETUP} + \text{NCS\_WR\_PULSE} + \text{NCS\_WR\_HOLD} \end{aligned}$$

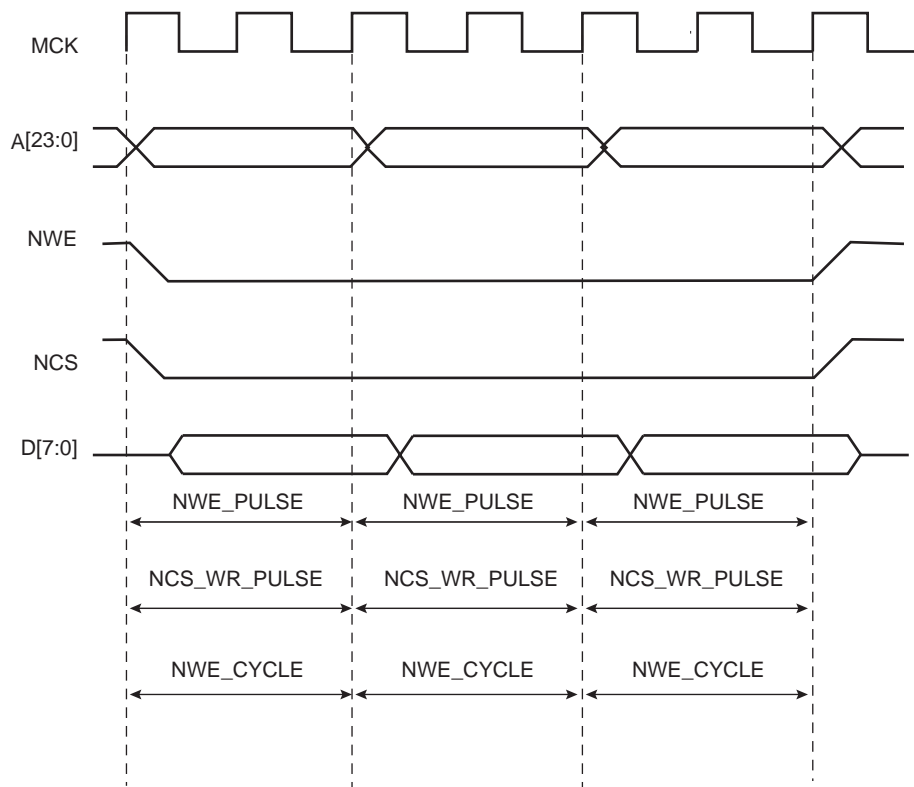
All NWE and NCS (write) timings are defined separately for each chip select as an integer number of Master Clock cycles. To ensure that the NWE and NCS timings are coherent, the user must define the total write cycle instead of the hold timing. This implicitly defines the NWE hold time and NCS (write) hold times as:

$$\begin{aligned} \text{NWE\_HOLD} &= \text{NWE\_CYCLE} - \text{NWE\_SETUP} - \text{NWE\_PULSE} \\ \text{NCS\_WR\_HOLD} &= \text{NWE\_CYCLE} - \text{NCS\_WR\_SETUP} - \text{NCS\_WR\_PULSE} \end{aligned}$$

### 27.9.3.4 Null Delay Setup and Hold

If null setup parameters are programmed for NWE and/or NCS, NWE and/or NCS remain active continuously in case of consecutive write cycles in the same memory (see [Figure 27-10](#)). However, for devices that perform write operations on the rising edge of NWE or NCS, such as SRAM, either a setup or a hold must be programmed.

**Figure 27-10. Null Setup and Hold Values of NCS and NWE in Write Cycle**



### 27.9.3.5 Null Pulse

Programming null pulse is not permitted. Pulse must be at least set to 1. A null value leads to unpredictable behavior.

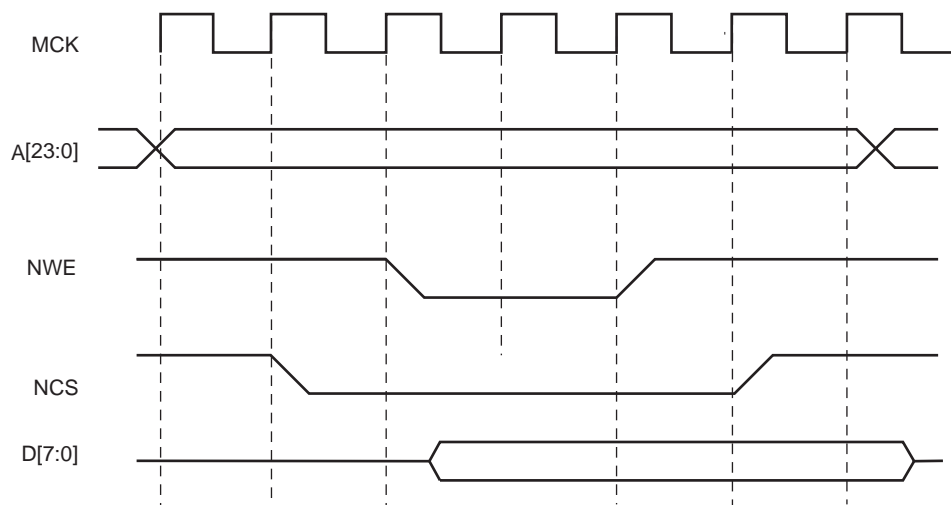
## 27.9.4 Write Mode

The **WRITE\_MODE** parameter in the **SMC\_MODE** register of the corresponding chip select indicates which signal controls the write operation.

### 27.9.4.1 Write is Controlled by NWE (**WRITE\_MODE** = 1):

[Figure 27-11](#) shows the waveforms of a write operation with **WRITE\_MODE** set to 1. The data is put on the bus during the pulse and hold steps of the NWE signal. The internal data buffers are switched to output mode after the **NWE\_SETUP** time, and until the end of the write cycle, regardless of the programmed waveform on NCS.

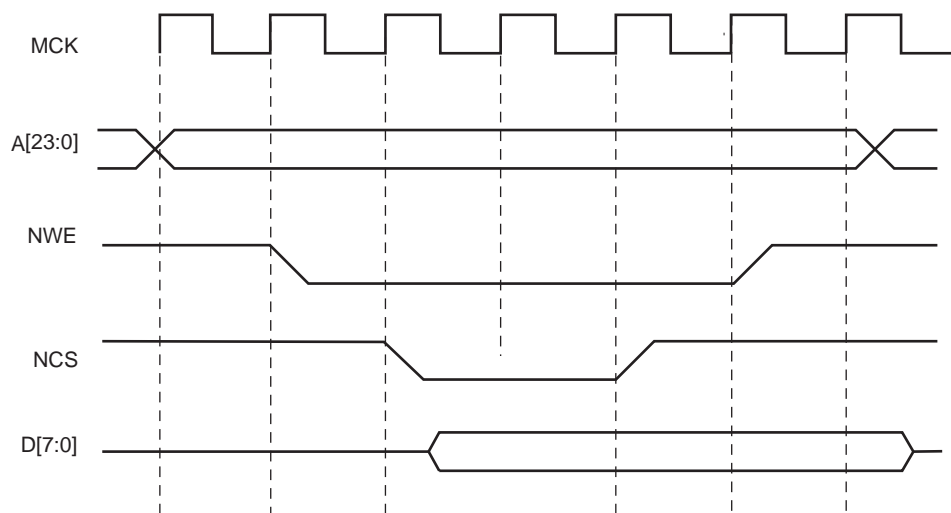
**Figure 27-11. WRITE\_MODE = 1. The write operation is controlled by NWE**



#### 27.9.4.2 Write is Controlled by NCS (WRITE\_MODE = 0)

Figure 27-12 shows the waveforms of a write operation with WRITE\_MODE set to 0. The data is put on the bus during the pulse and hold steps of the NCS signal. The internal data buffers are switched to output mode after the NCS\_WR\_SETUP time, and until the end of the write cycle, regardless of the programmed waveform on NWE.

**Figure 27-12. WRITE\_MODE = 0. The write operation is controlled by NCS**



### 27.9.5 Write Protected Registers

To prevent any single software error that may corrupt SMC behavior, the registers listed below can be write-protected by setting the WPEN bit in the SMC Write Protect Mode Register (SMC\_WPMR).

If a write access in a write-protected register is detected, then the WPVS flag in the SMC Write Protect Status Register (SMC\_WPSR) is set and the field WPVSR indicates in which register the write access has been attempted.

The WPVS flag is automatically reset after reading the SMC Write Protect Status Register (SMC\_WPSR).

List of the write-protected registers:

- [Section 27.16.1 "SMC Setup Register"](#)
- [Section 27.16.2 "SMC Pulse Register"](#)
- [Section 27.16.3 "SMC Cycle Register"](#)
- [Section 27.16.4 "SMC MODE Register"](#)

### 27.9.6 Coding Timing Parameters

All timing parameters are defined for one chip select and are grouped together in one SMC\_REGISTER according to their type.

The SMC\_SETUP register groups the definition of all setup parameters:

- NRD\_SETUP, NCS\_RD\_SETUP, NWE\_SETUP, NCS\_WR\_SETUP

The SMC\_PULSE register groups the definition of all pulse parameters:

- NRD\_PULSE, NCS\_RD\_PULSE, NWE\_PULSE, NCS\_WR\_PULSE

The SMC\_CYCLE register groups the definition of all cycle parameters:

- NRD\_CYCLE, NWE\_CYCLE

[Table 27-3](#) shows how the timing parameters are coded and their permitted range.

**Table 27-3. Coding and Range of Timing Parameters**

Coded Value	Number of Bits	Effective Value	Permitted Range	
			Coded Value	Effective Value
setup [5:0]	6	$128 \times \text{setup}[5] + \text{setup}[4:0]$	$0 \leq 31$	$0 \leq 128+31$
pulse [6:0]	7	$256 \times \text{pulse}[6] + \text{pulse}[5:0]$	$0 \leq 63$	$0 \leq 256+63$
cycle [8:0]	9	$256 \times \text{cycle}[8:7] + \text{cycle}[6:0]$	$0 \leq 127$	$0 \leq 256+127$ $0 \leq 512+127$ $0 \leq 768+127$

### 27.9.7 Reset Values of Timing Parameters

[Table 27-4](#) gives the default value of timing parameters at reset.

**Table 27-4. Reset Values of Timing Parameters**

Register	Reset Value	
SMC_SETUP		All setup timings are set to 1
SMC_PULSE		All pulse timings are set to 1
SMC_CYCLE		The read and write operation last 3 Master Clock cycles and provide one hold cycle
WRITE_MODE	1	Write is controlled with NWE
READ_MODE	1	Read is controlled with NRD

### 27.9.8 Usage Restriction

The SMC does not check the validity of the user-programmed parameters. If the sum of SETUP and PULSE parameters is larger than the corresponding CYCLE parameter, this leads to unpredictable behavior of the SMC.

For read operations:

Null but positive setup and hold of address and NRD and/or NCS can not be guaranteed at the memory interface because of the propagation delay of these signals through external logic and pads. If positive setup and hold values must be verified, then it is strictly recommended to program non-null values so as to cover possible skews between address, NCS and NRD signals.

For write operations:

If a null hold value is programmed on NWE, the SMC can guarantee a positive hold of address and NCS signal after the rising edge of NWE. This is true for WRITE\_MODE = 1 only. See [“Early Read Wait State” on page 463](#).

For read and write operations: a null value for pulse parameters is forbidden and may lead to unpredictable behavior.

In read and write cycles, the setup and hold time parameters are defined in reference to the address bus. For external devices that require setup and hold time between NCS and NRD signals (read), or between NCS and NWE signals (write), these setup and hold times must be converted into setup and hold times in reference to the address bus.

## 27.10 Scrambling/Unscrambling Function

The external data bus D[7:0] can be scrambled in order to prevent intellectual property data located in off-chip memories from being easily recovered by analyzing data at the package pin level of either microcontroller or memory device.

The scrambling and unscrambling are performed on-the-fly without additional wait states.

The scrambling method depends on two user-configurable key registers, SMC\_KEY1 and SMC\_KEY2. These key registers are only accessible in write mode.

The key must be securely stored in a reliable non-volatile memory in order to recover data from the off-chip memory. Any data scrambled with a given key cannot be recovered if the key is lost.

The scrambling/unscrambling function can be enabled or disabled by programming the SMC\_OCMS register.

When multiple chip selects are handled, it is possible to configure the scrambling function per chip select using the OCMS field in the SMC\_OCMS registers.

## 27.11 Automatic Wait States

Under certain circumstances, the SMC automatically inserts idle cycles between accesses to avoid bus contention or operation conflict.

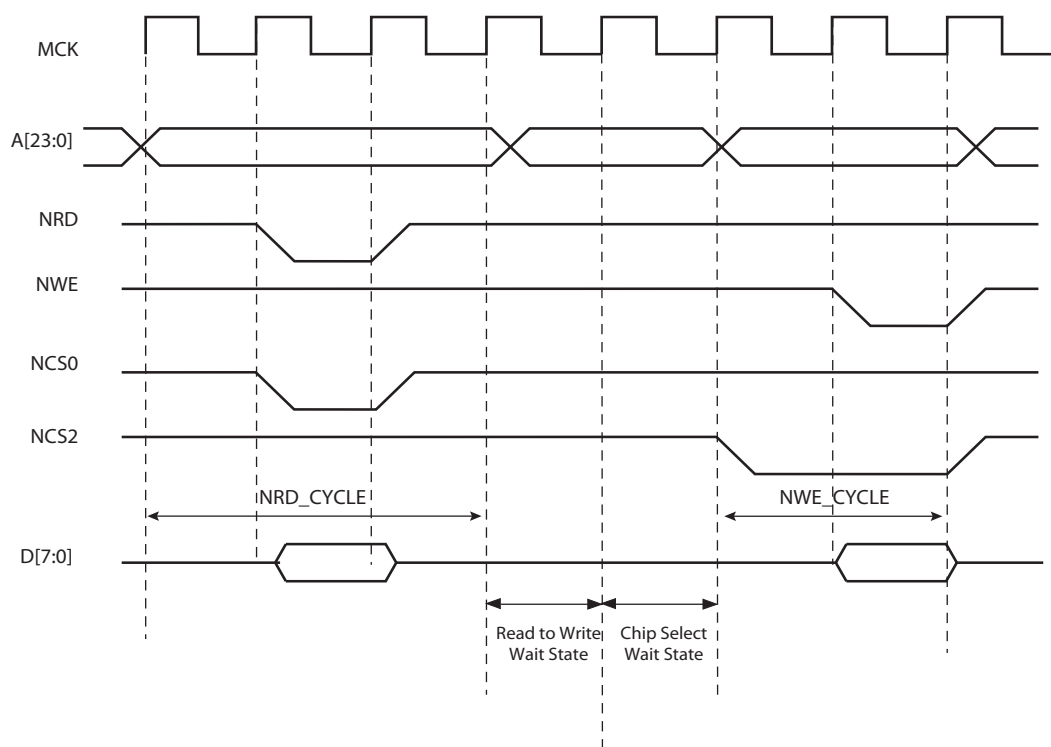
### 27.11.1 Chip Select Wait States

The SMC always inserts an idle cycle between 2 transfers on separate chip selects. This idle cycle ensures that there is no bus contention between the de-activation of one device and the activation of the next one.

During chip select wait state, all control lines are turned inactive: NWR, NCS[0..3], NRD lines are all set to 1.

[Figure 27-13](#) illustrates a chip select wait state between access on Chip Select 0 and Chip Select 2.

**Figure 27-13. Chip Select Wait State between a Read Access on NCS0 and a Write Access on NCS2**



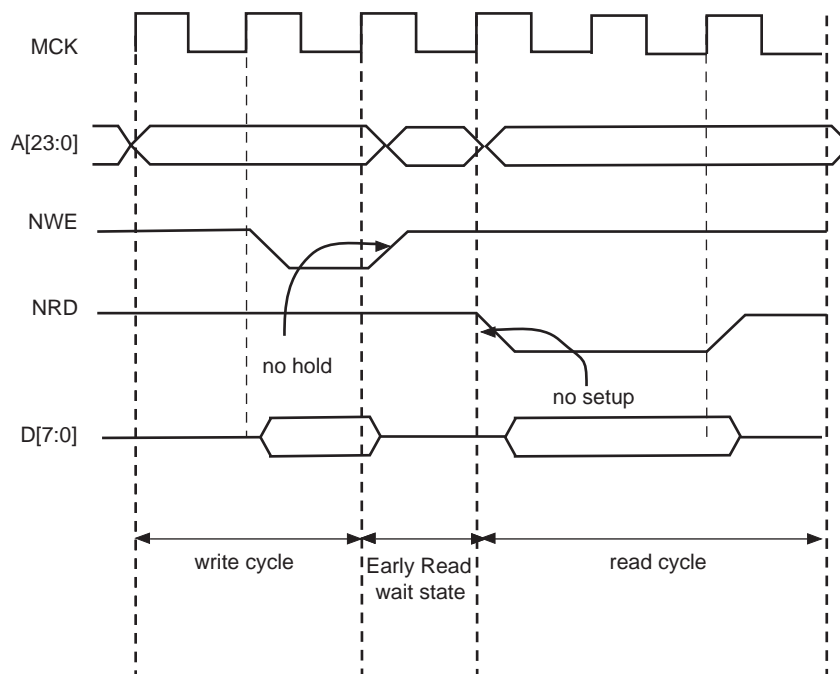
### 27.11.2 Early Read Wait State

In some cases, the SMC inserts a wait state cycle between a write access and a read access to allow time for the write cycle to end before the subsequent read cycle begins. This wait state is not generated in addition to a chip select wait state. The early read cycle thus only occurs between a write and read access to the same memory device (same chip select).

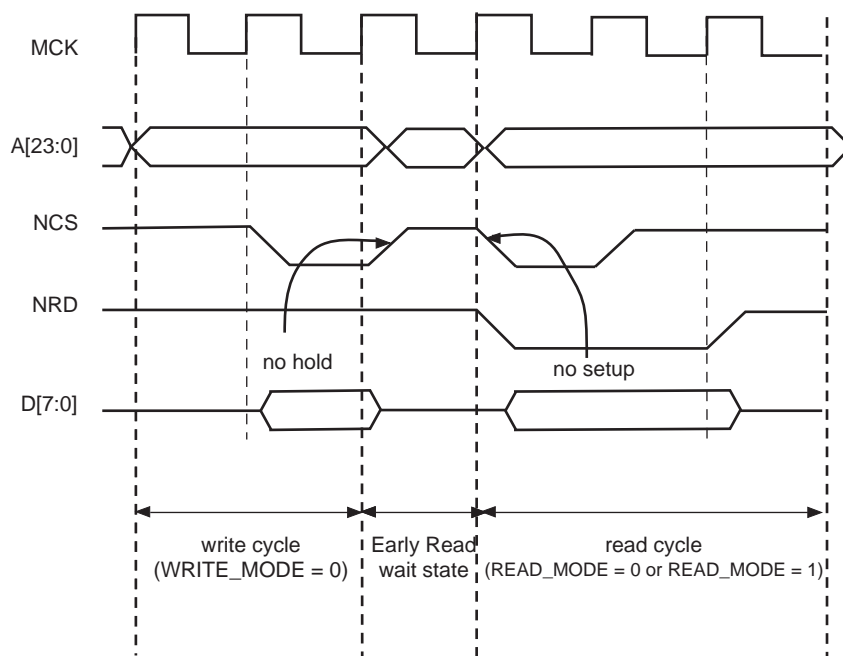
An early read wait state is automatically inserted if at least one of the following conditions is valid:

- if the write controlling signal has no hold time and the read controlling signal has no setup time ([Figure 27-14](#)).
- in NCS write controlled mode (`WRITE_MODE = 0`), if there is no hold timing on the NCS signal and the `NCS_RD_SETUP` parameter is set to 0, regardless of the read mode ([Figure 27-15](#)). The write operation must end with a NCS rising edge. Without an Early Read Wait State, the write operation could not complete properly.
- in NWE controlled mode (`WRITE_MODE = 1`) and if there is no hold timing (`NWE_HOLD = 0`), the feedback of the write control signal is used to control address, data, and chip select lines. If the external write control signal is not inactivated as expected due to load capacitances, an Early Read Wait State is inserted and address, data and control signals are maintained one more cycle. See [Figure 27-16](#).

**Figure 27-14. Early Read Wait State: Write with No Hold Followed by Read with No Setup**

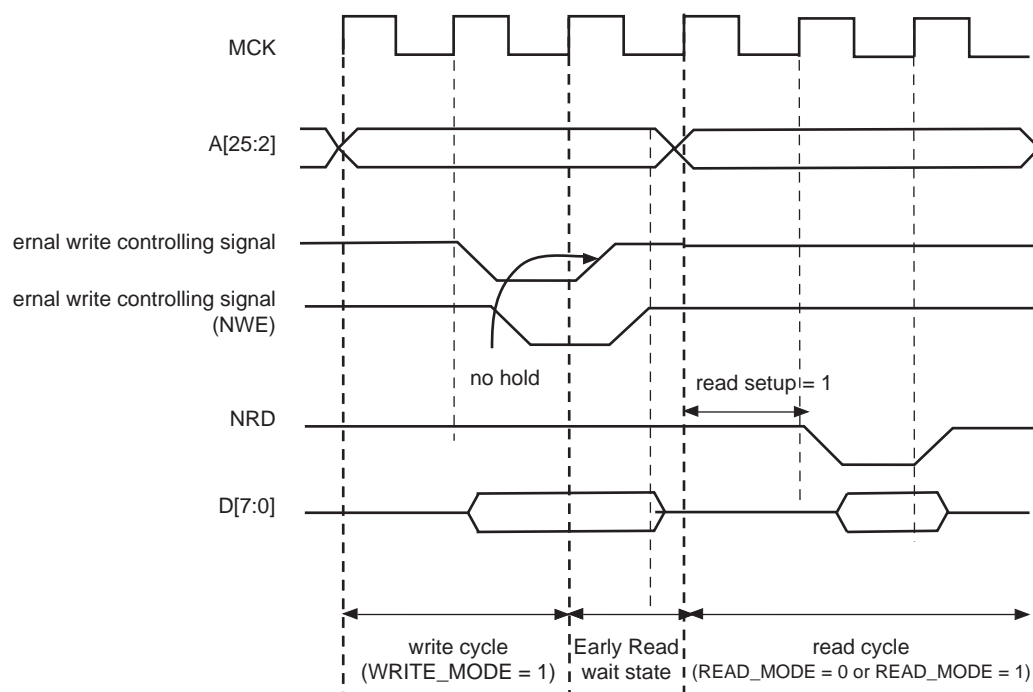


**Figure 27-15. Early Read Wait State: NCS Controlled Write with No Hold Followed by a Read with No NCS Setup**





**Figure 27-16. Early Read Wait State: NWE-controlled Write with No Hold Followed by a Read with one Set-up Cycle**



### 27.11.3 Reload User Configuration Wait State

The user may change any of the configuration parameters by writing the SMC user interface.

When detecting that a new user configuration has been written in the user interface, the SMC inserts a wait state before starting the next access. The so called “Reload User Configuration Wait State” is used by the SMC to load the new set of parameters to apply to next accesses.

The Reload Configuration Wait State is not applied in addition to the Chip Select Wait State. If accesses before and after re-programming the user interface are made to different devices (Chip Selects), then one single Chip Select Wait State is applied.

On the other hand, if accesses before and after writing the user interface are made to the same device, a Reload Configuration Wait State is inserted, even if the change does not concern the current Chip Select.

#### 27.11.3.1 User Procedure

To insert a Reload Configuration Wait State, the SMC detects a write access to any SMC\_MODE register of the user interface. If the user only modifies timing registers (SMC\_SETUP, SMC\_PULSE, SMC\_CYCLE registers) in the user interface, he must validate the modification by writing the SMC\_MODE, even if no change was made on the mode parameters.

The user must not change the configuration parameters of an SMC Chip Select (Setup, Pulse, Cycle, Mode) if accesses are performed on this CS during the modification. Any change of the Chip Select parameters, while fetching the code from a memory connected on this CS, may lead to unpredictable behavior. The instructions used to modify the parameters of an SMC Chip Select can be executed from the internal RAM or from a memory connected to another CS.

#### 27.11.3.2 Slow Clock Mode Transition

A Reload Configuration Wait State is also inserted when the Slow Clock Mode is entered or exited, after the end of the current transfer (see [“Slow Clock Mode” on page 476](#)).

#### 27.11.4 Read to Write Wait State

Due to an internal mechanism, a wait cycle is always inserted between consecutive read and write SMC accesses.

This wait cycle is referred to as a read to write wait state in this document.

This wait cycle is applied in addition to chip select and reload user configuration wait states when they are to be inserted. See [Figure 27-13 on page 463](#).

### 27.12 Data Float Wait States

Some memory devices are slow to release the external bus. For such devices, it is necessary to add wait states (data float wait states) after a read access:

- before starting a read access to a different external memory
- before starting a write access to the same device or to a different external one.

The Data Float Output Time ( $t_{DF}$ ) for each external memory device is programmed in the TDF\_CYCLES field of the SMC\_MODE register for the corresponding chip select. The value of TDF\_CYCLES indicates the number of data float wait cycles (between 0 and 15) before the external device releases the bus, and represents the time allowed for the data output to go to high impedance after the memory is disabled.

Data float wait states do not delay internal memory accesses. Hence, a single access to an external memory with long  $t_{DF}$  will not slow down the execution of a program from internal memory.

The data float wait states management depends on the READ\_MODE and the TDF\_MODE fields of the SMC\_MODE register for the corresponding chip select.

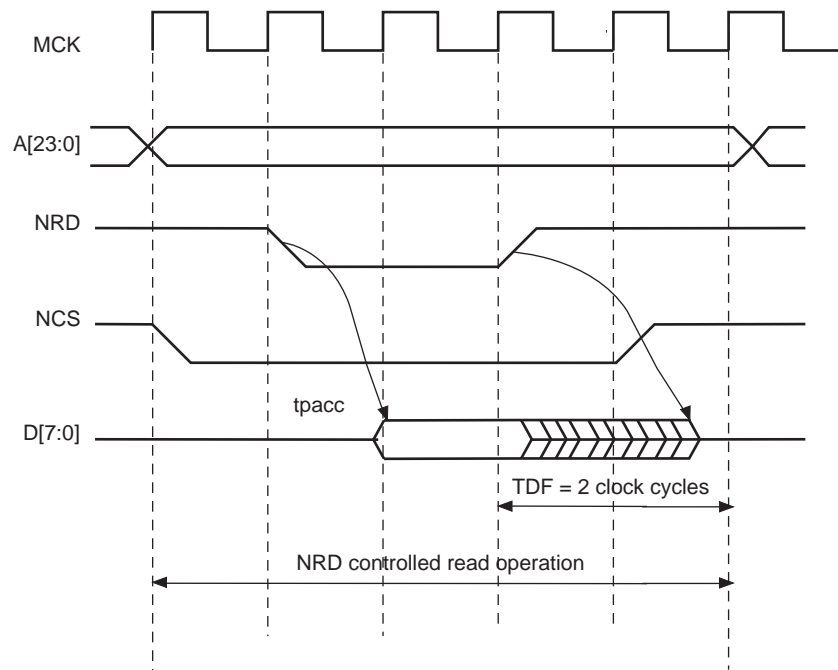
#### 27.12.1 READ\_MODE

Setting the READ\_MODE to 1 indicates to the SMC that the NRD signal is responsible for turning off the tri-state buffers of the external memory device. The Data Float Period then begins after the rising edge of the NRD signal and lasts TDF\_CYCLES MCK cycles.

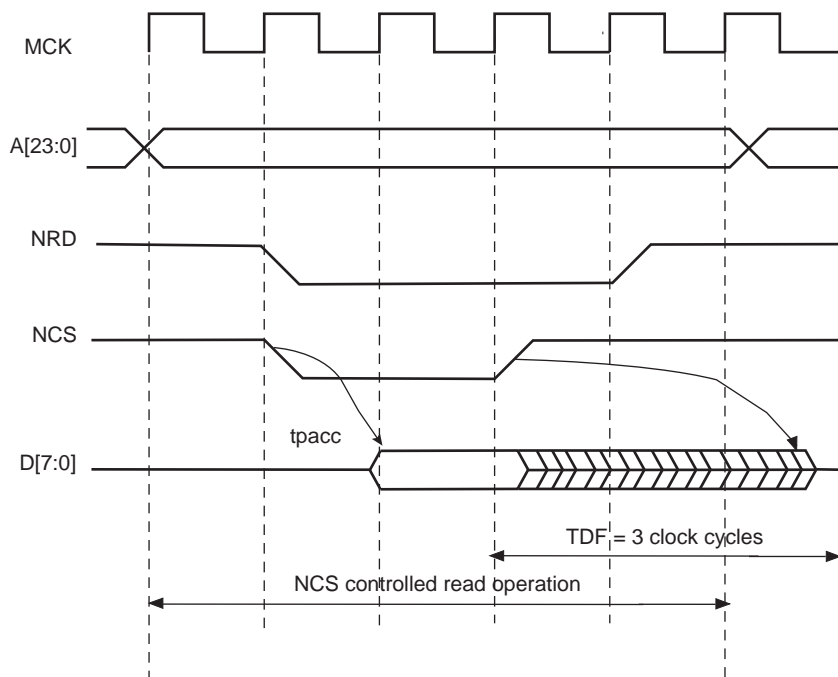
When the read operation is controlled by the NCS signal (READ\_MODE = 0), the TDF field gives the number of MCK cycles during which the data bus remains busy after the rising edge of NCS.

[Figure 27-17](#) illustrates the Data Float Period in NRD-controlled mode (READ\_MODE = 1), assuming a data float period of 2 cycles (TDF\_CYCLES = 2). [Figure 27-18](#) shows the read operation when controlled by NCS (READ\_MODE = 0) and the TDF\_CYCLES parameter equals 3.

**Figure 27-17. TDF Period in NRD Controlled Read Access (TDF = 2)**



**Figure 27-18. TDF Period in NCS Controlled Read Operation (TDF = 3)**



### 27.12.2 TDF Optimization Enabled (TDF\_MODE = 1)

When the TDF\_MODE of the SMC\_MODE register is set to 1 (TDF optimization is enabled), the SMC takes advantage of the setup period of the next access to optimize the number of wait states cycle to insert.

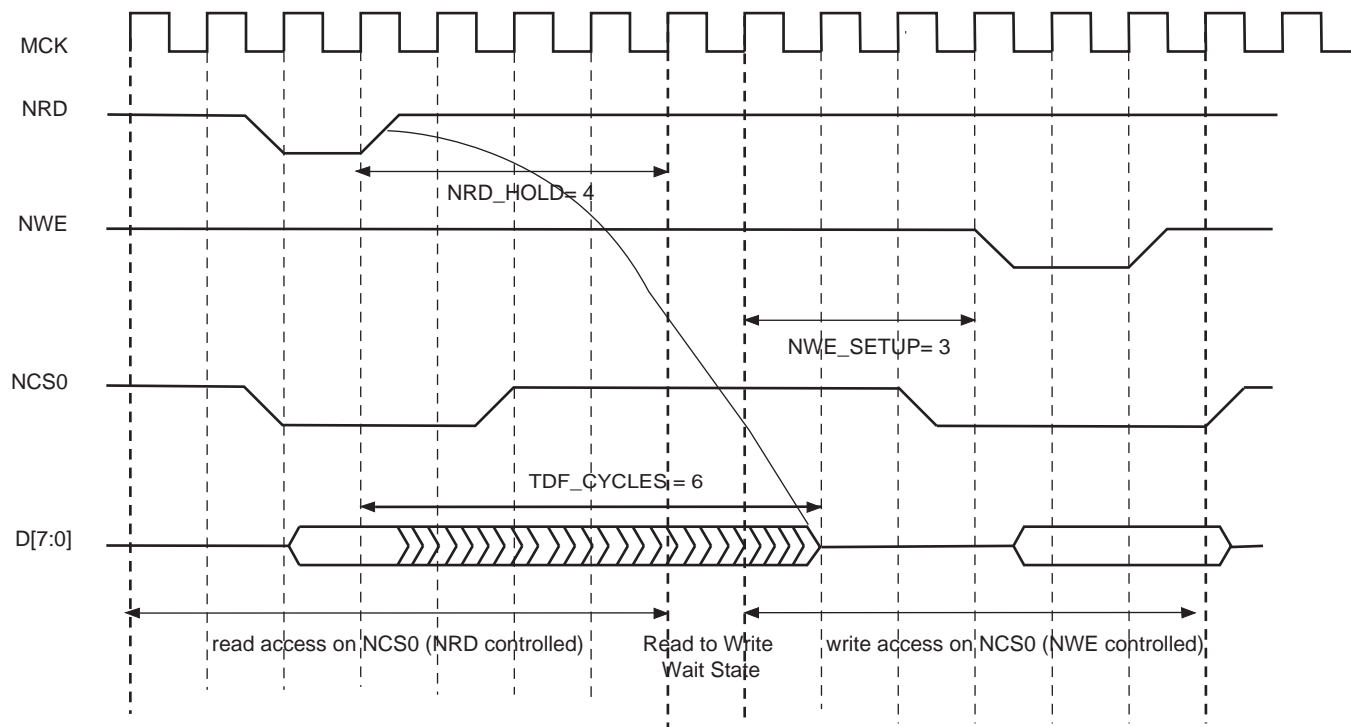
Figure 27-19 shows a read access controlled by NRD, followed by a write access controlled by NWE, on Chip Select 0. Chip Select 0 has been programmed with:

NRD\_HOLD = 4; READ\_MODE = 1 (NRD controlled)

NWE\_SETUP = 3; WRITE\_MODE = 1 (NWE controlled)

TDF\_CYCLES = 6; TDF\_MODE = 1 (optimization enabled).

**Figure 27-19. TDF Optimization: No TDF wait states are inserted if the TDF period is over when the next access begins**



### 27.12.3 TDF Optimization Disabled (TDF\_MODE = 0)

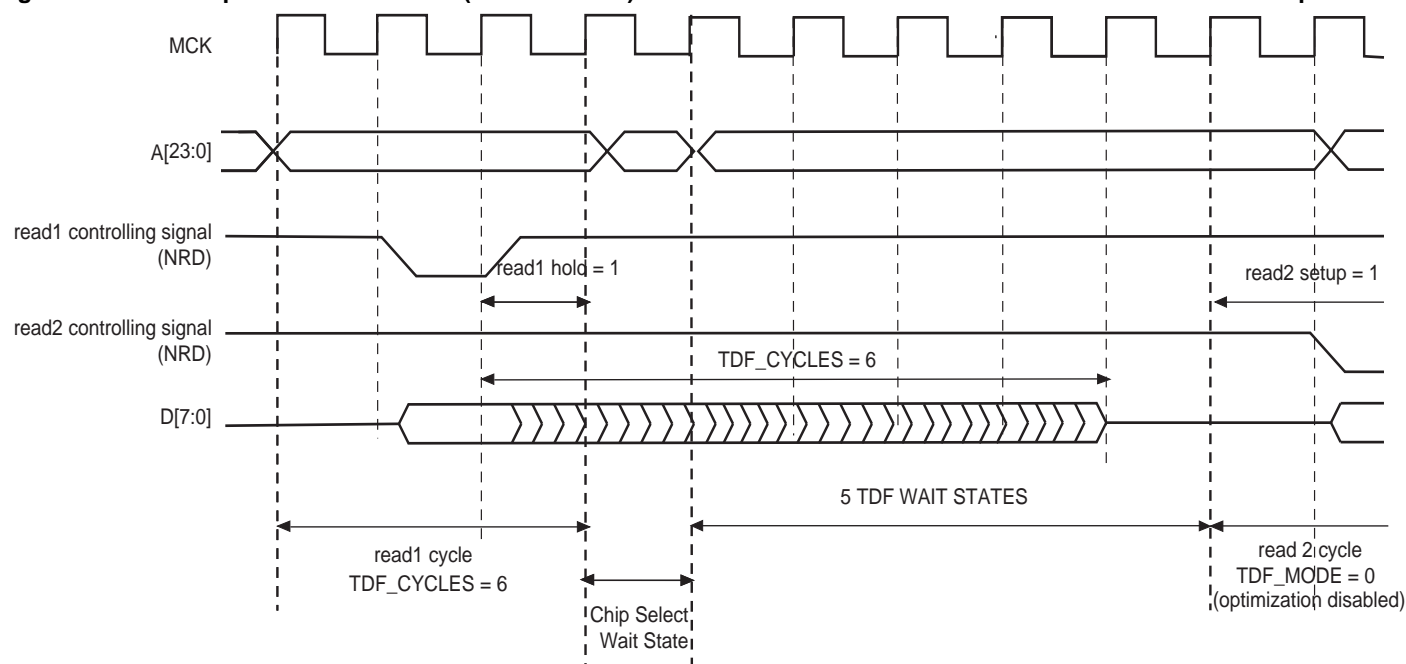
When optimization is disabled, tdf wait states are inserted at the end of the read transfer, so that the data float period is ended when the second access begins. If the hold period of the read1 controlling signal overlaps the data float period, no additional tdf wait states will be inserted.

Figure 27-20, Figure 27-21 and Figure 27-22 illustrate the cases:

- Read access followed by a read access on another chip select,
- Read access followed by a write access on another chip select,
- Read access followed by a write access on the same chip select,

with no TDF optimization.

**Figure 27-20. TDF Optimization Disabled (TDF Mode = 0). TDF wait states between 2 read accesses on different chip selects**



**Figure 27-21. TDF Mode = 0: TDF wait states between a read and a write access on different chip selects**

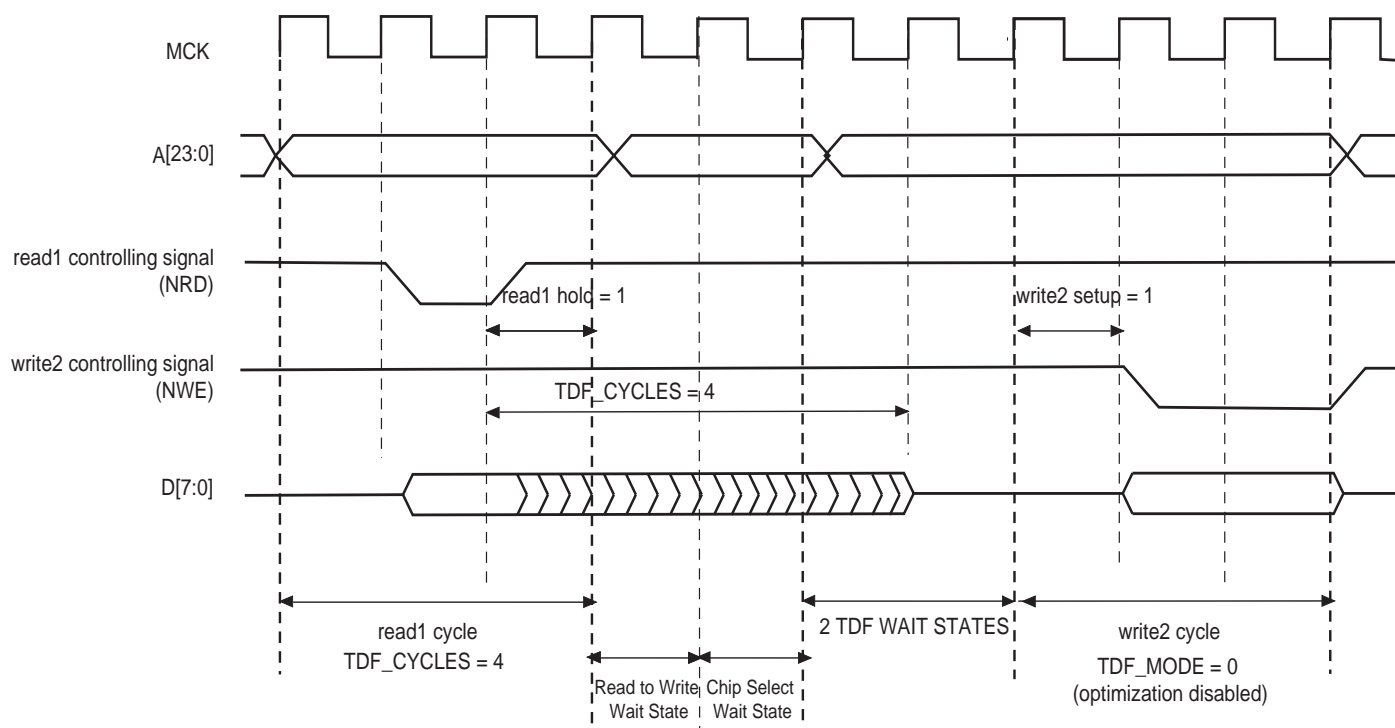
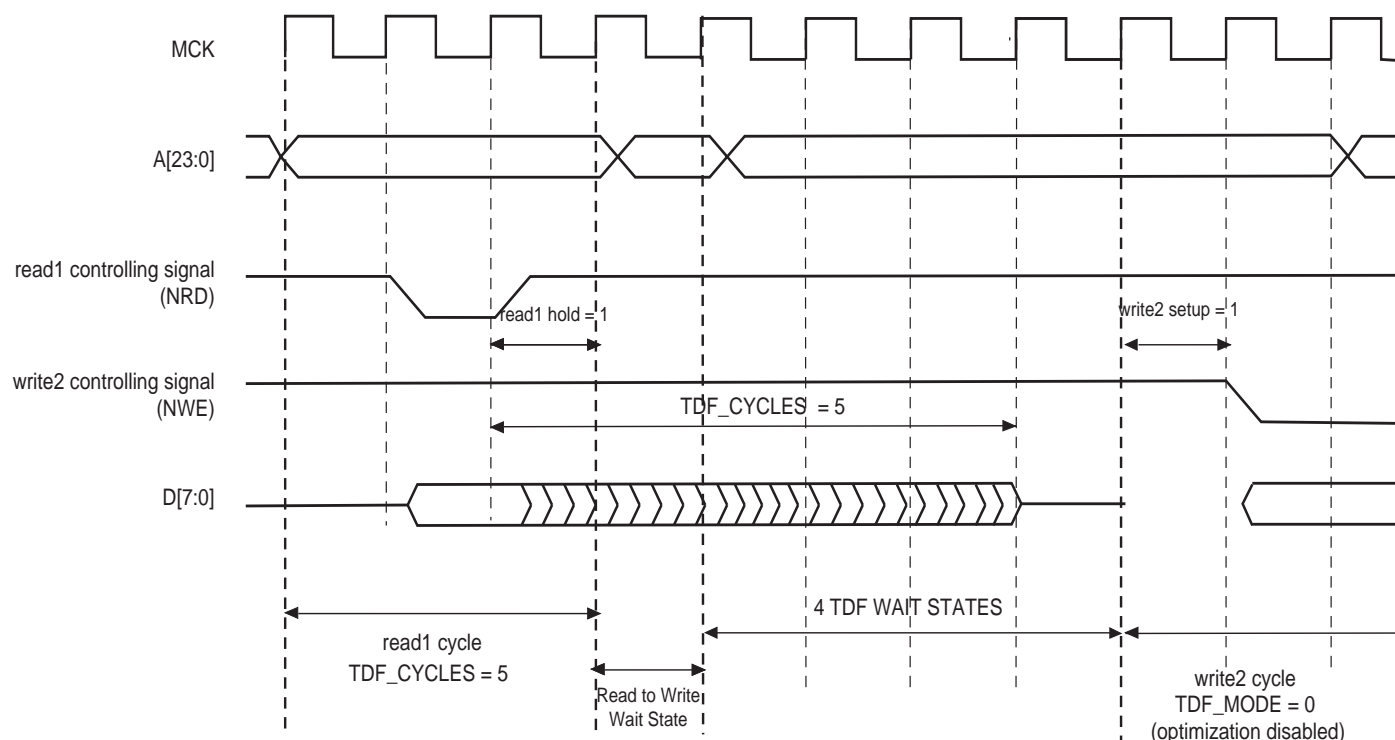


Figure 27-22. TDF Mode = 0: TDF wait states between read and write accesses on the same chip select



## 27.13 External Wait

Any access can be extended by an external device using the NWAIT input signal of the SMC. The EXNW\_MODE field of the SMC\_MODE register on the corresponding chip select must be set to either “10” (frozen mode) or “11” (ready mode). When the EXNW\_MODE is set to “00” (disabled), the NWAIT signal is simply ignored on the corresponding chip select. The NWAIT signal delays the read or write operation in regards to the read or write controlling signal, depending on the read and write modes of the corresponding chip select.

### 27.13.1 Restriction

When one of the EXNW\_MODE is enabled, it is mandatory to program at least one hold cycle for the read/write controlling signal. For that reason, the NWAIT signal cannot be used in Page Mode ([“Asynchronous Page Mode” on page 478](#)), or in Slow Clock Mode ([“Slow Clock Mode” on page 476](#)).

The NWAIT signal is assumed to be a response of the external device to the read/write request of the SMC. Then NWAIT is examined by the SMC only in the pulse state of the read or write controlling signal. The assertion of the NWAIT signal outside the expected period has no impact on SMC behavior.

### 27.13.2 Frozen Mode

When the external device asserts the NWAIT signal (active low), and after internal synchronization of this signal, the SMC state is frozen, i.e., SMC internal counters are frozen, and all control signals remain unchanged. When the resynchronized NWAIT signal is deasserted, the SMC completes the access, resuming the access from the point where it was stopped. See [Figure 27-23](#). This mode must be selected when the external device uses the NWAIT signal to delay the access and to freeze the SMC.

The assertion of the NWAIT signal outside the expected period is ignored as illustrated in [Figure 27-24](#).

**Figure 27-23. Write Access with NWAIT Assertion in Frozen Mode (EXNW\_MODE = 10)**

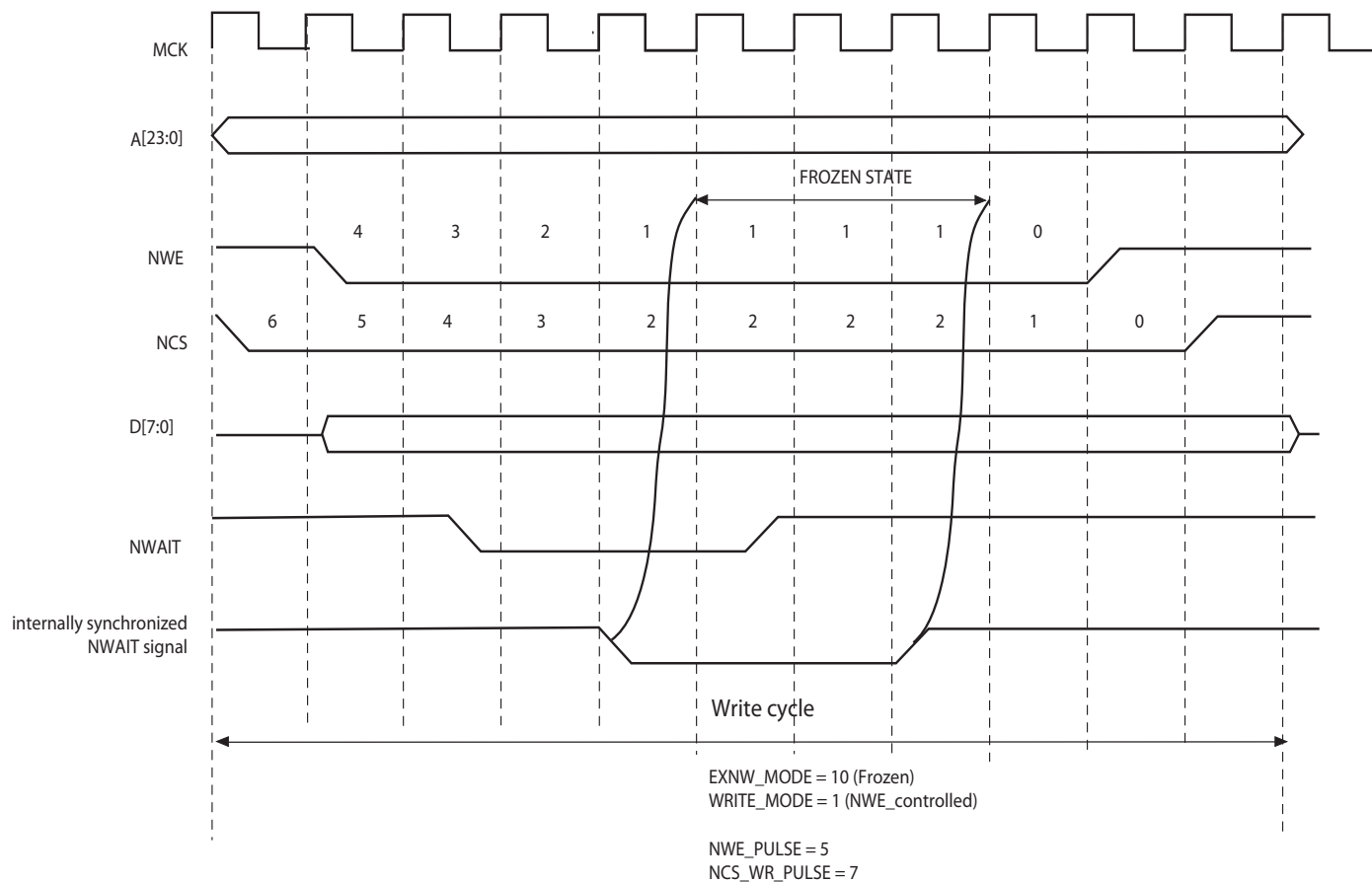
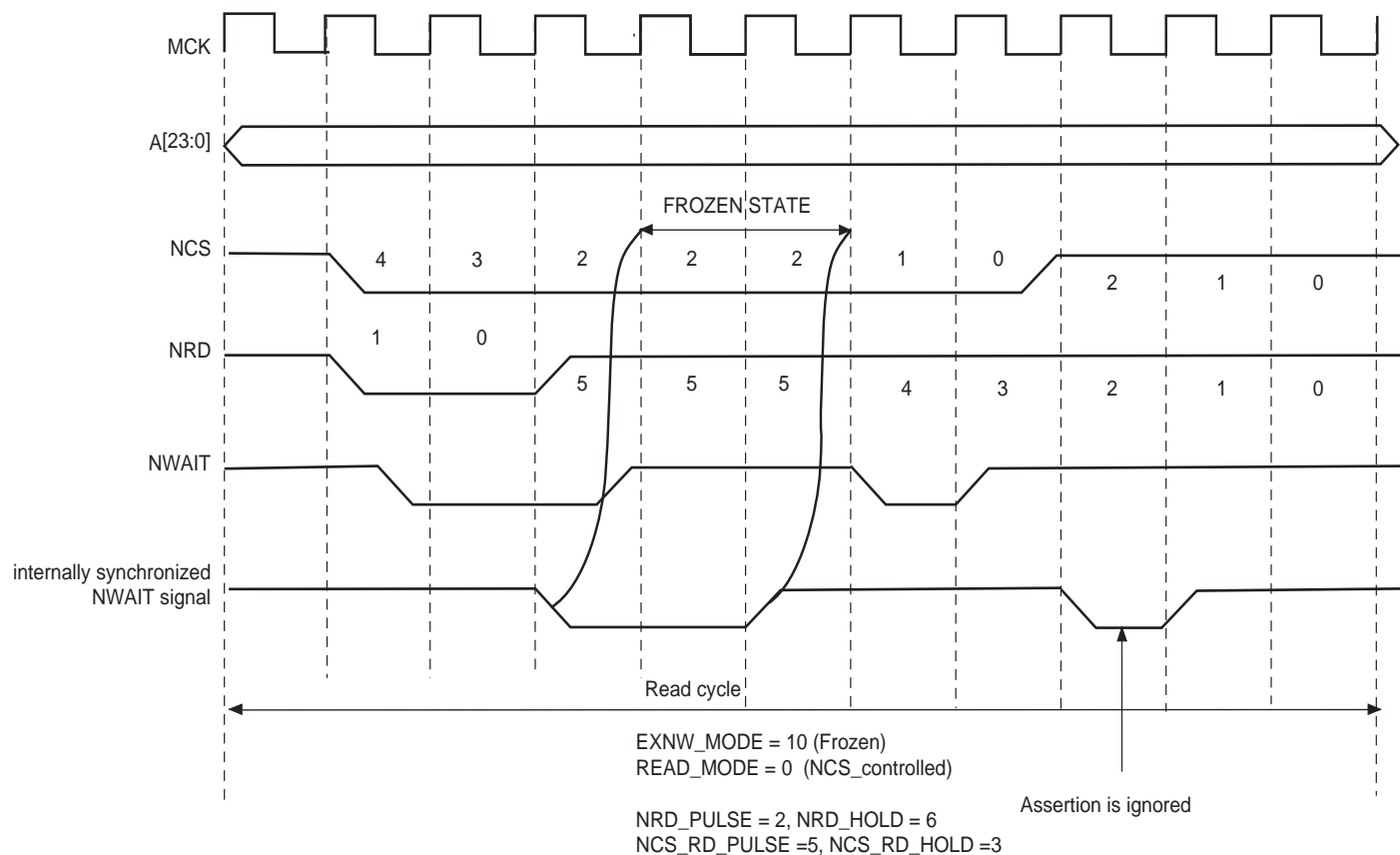


Figure 27-24. Read Access with NWAIT Assertion in Frozen Mode (EXNW\_MODE = 10)





### 27.13.3 Ready Mode

In Ready mode (EXNW\_MODE = 11), the SMC behaves differently. Normally, the SMC begins the access by down counting the setup and pulse counters of the read/write controlling signal. In the last cycle of the pulse phase, the resynchronized NWAIT signal is examined.

If asserted, the SMC suspends the access as shown in Figure 27-25 and Figure 27-26. After deassertion, the access is completed: the hold step of the access is performed.

This mode must be selected when the external device uses deassertion of the NWAIT signal to indicate its ability to complete the read or write operation.

If the NWAIT signal is deasserted before the end of the pulse, or asserted after the end of the pulse of the controlling read/write signal, it has no impact on the access length as shown in Figure 27-26.

**Figure 27-25. NWAIT Assertion in Write Access: Ready Mode (EXNW\_MODE = 11)**

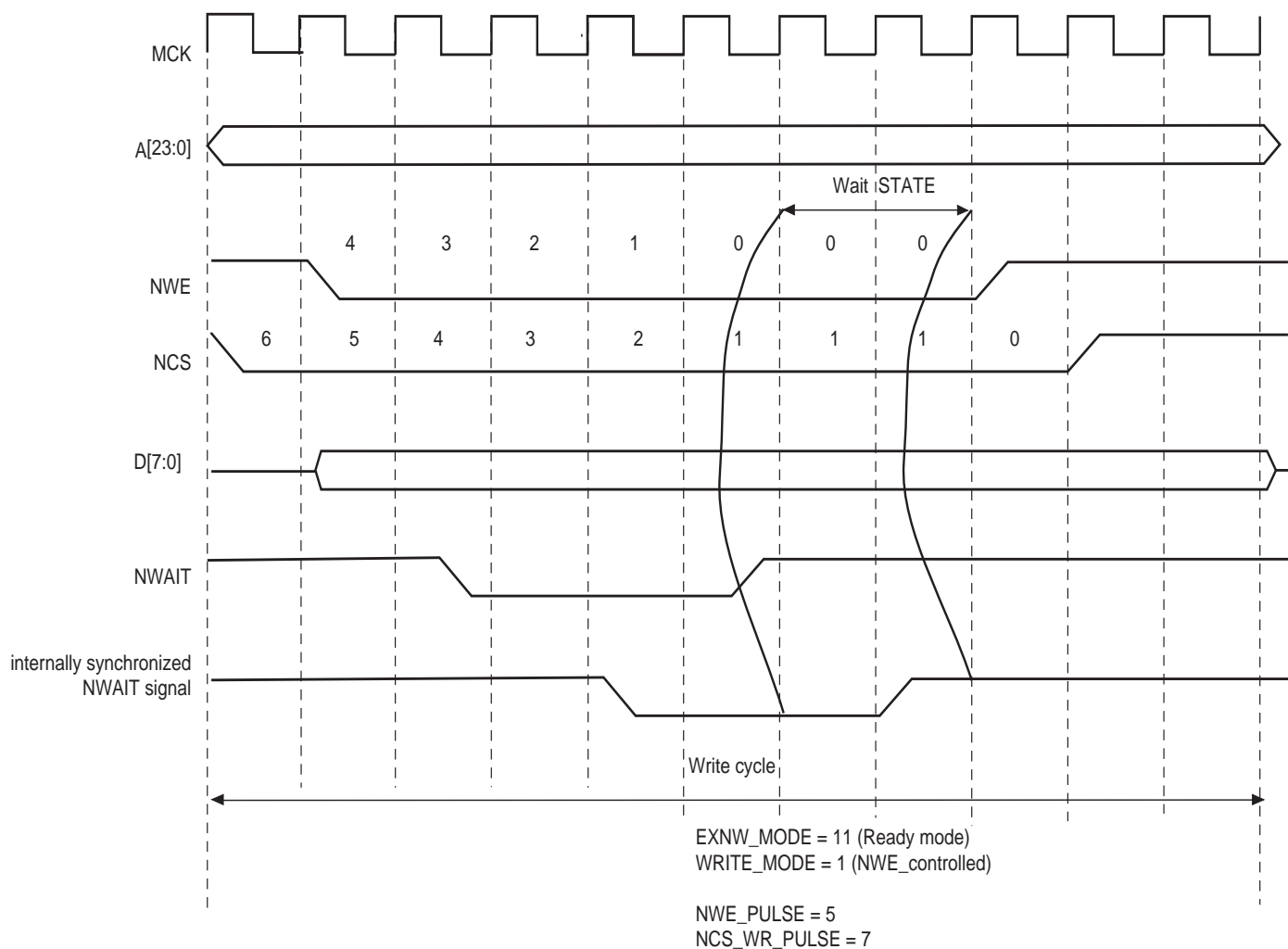
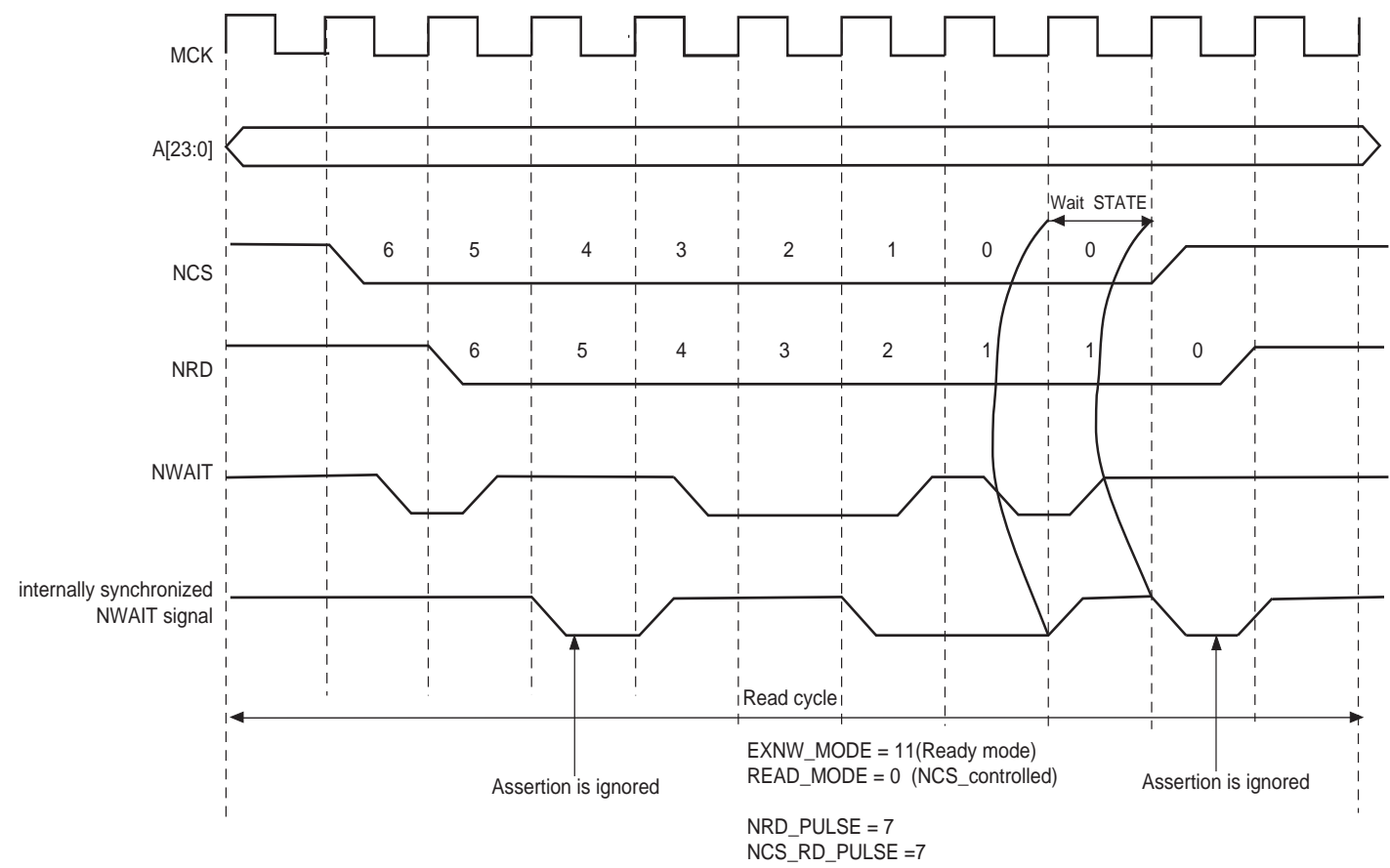


Figure 27-26. NWAIT Assertion in Read Access: Ready Mode (EXNW\_MODE = 11)



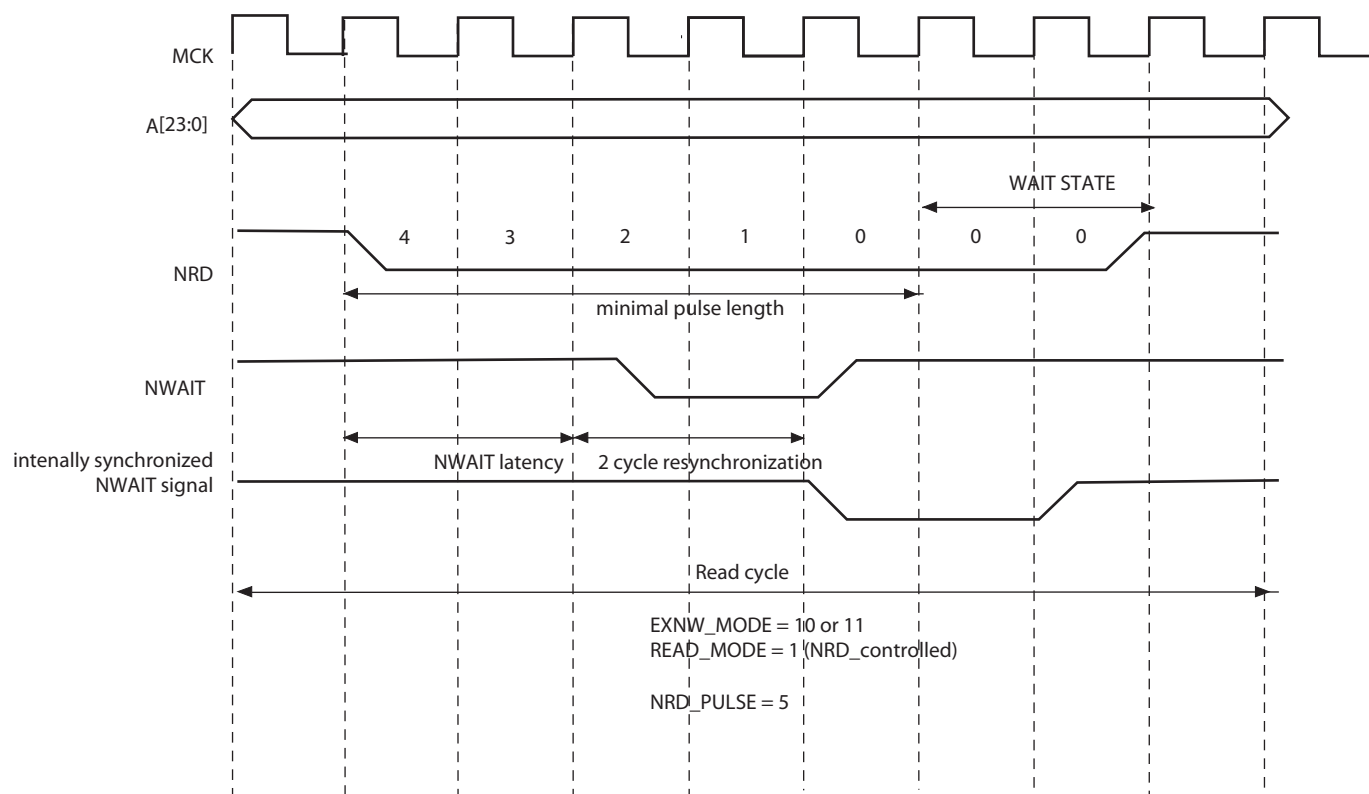
### 27.13.4 NWAIT Latency and Read/Write Timings

There may be a latency between the assertion of the read/write controlling signal and the assertion of the NWAIT signal by the device. The programmed pulse length of the read/write controlling signal must be at least equal to this latency plus the 2 cycles of resynchronization + 1 cycle. Otherwise, the SMC may enter the hold state of the access without detecting the NWAIT signal assertion. This is true in frozen mode as well as in ready mode. This is illustrated on [Figure 27-27](#).

When EXNW\_MODE is enabled (ready or frozen), the user must program a pulse length of the read and write controlling signal of at least:

minimal pulse length = NWAIT latency + 2 resynchronization cycles + 1 cycle

**Figure 27-27. NWAIT Latency**



## 27.14 Slow Clock Mode

The SMC is able to automatically apply a set of “slow clock mode” read/write waveforms when an internal signal driven by the Power Management Controller is asserted because MCK has been turned to a very slow clock rate (typically 32kHz clock rate). In this mode, the user-programmed waveforms are ignored and the slow clock mode waveforms are applied. This mode is provided so as to avoid reprogramming the User Interface with appropriate waveforms at very slow clock rate. When activated, the slow mode is active on all chip selects.

### 27.14.1 Slow Clock Mode Waveforms

Figure 27-28 illustrates the read and write operations in slow clock mode. They are valid on all chip selects. Table 27-5 indicates the value of read and write parameters in slow clock mode.

Figure 27-28. Read/Write Cycles in Slow Clock Mode

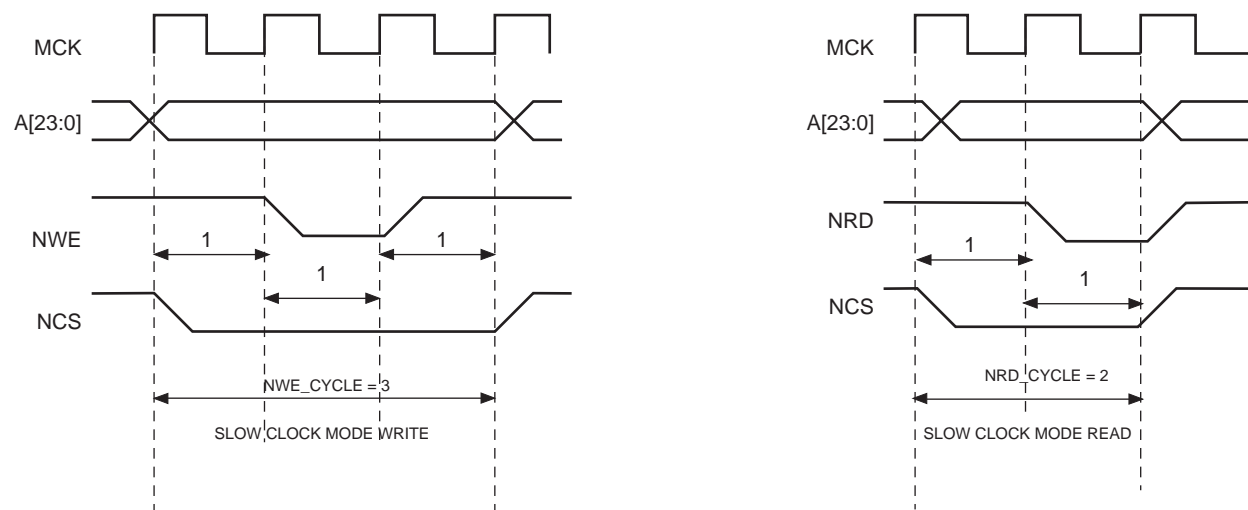


Table 27-5. Read and Write Timing Parameters in Slow Clock Mode

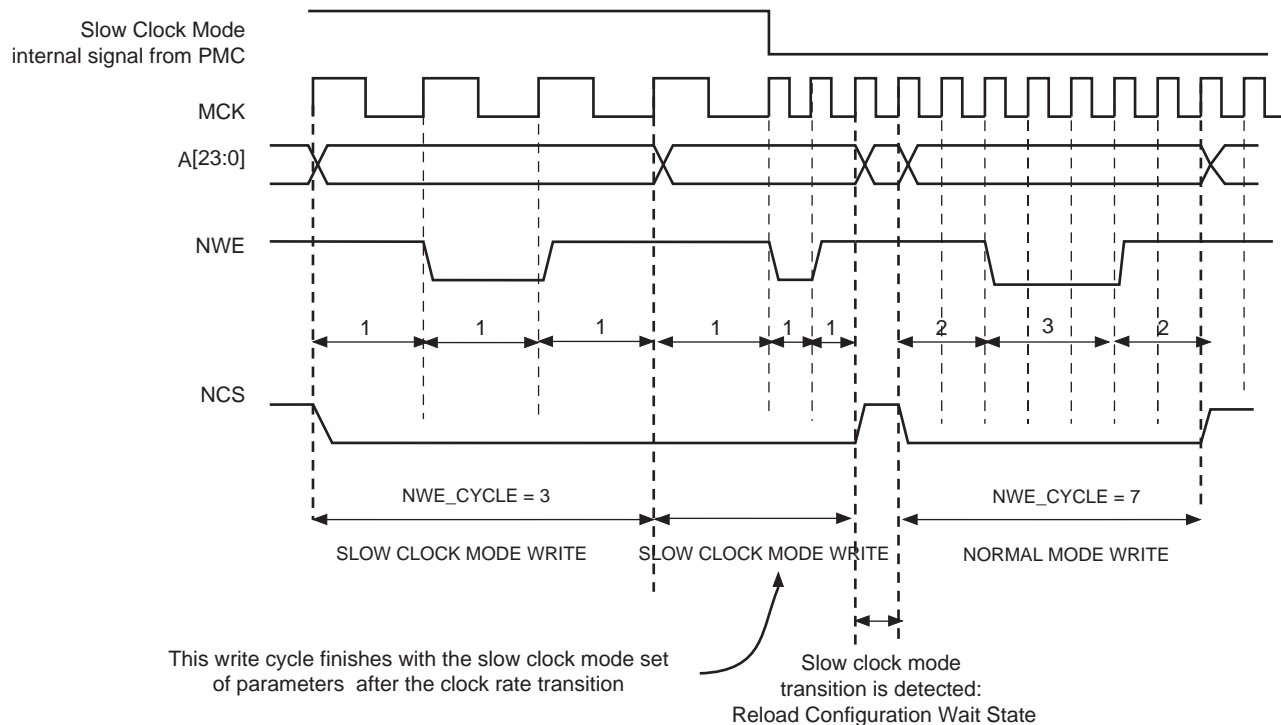
Read Parameters	Duration (cycles)	Write Parameters	Duration (cycles)
NRD_SETUP	1	NWE_SETUP	1
NRD_PULSE	1	NWE_PULSE	1
NCS_RD_SETUP	0	NCS_WR_SETUP	0
NCS_RD_PULSE	2	NCS_WR_PULSE	3
NRD_CYCLE	2	NWE_CYCLE	3

## 27.14.2 Switching from (to) Slow Clock Mode to (from) Normal Mode

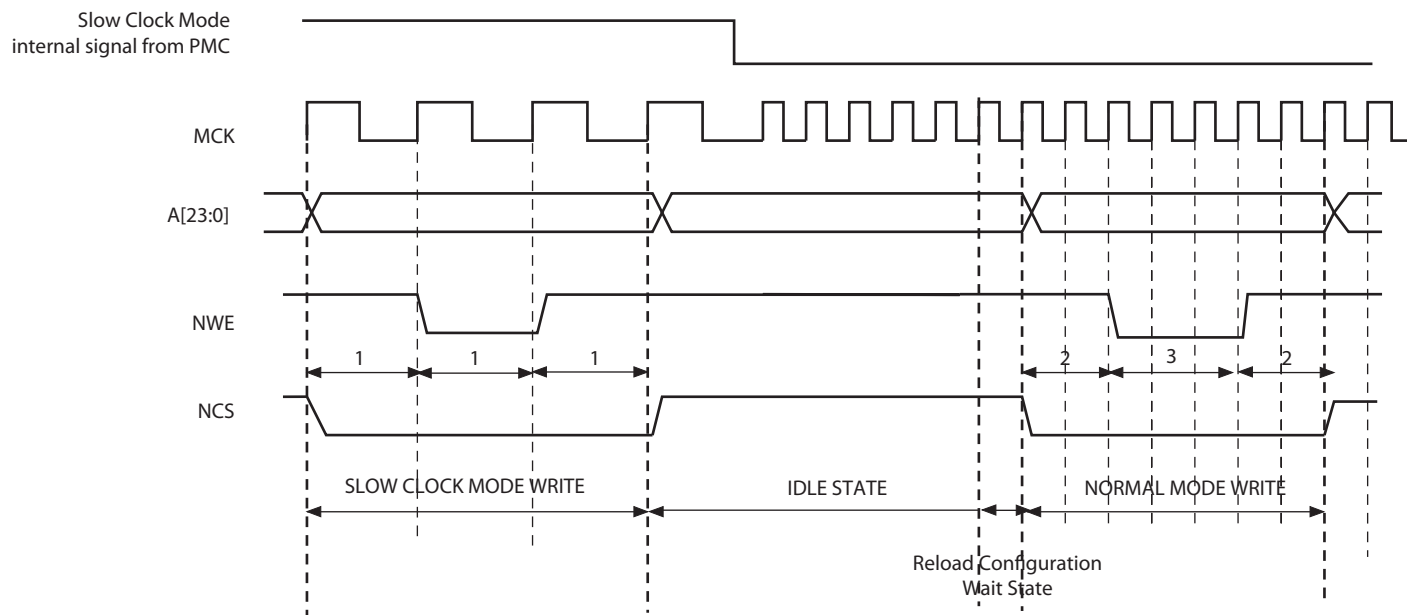
When switching from slow clock mode to the normal mode, the current slow clock mode transfer is completed at high clock rate, with the set of slow clock mode parameters. See [Figure 27-29 on page 477](#). The external device may not be fast enough to support such timings.

[Figure 27-30](#) illustrates the recommended procedure to properly switch from one mode to the other.

**Figure 27-29. Clock Rate Transition Occurs while the SMC is Performing a Write Operation**



**Figure 27-30. Recommended Procedure to Switch from Slow Clock Mode to Normal Mode or from Normal Mode to Slow Clock Mode**



## 27.15 Asynchronous Page Mode

The SMC supports asynchronous burst reads in page mode, providing that the page mode is enabled in the SMC\_MODE register (PMEN field). The page size must be configured in the SMC\_MODE register (PS field) to 4, 8, 16 or 32 bytes.

The page defines a set of consecutive bytes into memory. A 4-byte page (resp. 8-, 16-, 32-byte page) is always aligned to 4-byte boundaries (resp. 8-, 16-, 32-byte boundaries) of memory. The MSB of data address defines the address of the page in memory, the LSB of address define the address of the data in the page as detailed in [Table 27-6](#).

With page mode memory devices, the first access to one page ( $t_{pa}$ ) takes longer than the subsequent accesses to the page ( $t_{sa}$ ) as shown in [Figure 27-31](#). When in page mode, the SMC enables the user to define different read timings for the first access within one page, and next accesses within the page.

**Table 27-6. Page Address and Data Address within a Page**

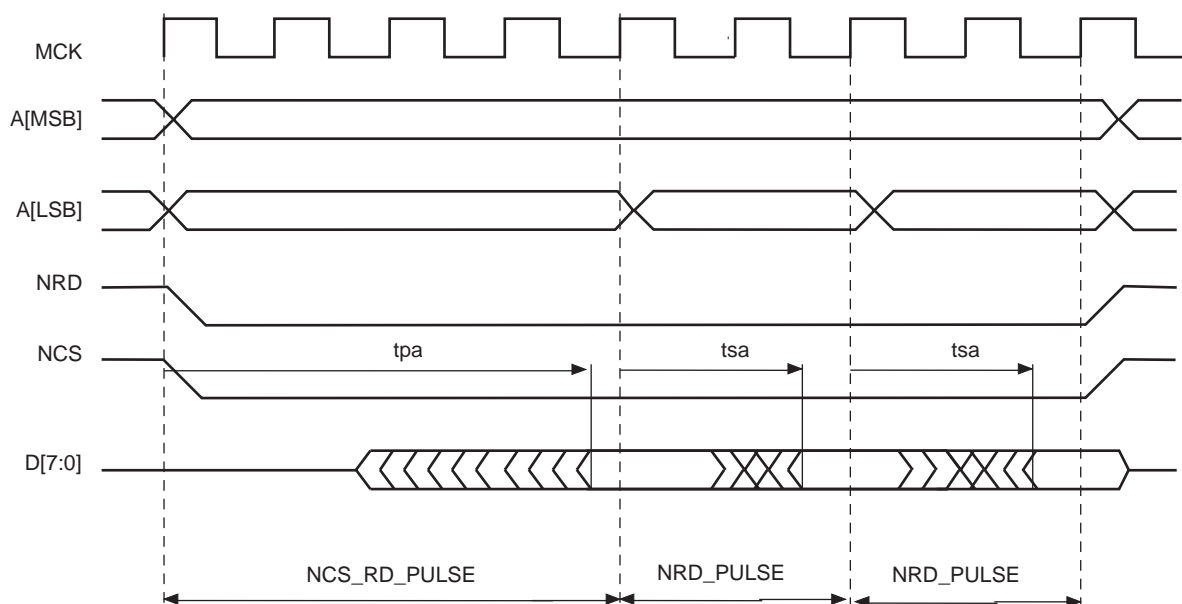
Page Size	Page Address <sup>(1)</sup>	Data Address in the Page
4 bytes	A[23:2]	A[1:0]
8 bytes	A[23:3]	A[2:0]
16 bytes	A[23:4]	A[3:0]
32 bytes	A[23:5]	A[4:0]

Note: 1. "A" denotes the address bus of the memory device.

### 27.15.1 Protocol and Timings in Page Mode

[Figure 27-31](#) shows the NRD and NCS timings in page mode access.

**Figure 27-31. Page Mode Read Protocol (Address MSB and LSB are defined in [Table 27-6](#))**



The NRD and NCS signals are held low during all read transfers, whatever the programmed values of the setup and hold timings in the User Interface may be. Moreover, the NRD and NCS timings are identical. The pulse length of the first access to the page is defined with the NCS\_RD\_PULSE field of the SMC\_PULSE register. The pulse length of subsequent accesses within the page are defined using the NRD\_PULSE parameter.

In page mode, the programming of the read timings is described in [Table 27-7](#):

**Table 27-7. Programming of Read Timings in Page Mode**

Parameter	Value	Definition
READ_MODE	'x'	No impact
NCS_RD_SETUP	'x'	No impact
NCS_RD_PULSE	$t_{pa}$	Access time of first access to the page
NRD_SETUP	'x'	No impact
NRD_PULSE	$t_{sa}$	Access time of subsequent accesses in the page
NRD_CYCLE	'x'	No impact

The SMC does not check the coherency of timings. It will always apply the NCS\_RD\_PULSE timings as page access timing ( $t_{pa}$ ) and the NRD\_PULSE for accesses to the page ( $t_{sa}$ ), even if the programmed value for  $t_{pa}$  is shorter than the programmed value for  $t_{sa}$ .

### 27.15.2 Page Mode Restriction

The page mode is not compatible with the use of the NWAIT signal. Using the page mode and the NWAIT signal may lead to unpredictable behavior.

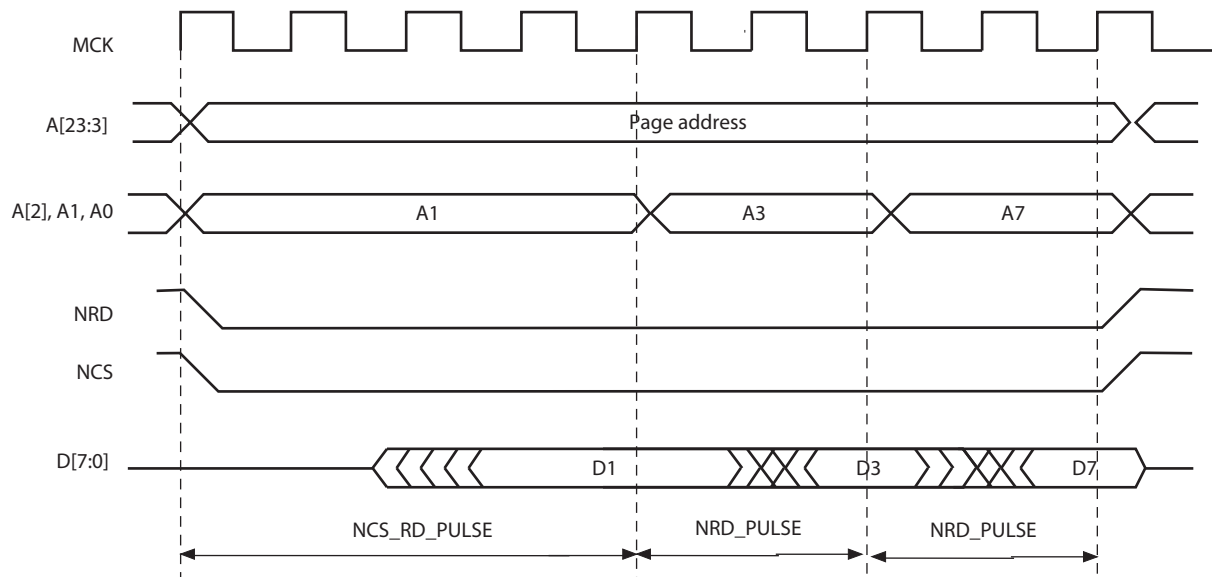
### 27.15.3 Sequential and Non-sequential Accesses

If the chip select and the MSB of addresses as defined in [Table 27-6](#) are identical, then the current access lies in the same page as the previous one, and no page break occurs.

Using this information, all data within the same page, sequential or not sequential, are accessed with a minimum access time ( $t_{sa}$ ). [Figure 27-32](#) illustrates access to an 8-bit memory device in page mode, with 8-byte pages. Access to D1 causes a page access with a long access time ( $t_{pa}$ ). Accesses to D3 and D7, though they are not sequential accesses, only require a short access time ( $t_{sa}$ ).

If the MSB of addresses are different, the SMC performs the access of a new page. In the same way, if the chip select is different from the previous access, a page break occurs. If two sequential accesses are made to the page mode memory, but separated by an other internal or external peripheral access, a page break occurs on the second access because the chip select of the device was deasserted between both accesses.

Figure 27-32. Access to Non-Sequential Data within the Same Page





## 27.16 Static Memory Controller (SMC) User Interface

The SMC is programmed using the registers listed in Table 27-8. For each chip select, a set of 4 registers is used to program the parameters of the external device connected on it. In Table 27-8, “CS\_number” denotes the chip select number. 16 bytes (0x10) are required per chip select.

The user must complete writing the configuration by writing any one of the SMC\_MODE registers.

**Table 27-8. Register Mapping**

Offset	Register	Name	Access	Reset
0x10 x CS_number + 0x00	SMC Setup Register	SMC_SETUP	Read-write	
0x10 x CS_number + 0x04	SMC Pulse Register	SMC_PULSE	Read-write	
0x10 x CS_number + 0x08	SMC Cycle Register	SMC_CYCLE	Read-write	
0x10 x CS_number + 0x0C	SMC Mode Register	SMC_MODE	Read-write	
0x80	SMC OCMS MODE Register	SMC_OCMS	Read-write	0x00000000
0x84	SMC OCMS KEY1 Register	SMC_KEY1	Write once	0x00000000
0x88	SMC OCMS KEY2 Register	SMC_KEY2	Write once	0x00000000
0xE4	SMC Write Protect Mode Register	SMC_WPMR	Read-write	0x00000000
0xE8	SMC Write Protect Status Register	SMC_WPSR	Read-only	0x00000000
0xEC-0xFC	Reserved	-	-	-

### 27.16.1 SMC Setup Register

**Name:** SMC\_SETUP[0..3]

**Address:** 0x400E0000 (0)[0], 0x400E0010 (0)[1], 0x400E0020 (0)[2], 0x400E0030 (0)[3], 0x4801C000 (1)[0], 0x4801C010 (1)[1], 0x4801C020 (1)[2], 0x4801C030 (1)[3]

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	NCS_RD_SETUP					
23	22	21	20	19	18	17	16
–	–	NRD_SETUP					
15	14	13	12	11	10	9	8
–	–	NCS_WR_SETUP					
7	6	5	4	3	2	1	0
–	–	NWE_SETUP					

- **NWE\_SETUP: NWE Setup Length**

The NWE signal setup length is defined as:

NWE setup length = (128\* NWE\_SETUP[5] + NWE\_SETUP[4:0]) clock cycles

- **NCS\_WR\_SETUP: NCS Setup Length in WRITE Access**

In write access, the NCS signal setup length is defined as:

NCS setup length = (128\* NCS\_WR\_SETUP[5] + NCS\_WR\_SETUP[4:0]) clock cycles

- **NRD\_SETUP: NRD Setup Length**

The NRD signal setup length is defined in clock cycles as:

NRD setup length = (128\* NRD\_SETUP[5] + NRD\_SETUP[4:0]) clock cycles

- **NCS\_RD\_SETUP: NCS Setup Length in READ Access**

In read access, the NCS signal setup length is defined as:

NCS setup length = (128\* NCS\_RD\_SETUP[5] + NCS\_RD\_SETUP[4:0]) clock cycles

## 27.16.2 SMC Pulse Register

**Name:** SMC\_PULSE[0..3]

**Address:** 0x400E0004 (0)[0], 0x400E0014 (0)[1], 0x400E0024 (0)[2], 0x400E0034 (0)[3], 0x4801C004 (1)[0], 0x4801C014 (1)[1], 0x4801C024 (1)[2], 0x4801C034 (1)[3]

**Access:** Read-write

31	30	29	28	27	26	25	24
–	NCS_RD_PULSE						
23	22	21	20	19	18	17	16
–	NRD_PULSE						
15	14	13	12	11	10	9	8
–	NCS_WR_PULSE						
7	6	5	4	3	2	1	0
–	NWE_PULSE						

- **NWE\_PULSE: NWE Pulse Length**

The NWE signal pulse length is defined as:

$\text{NWE pulse length} = (256 * \text{NWE\_PULSE}[6] + \text{NWE\_PULSE}[5:0]) \text{ clock cycles}$

The NWE pulse length must be at least 1 clock cycle.

- **NCS\_WR\_PULSE: NCS Pulse Length in WRITE Access**

In write access, the NCS signal pulse length is defined as:

$\text{NCS pulse length} = (256 * \text{NCS\_WR\_PULSE}[6] + \text{NCS\_WR\_PULSE}[5:0]) \text{ clock cycles}$

The NCS pulse length must be at least 1 clock cycle.

- **NRD\_PULSE: NRD Pulse Length**

In standard read access, the NRD signal pulse length is defined in clock cycles as:

$\text{NRD pulse length} = (256 * \text{NRD\_PULSE}[6] + \text{NRD\_PULSE}[5:0]) \text{ clock cycles}$

The NRD pulse length must be at least 1 clock cycle.

In page mode read access, the NRD\_PULSE parameter defines the duration of the subsequent accesses in the page.

- **NCS\_RD\_PULSE: NCS Pulse Length in READ Access**

In standard read access, the NCS signal pulse length is defined as:

$\text{NCS pulse length} = (256 * \text{NCS\_RD\_PULSE}[6] + \text{NCS\_RD\_PULSE}[5:0]) \text{ clock cycles}$

The NCS pulse length must be at least 1 clock cycle.

In page mode read access, the NCS\_RD\_PULSE parameter defines the duration of the first access to one page.

### 27.16.3 SMC Cycle Register

**Name:** SMC\_CYCLE[0..3]

**Address:** 0x400E0008 (0)[0], 0x400E0018 (0)[1], 0x400E0028 (0)[2], 0x400E0038 (0)[3], 0x4801C008 (1)[0], 0x4801C018 (1)[1], 0x4801C028 (1)[2], 0x4801C038 (1)[3]

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	NRD_CYCLE
23	22	21	20	19	18	17	16
NRD_CYCLE							
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	NWE_CYCLE
7	6	5	4	3	2	1	0
NWE_CYCLE							

- **NWE\_CYCLE: Total Write Cycle Length**

The total write cycle length is the total duration in clock cycles of the write cycle. It is equal to the sum of the setup, pulse and hold steps of the NWE and NCS signals. It is defined as:

Write cycle length = (NWE\_CYCLE[8:7]\*256 + NWE\_CYCLE[6:0]) clock cycles

- **NRD\_CYCLE: Total Read Cycle Length**

The total read cycle length is the total duration in clock cycles of the read cycle. It is equal to the sum of the setup, pulse and hold steps of the NRD and NCS signals. It is defined as:

Read cycle length = (NRD\_CYCLE[8:7]\*256 + NRD\_CYCLE[6:0]) clock cycles

## 27.16.4 SMC MODE Register

**Name:** SMC\_MODE[0..3]

**Address:** 0x400E000C (0)[0], 0x400E001C (0)[1], 0x400E002C (0)[2], 0x400E003C (0)[3], 0x4801C00C (1)[0], 0x4801C01C (1)[1], 0x4801C02C (1)[2], 0x4801C03C (1)[3]

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	PS		–	–	–	PMEN
23	22	21	20	19	18	17	16
–	–	–	TDF_MODE	TDF_CYCLES			
15	14	13	12	11	10	9	8
–	–	–	DBW	–	–	–	–
7	6	5	4	3	2	1	0
–	–	EXNW_MODE		–	–	WRITE_MODE	READ_MODE

### • READ\_MODE:

1: The read operation is controlled by the NRD signal.

- If TDF cycles are programmed, the external bus is marked busy after the rising edge of NRD.
- If TDF optimization is enabled (TDF\_MODE =1), TDF wait states are inserted after the setup of NRD.

0: The read operation is controlled by the NCS signal.

- If TDF cycles are programmed, the external bus is marked busy after the rising edge of NCS.
- If TDF optimization is enabled (TDF\_MODE =1), TDF wait states are inserted after the setup of NCS.

### • WRITE\_MODE

1: The write operation is controlled by the NWE signal.

- If TDF optimization is enabled (TDF\_MODE =1), TDF wait states will be inserted after the setup of NWE.

0: The write operation is controlled by the NCS signal.

- If TDF optimization is enabled (TDF\_MODE =1), TDF wait states will be inserted after the setup of NCS.

### • EXNW\_MODE: NWAIT Mode

The NWAIT signal is used to extend the current read or write signal. It is only taken into account during the pulse phase of the read and write controlling signal. When the use of NWAIT is enabled, at least one cycle hold duration must be programmed for the read and write controlling signal.

Value	Name	Description
0	DISABLED	Disabled
1		Reserved
2	FROZEN	Frozen Mode
3	READY	Ready Mode

- Disabled Mode: The NWAIT input signal is ignored on the corresponding Chip Select.
- Frozen Mode: If asserted, the NWAIT signal freezes the current read or write cycle. After deassertion, the read/write cycle is resumed from the point where it was stopped.
- Ready Mode: The NWAIT signal indicates the availability of the external device at the end of the pulse of the controlling read or write signal, to complete the access. If high, the access normally completes. If low, the access is extended until NWAIT returns high.

- **DBW: Data Bus Width**

Value	Name	Description
0	8_BIT	8-bit Data Bus
1	16_BIT	16-bit Data Bus

- **TDF\_CYCLES: Data Float Time**

This field gives the integer number of clock cycles required by the external device to release the data after the rising edge of the read controlling signal. The SMC always provide one full cycle of bus turnaround after the TDF\_CYCLES period. The external bus cannot be used by another chip select during TDF\_CYCLES + 1 cycles. From 0 up to 15 TDF\_CYCLES can be set.

- **TDF\_MODE: TDF Optimization**

1: TDF optimization is enabled.

- The number of TDF wait states is optimized using the setup period of the next read/write access.

0: TDF optimization is disabled.

- The number of TDF wait states is inserted before the next access begins.

- **PMEN: Page Mode Enabled**

1: Asynchronous burst read in page mode is applied on the corresponding chip select.

0: Standard read is applied.

- **PS: Page Size**

If page mode is enabled, this field indicates the size of the page in bytes.

Value	Name	Description
0	4_BYTE	4-byte page
1	8_BYTE	8-byte page
2	16_BYTE	16-byte page
3	32_BYTE	32-byte page

### 27.16.5 SMC OCMS Mode Register

Name: SMC\_OCMS

Address: 0x400E0080 (0), 0x4801C080 (1)

Access: Read-write

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	CS3SE	CS2SE	CS1SE	CS0SE
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	SMSE

- **CSxSE: Chip Select (x = 0 to 3) Scrambling Enable**

0: Disable Scrambling for CSx.

1: Enable Scrambling for CSx.

- **SMSE: Static Memory Controller Scrambling Enable**

0: Disable Scrambling for SMC access.

1: Enable Scrambling for SMC access.

### 27.16.6 SMC OCMS Key1 Register

Name: SMC\_KEY1

Address: 0x400E0084 (0), 0x4801C084 (1)

Access: Write Once

Reset: 0x00000000

31	30	29	28	27	26	25	24
KEY1							
23	22	21	20	19	18	17	16
KEY1							
15	14	13	12	11	10	9	8
KEY1							
7	6	5	4	3	2	1	0
KEY1							

- **KEY1: Off Chip Memory Scrambling (OCMS) Key Part 1**

When Off Chip Memory Scrambling is enabled setting the SMC\_OCMS and SMC\_TIMINGS registers in accordance, the data scrambling depends on KEY1 and KEY2 values.



### 27.16.7 SMC OCMS Key2 Register

Name: SMC\_KEY2

Address: 0x400E0088 (0), 0x4801C088 (1)

Access: Write Once

Reset: 0x00000000

31	30	29	28	27	26	25	24
KEY2							
23	22	21	20	19	18	17	16
KEY2							
15	14	13	12	11	10	9	8
KEY2							
7	6	5	4	3	2	1	0
KEY2							

- **KEY2: Off Chip Memory Scrambling (OCMS) Key Part 2**

When Off Chip Memory Scrambling is enabled setting the SMC\_OCMS and SMC\_TIMINGS registers in accordance, the data scrambling depends on KEY2 and KEY1 values.

### 27.16.8 SMC Write Protect Mode Register

**Name:** SMC\_WPMR

**Address:** 0x400E00E4 (0), 0x4801C0E4 (1)

**Access:** Read-write

**Reset:** See [Table 27-8](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	WPEN

- **WPEN: Write Protect Enable**

0 = Disables the Write Protect if WPKEY corresponds to 0x534D43 (“SMC” in ASCII).

1 = Enables the Write Protect if WPKEY corresponds to 0x534D43 (“SMC” in ASCII).

Protects the registers listed below:

- [Section 27.16.1 “SMC Setup Register”](#)
- [Section 27.16.2 “SMC Pulse Register”](#)
- [Section 27.16.3 “SMC Cycle Register”](#)
- [Section 27.16.4 “SMC MODE Register”](#)

- **WPKEY: Write Protect KEY**

Should be written at value 0x534D43 (“SMC” in ASCII). Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

### 27.16.9 SMC Write Protect Status Register

**Name:** SMC\_WPSR

**Address:** 0x400E00E8 (0), 0x4801C0E8 (1)

**Type:** Read-only

**Value:** See [Table 27-8](#)

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	WPVS

- **WPVS: Write Protect Enable**

0 = No Write Protect Violation has occurred since the last read of the SMC\_WPSR register.

1 = A Write Protect Violation occurred since the last read of the SMC\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protect Violation Source**

When WPVS is active, this field indicates the write-protected register (through address offset or code) in which a write access has been attempted.

**Note:** Reading SMC\_WPSR automatically clears all fields.

## 28. Peripheral DMA Controller (PDC)

### 28.1 Description

The Peripheral DMA Controller (PDC) transfers data between on-chip serial peripherals and the target memories. The link between the PDC and a serial peripheral is operated by the AHB to APB bridge.

The user interface of each PDC channel is integrated into the user interface of the peripheral it serves. The user interface of mono-directional channels (receive-only or transmit-only) contains two 32-bit memory pointers and two 16-bit counters, one set (pointer, counter) for the current transfer and one set (pointer, counter) for the next transfer. The bidirectional channel user interface contains four 32-bit memory pointers and four 16-bit counters. Each set (pointer, counter) is used by the current transmit, next transmit, current receive and next receive.

Using the PDC decreases processor overhead by reducing its intervention during the transfer. This lowers significantly the number of clock cycles required for a data transfer, improving microcontroller performance.

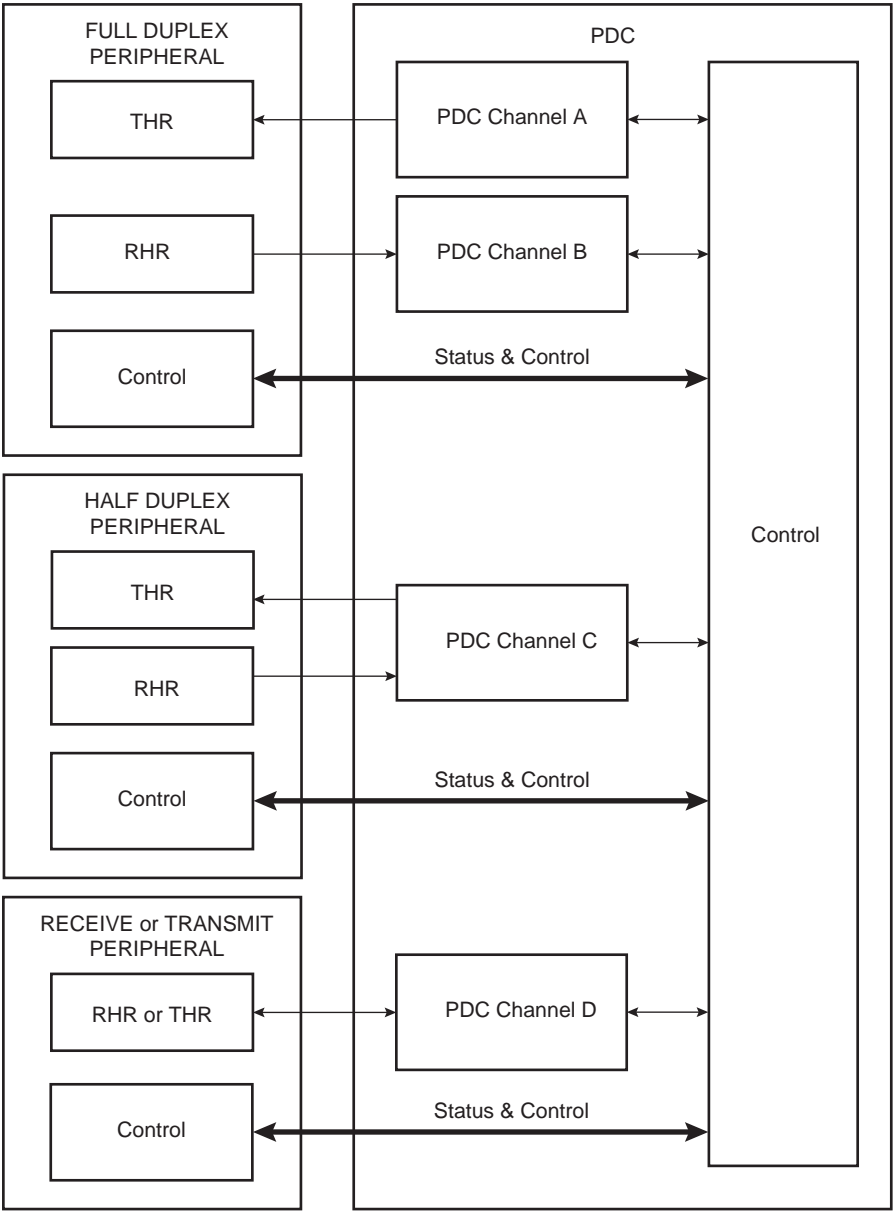
To launch a transfer, the peripheral triggers its associated PDC channels by using transmit and receive signals. When the programmed data is transferred, an end of transfer interrupt is generated by the peripheral itself.

### 28.2 Embedded Characteristics

- Performs Transfers to/from APB Communication Serial Peripherals
- Supports Half-duplex and Full-duplex Peripherals

28.3 Block Diagram

Figure 28-1. Block Diagram



## 28.4 Functional Description

### 28.4.1 Configuration

The PDC channel user interface enables the user to configure and control data transfers for each channel. The user interface of each PDC channel is integrated into the associated peripheral user interface.

The user interface of a serial peripheral, whether it is full- or half-duplex, contains four 32-bit pointers (RPR, RNPR, TPR, TNPR) and four 16-bit counter registers (RCR, RNCR, TCR, TNCR). However, the transmit and receive parts of each type are programmed differently: the transmit and receive parts of a full-duplex peripheral can be programmed at the same time, whereas only one part (transmit or receive) of a half-duplex peripheral can be programmed at a time.

32-bit pointers define the access location in memory for the current and next transfer, whether it is for read (transmit) or write (receive). 16-bit counters define the size of the current and next transfers. It is possible, at any moment, to read the number of transfers remaining for each channel.

The PDC has dedicated status registers which indicate if the transfer is enabled or disabled for each channel. The status for each channel is located in the associated peripheral status register. Transfers can be enabled and/or disabled by setting TXTEN/TXTDIS and RXTEN/RXTDIS in the peripheral's Transfer Control register.

At the end of a transfer, the PDC channel sends status flags to its associated peripheral. These flags are visible in the peripheral Status register (ENDRX, ENDTX, RXBUFF, and TXBUFE). Refer to [Section 28.4.3](#) and to the associated peripheral user interface.

The peripheral where a PDC transfer is configured must have its peripheral clock enabled. The peripheral clock must be also enabled to access the PDC register set associated to this peripheral.

### 28.4.2 Memory Pointers

Each full-duplex peripheral is connected to the PDC by a receive channel and a transmit channel. Both channels have 32-bit memory pointers that point to a receive area and to a transmit area, respectively, in the target memory.

Each half-duplex peripheral is connected to the PDC by a bidirectional channel. This channel has two 32-bit memory pointers, one for current transfer and the other for next transfer. These pointers point to transmit or receive data depending on the operating mode of the peripheral.

Depending on the type of transfer (byte, half-word or word), the memory pointer is incremented respectively by 1, 2 or 4 bytes.

If a memory pointer address changes in the middle of a transfer, the PDC channel continues operating using the new address.

### 28.4.3 Transfer Counters

Each channel has two 16-bit counters, one for the current transfer and the one for the next transfer. These counters define the size of data to be transferred by the channel. The current transfer counter is decremented first as the data addressed by the current memory pointer starts to be transferred. When the current transfer counter reaches zero, the channel checks its next transfer counter. If the value of the next counter is zero, the channel stops transferring data and sets the appropriate flag. If the next counter value is greater than zero, the values of the next pointer/next counter are copied into the current pointer/current counter and the channel resumes the transfer, whereas next pointer/next counter get zero/zero as values. At the end of this transfer, the PDC channel sets the appropriate flags in the Peripheral Status register.

The following list gives an overview of how status register flags behave depending on the counters' values:

- ENDRX flag is set when the PDC Receive Counter register (PERIPH\_RCR) reaches zero.
- RXBUFF flag is set when both PERIPH\_RCR and the PDC Receive Next Counter register (PERIPH\_RNCR) reach zero.
- ENDTX flag is set when the PDC Transmit Counter register (PERIPH\_TCR) reaches zero.
- TXBUFE flag is set when both PERIPH\_TCR and the PDC Transmit Next Counter register (PERIPH\_TNCR) reach zero.

These status flags are described in the Peripheral Status register (PERIPH\_PTSR).

#### 28.4.4 Data Transfers

The serial peripheral triggers its associated PDC channels' transfers using transmit enable (TXEN) and receive enable (RXEN) flags in the transfer control register integrated in the peripheral's user interface.

When the peripheral receives external data, it sends a Receive Ready signal to its PDC receive channel which then requests access to the Matrix. When access is granted, the PDC receive channel starts reading the peripheral Receive Holding register (RHR). The read data are stored in an internal buffer and then written to memory.

When the peripheral is about to send data, it sends a Transmit Ready to its PDC transmit channel which then requests access to the Matrix. When access is granted, the PDC transmit channel reads data from memory and transfers the data to the Transmit Holding register (THR) of its associated peripheral. The same peripheral sends data depending on its mechanism.

#### 28.4.5 PDC Flags and Peripheral Status Register

Each peripheral connected to the PDC sends out receive ready and transmit ready flags and the PDC returns flags to the peripheral. All these flags are only visible in the peripheral's Status register.

Depending on whether the peripheral is half- or full-duplex, the flags belong to either one single channel or two different channels.

##### 28.4.5.1 Receive Transfer End

The receive transfer end flag is set when PERIPH\_RCR reaches zero and the last data has been transferred to memory. This flag is reset by writing a non-zero value to PERIPH\_RCR or PERIPH\_RNCR.

##### 28.4.5.2 Transmit Transfer End

The transmit transfer end flag is set when PERIPH\_TCR reaches zero and the last data has been written to the peripheral THR.

This flag is reset by writing a non-zero value to PERIPH\_TCR or PERIPH\_TNCR.

##### 28.4.5.3 Receive Buffer Full

The receive buffer full flag is set when PERIPH\_RCR reaches zero, with PERIPH\_RNCR also set to zero and the last data transferred to memory.

This flag is reset by writing a non-zero value to PERIPH\_TCR or PERIPH\_TNCR.

##### 28.4.5.4 Transmit Buffer Empty

The transmit buffer empty flag is set when PERIPH\_TCR reaches zero, with PERIPH\_TNCR also set to zero and the last data written to peripheral THR.

This flag is reset by writing a non-zero value to PERIPH\_TCR or PERIPH\_TNCR.

## 28.5 Peripheral DMA Controller (PDC) User Interface

Table 28-1. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Receive Pointer Register	PERIPH <sup>(1)</sup> _RPR	Read/Write	0
0x04	Receive Counter Register	PERIPH_RCR	Read/Write	0
0x08	Transmit Pointer Register	PERIPH_TPR	Read/Write	0
0x0C	Transmit Counter Register	PERIPH_TCR	Read/Write	0
0x10	Receive Next Pointer Register	PERIPH_RNPR	Read/Write	0
0x14	Receive Next Counter Register	PERIPH_RNCR	Read/Write	0
0x18	Transmit Next Pointer Register	PERIPH_TNPR	Read/Write	0
0x1C	Transmit Next Counter Register	PERIPH_TNCR	Read/Write	0
0x20	Transfer Control Register	PERIPH_PTCR	Write-only	0
0x24	Transfer Status Register	PERIPH_PTSR	Read-only	0

Note: 1. PERIPH: Ten registers are mapped in the peripheral memory space at the same offset. These can be defined by the user depending on the function and the desired peripheral.



### 28.5.1 Receive Pointer Register

**Name:** PERIPH\_RPR

**Access:** Read/Write

31	30	29	28	27	26	25	24
RXPTR							
23	22	21	20	19	18	17	16
RXPTR							
15	14	13	12	11	10	9	8
RXPTR							
7	6	5	4	3	2	1	0
RXPTR							

- **RXPTR: Receive Pointer Register**

RXPTR must be set to receive buffer address.

When a half-duplex peripheral is connected to the PDC, RXPTR = TXPTR.

## 28.5.2 Receive Counter Register

**Name:** PERIPH\_RCR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
RXCTR							
7	6	5	4	3	2	1	0
RXCTR							

- **RXCTR: Receive Counter Register**

RXCTR must be set to receive buffer size.

When a half-duplex peripheral is connected to the PDC, RXCTR = TXCTR.

0: Stops peripheral data transfer to the receiver.

1 - 65535: Starts peripheral data transfer if the corresponding channel is active.

### 28.5.3 Transmit Pointer Register

**Name:** PERIPH\_TPR

**Access:** Read/Write

31	30	29	28	27	26	25	24
TXPTR							
23	22	21	20	19	18	17	16
TXPTR							
15	14	13	12	11	10	9	8
TXPTR							
7	6	5	4	3	2	1	0
TXPTR							

- **TXPTR: Transmit Counter Register**

TXPTR must be set to transmit buffer address.

When a half-duplex peripheral is connected to the PDC, RXPTR = TXPTR.

## 28.5.4 Transmit Counter Register

**Name:** PERIPH\_TCR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXCTR							
7	6	5	4	3	2	1	0
TXCTR							

- **TXCTR: Transmit Counter Register**

TXCTR must be set to transmit buffer size.

When a half-duplex peripheral is connected to the PDC, RXCTR = TXCTR.

0: Stops peripheral data transfer to the transmitter.

1- 65535: Starts peripheral data transfer if the corresponding channel is active.

### 28.5.5 Receive Next Pointer Register

**Name:** PERIPH\_RNPR

**Access:** Read/Write

31	30	29	28	27	26	25	24
RXNPTR							
23	22	21	20	19	18	17	16
RXNPTR							
15	14	13	12	11	10	9	8
RXNPTR							
7	6	5	4	3	2	1	0
RXNPTR							

- **RXNPTR: Receive Next Pointer**

RXNPTR contains the next receive buffer address.

When a half-duplex peripheral is connected to the PDC, RXNPTR = TXNPTR.

### 28.5.6 Receive Next Counter Register

**Name:** PERIPH\_RNCR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
RXNCTR							
7	6	5	4	3	2	1	0
RXNCTR							

- **RXNCTR: Receive Next Counter**

RXNCTR contains the next receive buffer size.

When a half-duplex peripheral is connected to the PDC, RXNCTR = TXNCTR.

## 28.5.7 Transmit Next Pointer Register

**Name:** PERIPH\_TNPR

**Access:** Read/Write

31	30	29	28	27	26	25	24
TXNPTR							
23	22	21	20	19	18	17	16
TXNPTR							
15	14	13	12	11	10	9	8
TXNPTR							
7	6	5	4	3	2	1	0
TXNPTR							

- **TXNPTR: Transmit Next Pointer**

TXNPTR contains the next transmit buffer address.

When a half-duplex peripheral is connected to the PDC, RXNPTR = TXNPTR.

## 28.5.8 Transmit Next Counter Register

**Name:** PERIPH\_TNCR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXNCTR							
7	6	5	4	3	2	1	0
TXNCTR							

- **TXNCTR: Transmit Counter Next**

TXNCTR contains the next transmit buffer size.

When a half-duplex peripheral is connected to the PDC, RXNCTR = TXNCTR.



## 28.5.9 Transfer Control Register

**Name:** PERIPH\_PTCR

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	TXTDIS	TXTEN
7	6	5	4	3	2	1	0
–	–	–	–	–	–	RXTDIS	RXTEN

- **RXTEN: Receiver Transfer Enable**

0: No effect.

1: Enables PDC receiver channel requests if RXTDIS is not set.

When a half-duplex peripheral is connected to the PDC, enabling the receiver channel requests automatically disables the transmitter channel requests. It is forbidden to set both TXTEN and RXTEN for a half-duplex peripheral.

- **RXTDIS: Receiver Transfer Disable**

0: No effect.

1: Disables the PDC receiver channel requests.

When a half-duplex peripheral is connected to the PDC, disabling the receiver channel requests also disables the transmitter channel requests.

- **TXTEN: Transmitter Transfer Enable**

0: No effect.

1: Enables the PDC transmitter channel requests.

When a half-duplex peripheral is connected to the PDC, it enables the transmitter channel requests only if RXTEN is not set. It is forbidden to set both TXTEN and RXTEN for a half-duplex peripheral.

- **TXTDIS: Transmitter Transfer Disable**

0: No effect.

1: Disables the PDC transmitter channel requests.

When a half-duplex peripheral is connected to the PDC, disabling the transmitter channel requests disables the receiver channel requests.

### 28.5.10 Transfer Status Register

**Name:** PERIPH\_PTSR

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	TXTEN
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	RXTEN

- **RXTEN: Receiver Transfer Enable**

0: PDC receiver channel requests are disabled.

1: PDC receiver channel requests are enabled.

- **TXTEN: Transmitter Transfer Enable**

0: PDC transmitter channel requests are disabled.

1: PDC transmitter channel requests are enabled.

## 29. Clock Generator

### 29.1 Description

The Clock Generator User Interface is embedded within the Power Management Controller and is described in [Section 30.17 "Power Management Controller \(PMC\) User Interface"](#). However, the Clock Generator registers are named CKGR\_.

### 29.2 Embedded Characteristics

The Clock Generator is made up of:

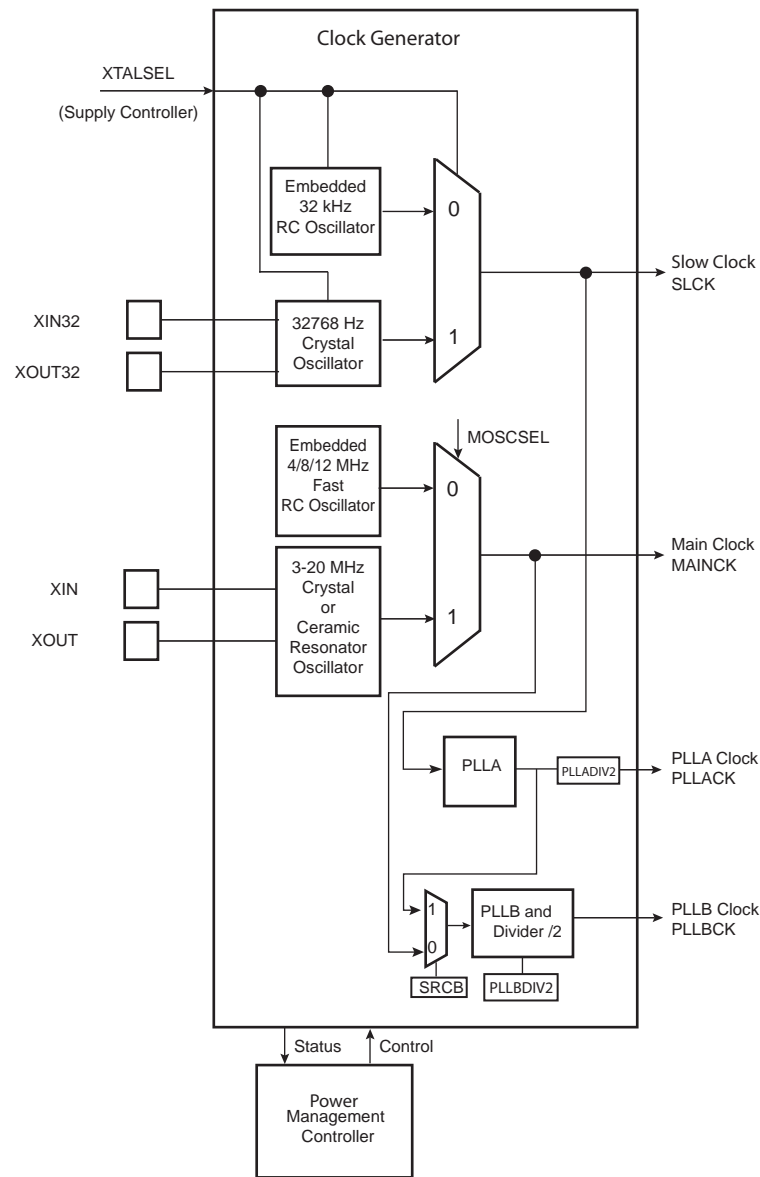
- A Low Power 32768 Hz Slow Clock Oscillator with bypass mode.
- A Low Power RC Oscillator
- A 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator, which can be bypassed.
- A factory programmed Fast RC Oscillator. Three output frequencies can be selected: 4/8/12 MHz. By default 4 MHz is selected.
- Two programmable PLLs, (PLLA input from 32 kHz, output clock range 8 MHz and PLLB input from 3 to 32 MHz, output clock range 80 to 240 MHz), capable of providing the clock MCK to the processor and to the peripherals.
- Write Protected Registers

It provides the following clocks:

- SLCK, the Slow Clock, which is the only permanent clock within the system.
- MAINCK is the output of the Main Clock Oscillator selection: either the Crystal or Ceramic Resonator-based Oscillator or 4/8/12 MHz Fast RC Oscillator.
- PLLACK is the output of the 8 MHz programmable PLL (PLLA).
- PLLBCK is the output of the Divider and 80 to 240 MHz programmable PLL (PLLB).

### 29.3 Block Diagram

Figure 29-1. Clock Generator Block Diagram



## 29.4 Slow Clock

The Supply Controller embeds a slow clock generator that is supplied with the VDDBU power supply. As soon as the VDDBU is supplied, both the crystal oscillator and the embedded RC oscillator are powered up, but only the embedded RC oscillator is enabled. This allows the slow clock to be valid in a short time (about 100  $\mu$ s).

The Slow Clock is generated either by the Slow Clock Crystal Oscillator or by the Slow Clock RC Oscillator.

The selection between the RC or the crystal oscillator is made by writing the XTALSEL bit in the Supply Controller Control Register (SUPC\_CR).

### 29.4.1 Slow Clock RC Oscillator

By default, the Slow Clock RC Oscillator is enabled and selected. The user has to take into account the possible drifts of the RC Oscillator. More details are given in the section “DC Characteristics” of the product datasheet.

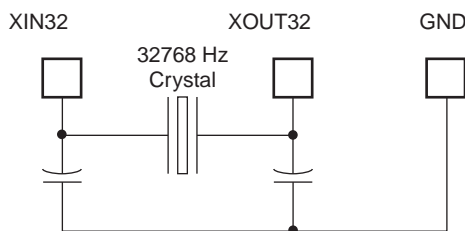
It can be disabled via the XTALSEL bit in the Supply Controller Control Register (SUPC\_CR).

### 29.4.2 Slow Clock Crystal Oscillator

The Clock Generator integrates a 32768 Hz low-power oscillator. In order to use this oscillator, the XIN32 and XOUT32 pins must be connected to a 32768 Hz crystal. Two external capacitors must be wired as shown in [Figure 29-2](#). More details are given in the section “DC Characteristics” of the product datasheet.

Note that the user is not obliged to use the Slow Clock Crystal and can use the RC oscillator instead.

**Figure 29-2. Typical Slow Clock Crystal Oscillator Connection**



The user can select the crystal oscillator to be the source of the slow clock, as it provides a more accurate frequency. The command is made by writing the Supply Controller Control Register (SUPC\_CR) with the XTALSEL bit at 1. This results in a sequence which enables the crystal oscillator and then disables the RC oscillator to save power. The switch of the slow clock source is glitch free. The OSCSEL bit of the Supply Controller Status Register (SUPC\_SR) or the OSCSEL bit of the PMC Status Register (PMC\_SR) tracks the oscillator frequency downstream. It must be read in order to be informed when the switch sequence, initiated when a new value is written in the XTALSEL bit of SUPC\_CR, is done.

Coming back on the RC oscillator is only possible by shutting down the VDDBU power supply. If the user does not need the crystal oscillator, the XIN32 and XOUT32 pins can be left unconnected.

The user can also set the crystal oscillator in bypass mode instead of connecting a crystal. In this case, the user has to provide the external clock signal on XIN32. The input characteristics of the XIN32 pin are given in the product electrical characteristics section. In order to set the bypass mode, the OSCBYPASS bit of the Supply Controller Mode Register (SUPC\_MR) needs to be set at 1.

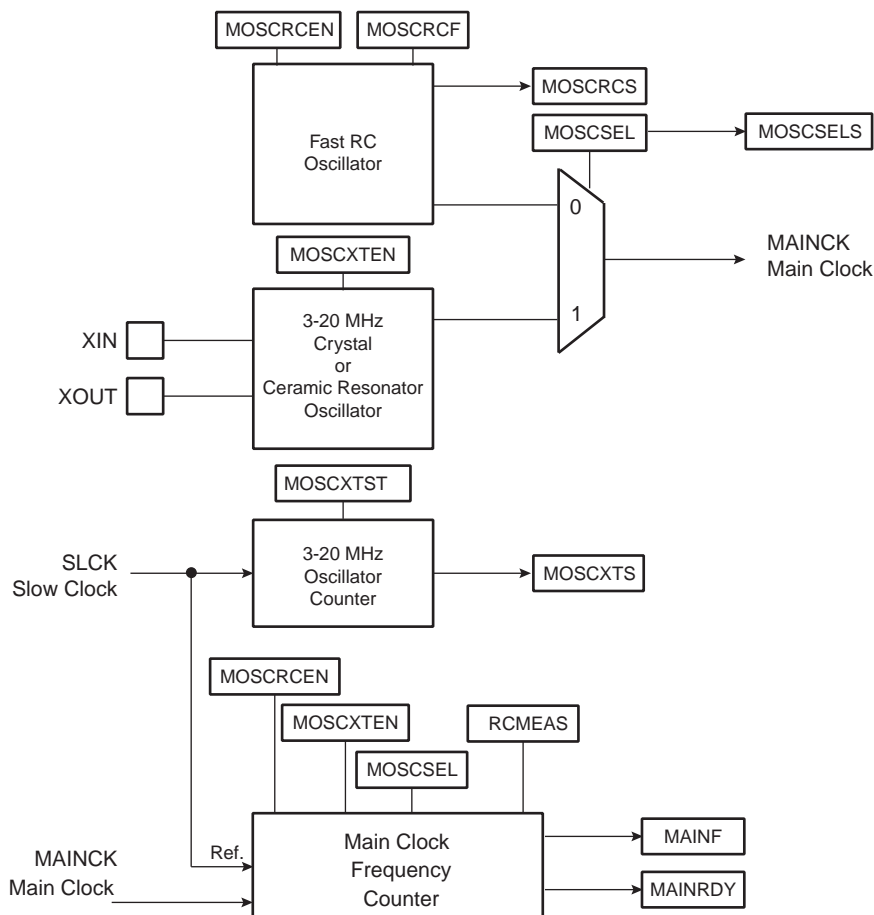
The user can set the Slow Clock Crystal Oscillator in bypass mode instead of connecting a crystal. In this case, the user has to provide the external clock signal on XIN32. The input characteristics of the XIN32 pin under these conditions are given in the product electrical characteristics section.

The programmer has to be sure to set the OSCBYPASS bit in the Supply Controller Mode Register (SUPC\_MR) and XTALSEL bit in the Supply Controller Control Register (SUPC\_CR).

## 29.5 Main Clock

Figure 29-3 shows the Main Clock block diagram.

Figure 29-3. Main Clock Block Diagram



The Main Clock has two sources:

- 4/8/12 MHz Fast RC Oscillator which starts very quickly and is used at start-up.
- 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator which can be bypassed.

### 29.5.1 Fast RC Oscillator

After reset, the 4/8/12 MHz Fast RC Oscillator is enabled with the 4 MHz frequency selected and it is selected as the source of MAINCK. MAINCK is the default clock selected to start up the system.

The Fast RC Oscillator frequencies are calibrated in production except the lowest frequency which is not calibrated.

Refer to the “DC Characteristics” section of the product datasheet.

The software can disable or enable the 4/8/12 MHz Fast RC Oscillator with the MOSRCEN bit in the Clock Generator Main Oscillator Register (CKGR\_MOR).

The user can also select the output frequency of the Fast RC Oscillator, either 4/8/12 MHz are available. It can be done through MOSRCF bits in CKGR\_MOR. When changing this frequency selection, the MOSCRCS bit in the Power Management Controller Status Register (PMC\_SR) is automatically cleared and MAINCK is stopped until the oscillator is stabilized. Once the oscillator is stabilized, MAINCK restarts and MOSCRCS is set.

When disabling the Main Clock by clearing the MOSRCEN bit in CKGR\_MOR, the MOSCRCS bit in the Power Management Controller Status Register (PMC\_SR) is automatically cleared, indicating the Main Clock is off.

Setting the MOSCRCS bit in the Power Management Controller Interrupt Enable Register (PMC\_IER) can trigger an interrupt to the processor.

It is recommended to disable the Main Clock as soon as the processor no longer uses it and runs out of SLCK, PLLACK. The CAL4, CAL8 and CAL12 values in the PMC Oscillator Calibration Register (PMC\_OCR) are the default values set by Atmel during production. These values are stored in a specific Flash memory area different from the main memory plane. These values cannot be modified by the user and cannot be erased by a Flash erase command or by the ERASE pin. Values written by the user's application in PMC\_OCR are reset after each power up or peripheral reset.

### 29.5.2 Fast RC Oscillator Clock Frequency Adjustment

It is possible for the user to adjust the main RC oscillator frequency through PMC\_OCR. By default, SEL4/8/12 are low, so the RC oscillator will be driven with Flash calibration bits which are programmed during chip production.

The user can adjust the trimming of the 4/8/12 MHz Fast RC Oscillator through this register in order to obtain more accurate frequency (to compensate derating factors such as temperature and voltage).

In order to calibrate the oscillator lower frequency, SEL4 must be set to 1 and a good frequency value must be configured in CAL4. Likewise, SEL8/12 must be set to 1 and a trim value must be configured in CAL8/12 in order to adjust the other frequencies of the oscillator.

It is possible to adjust the oscillator frequency while operating from this clock. For example, when running on lowest frequency it is possible to change the CAL4 value if SEL4 is set in PMC\_OCR.

It is possible to restart, at anytime, a measurement of the main frequency by means of the RCMEAS bit in Main Clock Frequency Register (CKGR\_MCFR). Thus, when MAINFRDY flag reads 1, another read access on Main Clock Frequency Register (CKGR\_MCFR) provides an image of the frequency of the main clock on MAINF field. The software can calculate the error with an expected frequency and correct the CAL4 (or CAL8/CAL12) field accordingly. This may be used to compensate frequency drift due to derating factors such as temperature and/or voltage.

### 29.5.3 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator

After reset, the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator is disabled and it is not selected as the source of MAINCK.

The user can select the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator to be the source of MAINCK, as it provides a more accurate frequency. The software enables or disables the main oscillator so as to reduce power consumption by clearing the MOSCXTEN bit in the Main Oscillator Register (CKGR\_MOR).

When disabling the main oscillator by clearing the MOSCXTEN bit in CKGR\_MOR, the MOSCXTS bit in PMC\_SR is automatically cleared, indicating the Main Clock is off.

When enabling the main oscillator, the user must initiate the main oscillator counter with a value corresponding to the start-up time of the oscillator. This start-up time depends on the crystal frequency connected to the oscillator.

When the MOSCXTEN bit and the MOSCXTST are written in CKGR\_MOR to enable the main oscillator, the XIN and XOUT pins are automatically switched into oscillator mode and MOSCXTS bit in the Power Management Controller Status Register (PMC\_SR) is cleared and the counter starts counting down on the slow clock divided by 8 from the MOSCXTST value. Since the MOSCXTST value is coded with 8 bits, the maximum start-up time is about 62 ms.

When the counter reaches 0, the MOSCXTS bit is set, indicating that the main clock is valid. Setting the MOSCXTS bit in PMC\_IMR can trigger an interrupt to the processor.

### 29.5.4 Main Clock Oscillator Selection

The user can select either the 4/8/12 MHz Fast RC Oscillator or the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator to be the source of Main Clock.

The advantage of the 4/8/12 MHz Fast RC Oscillator is that it provides fast start-up time, this is why it is selected by default (to start up the system) and when entering Wait Mode.

The advantage of the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator is that it is very accurate.

The selection is made by writing the MOSCSEL bit in the Main Oscillator Register (CKGR\_MOR). The switch of the Main Clock source is glitch free, so there is no need to run out of SLCK, PLLACK or PLLBCK in order to change the selection. The MOSCSELS bit of the Power Management Controller Status Register (PMC\_SR) allows knowing when the switch sequence is done.

Setting the MOSCSELS bit in PMC\_IMR can trigger an interrupt to the processor.

Enabling the Fast RC Oscillator (MOSCRGEN = 1) and changing the Fast RC Frequency (MOSCCRF) at the same time is not allowed.

The Fast RC must be enabled first and its frequency changed in a second step.

### 29.5.5 Software Sequence to Detect the Presence of Fast Crystal

The frequency meter carried on the CKGR\_MCFR register is operating on the selected main clock and not on the fast crystal clock nor on the fast RC Oscillator clock.

Therefore, to check for the presence of the fast crystal clock, it is necessary to have the main clock (MAINCK) driven by the fast crystal clock (MOSCSEL=1).

The following software sequence order must be followed:

- MCK must select the slow clock (CSS=0 in the PMC\_MCKR register).
- Wait for the MCKRDY flag in the PMC\_SR register to be 1.
- The fast crystal must be enabled by programming 1 in the MOSCXTEN field in the CKGR\_MOR register with the MOSCXTST field being programmed to the appropriate value (see the Electrical Characteristics chapter).
- Wait for the MOSCXTS flag to be 1 in the PMC\_SR register to get the end of a start-up period of the fast crystal oscillator.
- Then, MOSCSEL must be programmed to 1 in the CKGR\_MOR register to select fast main crystal oscillator for the main clock.
- MOSCSEL must be read until its value equals 1.
- Then the MOSCSELS status flag must be checked in the PMC\_SR register.

At this point, 2 cases may occur (either MOSCSELS = 0 or MOSCSELS = 1).

- If MOSCSELS = 1, there is a valid crystal connected and its frequency can be determined by initiating a frequency measure by programming RCMEAS in the CKGR\_MCFR register.
- If MOSCSELS = 0, there is no fast crystal clock (either no crystal connected or a crystal clock out of specification). A frequency measure can reinforce this status by initiating a frequency measure by programming RCMEAS in the CKGR\_MCFR register.
- If MOSCSELS=0, the selection of the main clock must be programmed back to the main RC oscillator by writing MOSCSEL to 0 prior to disabling the fast crystal oscillator.
- If MOSCSELS=0, the crystal oscillator can be disabled (MOSCXTEN=0 in the CKGR\_MOR register).

### 29.5.6 Main Clock Frequency Counter

The device features a Main Clock frequency counter that provides the frequency of the Main Clock.

The Main Clock frequency counter is reset and starts incrementing at the Main Clock speed after the next rising edge of the Slow Clock in the following cases:

- When the 4/8/12 MHz Fast RC Oscillator clock is selected as the source of Main Clock and when this oscillator becomes stable (i.e., when the MOSCRCS bit is set)
- When the 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator is selected as the source of Main Clock and when this oscillator becomes stable (i.e., when the MOSCXTS bit is set)
- When the Main Clock Oscillator selection is modified
- When the RCMEAS bit of CKGR\_MCFR is written to 1.

Then, at the 16th falling edge of Slow Clock, the MAINFRDY bit in the Clock Generator Main Clock Frequency Register (CKGR\_MCFR) is set and the counter stops counting. Its value can be read in the MAINF field of CKGR\_MCFR and

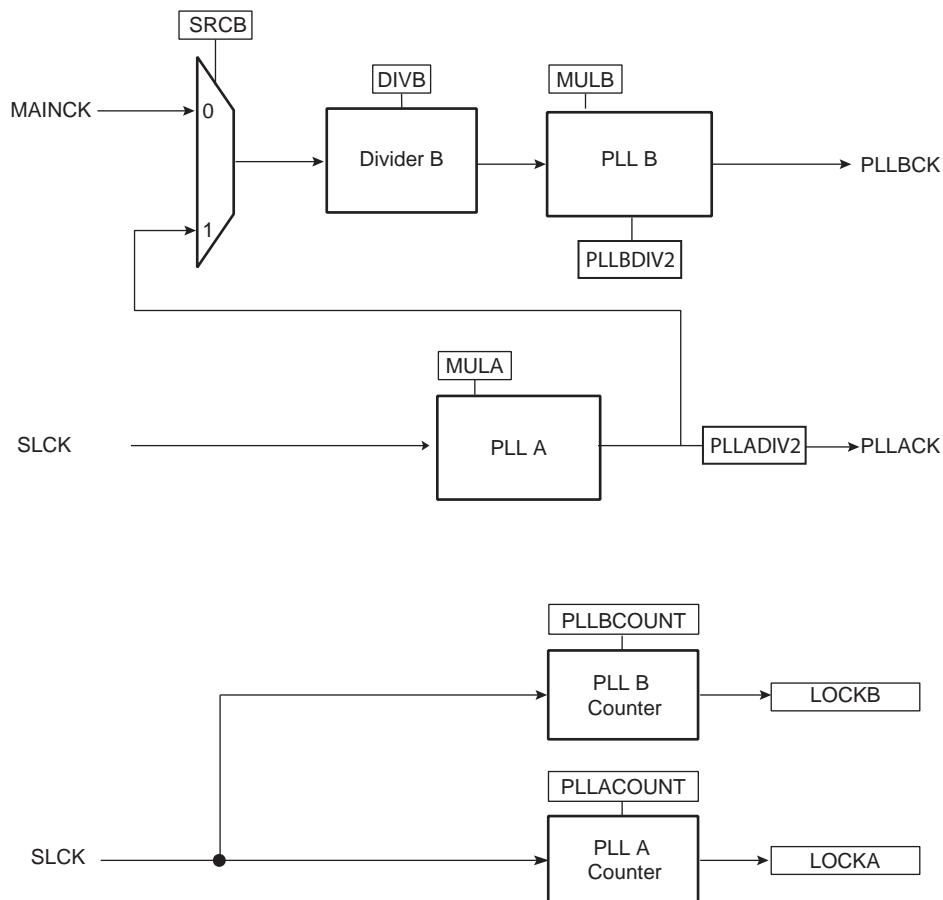


gives the number of Main Clock cycles during 16 periods of Slow Clock, so that the frequency of the 4/8/12 MHz Fast RC Oscillator or 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator can be determined.

## 29.6 Divider and PLL Block

The device features one Divider/two PLL Blocks that permit a wide range of frequencies to be selected on either the master clock, the processor clock or the programmable clock outputs. Figure 29-4 shows the block diagram of the dividers and PLL blocks.

Figure 29-4. Dividers and PLL Blocks Diagram



### 29.6.1 Divider and Phase Lock Loop Programming

The divider can be set between 1 and 255 in steps of 1. When a divider field (DIV) is set to 0, the output of the corresponding divider and the PLL output is a continuous signal at level 0. On reset, each DIV field is set to 0, thus the corresponding PLL input clock is set to 0.

The PLLs (PLLA, PLLB) allow multiplication of the SLCK clock source for PLLA or PLLA output clock or MAINCK divided output for PLLB. The PLL clock signal has a frequency that depends on the respective source signal frequency and on the parameters DIV (, DIVB) and MUL (MULA, MULB). The factor applied to the source signal frequency is  $(MUL + 1)/DIV$ . When MUL is written to 0 or PLLAEN=0, the PLL is disabled and its power consumption is saved. Re-enabling the PLL can be performed by writing a value higher than 0 in the MUL field and PLLAEN higher than 0.

To change the frequency of the PLLA, the PLLA must be first disabled by writing 0 in MULA field and 0 in PLLACOUNT field. Then, the PLLA can be configured to generate the new frequency by programming a new multiplier in MULA and the PLLACOUNT field in the same register access. See electrical characteristics to get the PLLACOUNT values covering the PLL transient time.

Whenever the PLL is re-enabled or one of its parameters is changed, the LOCK (LOCKA, LOCKB) bit in PMC\_SR is automatically cleared. The values written in the PLLCOUNT field (PLLACOUNT, PLLBCOUNT) in CKGR\_PLLR (CKGR\_PLLAR, CKGR\_PLLBR) are loaded in the PLL counter. The PLL counter then decrements at the speed of the Slow Clock until it reaches 0. At this time, the LOCK bit is set in PMC\_SR and can trigger an interrupt to the processor. The user has to load the number of Slow Clock cycles required to cover the PLL transient time into the PLLCOUNT field.

The PLL clock can be divided by 2 by writing the PLLDIV2 (PLLADIV2, PLLBDIV2) bit in PMC Master Clock Register (PMC\_MCKR).

The PLLADIV2 has no effect on PLLB clock input because the output of the PLLA is directly routed to PLLB input selection.

It is forbidden to change 4/8/12 MHz Fast RC Oscillator, or main selection in CKGR\_MOR register while Master Clock source is PLL and PLL reference clock is the Fast RC Oscillator.

The user must:

- Switch on the Main RC oscillator by writing 1 in CSS field of PMC\_MCKR.
- Change the frequency (MOSCRCF) or oscillator selection (MOSCSEL) in CKGR\_MOR.
- Wait for MOSCRCS (if frequency changes) or MOSCSELS (if oscillator selection changes) in PMC\_SR.
- Disable and then enable the PLL (LOCK in PMC\_IDR and PMC\_IER).
- Wait for LOCK flag in PMC\_SR.
- Switch back to PLL by writing the appropriate value to CSS field of PMC\_MCKR.

## 30. Power Management Controller (PMC)

### 30.1 Description

The Power Management Controller (PMC) optimizes power consumption by controlling all system and user peripheral clocks. The PMC enables/disables the clock inputs to many of the peripherals and the Cortex-M4 Processor.

The Supply Controller selects between the 32 kHz RC oscillator or the slow crystal oscillator. The unused oscillator is disabled automatically so that power consumption is optimized.

By default, at start-up the chip runs out of the Master Clock using the Fast RC Oscillator running at 4 MHz.

The user can trim the 8 and 12 MHz RC Oscillator frequencies by software.

### 30.2 Embedded Characteristics

The Power Management Controller provides the following clocks:

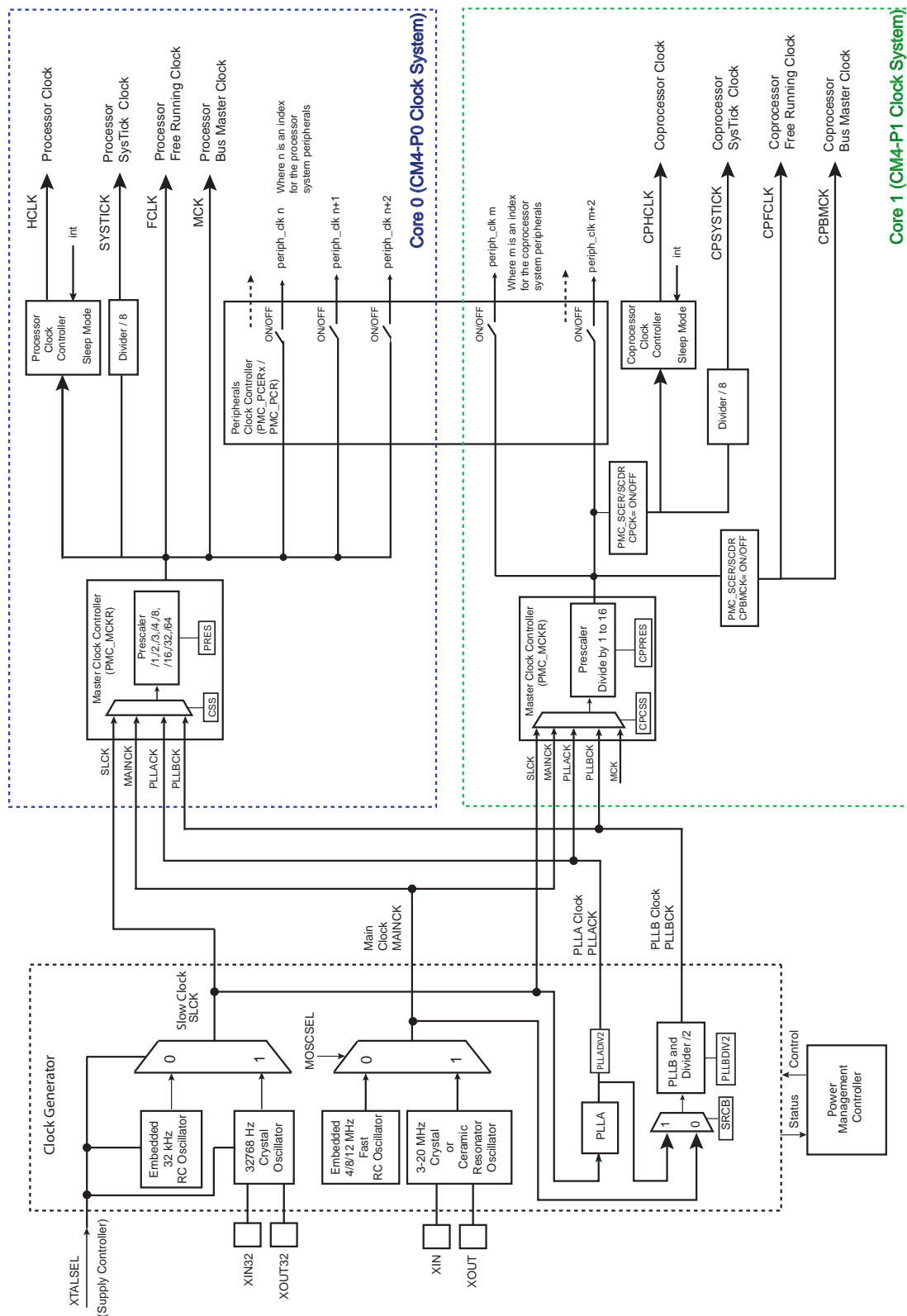
- MCK, the Master Clock, programmable from a few hundred Hz to the maximum operating frequency of the device. It is available to the modules running permanently, such as the Enhanced Embedded Flash Controller.
- Processor Clock (HCLK) and Coprocessor (second processor) Clock (CPHCLK), automatically switched off when entering the processor in Sleep Mode.
- Free running processor Clock (FCLK) and Free running Coprocessor Clock (CPFCLK)
- One SysTick external clock for each Cortex-M4 core
- Peripheral Clocks, typically MCK/CPMCK, provided to the embedded peripherals (USART, SPI, TWI, TC, etc.) and independently controllable. In order to reduce the number of clock names in a product, the Peripheral Clocks are named MCK in the product datasheet.
- Programmable Clock Outputs can be selected from the clocks provided by the clock generator and driven on the PCKx pins.
- Write Protected Registers

The Power Management Controller also provides the following operations on clocks:

- A main crystal oscillator clock failure detector.
- A 32768 kHz crystal oscillator frequency monitor.
- A frequency counter on main clock and an on-the-fly adjustable main RC oscillator frequency.

### 30.3 Block Diagram

Figure 30-1. General Clock Block Diagram



## 30.4 Master Clock Controller

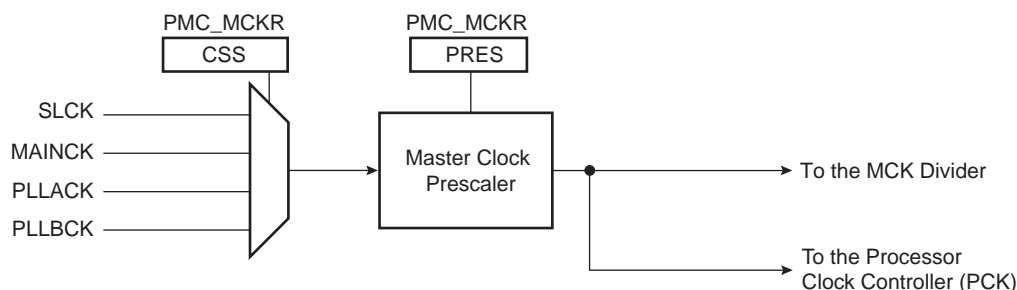
The Master Clock Controller provides selection and division of the Master Clock (MCK) and Coprocessor Master Clock (CPMCK). MCK is the clock provided to all the peripherals in the sub-system 0 and CPMCK is the clock provided to all peripherals in the sub-system 1). The Master Clock is selected from one of the clocks provided by the Clock Generator.

Selecting the Slow Clock provides a Slow Clock signal to the whole device. Selecting the Main Clock saves power consumption of the PLLs. The Master Clock Controller is made up of a clock selector and a prescaler.

The Master Clock selection is made by writing the CSS/CPCSS field (Clock Source Selection/Coprocessor Clock Source Selection) in PMC\_MCKR (Master Clock Register). The prescaler supports the division by a power of 2 of the selected clock between 1 and 64, and the division by 3. The PRES/PPRES field in PMC\_MCKR programs the prescaler.

Each time PMC\_MCKR is written to define a new Master Clock, the MCKRDY bit is cleared in PMC\_SR. It reads 0 until the Master Clock is established. Then, the MCKRDY bit is set and can trigger an interrupt to the processor. This feature is useful when switching from a highspeed clock to a lower one to inform the software when the change is actually done.

Figure 30-2. Master Clock Controller



## 30.5 Processor Clock Controller

The PMC features a Processor Clock Controller (HCLK) and Coprocessor Clock Controller (CPHCLK) that implements the Processor Sleep Mode. The Processor Clocks can be disabled by executing the WFI (WaitForInterrupt) or the WFE (WaitForEvent) processor instruction while the LPM bit is at 0 in the PMC Fast Start-up Mode Register (PMC\_FSMR).

The Processor Clock HCLK is enabled after a reset and is automatically re-enabled by any enabled interrupt. The Coprocessor Clock Controller CPHCLK is disabled after reset. It is up to the Master application to enable the CPHCLK clock. Similar to the Processor Clock HCLK, CPHCLK is automatically re-enabled by any enabled instruction after having executed a WFI instruction. The Processor Sleep Mode is achieved by disabling the Processor Clock, which is automatically re-enabled by any enabled fast or normal interrupt, or by the reset of the product.

When Processor Sleep Mode is entered, the current instruction is finished before the clock is stopped, but this does not prevent data transfers from other masters of the system bus.

## 30.6 SysTick Clock

The SysTick calibration value is fixed to 8000 which allows the generation of a time base of 1 ms with SysTick clock to the maximum frequency on MCK divided by 8.

## 30.7 Peripheral Clock Controller

The Power Management Controller controls the clocks of each embedded peripheral by means of the Peripheral Clock Controller. The user can individually enable and disable the Clock on the peripherals.

The user can also enable and disable these clocks by writing Peripheral Clock Enable 0 (PMC\_PCER0), Peripheral Clock Disable 0 (PMC\_PCDR0), Peripheral Clock Enable 1 (PMC\_PCER1) and Peripheral Clock Disable 1

(PMC\_PCDR1) registers. The status of the peripheral clock activity can be read in the Peripheral Clock Status Register (PMC\_PCSR0) and Peripheral Clock Status Register (PMC\_PCSR1).

If the peripherals located on the coprocessor system bus require data exchange with the co-processor or the main processor, the CPBMCK clock must be enabled prior to enable any co-processor peripheral clock.

When a peripheral clock is disabled, the clock is immediately stopped. The peripheral clocks are automatically disabled after a reset.

In order to stop a peripheral, it is recommended that the system software wait until the peripheral has executed its last programmed operation before disabling the clock. This is to avoid data corruption or erroneous behavior of the system.

The bit number within the Peripheral Clock Control registers (PMC\_PCER0-1, PMC\_PCDR0-1, and PMC\_PCSR0-1) is the Peripheral Identifier defined at the product level. The bit number corresponds to the interrupt source number assigned to the peripheral.

## 30.8 Free Running Processor Clock

The Free Running Processor Clock (FCLK) together with the Free Running Coprocessor Master Clock (CPFCLK) used for sampling interrupts and clocking debug blocks ensures that interrupts can be sampled, and sleep events can be traced, while the processor(s) is(are) sleeping. It is connected to Master Clock (MCK)/Coproprocessor Master Clock (CPMCK).

## 30.9 Programmable Clock Output Controller

The PMC controls 3 signals to be output on external pins, PCKx. Each signal can be independently programmed via the Programmable Clock Registers (PMC\_PCKx).

PCKx can be independently selected between the Slow Clock (SLCK), the Main Clock (MAINCK), the PLLA Clock (PLLACK), the PLLB Clock (PLLBCK), and the Master Clock (MCK) by writing the CSS field in PMC\_PCKx. Each output signal can also be divided by a power of 2 between 1 and 64 by writing the PRES (Prescaler) field in PMC\_PCKx.

Each output signal can be enabled and disabled by writing 1 in the corresponding bit, PCKx of PMC\_SCER and PMC\_SCDR, respectively. Status of the active programmable output clocks are given in the PCKx bits of PMC\_SCSR (System Clock Status Register).

Moreover, like the PCK, a status bit in PMC\_SR indicates that the Programmable Clock is actually what has been programmed in the Programmable Clock registers.

As the Programmable Clock Controller does not manage with glitch prevention when switching clocks, it is strongly recommended to disable the Programmable Clock before any configuration change and to re-enable it after the change is actually performed.

## 30.10 Main Processor Fast Start-up

The device allows the main processor to restart in less than 10 microseconds while the device is in Wait Mode.

The system enters Wait Mode either by writing the WAITMODE bit at 1 in the PMC Clock Generator Main Oscillator Register (CKGR\_MOR), or by executing the WaitForEvent (WFE) instruction of the processor while the LPM bit is at 1 in the PMC Fast Start-up Mode Register (PMC\_FSMR). Waiting for the MOSCRGEN bit to be cleared is strongly recommended to ensure that the core will not execute undesired instructions.

In case of a dual core activity, it is recommended to check the coprocessor state before instructing the main processor to enter the Wait Mode.

**Important:** Prior to instructing the system to enter the Wait Mode, the internal sources of wake-up must be cleared. It must be verified that none of the enabled external wake-up inputs (WKUP) hold an active polarity.

A Fast Start-up is enabled upon the detection of a programmed level on one of the 16 wake-up inputs (WKUP) or upon an active alarm from the RTC, RTT. The polarity of the 16 wake-up inputs is programmable by writing the PMC Fast Start-up Polarity Register (PMC\_FSPR).

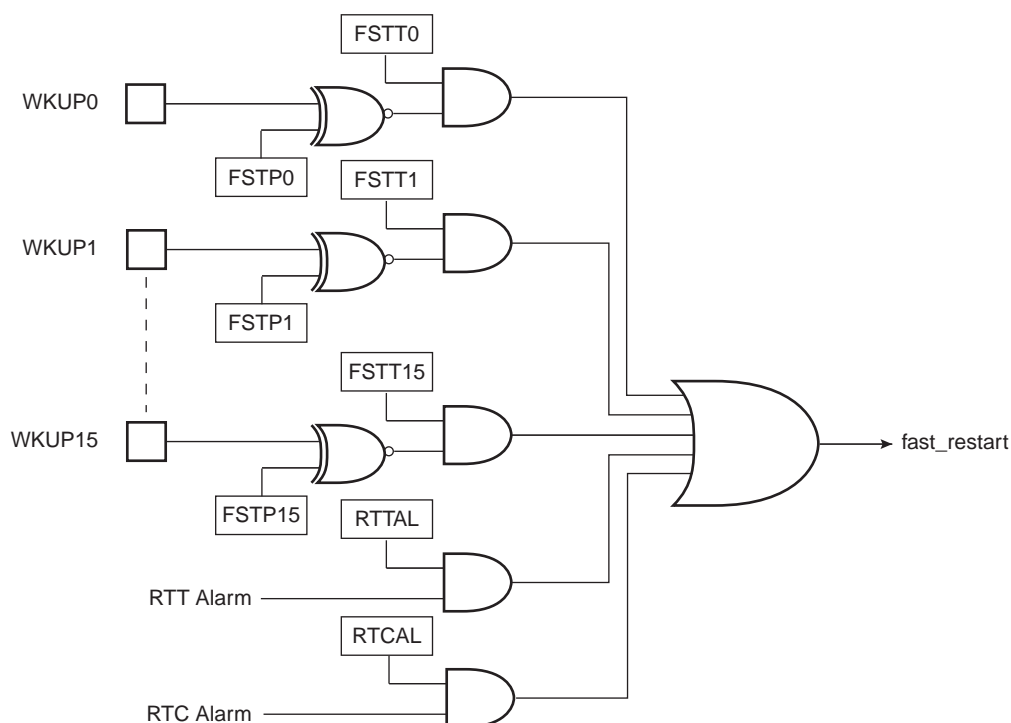
The Fast Restart circuitry, as shown in Figure 30-3, is fully asynchronous and provides a fast start-up signal to the Power Management Controller. As soon as the fast start-up signal is asserted, the embedded 4/8/12 MHz Fast RC Oscillator restarts automatically.

When entering the Wait Mode, the embedded flash can be placed in low power mode depending on the configuration of the FLPM in PMC\_FSMR register. This bitfield can be programmed anytime and will be taken into account at the next time the system enters Wait Mode.

The power consumption reduction is optimal when configuring 1 (deep power down mode) in FLPM. If 0 is programmed (standby mode), the power consumption is slightly higher as compared to the deep power down mode.

When programming 2 in FLPM, the Wait Mode flash power consumption is equivalent to the active mode when there is no read access on the flash.

**Figure 30-3. Fast Start-up Circuitry**



Each wake-up input pin and alarm can be enabled to generate a Fast Start-up event by writing 1 to the corresponding bit in the Fast Start-up Mode Register (PMC\_FSMR).

The user interface does not provide any status for Fast Start-up, but the user can easily recover this information by reading the PIO Controller and the status registers of the RTC, RTT.

### 30.11 Coprocessor Sleep Mode

The coprocessor enters Sleep Mode by executing the WaitForInterrupt (WFI) instruction of the coprocessor. Any enabled interrupt can wake the processor up.

### 30.12 Main Crystal Clock Failure Detector

The clock failure detector monitors the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator to identify an eventual defect of this oscillator (for example, if the crystal is unconnected).

The clock failure detector can be enabled or disabled by means of the CFDEN bit in the PMC Clock Generator Main Oscillator Register (CKGR\_MOR). After reset, the detector is disabled. However, if the 3 to 20 MHz Crystal or Ceramic Resonator-based Oscillator is disabled, the clock failure detector is disabled too.

The slow RC oscillator must be enabled. A failure is detected by means of a counter incrementing on the 3 to 20 MHz Crystal oscillator or Ceramic Resonator-based oscillator clock edge and timing logic clocked on the slow clock RC oscillator controlling the counter. The counter is cleared when the slow clock RC oscillator signal is low and enabled when the slow clock RC oscillator is high. Thus the failure detection time is 1 slow clock RC oscillator clock period. If, during the high level period of the slow clock RC oscillator, less than 8 fast crystal oscillator clock periods have been counted, then a failure is declared.

If a failure of the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator clock is detected, the CFDEV flag is set in the PMC Status Register (PMC\_SR), and generates an interrupt if it is not masked. The interrupt remains active until a read operation in the PMC\_SR register. The user can know the status of the clock failure detector at any time by reading the CFDS bit in the PMC\_SR register.

If the 3 to 20 MHz Crystal or Ceramic Resonator-based oscillator clock is selected as the source clock of MAINCK (MOSCSEL = 1), and if the Master Clock Source is PLLACK or PLLBCK (CSS = 2), a clock failure detection automatically forces MAINCK to be the source clock for the master clock (MCK). Then, regardless of the PMC configuration, a clock failure detection automatically forces the 4/8/12 MHz Fast RC Oscillator to be the source clock for MAINCK. If the Fast RC Oscillator is disabled when a clock failure detection occurs, it is automatically re-enabled by the clock failure detection mechanism.

It takes 2 slow clock RC oscillator cycles to detect and switch from the 3 to 20 MHz Crystal, or Ceramic Resonator-based Oscillator, to the 4/8/12 MHz Fast RC Oscillator if the Master Clock source is Main Clock, or 3 slow clock RC oscillator cycles if the Master Clock source is PLLACK or PLLBCK.

The user can know the status of the clock failure detector at any time by reading the FOS bit in the PMC\_SR register.

This fault output remains active until the defect is detected and until it is cleared by the bit FOCLR in the PMC Fault Output Clear Register (PMC\_FOCR).

### 30.13 Slow Crystal Clock Frequency Monitor

The frequency of the slow clock crystal oscillator can be monitored by means of logic driven by the main RC oscillator known as a reliable clock source. This function is enabled by configuring the XT32KFME bit of the Main Oscillator Register (CKGR\_MOR).

An error flag (XT32KERR in PMC\_SR) is asserted when the slow clock crystal oscillator frequency is out of the +/- 10% nominal frequency value (i.e. 32768 kHz). The error flag can be cleared only if the slow clock frequency monitoring is disabled.

When the main RC oscillator frequency is 4 MHz, the accuracy of the measurement is +/-40% as this frequency is not trimmed during production. Therefore, +/-10% accuracy is obtained only if the RC oscillator frequency is configured for 8 or 12 MHz.

The monitored clock frequency is declared invalid if at least 4 consecutive clock period measurement results are over the nominal period +/-10%.

Due to the possible frequency variation of the embedded main RC oscillator acting as reference clock for the monitor logic, any slow clock crystal frequency deviation over +/-10% of the nominal frequency is systematically reported as an error by means of XT32KERR in PMC\_SR. Between -1% and -10% and +1% and +10%, the error is not systematically reported.

Thus only a crystal running at 32768 kHz frequency ensures that the error flag will not be asserted. The permitted drift of the crystal is 10000ppm (1%), which allows any standard crystal to be used.

If the main RC frequency needs to be changed while the slow clock frequency monitor is operating, the monitoring must be stopped prior to change the main RC frequency. Then it can be re-enabled as soon as MOSCRCS is set in PMC\_SR register.

The error flag can be defined as an interrupt source of the PMC by setting the XT32KERR bit of PMC\_IER.



## 30.14 Programming Sequence

### 1. Enabling the Main Oscillator:

The main oscillator is enabled by setting the MOSCXTEN field in the Main Oscillator Register (CKGR\_MOR). The user can define a start-up time. This can be achieved by writing a value in the MOSCXTST field in CKGR\_MOR. Once this register has been correctly configured, the user must wait for MOSCXTS field in the PMC\_SR register to be set. This can be done either by polling the status register, or by waiting the interrupt line to be raised if the associated interrupt to MOSCXTS has been enabled in the PMC\_IER register.

### 2. Checking the Main Oscillator Frequency (Optional):

In some situations the user may need an accurate measure of the main clock frequency. This measure can be accomplished via the Main Clock Frequency Register (CKGR\_MCFR).

Once the MAINFRDY field is set in CKGR\_MCFR, the user may read the MAINF field in CKGR\_MCFR by performing another CKGR\_MCFR read access. This provides the number of main clock cycles within sixteen slow clock cycles.

### 3. Setting PLL and Divider:

All parameters needed to configure PLLA and the divider are located in CKGR\_PLLAxR.

The DIV field is used to control the divider itself. It must be set to 1 when PLL is used. By default, DIV parameter is set to 0 which means that the divider is turned off.

The MUL field is the PLL multiplier factor. This parameter can be programmed between 0 and 254. If MUL is set to 0, PLL will be turned off, otherwise the PLL output frequency is PLL input frequency multiplied by (MUL + 1).

The PLLCOUNT field specifies the number of slow clock cycles before the LOCK bit is set in PMC\_SR, after CKGR\_PLLA(B)R has been written.

Once the CKGR\_PLL register has been written, the user must wait for the LOCK bit to be set in the PMC\_SR. This can be done either by polling the status register or by waiting the interrupt line to be raised if the associated interrupt to LOCK has been enabled in PMC\_IER. All parameters in CKGR\_PLLA(B)R can be programmed in a single write operation. If at some stage one of the following parameters, MUL or DIV is modified, the LOCK bit will go low to indicate that PLL is not ready yet. When PLL is locked, LOCK will be set again. The user is constrained to wait for LOCK bit to be set before using the PLL output clock.

### 4. Selection of Master Clock and Processor Clock

The Master Clock and the Processor Clock are configurable via the Master Clock Register (PMC\_MCKR).

The CSS field is used to select the Master Clock divider source. By default, the selected clock source is main clock.

The PRES field is used to control the Master Clock prescaler. The user can choose between different values (1, 2, 3, 4, 8, 16, 32, 64). Master Clock output is prescaler input divided by PRES parameter. By default, PRES parameter is set to 1 which means that master clock is equal to main clock.

Once PMC\_MCKR has been written, the user must wait for the MCKRDY bit to be set in PMC\_SR. This can be done either by polling the status register or by waiting for the interrupt line to be raised if the associated interrupt to MCKRDY has been enabled in the PMC\_IER register.

The PMC\_MCKR must not be programmed in a single write operation. The preferred programming sequence for PMC\_MCKR is as follows:

- If a new value for CSS field corresponds to PLL Clock,
  - Program the PRES field in PMC\_MCKR.
  - Wait for the MCKRDY bit to be set in PMC\_SR.
  - Program the CSS field in PMC\_MCKR.
  - Wait for the MCKRDY bit to be set in PMC\_SR.
- If a new value for CSS field corresponds to Main Clock or Slow Clock,
  - Program the CSS field in PMC\_MCKR.
  - Wait for the MCKRDY bit to be set in the PMC\_SR.

- Program the PRES field in PMC\_MCKR.
- Wait for the MCKRDY bit to be set in PMC\_SR.

If at some stage one of the following parameters, CSS or PRES is modified, the MCKRDY bit will go low to indicate that the Master Clock and the Processor Clock are not ready yet. The user must wait for MCKRDY bit to be set again before using the Master and Processor Clocks.

Note: IF PLLx clock was selected as the Master Clock and the user decides to modify it by writing in CKGR\_PLLR, the MCKRDY flag will go low while PLL is unlocked. Once PLL is locked again, LOCK goes high and MCKRDY is set. While PLL is unlocked, the Master Clock selection is automatically changed to Slow Clock. For further information, see [Section 30.15.2 “Clock Switching Waveforms” on page 524](#).

Code Example:

```
write_register(PMC_MCKR, 0x00000001)
wait (MCKRDY=1)
write_register(PMC_MCKR, 0x00000011)
wait (MCKRDY=1)
```

The Master Clock is main clock divided by 2.

The Processor Clock is the Master Clock.

## 5. Selection of Programmable Clocks

Programmable clocks are controlled via registers, PMC\_SCER, PMC\_SCDR and PMC\_SCSR.

Programmable clocks can be enabled and/or disabled via PMC\_SCER and PMC\_SCDR. 3 Programmable clocks can be enabled or disabled. The PMC\_SCSR provides a clear indication as to which Programmable clock is enabled. By default all Programmable clocks are disabled.

Programmable Clock Registers (PMC\_PCKx) are used to configure Programmable clocks.

The CSS field is used to select the Programmable clock divider source. Four clock options are available: main clock, slow clock, PLLACK, PLLBCK. By default, the clock source selected is slow clock.

The PRES field is used to control the Programmable clock prescaler. It is possible to choose between different values (1, 2, 4, 8, 16, 32, 64). Programmable clock output is prescaler input divided by PRES parameter. By default, the PRES parameter is set to 0 which means that master clock is equal to slow clock.

Once PMC\_PCKx has been programmed, The corresponding Programmable clock must be enabled and the user is constrained to wait for the PCKRDYx bit to be set in PMC\_SR. This can be done either by polling the status register or by waiting the interrupt line to be raised, if the associated interrupt to PCKRDYx has been enabled in the PMC\_IER register. All parameters in PMC\_PCKx can be programmed in a single write operation.

If the CSS and PRES parameters are to be modified, the corresponding Programmable clock must be disabled first. The parameters can then be modified. Once this has been done, the user must re-enable the Programmable clock and wait for the PCKRDYx bit to be set.

## 6. Enabling Peripheral Clocks

Once all of the previous steps have been completed, the peripheral clocks can be enabled and/or disabled via registers PMC\_PCER0, PMC\_PCER, PMC\_PCDR0 and PMC\_PCDR.

## 30.15 Clock Switching Details

### 30.15.1 Master Clock Switching Timings

Table 30-1 and give the worst case timings required for the Master Clock to switch from one selected clock to another one. This is in the event that the prescaler is de-activated. When the prescaler is activated, an additional time of 64 clock cycles of the newly selected clock has to be added.

**Table 30-1. Clock Switching Timings (Worst Case)**

<b>To</b>	<b>From</b>	<b>Main Clock</b>	<b>SLCK</b>	<b>PLL Clock</b>
Main Clock		–	4 x SLCK + 2.5 x Main Clock	3 x PLL Clock + 4 x SLCK + 1 x Main Clock
SLCK		0.5 x Main Clock + 4.5 x SLCK	–	3 x PLL Clock + 5 x SLCK
PLL Clock		0.5 x Main Clock + 4 x SLCK + PLLCOUNT x SLCK + 2.5 x PLLx Clock	2.5 x PLL Clock + 5 x SLCK + PLLCOUNT x SLCK	2.5 x PLL Clock + 4 x SLCK + PLLCOUNT x SLCK

- Notes: 1. PLL designates either the PLLA or the PLLB Clock.  
2. PLLCOUNT designates either PLLACOUNT or PLLBCOUNT.

**Table 30-2. Clock Switching Timings between Two PLLs (Worst Case)**

<b>To</b>	<b>From</b>	<b>PLLA Clock</b>	<b>PLLB Clock</b>
PLLA Clock		2.5 x PLLA Clock + 4 x SLCK + PLLACOUNT x SLCK	3 x PLLA Clock + 4 x SLCK + 1.5 x PLLA Clock
PLLB Clock		3 x PLLB Clock + 4 x SLCK + 1.5 x PLLB Clock	2.5 x PLLB Clock + 4 x SLCK + PLLBCOUNT x SLCK

30.15.2 Clock Switching Waveforms

Figure 30-4. Switch Master Clock from Slow Clock to PLLx Clock

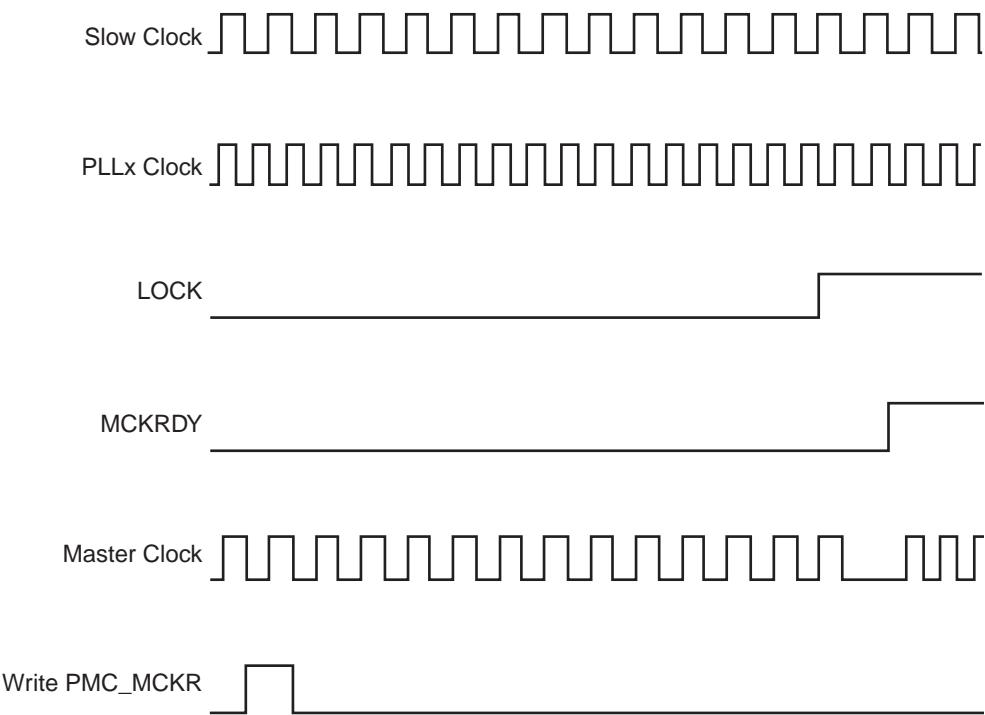
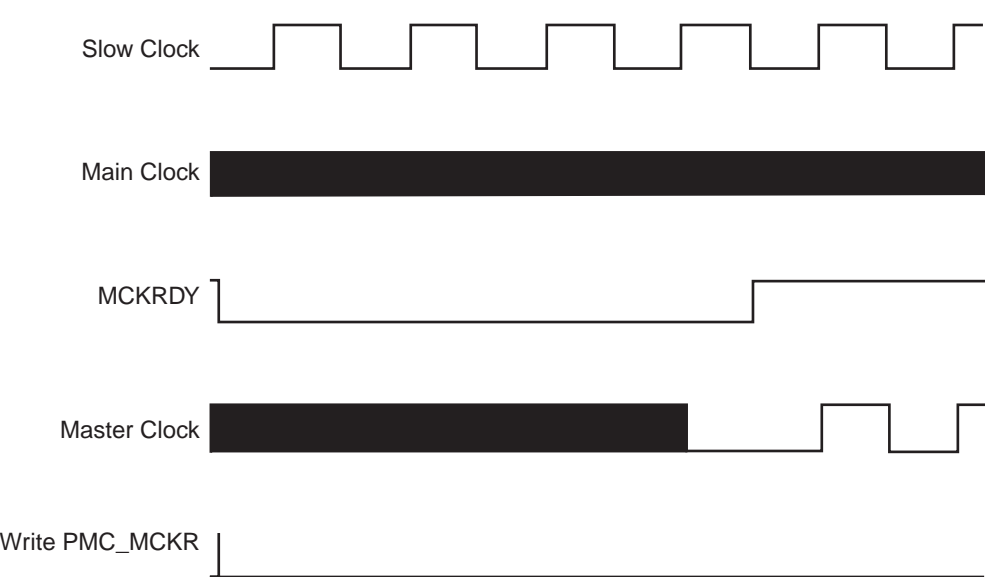
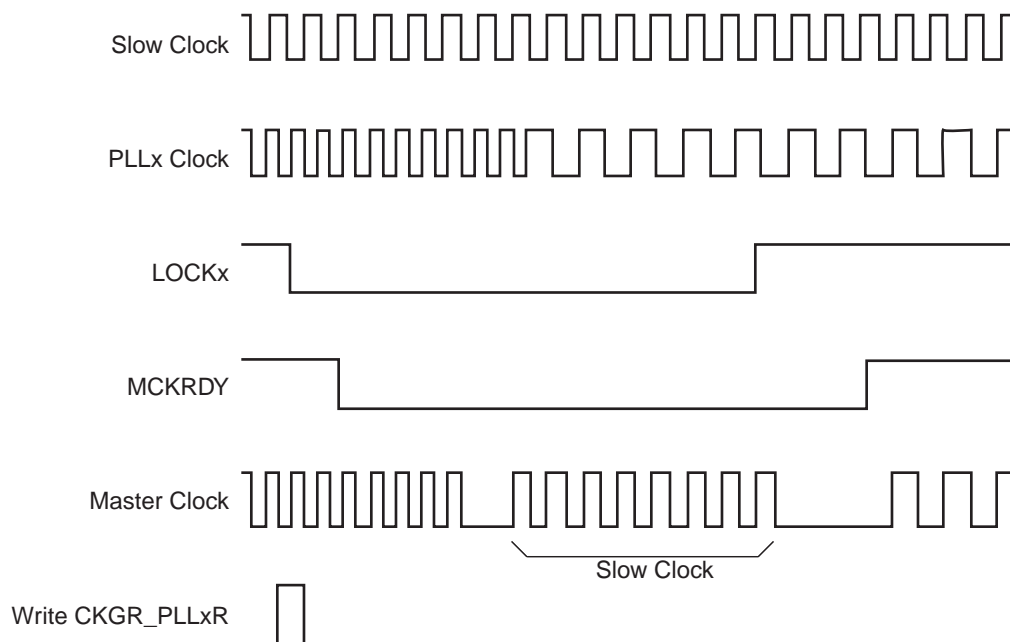


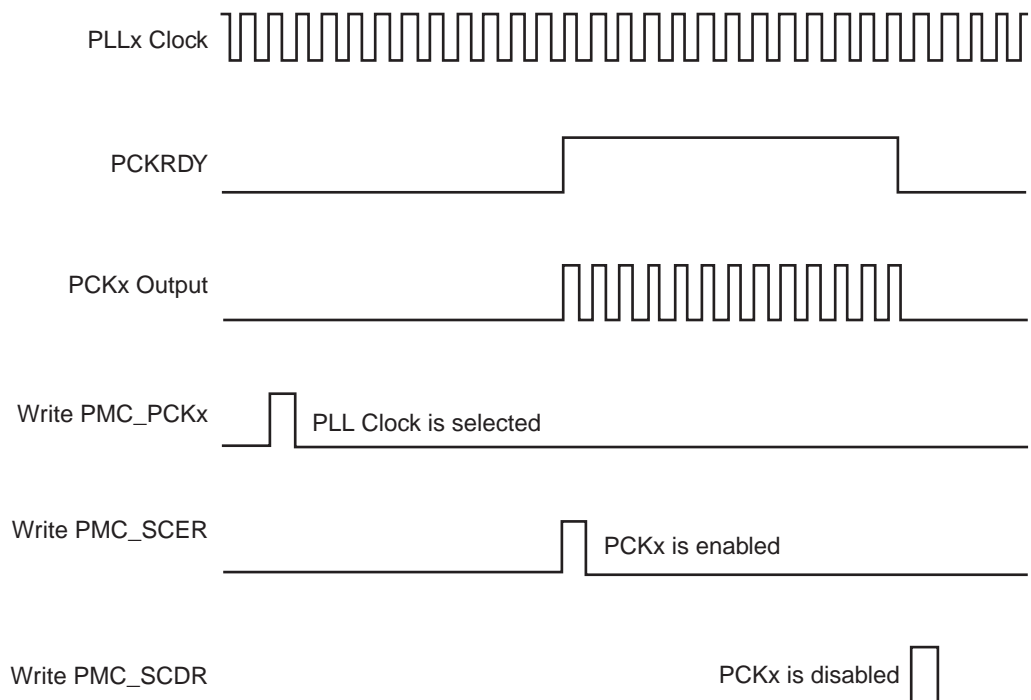
Figure 30-5. Switch Master Clock from Main Clock to Slow Clock



**Figure 30-6. Change PLLx Programming**



**Figure 30-7. Programmable Clock Output Programming**



## 30.16 Write Protection Registers

To prevent any single software error that may corrupt PMC behavior, certain address spaces can be write protected by setting the WPEN bit in the “[PMC Write Protect Mode Register](#)” (PMC\_WPMR).

If a write access to the protected registers is detected, then the WPVS flag in the PMC Write Protect Status Register (PMC\_WPSR) is set and the field WPVSRC indicates in which register the write access has been attempted.

The WPVS flag is reset by writing the PMC Write Protect Mode Register (PMC\_WPMR) with the appropriate access key, WPKEY.

The protected registers are:

- “[PMC System Clock Enable Register](#)”
- “[PMC System Clock Disable Register](#)”
- “[PMC Peripheral Clock Enable Register 0](#)”
- “[PMC Peripheral Clock Disable Register 0](#)”
- “[PMC Clock Generator Main Oscillator Register](#)”
- “[PMC Clock Generator PLLA Register](#)”
- “[PMC Clock Generator PLLB Register](#)”
- “[PMC Master Clock Register](#)”
- “[PMC Programmable Clock Register](#)”
- “[PMC Fast Start-up Mode Register](#)”
- “[PMC Fast Start-up Polarity Register](#)”
- “[PMC Coprocessor Fast Start-up Mode Register](#)”
- “[PMC Peripheral Clock Enable Register 1](#)”
- “[PMC Peripheral Clock Disable Register 1](#)”
- “[PMC Oscillator Calibration Register](#)”

## 30.17 Power Management Controller (PMC) User Interface

**Table 30-3. Register Mapping**

Offset	Register	Name	Access	Reset
0x0000	System Clock Enable Register	PMC_SCER	Write-only	–
0x0004	System Clock Disable Register	PMC_SCDR	Write-only	–
0x0008	System Clock Status Register	PMC_SCSR	Read-only	0x0000_0001
0x000C	Reserved	–	–	–
0x0010	Peripheral Clock Enable Register 0	PMC_PCER0	Write-only	–
0x0014	Peripheral Clock Disable Register 0	PMC_PCDR0	Write-only	–
0x0018	Peripheral Clock Status Register 0	PMC_PCSR0	Read-only	0x0000_0000
0x0020	Main Oscillator Register	CKGR_MOR	Read-write	0x0000_0008
0x0024	Main Clock Frequency Register	CKGR_MCFR	Read-write	0x0000_0000
0x0028	PLLA Register	CKGR_PLLAR	Read-write	0x0000_3F00
0x002C	PLLB Register	CKGR_PLLBR	Read-write	0x0000_3F00
0x0030	Master Clock Register	PMC_MCKR	Read-write	0x0000_0001
0x0034 - 0x003C	Reserved	–	–	–
0x0040	Programmable Clock 0 Register	PMC_PCK0	Read-write	0x0000_0000
0x0044	Programmable Clock 1 Register	PMC_PCK1	Read-write	0x0000_0000
0x0048	Programmable Clock 2 Register	PMC_PCK2	Read-write	0x0000_0000
0x004C - 0x005C	Reserved	–	–	–
0x0060	Interrupt Enable Register	PMC_IER	Write-only	–
0x0064	Interrupt Disable Register	PMC_IDR	Write-only	–
0x0068	Status Register	PMC_SR	Read-only	0x0001_0008
0x006C	Interrupt Mask Register	PMC_IMR	Read-only	0x0000_0000
0x0070	Fast Start-up Mode Register	PMC_FSMR	Read-write	0x0000_0000
0x0074	Fast Start-up Polarity Register	PMC_FSPR	Read-write	0x0000_0000
0x0078	Fault Output Clear Register	PMC_FOCR	Write-only	–
0x007C	Coprocessor Fast Start-up Mode Register	PMC_CPFMR	Read-write	0x0000_0000
0x0080- 0x00E0	Reserved	–	–	–
0x00E4	Write Protect Mode Register	PMC_WPMR	Read-write	0x0
0x00E8	Write Protect Status Register	PMC_WPSR	Read-only	0x0
0x00EC-0x00FC	Reserved	–	–	–
0x0100	Peripheral Clock Enable Register 1	PMC_PCER1	Write-only	–
0x0104	Peripheral Clock Disable Register 1	PMC_PCDR1	Write-only	–
0x0108	Peripheral Clock Status Register 1	PMC_PCSR1	Read-only	0x0000_0000
0x010C	Reserved	–	–	–
0x0110	Oscillator Calibration Register	PMC_OCR	Read-write	0x0040_4040
0x130	PLL Maximum Multiplier Value Register	PMC_PMMR	Read-Write	0x07FF07FF

Note: If an offset is not listed in the table it must be considered as “reserved”.

### 30.17.1 PMC System Clock Enable Register

**Name:** PMC\_SCER

**Address:** 0x400E0400

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
CPKEY				–	–	CPBMCK	CPCK
15	14	13	12	11	10	9	8
–	–	–	–	–	PCK2	PCK1	PCK0
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)” .

- **PCKx: Programmable Clock x Output Enable**

0 = No effect.

1 = Enables the corresponding Programmable Clock output.

- **CPCK: Coprocessor (Second Processor) Clocks Enable**

0 = No effect.

1 = Enables the corresponding Coprocessor Clocks (CPHCLK, CPSYSTICK) if CPKEY = 0xA

- **CPBMCK: Coprocessor Bus Master Clocks Enable**

0 = No effect.

1 = Enables the corresponding Coprocessor Bus Master Clock (CPBMCK, CPFCLK) if CPKEY = 0xA

**Note:** Enabling CPBMCK must be performed prior or at the same time as CPCK is programmed to 1 in PMC\_SCER register or prior communication with one the peripherals of the coprocessor system bus.

- **CPKEY: Coprocessor Clocks Enable Key**

Value	Name	Description
0xA	PASSWD	This field must be written to 0xA in order to validate CPCK field.



### 30.17.2 PMC System Clock Disable Register

**Name:** PMC\_SCDR

**Address:** 0x400E0404

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
CPKEY				–	–	CPBMCK	CPCK
15	14	13	12	11	10	9	8
–	–	–	–	–	PCK2	PCK1	PCK0
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)” .

- **PCKx: Programmable Clock x Output Disable**

0 = No effect.

1 = Disables the corresponding Programmable Clock output.

- **CPCK: Coprocessor Clocks Disable**

0 = No effect.

1 = Enables the corresponding Coprocessor Clocks (CPHCLK,CPFCLK,CPSYSTICK) if CPKEY = 0xA.

- **CPBMCK: Coprocessor Bus Master Clocks Disable**

0 = No effect.

1 = Disables the corresponding Coprocessor Bus Master Clock (CPBMCK,CPFCLK) if CPKEY = 0xA

Note: Disabling CPBMCK must not be performed if CPCK is 1 in PMC\_SCSR register.

- **CPKEY: Coprocessor Clocks Disable Key**

Value	Name	Description
0xA	PASSWD	This field must be written to 0xA in order to validate CPCK field.

### 30.17.3 PMC System Clock Status Register

**Name:** PMC\_SCSR

**Address:** 0x400E0408

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	CPBMCK	CPCK
15	14	13	12	11	10	9	8
–	–	–	–	–	PCK2	PCK1	PCK0
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

- **PCKx: Programmable Clock x Output Status**

0 = The corresponding Programmable Clock output is disabled.

1 = The corresponding Programmable Clock output is enabled.

- **CPCK: Coprocessor (Second Processor) Clocks Status**

0 = Coprocessor Clocks (CPHCLK,CPSYSTICK) are disabled (value after reset).

1 = Coprocessor Clocks (CPHCLK,CPSYSTICK) are enabled.

- **CPBMCK: Coprocessor Bus Master Clock Status**

0 = Coprocessor Clocks (CPBMCK,CPFCLK) are disabled (value after reset).

1 = Coprocessor Clocks (CPBMCK,CPFCLK) are enabled.

### 30.17.4 PMC Peripheral Clock Enable Register 0

**Name:** PMC\_PCER0

**Address:** 0x400E0410

**Access:** Write-only

31	30	29	28	27	26	25	24
PID31	–	PID29	PID28	PID27	PID26	PID25	PID24
23	22	21	20	19	18	17	16
PID23	–	PID21	PID20	PID19	PID18	PID17	PID16
15	14	13	12	11	10	9	8
PID15	PID14	PID13	PID12	PID11	PID10	PID9	PID8
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **PIDx: Peripheral Clock x Enable**

0 = No effect.

1 = Enables the corresponding peripheral clock.

**Note:** To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet. Other peripherals can be enabled in PMC\_PCER1 ([Section 30.17.23 “PMC Peripheral Clock Enable Register 1”](#)).

**Note:** Programming the control bits of the Peripheral ID that are not implemented has no effect on the behavior of the PMC.

### 30.17.5 PMC Peripheral Clock Disable Register 0

**Name:** PMC\_PCDR0

**Address:** 0x400E0414

**Access:** Write-only

31	30	29	28	27	26	25	24
PID31	–	PID29	PID28	PID27	PID26	PID25	PID24
23	22	21	20	19	18	17	16
PID23	–	PID21	PID20	PID19	PID18	PID17	PID16
15	14	13	12	11	10	9	8
PID15	PID14	PID13	PID12	PID11	PID10	PID9	PID8
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **PIDx: Peripheral Clock x Disable**

0 = No effect.

1 = Disables the corresponding peripheral clock.

**Note:** To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet. Other peripherals can be disabled in PMC\_PCDR1 ([Section 30.17.24 “PMC Peripheral Clock Disable Register 1”](#)).

### 30.17.6 PMC Peripheral Clock Status Register 0

**Name:** PMC\_PCSR0

**Address:** 0x400E0418

**Access:** Read-only

31	30	29	28	27	26	25	24
PID31	–	PID29	PID28	PID27	PID26	PID25	PID24
23	22	21	20	19	18	17	16
PID23	–	PID21	PID20	PID19	PID18	PID17	PID16
15	14	13	12	11	10	9	8
PID15	PID14	PID13	PID12	PID11	PID10	PID9	PID8
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

- **PIDx: Peripheral Clock x Status**

0 = The corresponding peripheral clock is disabled.

1 = The corresponding peripheral clock is enabled.

**Note:** To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet. Other peripherals status can be read in PMC\_PCSR1 ([Section 30.17.25 “PMC Peripheral Clock Status Register 1”](#)).

### 30.17.7 PMC Clock Generator Main Oscillator Register

**Name:** CKGR\_MOR

**Address:** 0x400E0420

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	XT32KFME	CFDEN	MOSCSEL
23	22	21	20	19	18	17	16
KEY							
15	14	13	12	11	10	9	8
MOSCXTST							
7	6	5	4	3	2	1	0
–	MOSCRCF			MOSCRCE	WAITMODE	MOSCXTBY	MOSCXTEN

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)”.

- **MOSCXTEN: Main Crystal Oscillator Enable**

A crystal must be connected between XIN and XOUT.

0 = The Main Crystal Oscillator is disabled.

1 = The Main Crystal Oscillator is enabled. MOSCXTBY must be set to 0.

When MOSCXTEN is set, the MOSCXTS flag is set once the Main Crystal Oscillator start-up time is achieved.

- **MOSCXTBY: Main Crystal Oscillator Bypass**

0 = No effect.

1 = The Main Crystal Oscillator is bypassed. MOSCXTEN must be set to 0. An external clock must be connected on XIN.

When MOSCXTBY is set, the MOSCXTS flag in PMC\_SR is automatically set.

Clearing MOSCXTEN and MOSCXTBY bits allows resetting the MOSCXTS flag.

- **WAITMODE: Wait Mode Command**

0 = No effect.

1 = Enters the device in Wait Mode.

Note: The WAITMODE bit is write-only.

- **MOSCRCE: Main On-Chip RC Oscillator Enable**

0 = The Main On-Chip RC Oscillator is disabled.

1 = The Main On-Chip RC Oscillator is enabled.

When MOSCRCE is set, the MOSCRCS flag is set once the Main On-Chip RC Oscillator start-up time is achieved.

- **MOSCRCF: Main On-Chip RC Oscillator Frequency Selection**

At start-up, the Main On-Chip RC Oscillator frequency is 4 MHz.

Value	Name	Description
0x0	4_MHz	The Fast RC Oscillator Frequency is at 4 MHz (default)
0x1	8_MHz	The Fast RC Oscillator Frequency is at 8 MHz
0x2	12_MHz	The Fast RC Oscillator Frequency is at 12 MHz

Note: MOSCRCF must be changed only if MOSCRCS is set in the PMC\_SR register. Therefore MOSCRCF and MOSCRCE cannot be changed at the same time.

- **MOSCXTST: Main Crystal Oscillator Start-up Time**

Specifies the number of Slow Clock cycles multiplied by 8 for the Main Crystal Oscillator start-up time.

- **KEY: Write Access Password**

Value	Name	Description
0x37	PASSWD	Writing any other value in this field aborts the write operation. Always reads as 0.

- **MOSCSEL: Main Oscillator Selection**

0 = The Main On-Chip RC Oscillator is selected.

1 = The Main Crystal Oscillator is selected.

- **CFDEN: Clock Failure Detector Enable**

0 = The Clock Failure Detector is disabled.

1 = The Clock Failure Detector is enabled.

Note: 1. The slow RC oscillator must be enabled when the CFDEN is enabled.

- **XT32KFME: Slow Crystal Oscillator Frequency Monitoring Enable**

0 = The 32768 Hz Crystal Oscillator Frequency Monitoring is disabled.

1 = The 32768 Hz Crystal Oscillator Frequency Monitoring is enabled.

### 30.17.8 PMC Clock Generator Main Clock Frequency Register

**Name:** CKGR\_MCFR

**Address:** 0x400E0424

**Access:** Read-Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	RCMEAS	–	–	–	MAINFRDY
15	14	13	12	11	10	9	8
MAINF							
7	6	5	4	3	2	1	0
MAINF							

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)” .

- **MAINF: Main Clock Frequency**

Gives the number of Main Clock cycles within 16 Slow Clock periods.

- **MAINFRDY: Main Clock Ready**

0 = MAINF value is not valid or the Main Oscillator is disabled or a measure has just been started by means of RCMEAS.

1 = The Main Oscillator has been enabled previously and MAINF value is available.

**Note:** To ensure that a correct value is read on the MAINF bitfield, the MAINFRDY flag must be read at 1 then another read access must be performed on the register to get a stable value on the MAINF bitfield.

- **RCMEAS: RC Oscillator Frequency Measure (write-only)**

0 = No effect.

1 = Restarts measuring of the main RC frequency. MAINF will carry the new frequency as soon as a low to high transition occurs on the MAINFRDY flag.

The measure is performed on the main frequency (i.e. not limited to RC oscillator only), but if the main clock frequency source is the fast crystal oscillator, the restart of measuring is not needed because of the well known stability of crystal oscillators.



### 30.17.9 PMC Clock Generator PLLA Register

**Name:** CKGR\_PLLAR

**Address:** 0x400E0428

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	MULA		
23	22	21	20	19	18	17	16
MULA							
15	14	13	12	11	10	9	8
–	–	PLLACOUNT					
7	6	5	4	3	2	1	0
PLLAEN							

Possible limitations on PLLA input frequencies and multiplier factors should be checked before using the PMC.

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **PLLAEN: PLLA Control**

0 = PLLA is disabled.

1 = PLLA is enabled

2 up to 255 = forbidden.

- **PLLACOUNT: PLLA Counter**

Specifies the number of Slow Clock cycles before the LOCKA bit is set in PMC\_SR after CKGR\_PLLAR is written.

- **MULA: PLLA Multiplier**

0 = The PLLA is deactivated (PLLA also disabled if DIVA = 0).

200 up to 254 = The PLLA Clock frequency is the PLLA input frequency multiplied by MULA + 1.

To change the PLLA frequency, read [Section 29.6.1 “Divider and Phase Lock Loop Programming”](#) on page 513.

### 30.17.10 PMC Clock Generator PLLB Register

**Name:** CKGR\_PLLBR

**Address:** 0x400E042C

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	SRCB	–	–	MULB		
23	22	21	20	19	18	17	16
MULB							
15	14	13	12	11	10	9	8
–	–	PLLBCOUNT					
7	6	5	4	3	2	1	0
DIVB							

Possible limitations on PLLB input frequencies and multiplier factors should be checked before using the PMC.

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)”.

- **DIVB: PLLB Front-End Divider**

0 = Divider output is stuck at 0 and PLLB is disabled.

1 = Divider is bypassed (divide by 1)

2 up to 255 = clock is divided by DIVB

- **PLLBCOUNT: PLLB Counter**

Specifies the number of Slow Clock cycles before the LOCKB bit is set in PMC\_SR after CKGR\_PLLBR is written.

- **MULB: PLLB Multiplier**

0 = The PLLB is deactivated (PLLB also disabled if DIVB = 0).

1 up to 62 = The PLLB Clock frequency is the PLLB input frequency multiplied by MULB + 1.

- **SRCB: Source for PLLB**

Value	Name	Description
0	MAINCK_IN_PLLB	The PLLB input clock is Main Clock
1	PLLA_IN_PLLB	The PLLB input clock is PLLA output. MAIN CK is PLLA output

### 30.17.11 PMC Master Clock Register

**Name:** PMC\_MCKR

**Address:** 0x400E0430

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
CPPRES				–	CPCSS		
15	14	13	12	11	10	9	8
–	–	PLLBDIV2	PLLADIV2	–	–	–	–
7	6	5	4	3	2	1	0
–	PRES			–	–	CSS	

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)”.

#### • CSS: Master Clock Source Selection

Value	Name	Description
0	SLOW_CLK	Slow Clock is selected
1	MAIN_CLK	Main Clock is selected
2	PLLA_CLK	PLLA Clock is selected
3	PLLB_CLK	PLLBClock is selected

#### • PRES: Processor Clock Prescaler

Value	Name	Description
0	CLK_1	Selected clock
1	CLK_2	Selected clock divided by 2
2	CLK_4	Selected clock divided by 4
3	CLK_8	Selected clock divided by 8
4	CLK_16	Selected clock divided by 16
5	CLK_32	Selected clock divided by 32
6	CLK_64	Selected clock divided by 64
7	CLK_3	Selected clock divided by 3

#### • PLLADIV2: PLLA Divisor by 2

PLLADIV2	PLLA Clock Division
0	PLLA clock frequency is divided by 1.
1	PLLA clock frequency is divided by 2.

- **PLLBDIV2 PLLB Divisor by 2**

PLLBDIV2	PLLB Clock Division
0	PLLB clock frequency is divided by 1.
1	PLLB clock frequency is divided by 2.

- **CPCSS: Coprocessor Master Clock Source Selection**

Value	Name	Description
0	SLOW_CLK	Slow Clock is selected
1	MAIN_CLK	Main Clock is selected
2	PLLA_CLK	PLLA Clock is selected
3	PLLB_CLK	PLLB Clock is selected
4	MCK	Master Clock is selected

- **CPPRES: Coprocessor Programmable Clock Prescaler**

0 up to 15 = The selected clock is divided by CPPRES+1.

### 30.17.12 PMC Programmable Clock Register

**Name:** PMC\_PCKx

**Address:** 0x400E0440

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	PRES			–	CSS		

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **CSS: Master Clock Source Selection**

Value	Name	Description
0	SLOW_CLK	Slow Clock is selected
1	MAIN_CLK	Main Clock is selected
2	PLLA_CLK	PLLA Clock is selected
3	PLLB_CLK	PLLB Clock is selected
4	MCK	Master Clock is selected

- **PRES: Programmable Clock Prescaler**

Value	Name	Description
0	CLK_1	Selected clock
1	CLK_2	Selected clock divided by 2
2	CLK_4	Selected clock divided by 4
3	CLK_8	Selected clock divided by 8
4	CLK_16	Selected clock divided by 16
5	CLK_32	Selected clock divided by 32
6	CLK_64	Selected clock divided by 64

### 30.17.13 PMC Interrupt Enable Register

**Name:** PMC\_IER

**Address:** 0x400E0460

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	XT32KERR	–	–	CFDEV	MOSCRCS	MOSCSELS
15	14	13	12	11	10	9	8
–	–	–	–	–	PCKRDY2	PCKRDY1	PCKRDY0
7	6	5	4	3	2	1	0
–	–	–	–	MCKRDY	LOCKB	LOCKA	MOSCXTS

- **MOSCXTS:** Main Crystal Oscillator Status Interrupt Enable
- **LOCKA:** PLLA Lock Interrupt Enable
- **LOCKB:** PLLB Lock Interrupt Enable
- **MCKRDY:** Master Clock Ready Interrupt Enable
- **PCKRDYx:** Programmable Clock Ready x Interrupt Enable
- **MOSCSELS:** Main Oscillator Selection Status Interrupt Enable
- **MOSCRCS:** Main On-Chip RC Status Interrupt Enable
- **CFDEV:** Clock Failure Detector Event Interrupt Enable
- **XT32KERR:** Slow Crystal Oscillator Error Interrupt Enable

### 30.17.14 PMC Interrupt Disable Register

**Name:** PMC\_IDR

**Address:** 0x400E0464

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	XT32KERR	–	–	CFDEV	MOSCRCS	MOSCSELS
15	14	13	12	11	10	9	8
–	–	–	–	–	PCKRDY2	PCKRDY1	PCKRDY0
7	6	5	4	3	2	1	0
–	–	–	–	MCKRDY	LOCKB	LOCKA	MOSCXTS

- **MOSCXTS:** Main Crystal Oscillator Status Interrupt Disable
- **LOCKA:** PLLA Lock Interrupt Disable
- **LOCKB:** PLLB Lock Interrupt Disable
- **MCKRDY:** Master Clock Ready Interrupt Disable
- **PCKRDYx:** Programmable Clock Ready x Interrupt Disable
- **MOSCSELS:** Main Oscillator Selection Status Interrupt Disable
- **MOSCRCS:** Main On-Chip RC Status Interrupt Disable
- **CFDEV:** Clock Failure Detector Event Interrupt Disable
- **XT32KERR:** Slow Crystal Oscillator Error Interrupt Disable

### 30.17.15 PMC Status Register

**Name:** PMC\_SR

**Address:** 0x400E0468

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	XT32KERR	FOS	CFDS	CFDEV	MOSCRCS	MOSCSELS
15	14	13	12	11	10	9	8
–	–	–	–	–	PCKRDY2	PCKRDY1	PCKRDY0
7	6	5	4	3	2	1	0
OSCSELS	–	–	–	MCKRDY	LOCKB	LOCKA	MOSCXTS

- **MOSCXTS: Main XTAL Oscillator Status**

0 = Main XTAL oscillator is not stabilized.

1 = Main XTAL oscillator is stabilized.

- **LOCKA: PLLA Lock Status**

0 = PLLA is not locked

1 = PLLA is locked.

- **LOCKB: PLLB Lock Status**

0 = PLLB is not locked

1 = PLLB is locked.

- **MCKRDY: Master Clock Status**

0 = Master Clock is not ready.

1 = Master Clock is ready.

- **OSCSELS: Slow Clock Oscillator Selection**

0 = Internal slow clock RC oscillator is selected.

1 = External slow clock 32 kHz oscillator is selected.

- **PCKRDYx: Programmable Clock Ready Status**

0 = Programmable Clock x is not ready.

1 = Programmable Clock x is ready.

- **MOSCSELS: Main Oscillator Selection Status**

0 = Selection is in progress.

1 = Selection is done.

- **MOSCRCS: Main On-Chip RC Oscillator Status**

0 = Main on-chip RC oscillator is not stabilized.

1 = Main on-chip RC oscillator is stabilized.



- **CFDEV: Clock Failure Detector Event**

0 = No clock failure detection of the fast crystal oscillator clock has occurred since the last read of PMC\_SR.

1 = At least one clock failure detection of the fast crystal oscillator clock has occurred since the last read of PMC\_SR.

- **CFDS: Clock Failure Detector Status**

0 = A clock failure of the fast crystal oscillator clock is not detected.

1 = A clock failure of the fast crystal oscillator clock is detected.

- **FOS: Clock Failure Detector Fault Output Status**

0 = The fault output of the clock failure detector is inactive.

1 = The fault output of the clock failure detector is active.

- **XT32KERR: Slow Crystal Oscillator Error**

0 = The frequency of the slow crystal oscillator is correct (32768 Hz +/- 1%) or the monitoring is disabled.

1 = The frequency of the slow crystal oscillator is incorrect or has been incorrect for an elapsed period of time since the monitoring has been enabled.

### 30.17.16 PMC Interrupt Mask Register

**Name:** PMC\_IMR

**Address:** 0x400E046C

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	XT32KERR	–	–	CFDEV	MOSCRCS	MOSCSELS
15	14	13	12	11	10	9	8
–	–	–	–	–	PCKRDY2	PCKRDY1	PCKRDY0
7	6	5	4	3	2	1	0
–	–	–	–	MCKRDY	LOCKB	LOCKA	MOSCXTS

- **MOSCXTS:** Main Crystal Oscillator Status Interrupt Mask
- **LOCKA:** PLLA Lock Interrupt Mask
- **LOCKB:** PLLB Lock Interrupt Mask
- **MCKRDY:** Master Clock Ready Interrupt Mask
- **PCKRDYx:** Programmable Clock Ready x Interrupt Mask
- **MOSCSELS:** Main Oscillator Selection Status Interrupt Mask
- **MOSCRCS:** Main On-Chip RC Status Interrupt Mask
- **CFDEV:** Clock Failure Detector Event Interrupt Mask
- **XT32KERR:** Slow Crystal Oscillator Error Interrupt Mask

### 30.17.17 PMC Fast Start-up Mode Register

**Name:** PMC\_FSMR

**Address:** 0x400E0470

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	FLPM		LPM	–	–	RTCAL	RTTAL
15	14	13	12	11	10	9	8
FSTT15	FSTT14	FSTT13	FSTT12	FSTT11	FSTT10	FSTT9	FSTT8
7	6	5	4	3	2	1	0
FSTT7	FSTT6	FSTT5	FSTT4	FSTT3	FSTT2	FSTT1	FSTT0

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)”.

- **FSTT0 - FSTT15: Fast Start-up Input Enable 0 to 15**

0 = The corresponding wake-up input has no effect on the Power Management Controller.

1 = The corresponding wake-up input enables a fast restart signal to the Power Management Controller.

- **RTTAL: RTT Alarm Enable**

0 = The RTT alarm has no effect on the Power Management Controller.

1 = The RTT alarm enables a fast restart signal to the Power Management Controller.

- **RTCAL: RTC Alarm Enable**

0 = The RTC alarm has no effect on the Power Management Controller.

1 = The RTC alarm enables a fast restart signal to the Power Management Controller.

- **LPM: Low Power Mode**

0 = The WaitForInterrupt (WFI) or the WaitForEvent (WFE) instruction of the processor makes the processor enter Sleep Mode.

1 = The WaitForEvent (WFE) instruction of the processor makes the system to enter in Wait Mode.

- **FLPM: Flash Low Power Mode**

Value	Name	Description
0	FLASH_STANDBY	Flash is in Standby Mode when system enters Wait Mode
1	FLASH_DEEP_POWERDOWN	Flash is in deep power down mode when system enters Wait Mode
2	FLASH_IDLE	Idle mode

### 30.17.18 PMC Fast Start-up Polarity Register

**Name:** PMC\_FSPR

**Address:** 0x400E0474

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
FSTP15	FSTP14	FSTP13	FSTP12	FSTP11	FSTP10	FSTP9	FSTP8
7	6	5	4	3	2	1	0
FSTP7	FSTP6	FSTP5	FSTP4	FSTP3	FSTP2	FSTP1	FSTP0

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **FSTPx: Fast Start-up Input Polarityx**

Defines the active polarity of the corresponding wake-up input. If the corresponding wake-up input is enabled and at the FSTP level, it enables a fast restart signal.

### 30.17.19 PMC Coprocessor Fast Start-up Mode Register

**Name:** PMC\_CPFSMR

**Address:** 0x400E047C

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	RTCAL	RTTAL
15	14	13	12	11	10	9	8
FSTT15	FSTT14	FSTT13	FSTT12	FSTT11	FSTT10	FSTT9	FSTT8
7	6	5	4	3	2	1	0
FSTT7	FSTT6	FSTT5	FSTT4	FSTT3	FSTT2	FSTT1	FSTT0

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **FSTT0 - FSTT15: Fast Start-up Input Enable 0 to 15**

0 = The corresponding wake-up input has no effect on the Power Management Controller.

1 = The corresponding wake-up input enables a fast restart signal to the Power Management Controller.

- **RTTAL: RTT Alarm Enable**

0 = The RTT alarm has no effect on the Power Management Controller.

1 = The RTT alarm enables a fast restart signal to the Power Management Controller.

- **RTCAL: RTC Alarm Enable**

0 = The RTC alarm has no effect on the Power Management Controller.

1 = The RTC alarm enables a fast restart signal to the Power Management Controller.

### 30.17.20 PMC Fault Output Clear Register

**Name:** PMC\_FOCR

**Address:** 0x400E0478

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	FOCLR

- **FOCLR: Fault Output Clear**

Clears the clock failure detector fault output.

### 30.17.21 PMC Write Protect Mode Register

**Name:** PMC\_WPMR

**Address:** 0x400E04E4

**Access:** Read-write

**Reset:** See [Table 30-3](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

- **WPEN: Write Protect Enable**

0 = Disables the Write Protect if WPKEY corresponds to 0x504D43 (“PMC” in ASCII).

1 = Enables the Write Protect if WPKEY corresponds to 0x504D43 (“PMC” in ASCII).

Protects the registers:

- “PMC System Clock Enable Register”
- “PMC System Clock Disable Register”
- “PMC Peripheral Clock Enable Register 0”
- “PMC Peripheral Clock Disable Register 0”
- “PMC Clock Generator Main Oscillator Register”
- “PMC Clock Generator PLLA Register”
- “PMC Master Clock Register”
- “PMC Programmable Clock Register”
- “PMC Fast Start-up Mode Register”
- “PMC Fast Start-up Polarity Register”
- “PMC Coprocessor Fast Start-up Mode Register”
- “PMC Peripheral Clock Enable Register 1”
- “PMC Peripheral Clock Disable Register 1”
- “PMC Oscillator Calibration Register”

- **WPKEY: Write Protect KEY**

Value	Name	Description
0x504D43	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

### 30.17.22 PMC Write Protect Status Register

**Name:** PMC\_WPSR

**Address:** 0x400E04E8

**Access:** Read-only

**Reset:** See [Table 30-3](#)

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

- **WPVS: Write Protect Violation Status**

0 = No Write Protect Violation has occurred since the last read of the PMC\_WPSR register.

1 = A Write Protect Violation has occurred since the last read of the PMC\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protect Violation Source**

When WPVS is active, this field indicates the write-protected register (through address offset or code) in which a write access has been attempted.

Reading PMC\_WPSR automatically clears all fields.



### 30.17.23 PMC Peripheral Clock Enable Register 1

**Name:** PMC\_PCER1

**Address:** 0x400E0500

**Access:** Write-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	PID43	PID42	PID41	PID40
7	6	5	4	3	2	1	0
PID39	PID38	PID37	PID36	PID35	PID34	PID33	PID32

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **PIDx: Peripheral Clock x Enable**

0 = No effect.

1 = Enables the corresponding peripheral clock.

Notes: 1. To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet.

2. Programming the control bits of the Peripheral ID that are not implemented has no effect on the behavior of the PMC.

### 30.17.24 PMC Peripheral Clock Disable Register 1

**Name:** PMC\_PCDR1

**Address:** 0x400E0504

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	PID43	PID42	PID41	PID40
7	6	5	4	3	2	1	0
PID39	PID38	PID37	PID36	PID35	PID34	PID33	PID32

This register can only be written if the WPEN bit is cleared in [“PMC Write Protect Mode Register”](#).

- **PIDx: Peripheral Clock x Disable**

0 = No effect.

1 = Disables the corresponding peripheral clock.

**Note:** To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet.

### 30.17.25 PMC Peripheral Clock Status Register 1

**Name:** PMC\_PCSR1

**Address:** 0x400E0508

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	PID43	PID42	PID41	PID40
7	6	5	4	3	2	1	0
PID39	PID38	PID37	PID36	PID35	PID34	PID33	PID32

- **PIDx: Peripheral Clock x Status**

0 = The corresponding peripheral clock is disabled.

1 = The corresponding peripheral clock is enabled.

**Note:** To get PIDx, refer to identifiers as defined in the section “Peripheral Identifiers” in the product datasheet.

### 30.17.26 PMC Oscillator Calibration Register

**Name:** PMC\_OCR

**Address:** 0x400E0510

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
SEL12	CAL12						
15	14	13	12	11	10	9	8
SEL8	CAL8						
7	6	5	4	3	2	1	0
SEL4	CAL4						

This register can only be written if the WPEN bit is cleared in “[PMC Write Protect Mode Register](#)”.

- **CAL4: RC Oscillator Calibration bits for 4 MHz**

Calibration bits applied to the RC Oscillator when SEL4 is set.

- **SEL4: Selection of RC Oscillator Calibration bits for 4 MHz**

0 = Default value stored in Flash memory.

1 = Value written by user in CAL4 field of this register.

- **CAL8: RC Oscillator Calibration bits for 8 MHz**

Calibration bits applied to the RC Oscillator when SEL8 is set.

- **SEL8: Selection of RC Oscillator Calibration bits for 8 MHz**

0 = Factory determined value stored in Flash memory.

1 = Value written by user in CAL8 field of this register.

- **CAL12: RC Oscillator Calibration bits for 12 MHz**

Calibration bits applied to the RC Oscillator when SEL12 is set.

- **SEL12: Selection of RC Oscillator Calibration bits for 12 MHz**

0 = Factory determined value stored in Flash memory.

1 = Value written by user in CAL12 field of this register.

## 31. Chip Identifier (CHIPID)

### 31.1 Description

Chip Identifier (CHIPID) registers permit recognition of the device and its revision. These registers provide the sizes and types of the on-chip memories, as well as the set of embedded peripherals.

Two chip identifier registers are embedded: CHIPID\_CIDR (Chip ID Register) and CHIPID\_EXID (Extension ID). Both registers contain a hard-wired value that is read-only. The first register contains the following fields:

- EXT - shows the use of the extension identifier register
- NVPTYP and NVPSIZ - identifies the type of embedded non-volatile memory and its size
- ARCH - identifies the set of embedded peripherals
- SRAMSIZ - indicates the size of the embedded SRAM
- EPROC - indicates the embedded ARM processor
- VERSION - gives the revision of the silicon

The second register is device-dependent and reads 0 if the bit EXT is 0.

### 31.2 Embedded Characteristics

- Chip ID Registers
  - Identification of the Device Revision, Sizes of the Embedded Memories, Set of Peripherals, Embedded Processor

Table 31-1. SAM4C Chip IDs Registers

Chip Name	CHIPID_CIDR	CHIPID_EXID
SAM4C16C (Rev A)	0xA64C_0CE0	0x0
SAM4C8C (Rev A)	0xA64C_0AE0	0x0

## 31.3 Chip Identifier (CHIPID) User Interface

Table 31-2. Register Mapping

Offset	Register	Name	Access	Reset
0x0	Chip ID Register	CHIPID_CIDR	Read-only	–
0x4	Chip ID Extension Register	CHIPID_EXID	Read-only	–

### 31.3.1 Chip ID Register

**Name:** CHIPID\_CIDR

**Address:** 0x400E0740

**Access:** Read-only

31	30	29	28	27	26	25	24
EXT	NVPTYP				ARCH		
23	22	21	20	19	18	17	16
ARCH				SRAMSIZ			
15	14	13	12	11	10	9	8
NVPSIZ2				NVPSIZ			
7	6	5	4	3	2	1	0
EPROC			VERSION				

- **VERSION: Version of the Device**

Current version of the device.

- **EPROC: Embedded Processor**

Value	Name	Description
1	ARM946ES	ARM946ES
2	ARM7TDMI	ARM7TDMI
3	CM3	Cortex-M3
4	ARM920T	ARM920T
5	ARM926EJS	ARM926EJS
6	CA5	Cortex-A5
7	CM4	Cortex-M4

- **NVPSIZ: Nonvolatile Program Memory Size**

Value	Name	Description
0	NONE	None
1	8K	8 Kbytes
2	16K	16 Kbytes
3	32K	32 Kbytes
4	–	Reserved
5	64K	64 Kbytes
6	–	Reserved
7	128K	128 Kbytes
8	–	Reserved
9	256K	256 Kbytes
10	512K	512 Kbytes
11	–	Reserved
12	1024K	1024 Kbytes

Value	Name	Description
13	–	Reserved
14	2048K	2048 Kbytes
15	–	Reserved

- **NVPSIZ2: Second Nonvolatile Program Memory Size**

Value	Name	Description
0	NONE	None
1	8K	8 Kbytes
2	16K	16 Kbytes
3	32K	32 Kbytes
4	–	Reserved
5	64K	64 Kbytes
6	–	Reserved
7	128K	128 Kbytes
8	–	Reserved
9	256K	256 Kbytes
10	512K	512 Kbytes
11	–	Reserved
12	1024K	1024 Kbytes
13	–	Reserved
14	2048K	2048 Kbytes
15	–	Reserved

- **SRAMSIZ: Internal SRAM Size**

Value	Name	Description
0	48K	48 Kbytes
1	192K	192 Kbytes
2	2K	2 Kbytes
3	6K	6 Kbytes
4	24K	24 Kbytes
5	4K	4 Kbytes
6	80K	80 Kbytes
7	160K	160 Kbytes
8	8K	8 Kbytes
9	16K	16 Kbytes
10	32K	32 Kbytes
11	64K	64 Kbytes
12	128K	128 Kbytes



Value	Name	Description
13	256K	256 Kbytes
14	96K	96 Kbytes
15	512K	512 Kbytes

- **ARCH: Architecture Identifier**

Value	Name	Description
0x64	SAM4CxxC	SAM4CxC (100-pin version)

- **NVPTYP: Nonvolatile Program Memory Type**

Value	Name	Description
0	ROM	ROM
1	ROMLESS	ROMless or on-chip Flash
4	SRAM	SRAM emulating ROM
2	FLASH	Embedded Flash Memory
3	ROM_FLASH	ROM and Embedded Flash Memory <ul style="list-style-type: none"> <li>• NVPSIZ is ROM size</li> <li>• NVPSIZ2 is Flash size</li> </ul>

- **EXT: Extension Flag**

0 = Chip ID has a single register definition without extension.

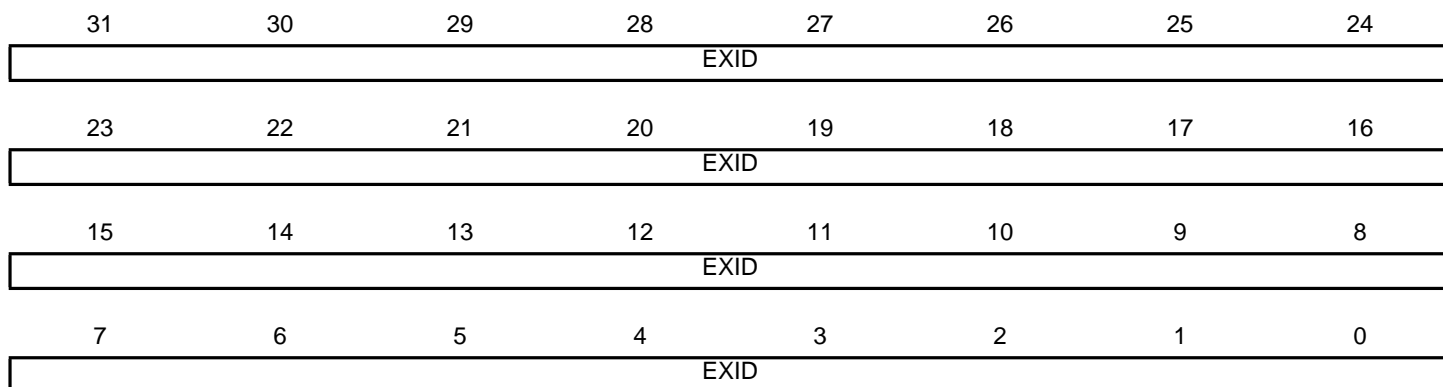
1 = An extended Chip ID exists.

### 31.3.2 Chip ID Extension Register

**Name:** CHIPID\_EXID

**Address:** 0x400E0744

**Access:** Read-only



- **EXID: Chip ID Extension**

Reads 0 if the EXT bit in CHIPID\_CIDR is 0.

Value	Name	Description
0x0	SAM4C	Dual Core

## 32. Parallel Input/Output Controller (PIO)

### 32.1 Description

The Parallel Input/Output Controller (PIO) manages up to 32 fully programmable input/output lines. Each I/O line may be dedicated as a general-purpose I/O or be assigned to a function of an embedded peripheral. This assures effective optimization of the pins of the product.

Each I/O line is associated with a bit number in all of the 32-bit registers of the 32-bit wide user interface.

Each I/O line of the PIO Controller features:

- An input change interrupt enabling level change detection on any I/O line.
- Additional Interrupt modes enabling rising edge, falling edge, low-level or high-level detection on any I/O line.
- A glitch filter providing rejection of glitches lower than one-half of PIO clock cycle.
- A debouncing filter providing rejection of unwanted pulses from key or push button operations.
- Multi-drive capability similar to an open drain I/O line.
- Control of the pull-up and pull-down of the I/O line.
- Input visibility and output control.

The PIO Controller also features a synchronous output providing up to 32 bits of data output in a single write operation.

### 32.2 Embedded Characteristics

- Up to 32 Programmable I/O Lines
- Fully Programmable through Set/Clear Registers
- Multiplexing of Four Peripheral Functions per I/O Line
- For each I/O Line (Whether Assigned to a Peripheral or Used as General Purpose I/O)
  - Input Change Interrupt
  - Programmable Glitch Filter
  - Programmable Debouncing Filter
  - Multi-drive Option Enables Driving in Open Drain
  - Programmable Pull-Up on Each I/O Line
  - Pin Data Status Register, Supplies Visibility of the Level on the Pin at Any Time
  - Additional Interrupt Modes on a Programmable Event: Rising Edge, Falling Edge, Low-Level or High-Level
- Synchronous Output, Provides Set and Clear of Several I/O Lines in a Single Write
- Write Protect Registers
- Programmable Schmitt Trigger Inputs
- Programmable I/O Drive

32.3 Block Diagram

Figure 32-1. Block Diagram

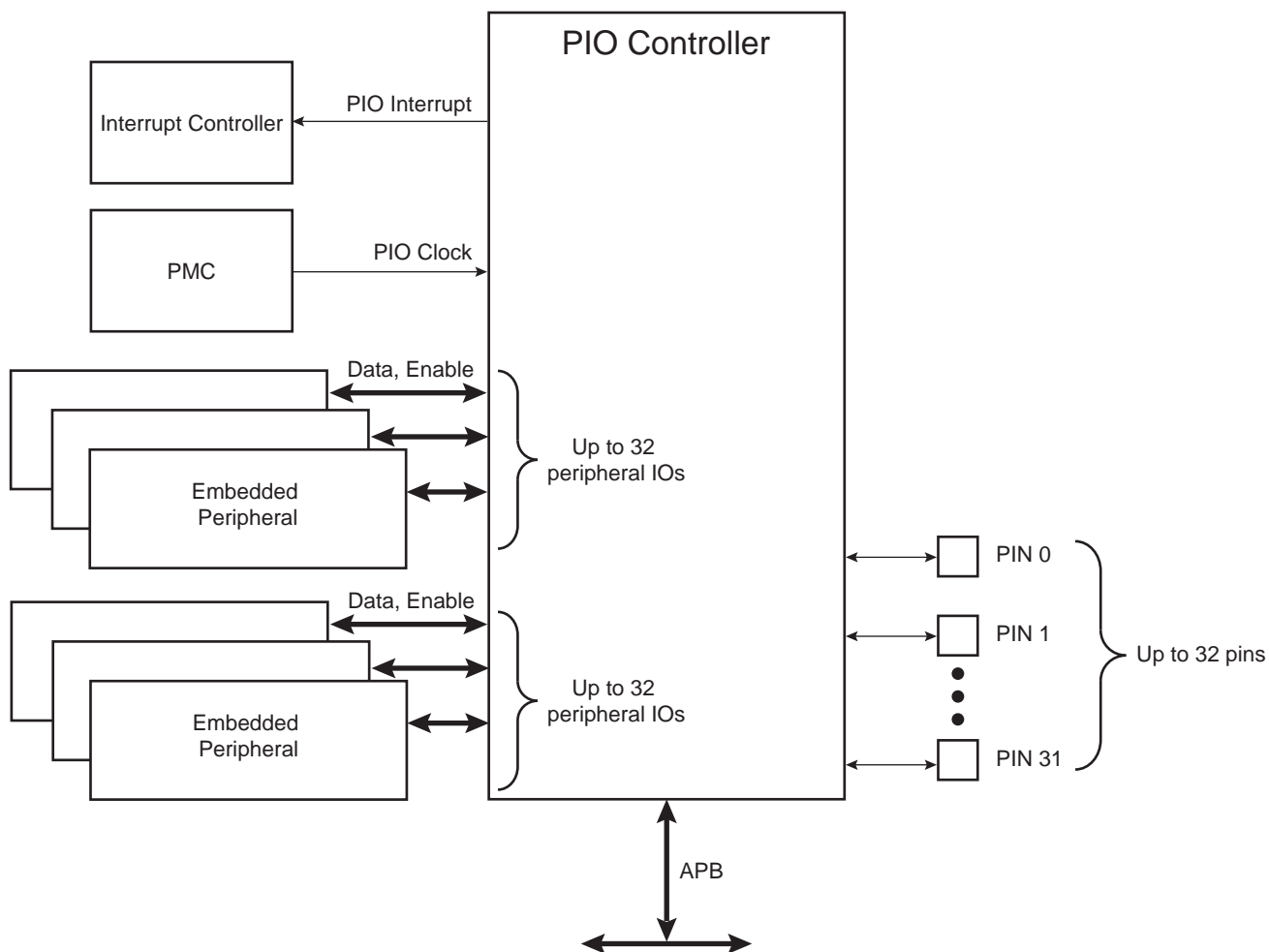
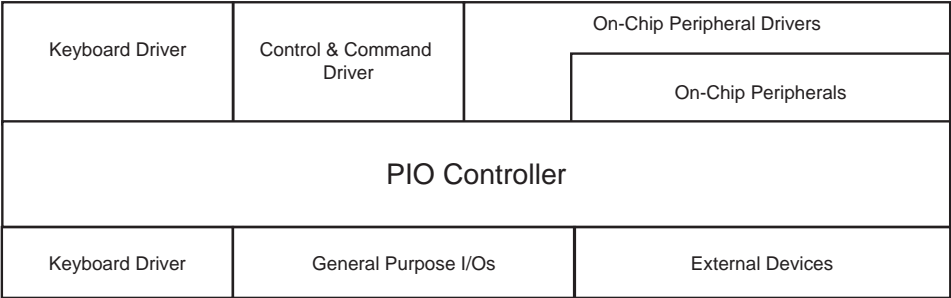


Figure 32-2. Application Block Diagram



## 32.4 Product Dependencies

### 32.4.1 Pin Multiplexing

Each pin is configurable, depending on the product, as either a general-purpose I/O line only, or as an I/O line multiplexed with one or two peripheral I/Os. As the multiplexing is hardware defined and thus product-dependent, the hardware designer and programmer must carefully determine the configuration of the PIO Controllers required by their application. When an I/O line is general-purpose only, i.e. not multiplexed with any peripheral I/O, programming of the PIO Controller regarding the assignment to a peripheral has no effect and only the PIO Controller can control how the pin is driven by the product.

### 32.4.2 External Interrupt Lines

The interrupt signals FIQ and IRQ0 to IRQn are generally multiplexed through the PIO Controllers. However, it is not necessary to assign the I/O line to the interrupt function as the PIO Controller has no effect on inputs and the interrupt lines (FIQ or IRQs) are used only as inputs.

### 32.4.3 Power Management

The Power Management Controller controls the PIO Controller clock in order to save power. Writing any of the registers of the user interface does not require the PIO Controller clock to be enabled. This means that the configuration of the I/O lines does not require the PIO Controller clock to be enabled.

However, when the clock is disabled, not all of the features of the PIO Controller are available, including glitch filtering. Note that the input change interrupt, the interrupt modes on a programmable event and the read of the pin level require the clock to be validated.

After a hardware reset, the PIO clock is disabled by default.

The user must configure the Power Management Controller before any access to the input line information.

### 32.4.4 Interrupt Generation

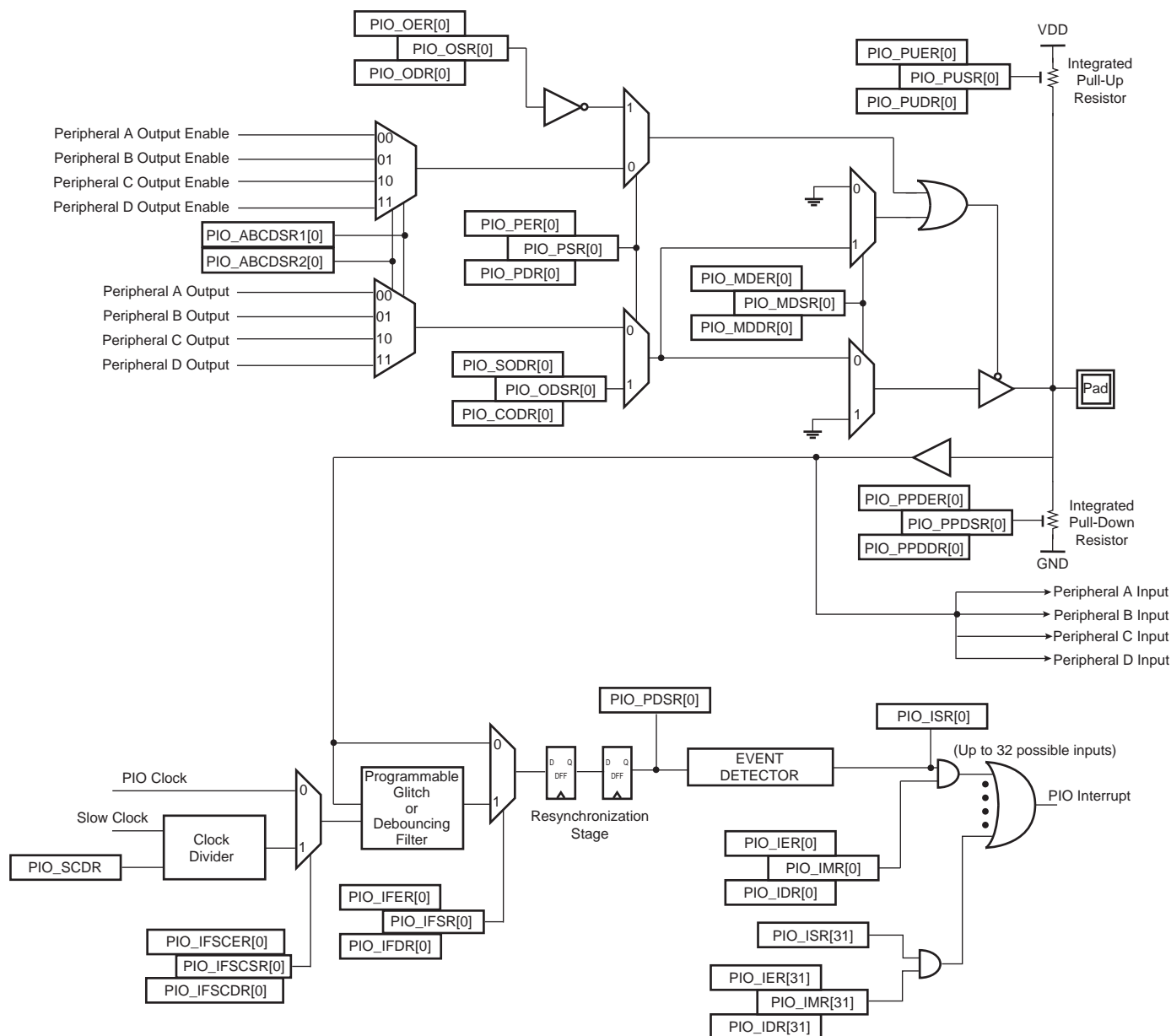
For interrupt handling, the PIO Controllers are considered as user peripherals. This means that the PIO Controller interrupt lines are connected among the interrupt sources. Refer to the PIO Controller peripheral identifier in the product description to identify the interrupt sources dedicated to the PIO Controllers. Using the PIO Controller requires the Interrupt Controller to be programmed first.

The PIO Controller interrupt can be generated only if the PIO Controller clock is enabled.

## 32.5 Functional Description

The PIO Controller features up to 32 fully-programmable I/O lines. Most of the control logic associated to each I/O is represented in Figure 32-3. In this description each signal shown represents one of up to 32 possible indexes.

Figure 32-3. I/O Line Control Logic



### 32.5.1 Pull-up and Pull-down Resistor Control

Each I/O line is designed with an embedded pull-up resistor and an embedded pull-down resistor. The pull-up resistor can be enabled or disabled by writing to the Pull-up Enable register (PIO\_PUER) or Pull-up Disable register (PIO\_PUDR), respectively. Writing to these registers results in setting or clearing the corresponding bit in the Pull-up Status register (PIO\_PUSR). Reading a one in PIO\_PUSR means the pull-up is disabled and reading a zero means the pull-up is enabled. The pull-down resistor can be enabled or disabled by writing the Pull-down Enable register (PIO\_PPDER) or the Pull-down Disable register (PIO\_PPDDR), respectively. Writing in these registers results in setting or clearing the corresponding bit in the Pull-down Status register (PIO\_PPDSR). Reading a one in PIO\_PPDSR means the pull-up is disabled and reading a zero means the pull-down is enabled.

Enabling the pull-down resistor while the pull-up resistor is still enabled is not possible. In this case, the write of PIO\_PPDER for the relevant I/O line is discarded. Likewise, enabling the pull-up resistor while the pull-down resistor is still enabled is not possible. In this case, the write of PIO\_PUER for the relevant I/O line is discarded.

Control of the pull-up resistor is possible regardless of the configuration of the I/O line.

After reset, all of the pull-ups are enabled, i.e. PIO\_PUSR resets at the value 0x0, and all the pull-downs are disabled, i.e. PIO\_PPDSR resets at the value 0xFFFFFFFF.

### 32.5.2 I/O Line or Peripheral Function Selection

When a pin is multiplexed with one or two peripheral functions, the selection is controlled with the Enable register (PIO\_PER) and the Disable register (PIO\_PDR). The Status register (PIO\_PSR) is the result of the set and clear registers and indicates whether the pin is controlled by the corresponding peripheral or by the PIO Controller. A value of zero indicates that the pin is controlled by the corresponding on-chip peripheral selected in the ABCD Select registers (PIO\_ABCDSR1 and PIO\_ABCDSR2). A value of one indicates the pin is controlled by the PIO Controller.

If a pin is used as a general-purpose I/O line (not multiplexed with an on-chip peripheral), PIO\_PER and PIO\_PDR have no effect and PIO\_PSR returns a one for the corresponding bit.

After reset, the I/O lines are controlled by the PIO Controller, i.e. PIO\_PSR resets at one. However, in some events, it is important that PIO lines are controlled by the peripheral (as in the case of memory chip select lines that must be driven inactive after reset, or for address lines that must be driven low for booting out of an external memory). Thus, the reset value of PIO\_PSR is defined at the product level and depends on the multiplexing of the device.

### 32.5.3 Peripheral A or B or C or D Selection

The PIO Controller provides multiplexing of up to four peripheral functions on a single pin. The selection is performed by writing PIO\_ABCDSR1 and PIO\_ABCDSR2.

For each pin:

- The corresponding bit at level zero in PIO\_ABCDSR1 and the corresponding bit at level zero in PIO\_ABCDSR2 means peripheral A is selected.
- The corresponding bit at level one in PIO\_ABCDSR1 and the corresponding bit at level zero in PIO\_ABCDSR2 means peripheral B is selected.
- The corresponding bit at level zero in PIO\_ABCDSR1 and the corresponding bit at level one in PIO\_ABCDSR2 means peripheral C is selected.
- The corresponding bit at level one in PIO\_ABCDSR1 and the corresponding bit at level zero in PIO\_ABCDSR2 means peripheral D is selected.

Note that multiplexing of peripheral lines A, B, C and D only affects the output line. The peripheral input lines are always connected to the pin input.

Writing in PIO\_ABCDSR1 and PIO\_ABCDSR2 manages the multiplexing regardless of the configuration of the pin. However, assignment of a pin to a peripheral function requires a write in PIO\_ABCDSR1 and PIO\_ABCDSR2 in addition to a write in PIO\_PDR.

After reset, PIO\_ABCDSR1 and PIO\_ABCDSR2 are zero, thus indicating that all the PIO lines are configured on peripheral A. However, peripheral A generally does not drive the pin as the PIO Controller resets in I/O line mode.

### 32.5.4 Output Control

When the I/O line is assigned to a peripheral function, i.e., the corresponding bit in PIO\_PSR is at zero, the drive of the I/O line is controlled by the peripheral. Peripheral A or B or C or D depending on the value in PIO\_ABCDSR1 and PIO\_ABCDSR2 determines whether the pin is driven or not.

When the I/O line is controlled by the PIO Controller, the pin can be configured to be driven. This is done by writing the Output Enable register (PIO\_OER) and Output Disable register (PIO\_ODR). The results of these write operations are detected in the Output Status register (PIO\_OSR). When a bit in this register is at zero, the corresponding I/O line is used as an input only. When the bit is at one, the corresponding I/O line is driven by the PIO Controller.

The level driven on an I/O line can be determined by writing in the Set Output Data register (PIO\_SODR) and the Clear Output Data register (PIO\_CODR). These write operations, respectively, set and clear the Output Data Status register (PIO\_ODSR), which represents the data driven on the I/O lines. Writing in PIO\_OER and PIO\_ODR manages PIO\_OSR whether the pin is configured to be controlled by the PIO Controller or assigned to a peripheral function. This enables configuration of the I/O line prior to setting it to be managed by the PIO Controller.

Similarly, writing in PIO\_SODR and PIO\_CODR affects PIO\_ODSR. This is important as it defines the first level driven on the I/O line.

### 32.5.5 Synchronous Data Output

Clearing one or more PIO line(s) and setting another one or more PIO line(s) synchronously cannot be done by using PIO\_SODR and PIO\_CODR registers. It requires two successive write operations into two different registers. To overcome this, the PIO Controller offers a direct control of PIO outputs by single write access to PIO\_ODSR. Only bits unmasked by the Output Write Status register (PIO\_OWSR) are written. The mask bits in PIO\_OWSR are set by writing to the Output Write Enable register (PIO\_OWER) and cleared by writing to the Output Write Disable register (PIO\_OWDR).

After reset, the synchronous data output is disabled on all the I/O lines as PIO\_OWSR resets at 0x0.

### 32.5.6 Multi-Drive Control (Open Drain)

Each I/O can be independently programmed in open drain by using the multi-drive feature. This feature permits several drivers to be connected on the I/O line which is driven low only by each device. An external pull-up resistor (or enabling of the internal one) is generally required to guarantee a high level on the line.

The multi-drive feature is controlled by the Multi-driver Enable register (PIO\_MDER) and the Multi-driver Disable register (PIO\_MDDR). The multi-drive can be selected whether the I/O line is controlled by the PIO Controller or assigned to a peripheral function. The Multi-driver Status register (PIO\_MDSR) indicates the pins that are configured to support external drivers.

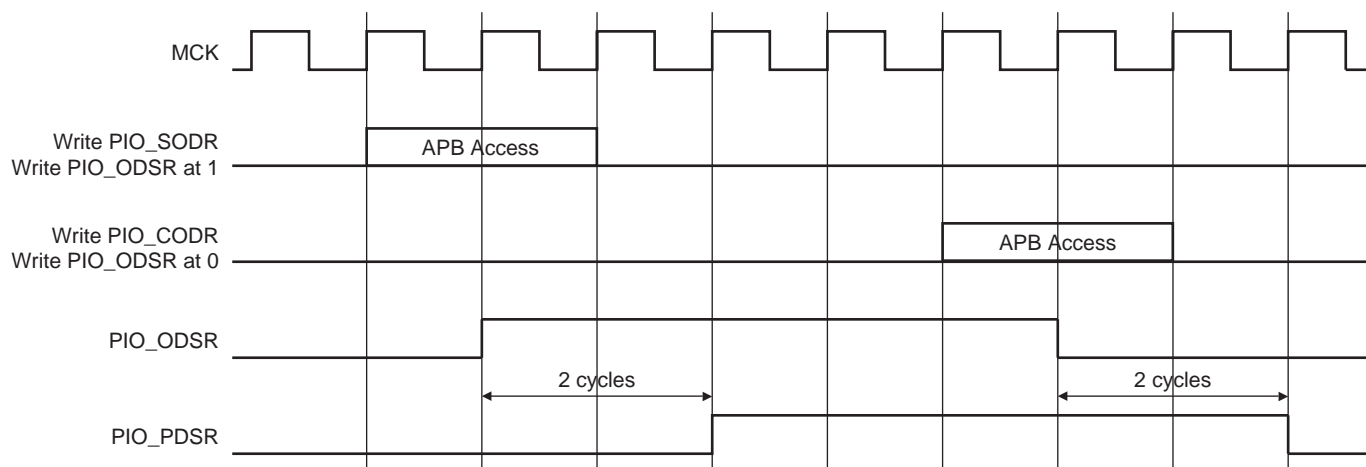
After reset, the multi-drive feature is disabled on all pins, i.e. PIO\_MDSR resets at value 0x0.

### 32.5.7 Output Line Timings

Figure 32-4 shows how the outputs are driven either by writing PIO\_SODR or PIO\_CODR, or by directly writing PIO\_ODSR. This last case is valid only if the corresponding bit in PIO\_OWSR is set. Figure 32-4 also shows when the feedback in the Pin Data Status register (PIO\_PDSR) is available.



**Figure 32-4. Output Line Timings**



### 32.5.8 Inputs

The level on each I/O line can be read through PIO\_PDSR. This register indicates the level of the I/O lines regardless of their configuration, whether uniquely as an input, or driven by the PIO Controller, or driven by a peripheral.

Reading the I/O line levels requires the clock of the PIO Controller to be enabled, otherwise PIO\_PDSR reads the levels present on the I/O line at the time the clock was disabled.

### 32.5.9 Input Glitch and Debouncing Filters

Optional input glitch and debouncing filters are independently programmable on each I/O line.

The glitch filter can filter a glitch with a duration of less than 1/2 master clock (MCK) and the debouncing filter can filter a pulse of less than 1/2 period of a programmable divided slow clock.

The selection between glitch filtering or debounce filtering is done by writing in the PIO Input Filter Slow Clock Disable register (PIO\_IFSCDR) and the PIO Input Filter Slow Clock Enable register (PIO\_IFSCER). Writing PIO\_IFSCDR and PIO\_IFSCER, respectively, sets and clears bits in the Input Filter Slow Clock Status register (PIO\_IFSCSR).

The current selection status can be checked by reading the register PIO\_IFSCSR.

- If PIO\_IFSCSR[i] = 0: The glitch filter can filter a glitch with a duration of less than 1/2 master clock period.
- If PIO\_IFSCSR[i] = 1: The debouncing filter can filter a pulse with a duration of less than 1/2 programmable divided slow clock period.

For the debouncing filter, the period of the divided slow clock is performed by writing in the DIV field of the Slow Clock Divider register (PIO\_SCDR).

$$T_{div\_slck} = ((DIV+1)*2).T_{slow\_clock}$$

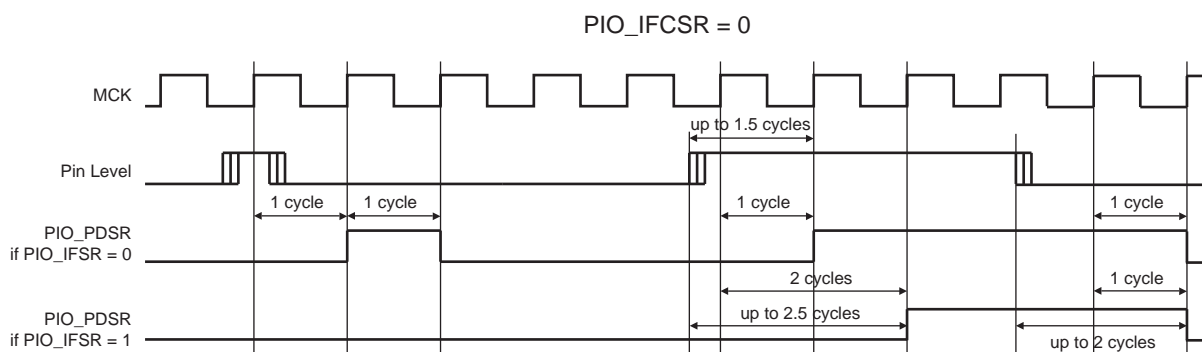
When the glitch or debouncing filter is enabled, a glitch or pulse with a duration of less than 1/2 selected clock cycle (selected clock represents MCK or divided slow clock depending on PIO\_IFSCDR and PIO\_IFSCER programming) is automatically rejected, while a pulse with a duration of one selected clock (MCK or divided slow clock) cycle or more is accepted. For pulse durations between 1/2 selected clock cycle and one selected clock cycle, the pulse may or may not be taken into account, depending on the precise timing of its occurrence. Thus for a pulse to be visible, it must exceed one selected clock cycle, whereas for a glitch to be reliably filtered out, its duration must not exceed 1/2 selected clock cycle.

The filters also introduce some latencies, illustrated in [Figure 32-5](#) and [Figure 32-6](#).

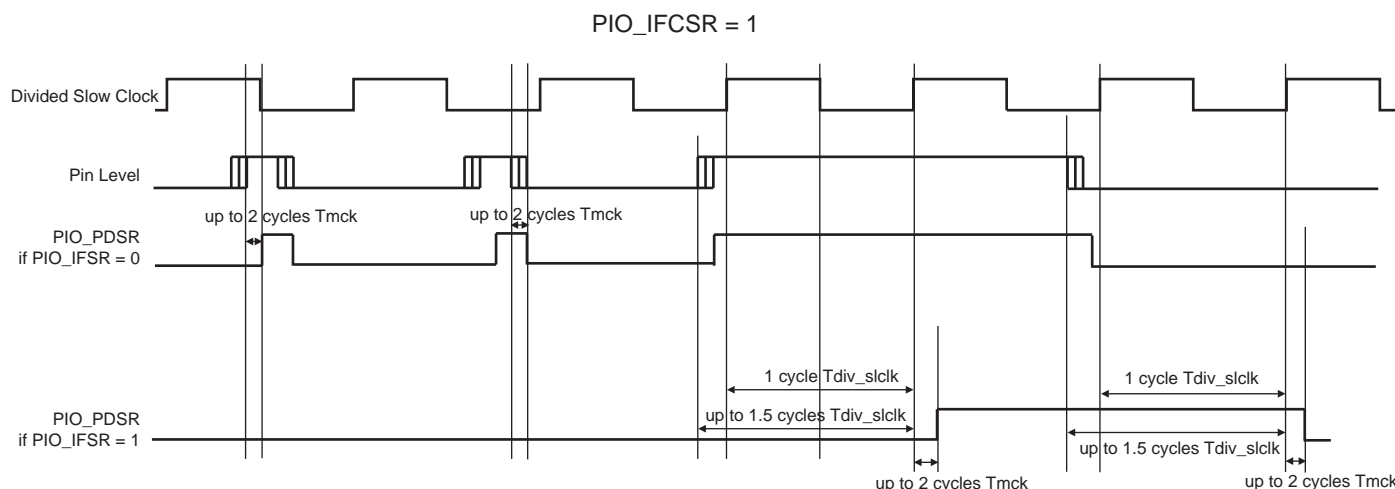
The glitch filters are controlled by the Input Filter Enable register (PIO\_IFER), the Input Filter Disable register (PIO\_IFDR) and the Input Filter Status register (PIO\_IFSR). Writing PIO\_IFER and PIO\_IFDR respectively sets and clears bits in PIO\_IFSR. This last register enables the glitch filter on the I/O lines.

When the glitch and/or debouncing filter is enabled, it does not modify the behavior of the inputs on the peripherals. It acts only on the value read in PIO\_PDSR and on the input change interrupt detection. The glitch and debouncing filters require that the PIO Controller clock is enabled.

**Figure 32-5. Input Glitch Filter Timing**



**Figure 32-6. Input Debouncing Filter Timing**



### 32.5.10 Input Edge/Level Interrupt

The PIO Controller can be programmed to generate an interrupt when it detects an edge or a level on an I/O line. The Input Edge/Level interrupt is controlled by writing the Interrupt Enable register (PIO\_IER) and the Interrupt Disable register (PIO\_IDR), which enable and disable the input change interrupt respectively by setting and clearing the corresponding bit in the Interrupt Mask register (PIO\_IMR). As input change detection is possible only by comparing two successive samplings of the input of the I/O line, the PIO Controller clock must be enabled. The Input Change interrupt is available regardless of the configuration of the I/O line, i.e. configured as an input only, controlled by the PIO Controller or assigned to a peripheral function.

By default, the interrupt can be generated at any time an edge is detected on the input.

Some additional interrupt modes can be enabled/disabled by writing in the Additional Interrupt Modes Enable register (PIO\_AIMER) and Additional Interrupt Modes Disable register (PIO\_AIMDR). The current state of this selection can be read through the Additional Interrupt Modes Mask register (PIO\_AIMMR).

These additional modes are:

- Rising edge detection
- Falling edge detection

- Low-level detection
- High-level detection

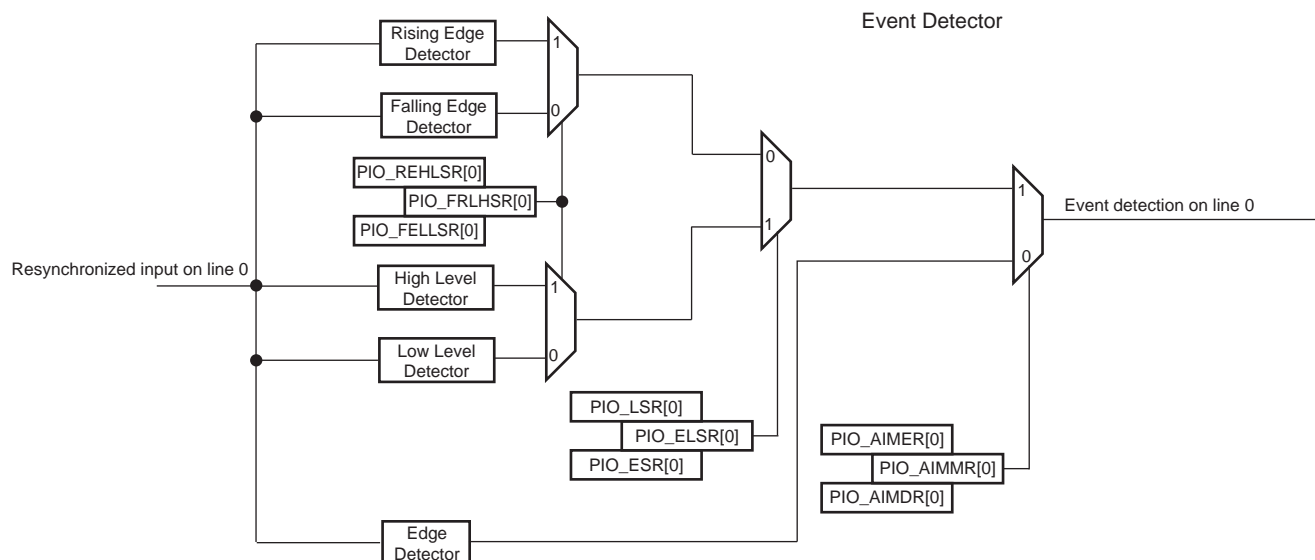
In order to select an additional interrupt mode:

- The type of event detection (edge or level) must be selected by writing in the Edge Select register (PIO\_ESR) and Level Select register (PIO\_LSR) which , respectively, the edge and level detection. The current status of this selection is accessible through the Edge/Level Status register (PIO\_ELSR).
- The polarity of the event detection (rising/falling edge or high/low-level) must be selected by writing in the Falling Edge /Low-Level Select register (PIO\_FELLSR) and Rising Edge/High-Level Select register (PIO\_REHLSR) which allow to select falling or rising edge (if edge is selected in PIO\_ELSR) edge or high- or low-level detection (if level is selected in PIO\_ELSR). The current status of this selection is accessible through the Fall/Rise - Low/High Status register (PIO\_FRLHSR).

When an input edge or level is detected on an I/O line, the corresponding bit in the Interrupt Status register (PIO\_ISR) is set. If the corresponding bit in PIO\_IMR is set, the PIO Controller interrupt line is asserted. The interrupt signals of the 32 channels are ORed-wired together to generate a single interrupt signal to the interrupt controller.

When the software reads PIO\_ISR, all the interrupts are automatically cleared. This signifies that all the interrupts that are pending when PIO\_ISR is read must be handled. When an Interrupt is enabled on a “level”, the interrupt is generated as long as the interrupt source is not cleared, even if some read accesses in PIO\_ISR are performed.

**Figure 32-7. Event Detector on Input Lines (Figure Represents Line 0)**



### 32.5.10.1 Example

If generating an interrupt is required on the lines below, the configuration required is described in [Section 32.5.10.2 "Interrupt Mode Configuration"](#), [Section 32.5.10.3 "Edge or Level Detection Configuration"](#) and [Section 32.5.10.4 "Falling/Rising Edge or Low/High-Level Detection Configuration"](#):

- Rising edge on PIO line 0
- Falling edge on PIO line 1
- Rising edge on PIO line 2
- Low-level on PIO line 3
- High-level on PIO line 4
- High-level on PIO line 5
- Falling edge on PIO line 6
- Rising edge on PIO line 7
- Any edge on the other lines

the configuration required is described below.

### 32.5.10.2 Interrupt Mode Configuration

All the interrupt sources are enabled by writing 32'hFFFF\_FFFF in PIO\_IER.

Then the additional interrupt mode is enabled for lines 0 to 7 by writing 32'h0000\_00FF in PIO\_AIMER.

### 32.5.10.3 Edge or Level Detection Configuration

Lines 3, 4 and 5 are configured in level detection by writing 32'h0000\_0038 in PIO\_LSR.

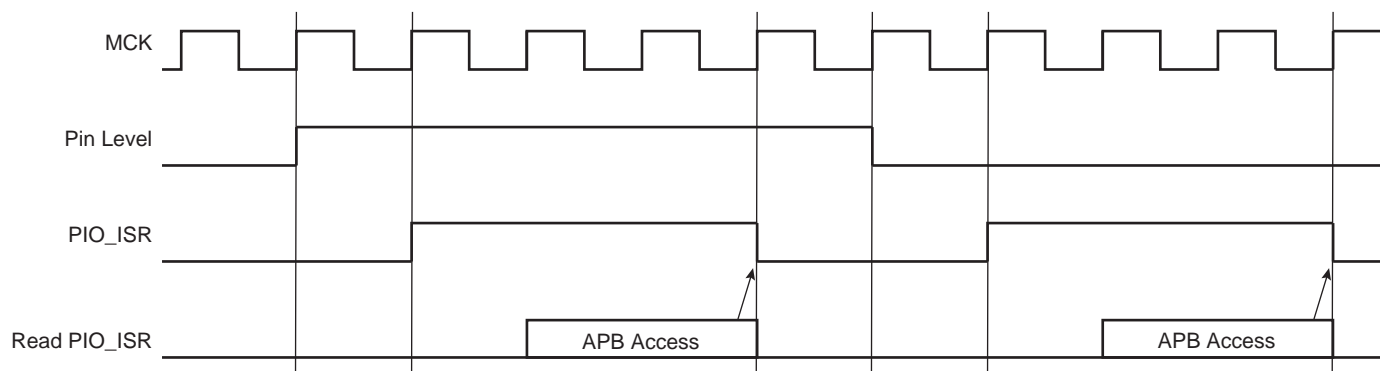
The other lines are configured in edge detection by default, if they have not been previously configured. Otherwise, lines 0, 1, 2, 6 and 7 must be configured in edge detection by writing 32'h0000\_00C7 in PIO\_ESR.

### 32.5.10.4 Falling/Rising Edge or Low/High-Level Detection Configuration

Lines 0, 2, 4, 5 and 7 are configured in rising edge or high-level detection by writing 32'h0000\_00B5 in PIO\_REHLSR.

The other lines are configured in falling edge or low-level detection by default if they have not been previously configured. Otherwise, lines 1, 3 and 6 must be configured in falling edge/low-level detection by writing 32'h0000\_004A in PIO\_FELLSR.

**Figure 32-8. Input Change Interrupt Timings When No Additional Interrupt Modes**



**Figure 32-9.**

### 32.5.11 Programmable I/O Drive

It is possible to configure the I/O drive for pads PA0 to PA31. For any details, refer to the product electrical characteristics.

### 32.5.12 Programmable Schmitt Trigger

It is possible to configure each input for the Schmitt trigger. By default the Schmitt trigger is active. Disabling the Schmitt trigger is requested when using the QTouch™ Library.

### 32.5.13 Register Write Protection

To prevent any single software error from corrupting PIO behavior, certain registers in the address space can be write-protected by setting the WPEN bit in the [“PIO Write Protection Mode Register”](#) (PIO\_WPMR).

If a write access to a write-protected register is detected, the WPVS flag in the [“PIO Write Protection Status Register”](#) (PIO\_WPSR) is set and the field WPVSR indicates the register in which the write access has been attempted.

The WPVS bit is automatically cleared after reading the PIO\_WPSR.

The following registers can be write-protected:

- [“PIO Enable Register” on page 578](#)
- [“PIO Disable Register” on page 578](#)
- [“PIO Output Enable Register” on page 580](#)
- [“PIO Output Disable Register” on page 580](#)
- [“PIO Input Filter Enable Register” on page 582](#)
- [“PIO Input Filter Disable Register” on page 582](#)
- [“PIO Multi-driver Enable Register” on page 590](#)
- [“PIO Multi-driver Disable Register” on page 590](#)
- [“PIO Pull-Up Disable Register” on page 592](#)
- [“PIO Pull-Up Enable Register” on page 592](#)
- [“PIO Peripheral ABCD Select Register 1” on page 594](#)
- [“PIO Peripheral ABCD Select Register 2” on page 595](#)
- [“PIO Output Write Enable Register” on page 601](#)
- [“PIO Output Write Disable Register” on page 601](#)
- [“PIO Pad Pull-Down Disable Register” on page 599](#)
- [“PIO Pad Pull-Down Status Register” on page 600](#)

## 32.6 I/O Lines Programming Example

The programming example shown in [Table 32-1](#) is used to obtain the following configuration.

- 4-bit output port on I/O lines 0 to 3, (should be written in a single write operation), open-drain, with pull-up resistor
- Four output signals on I/O lines 4 to 7 (to drive LEDs for example), driven high and low, no pull-up resistor, no pull-down resistor
- Four input signals on I/O lines 8 to 11 (to read push-button states for example), with pull-up resistors, glitch filters and input change interrupts
- Four input signals on I/O line 12 to 15 to read an external device status (polled, thus no input change interrupt), no pull-up resistor, no glitch filter
- I/O lines 16 to 19 assigned to peripheral A functions with pull-up resistor
- I/O lines 20 to 23 assigned to peripheral B functions with pull-down resistor
- I/O line 24 to 27 assigned to peripheral C with Input Change Interrupt, no pull-up resistor and no pull-down resistor
- I/O line 28 to 31 assigned to peripheral D, no pull-up resistor and no pull-down resistor

**Table 32-1. Programming Example**

Register	Value to be Written
PIO_PER	0x0000_FFFF
PIO_PDR	0xFFFF_0000
PIO_OER	0x0000_00FF
PIO_ODR	0xFFFF_FF00
PIO_IFER	0x0000_0F00
PIO_IFDR	0xFFFF_F0FF
PIO_SODR	0x0000_0000
PIO_CODR	0x0FFF_FFFF
PIO_IER	0x0F00_0F00
PIO_IDR	0xF0FF_F0FF
PIO_MDER	0x0000_000F
PIO_MDDR	0xFFFF_FFF0
PIO_PUDR	0xFFFF0_00F0
PIO_PUER	0x000F_FF0F
PIO_PPDDR	0xFF0F_FFFF
PIO_PPDER	0x00F0_0000
PIO_ABCDSR1	0xF0F0_0000
PIO_ABCDSR2	0xFF00_0000
PIO_OWER	0x0000_000F
PIO_OWDR	0x0FFF_FFF0

## 32.7 Parallel Input/Output Controller (PIO) User Interface

Each I/O line controlled by the PIO Controller is associated with a bit in each of the PIO Controller User Interface registers. Each register is 32 bits wide. If a parallel I/O line is not defined, writing to the corresponding bits has no effect. Undefined bits read zero. If the I/O line is not multiplexed with any peripheral, the I/O line is controlled by the PIO Controller and PIO\_PSR returns one systematically.

**Table 32-2. Register Mapping**

Offset	Register	Name	Access	Reset
0x0000	PIO Enable Register	PIO_PER	Write-only	–
0x0004	PIO Disable Register	PIO_PDR	Write-only	–
0x0008	PIO Status Register	PIO_PSR	Read-only	(1)
0x000C	Reserved	–	–	–
0x0010	Output Enable Register	PIO_OER	Write-only	–
0x0014	Output Disable Register	PIO_ODR	Write-only	–
0x0018	Output Status Register	PIO_OSR	Read-only	0x0000 0000
0x001C	Reserved	–	–	–
0x0020	Glitch Input Filter Enable Register	PIO_IFER	Write-only	–
0x0024	Glitch Input Filter Disable Register	PIO_IFDR	Write-only	–
0x0028	Glitch Input Filter Status Register	PIO_IFSR	Read-only	0x0000 0000
0x002C	Reserved	–	–	–
0x0030	Set Output Data Register	PIO_SODR	Write-only	–
0x0034	Clear Output Data Register	PIO_CODR	Write-only	–
0x0038	Output Data Status Register	PIO_ODSR	Read-only or <sup>(2)</sup> Read/Write	–
0x003C	Pin Data Status Register	PIO_PDSR	Read-only	(3)
0x0040	Interrupt Enable Register	PIO_IER	Write-only	–
0x0044	Interrupt Disable Register	PIO_IDR	Write-only	–
0x0048	Interrupt Mask Register	PIO_IMR	Read-only	0x00000000
0x004C	Interrupt Status Register <sup>(4)</sup>	PIO_ISR	Read-only	0x00000000
0x0050	Multi-driver Enable Register	PIO_MDER	Write-only	–
0x0054	Multi-driver Disable Register	PIO_MDDR	Write-only	–
0x0058	Multi-driver Status Register	PIO_MDSR	Read-only	0x00000000
0x005C	Reserved	–	–	–
0x0060	Pull-up Disable Register	PIO_PUDR	Write-only	–
0x0064	Pull-up Enable Register	PIO_PUER	Write-only	–
0x0068	Pad Pull-up Status Register	PIO_PUSR	Read-only	(1)
0x006C	Reserved	–	–	–

**Table 32-2. Register Mapping (Continued)**

Offset	Register	Name	Access	Reset
0x0070	Peripheral Select Register 1	PIO_ABCDSR1	Read/Write	0x00000000
0x0074	Peripheral Select Register 2	PIO_ABCDSR2	Read/Write	0x00000000
0x0078 to 0x007C	Reserved	–	–	–
0x0080	Input Filter Slow Clock Disable Register	PIO_IFSCDR	Write-only	–
0x0084	Input Filter Slow Clock Enable Register	PIO_IFSCER	Write-only	–
0x0088	Input Filter Slow Clock Status Register	PIO_IFSCSR	Read-only	0x00000000
0x008C	Slow Clock Divider Debouncing Register	PIO_SCDR	Read/Write	0x00000000
0x0090	Pad Pull-down Disable Register	PIO_PPDDR	Write-only	–
0x0094	Pad Pull-down Enable Register	PIO_PPDER	Write-only	–
0x0098	Pad Pull-down Status Register	PIO_PPDSR	Read-only	(1)
0x009C	Reserved	–	–	–
0x00A0	Output Write Enable	PIO_OWER	Write-only	–
0x00A4	Output Write Disable	PIO_OWDR	Write-only	–
0x00A8	Output Write Status Register	PIO_OWSR	Read-only	0x00000000
0x00AC	Reserved	–	–	–
0x00B0	Additional Interrupt Modes Enable Register	PIO_AIMER	Write-only	–
0x00B4	Additional Interrupt Modes Disable Register	PIO_AIMDR	Write-only	–
0x00B8	Additional Interrupt Modes Mask Register	PIO_AIMMR	Read-only	0x00000000
0x00BC	Reserved	–	–	–
0x00C0	Edge Select Register	PIO_ESR	Write-only	–
0x00C4	Level Select Register	PIO_LSR	Write-only	–
0x00C8	Edge/Level Status Register	PIO_ELSR	Read-only	0x00000000
0x00CC	Reserved	–	–	–
0x00D0	Falling Edge/Low-Level Select Register	PIO_FELLSR	Write-only	–
0x00D4	Rising Edge/ High-Level Select Register	PIO_REHLSR	Write-only	–
0x00D8	Fall/Rise - Low/High Status Register	PIO_FRLHSR	Read-only	0x00000000
0x00DC	Reserved	–	–	–
0x00E0	Reserved	–	–	–
0x00E4	Write Protection Mode Register	PIO_WPMR	Read/Write	0x0
0x00E8	Write Protection Status Register	PIO_WPSR	Read-only	0x0
0x00EC to 0x00F8	Reserved	–	–	–
0x0100	Schmitt Trigger Register	PIO_SCHMITT	Read/Write	0x00000000
0x0104-0x010C	Reserved	–	–	–
0x0110	Reserved	–	–	–
0x0114	Reserved	–	–	–



**Table 32-2. Register Mapping (Continued)**

Offset	Register	Name	Access	Reset
0x0118	I/O Drive Register 1	PIO_DRIVER1	Read/Write	0x00000000
0x011C	I/O Drive Register 2	PIO_DRIVER2	Read/Write	0x00000000
0x0120 to 0x014C	Reserved	—	—	—

- Notes:
1. Reset value depends on the product implementation.
  2. PIO\_ODSR is Read-only or Read/Write depending on PIO\_OWSR I/O lines.
  3. Reset value of PIO\_PDSR depends on the level of the I/O lines. Reading the I/O line levels requires the clock of the PIO Controller to be enabled, otherwise PIO\_PDSR reads the levels present on the I/O line at the time the clock was disabled.
  4. PIO\_ISR is reset at 0x0. However, the first read of the register may read a different value as input changes may have occurred.
  5. If an offset is not listed in the table it must be considered as reserved.

### 32.7.1 PIO Enable Register

**Name:** PIO\_PER

**Address:** 0x400E0E00 (PIOA), 0x400E1000 (PIOB), 0x4800C000 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: PIO Enable**

0: No effect.

1: Enables the PIO to control the corresponding pin (disables peripheral control of the pin).

### 32.7.2 PIO Disable Register

**Name:** PIO\_PDR

**Address:** 0x400E0E04 (PIOA), 0x400E1004 (PIOB), 0x4800C004 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: PIO Disable**

0: No effect.

1: Disables the PIO from controlling the corresponding pin (enables peripheral control of the pin).

### 32.7.3 PIO Status Register

**Name:** PIO\_PSR

**Address:** 0x400E0E08 (PIOA), 0x400E1008 (PIOB), 0x4800C008 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: PIO Status**

0: PIO is inactive on the corresponding I/O line (peripheral is active).

1: PIO is active on the corresponding I/O line (peripheral is inactive).

### 32.7.4 PIO Output Enable Register

**Name:** PIO\_OER

**Address:** 0x400E0E10 (PIOA), 0x400E1010 (PIOB), 0x4800C010 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Output Enable**

0: No effect.

1: Enables the output on the I/O line.

### 32.7.5 PIO Output Disable Register

**Name:** PIO\_ODR

**Address:** 0x400E0E14 (PIOA), 0x400E1014 (PIOB), 0x4800C014 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Output Disable**

0: No effect.

1: Disables the output on the I/O line.

### 32.7.6 PIO Output Status Register

**Name:** PIO\_OSR

**Address:** 0x400E0E18 (PIOA), 0x400E1018 (PIOB), 0x4800C018 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Output Status**

0: The I/O line is a pure input.

1: The I/O line is enabled in output.

### 32.7.7 PIO Input Filter Enable Register

**Name:** PIO\_IFER

**Address:** 0x400E0E20 (PIOA), 0x400E1020 (PIOB), 0x4800C020 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Input Filter Enable**

0: No effect.

1: Enables the input glitch filter on the I/O line.

### 32.7.8 PIO Input Filter Disable Register

**Name:** PIO\_IFDR

**Address:** 0x400E0E24 (PIOA), 0x400E1024 (PIOB), 0x4800C024 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Input Filter Disable**

0: No effect.

1: Disables the input glitch filter on the I/O line.

### 32.7.9 PIO Input Filter Status Register

**Name:** PIO\_IFSR

**Address:** 0x400E0E28 (PIOA), 0x400E1028 (PIOB), 0x4800C028 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Input Filter Status**

0: The input glitch filter is disabled on the I/O line.

1: The input glitch filter is enabled on the I/O line.

### 32.7.10 PIO Set Output Data Register

**Name:** PIO\_SODR

**Address:** 0x400E0E30 (PIOA), 0x400E1030 (PIOB), 0x4800C030 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Set Output Data**

0: No effect.

1: Sets the data to be driven on the I/O line.

### 32.7.11 PIO Clear Output Data Register

**Name:** PIO\_CODR

**Address:** 0x400E0E34 (PIOA), 0x400E1034 (PIOB), 0x4800C034 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Clear Output Data**

0: No effect.

1: Clears the data to be driven on the I/O line.



### 32.7.12 PIO Output Data Status Register

**Name:** PIO\_ODSR

**Address:** 0x400E0E38 (PIOA), 0x400E1038 (PIOB), 0x4800C038 (PIOC)

**Access:** Read-only or Read/Write

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Output Data Status**

0: The data to be driven on the I/O line is 0.

1: The data to be driven on the I/O line is 1.

### 32.7.13 PIO Pin Data Status Register

**Name:** PIO\_PDSR

**Address:** 0x400E0E3C (PIOA), 0x400E103C (PIOB), 0x4800C03C (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Output Data Status**

0: The I/O line is at level 0.

1: The I/O line is at level 1.

### 32.7.14 PIO Interrupt Enable Register

**Name:** PIO\_IER

**Address:** 0x400E0E40 (PIOA), 0x400E1040 (PIOB), 0x4800C040 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Input Change Interrupt Enable**

0: No effect.

1: Enables the Input Change interrupt on the I/O line.

### 32.7.15 PIO Interrupt Disable Register

**Name:** PIO\_IDR

**Address:** 0x400E0E44 (PIOA), 0x400E1044 (PIOB), 0x4800C044 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Input Change Interrupt Disable**

0: No effect.

1: Disables the Input Change interrupt on the I/O line.

### 32.7.16 PIO Interrupt Mask Register

**Name:** PIO\_IMR

**Address:** 0x400E0E48 (PIOA), 0x400E1048 (PIOB), 0x4800C048 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Input Change Interrupt Mask**

0: Input Change interrupt is disabled on the I/O line.

1: Input Change interrupt is enabled on the I/O line.

### 32.7.17 PIO Interrupt Status Register

**Name:** PIO\_ISR

**Address:** 0x400E0E4C (PIOA), 0x400E104C (PIOB), 0x4800C04C (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Input Change Interrupt Status**

0: No Input Change has been detected on the I/O line since PIO\_ISR was last read or since reset.

1: At least one Input Change has been detected on the I/O line since PIO\_ISR was last read or since reset.

### 32.7.18 PIO Multi-driver Enable Register

**Name:** PIO\_MDER

**Address:** 0x400E0E50 (PIOA), 0x400E1050 (PIOB), 0x4800C050 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Multi-Drive Enable**

0: No effect.

1: Enables multi-drive on the I/O line.

### 32.7.19 PIO Multi-driver Disable Register

**Name:** PIO\_MDDR

**Address:** 0x400E0E54 (PIOA), 0x400E1054 (PIOB), 0x4800C054 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Multi-Drive Disable**

0: No effect.

1: Disables multi-drive on the I/O line.

### 32.7.20 PIO Multi-driver Status Register

**Name:** PIO\_MDSR

**Address:** 0x400E0E58 (PIOA), 0x400E1058 (PIOB), 0x4800C058 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Multi-Drive Status**

0: The multi-drive is disabled on the I/O line. The pin is driven at high- and low-level.

1: The multi-drive is enabled on the I/O line. The pin is driven at low-level only.

### 32.7.21 PIO Pull-Up Disable Register

**Name:** PIO\_PUDR

**Address:** 0x400E0E60 (PIOA), 0x400E1060 (PIOB), 0x4800C060 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Pull-Up Disable**

0: No effect.

1: Disables the pull-up resistor on the I/O line.

### 32.7.22 PIO Pull-Up Enable Register

**Name:** PIO\_PUER

**Address:** 0x400E0E64 (PIOA), 0x400E1064 (PIOB), 0x4800C064 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Pull-Up Enable**

0: No effect.

1: Enables the pull-up resistor on the I/O line.



### 32.7.23 PIO Pull-Up Status Register

**Name:** PIO\_PUSR

**Address:** 0x400E0E68 (PIOA), 0x400E1068 (PIOB), 0x4800C068 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Pull-Up Status**

0: Pull-up resistor is enabled on the I/O line.

1: Pull-up resistor is disabled on the I/O line.

### 32.7.24 PIO Peripheral ABCD Select Register 1

**Name:** PIO\_ABCDSR1

**Access:** Read/Write

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Peripheral Select**

If the same bit is set to 0 in PIO\_ABCDSR2:

0: Assigns the I/O line to the Peripheral A function.

1: Assigns the I/O line to the Peripheral B function.

If the same bit is set to 1 in PIO\_ABCDSR2:

0: Assigns the I/O line to the Peripheral C function.

1: Assigns the I/O line to the Peripheral D function.

### 32.7.25 PIO Peripheral ABCD Select Register 2

**Name:** PIO\_ABCDSR2

**Access:** Read/Write

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#) .

- **P0-P31: Peripheral Select.**

If the same bit is set to 0 in PIO\_ABCDSR1:

0: Assigns the I/O line to the Peripheral A function.

1: Assigns the I/O line to the Peripheral C function.

If the same bit is set to 1 in PIO\_ABCDSR1:

0: Assigns the I/O line to the Peripheral B function.

1: Assigns the I/O line to the Peripheral D function.

### 32.7.26 PIO Input Filter Slow Clock Disable Register

**Name:** PIO\_IFSCDR

**Address:** 0x400E0E80 (PIOA), 0x400E1080 (PIOB), 0x4800C080 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: PIO Clock Glitch Filtering Select**

0: No effect.

1: The glitch filter is able to filter glitches with a duration  $< T_{mck}/2$ .

### 32.7.27 PIO Input Filter Slow Clock Enable Register

**Name:** PIO\_IFSCER

**Address:** 0x400E0E84 (PIOA), 0x400E1084 (PIOB), 0x4800C084 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Debouncing Filtering Select**

0: No effect.

1: The debouncing filter is able to filter pulses with a duration  $< T_{div\_slck}/2$ .

### 32.7.28 PIO Input Filter Slow Clock Status Register

**Name:** PIO\_IFSCSR

**Address:** 0x400E0E88 (PIOA), 0x400E1088 (PIOB), 0x4800C088 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Glitch or Debouncing Filter Selection Status**

0: The glitch filter is able to filter glitches with a duration  $< T_{mck2}$ .

1: The debouncing filter is able to filter pulses with a duration  $< T_{div\_sclk}/2$ .

32.7.29 PIO Slow Clock Divider Debouncing Register

**Name:** PIO\_SCDR  
**Address:** 0x400E0E8C (PIOA), 0x400E108C (PIOB), 0x4800C08C (PIOC)  
**Access:** Read/Write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	DIV					
7	6	5	4	3	2	1	0
DIV							

• **DIV: Slow Clock Divider Selection for Debouncing**

$T_{div\_slclk} = 2 \cdot (DIV + 1) \cdot T_{slow\_clock}$

### 32.7.30 PIO Pad Pull-Down Disable Register

**Name:** PIO\_PPDDR

**Address:** 0x400E0E90 (PIOA), 0x400E1090 (PIOB), 0x4800C090 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Pull-Down Disable**

0: No effect.

1: Disables the pull-down resistor on the I/O line.

### 32.7.31 PIO Pad Pull-Down Enable Register

**Name:** PIO\_PPDER

**Address:** 0x400E0E94 (PIOA), 0x400E1094 (PIOB), 0x4800C094 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Pull-Down Enable**

0: No effect.

1: Enables the pull-down resistor on the I/O line.

### 32.7.32 PIO Pad Pull-Down Status Register

**Name:** PIO\_PPDSR

**Address:** 0x400E0E98 (PIOA), 0x400E1098 (PIOB), 0x4800C098 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Pull-Down Status**

0: Pull-down resistor is enabled on the I/O line.

1: Pull-down resistor is disabled on the I/O line.



### 32.7.33 PIO Output Write Enable Register

**Name:** PIO\_OWER

**Address:** 0x400E0EA0 (PIOA), 0x400E10A0 (PIOB), 0x4800C0A0 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Output Write Enable**

0: No effect.

1: Enables writing PIO\_ODSR for the I/O line.

### 32.7.34 PIO Output Write Disable Register

**Name:** PIO\_OWDR

**Address:** 0x400E0EA4 (PIOA), 0x400E10A4 (PIOB), 0x4800C0A4 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

This register can only be written if the WPEN bit is cleared in [“PIO Write Protection Mode Register”](#).

- **P0-P31: Output Write Disable**

0: No effect.

1: Disables writing PIO\_ODSR for the I/O line.

### 32.7.35 PIO Output Write Status Register

**Name:** PIO\_OWSR

**Address:** 0x400E0EA8 (PIOA), 0x400E10A8 (PIOB), 0x4800C0A8 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Output Write Status**

0: Writing PIO\_ODSR does not affect the I/O line.

1: Writing PIO\_ODSR affects the I/O line.

### 32.7.36 PIO Additional Interrupt Modes Enable Register

**Name:** PIO\_AIMER

**Address:** 0x400E0EB0 (PIOA), 0x400E10B0 (PIOB), 0x4800C0B0 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Additional Interrupt Modes Enable**

0: No effect.

1: The interrupt source is the event described in PIO\_ELSR and PIO\_FRLHSR.

### 32.7.37 PIO Additional Interrupt Modes Disable Register

**Name:** PIO\_AIMDR

**Address:** 0x400E0EB4 (PIOA), 0x400E10B4 (PIOB), 0x4800C0B4 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Additional Interrupt Modes Disable**

0: No effect.

1: The interrupt mode is set to the default interrupt mode (both-edge detection).

### 32.7.38 PIO Additional Interrupt Modes Mask Register

**Name:** PIO\_AIMMR

**Address:** 0x400E0EB8 (PIOA), 0x400E10B8 (PIOB), 0x4800C0B8 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Peripheral CD Status**

0: The interrupt source is a both-edge detection event.

1: The interrupt source is described by the registers PIO\_ELSR and PIO\_FRLHSR.

### 32.7.39 PIO Edge Select Register

**Name:** PIO\_ESR

**Address:** 0x400E0EC0 (PIOA), 0x400E10C0 (PIOB), 0x4800C0C0 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Edge Interrupt Selection**

0: No effect.

1: The interrupt source is an edge-detection event.

### 32.7.40 PIO Level Select Register

**Name:** PIO\_LSR

**Address:** 0x400E0EC4 (PIOA), 0x400E10C4 (PIOB), 0x4800C0C4 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Level Interrupt Selection**

0: No effect.

1: The interrupt source is a level-detection event.

### 32.7.41 PIO Edge/Level Status Register

**Name:** PIO\_ELSR

**Address:** 0x400E0EC8 (PIOA), 0x400E10C8 (PIOB), 0x4800C0C8 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Edge/Level Interrupt Source Selection**

0: The interrupt source is an edge-detection event.

1: The interrupt source is a level-detection event.

### 32.7.42 PIO Falling Edge/Low-Level Select Register

**Name:** PIO\_FELLSR

**Address:** 0x400E0ED0 (PIOA), 0x400E10D0 (PIOB), 0x4800C0D0 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Falling Edge/Low-Level Interrupt Selection**

0: No effect.

1: The interrupt source is set to a falling edge detection or low-level detection event, depending on PIO\_ELSR.

### 32.7.43 PIO Rising Edge/High-Level Select Register

**Name:** PIO\_REHLSR

**Address:** 0x400E0ED4 (PIOA), 0x400E10D4 (PIOB), 0x4800C0D4 (PIOC)

**Access:** Write-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Rising Edge /High-Level Interrupt Selection**

0: No effect.

1: The interrupt source is set to a rising edge detection or high-level detection event, depending on PIO\_ELSR.

### 32.7.44 PIO Fall/Rise - Low/High Status Register

**Name:** PIO\_FRLHSR

**Address:** 0x400E0ED8 (PIOA), 0x400E10D8 (PIOB), 0x4800C0D8 (PIOC)

**Access:** Read-only

31	30	29	28	27	26	25	24
P31	P30	P29	P28	P27	P26	P25	P24
23	22	21	20	19	18	17	16
P23	P22	P21	P20	P19	P18	P17	P16
15	14	13	12	11	10	9	8
P15	P14	P13	P12	P11	P10	P9	P8
7	6	5	4	3	2	1	0
P7	P6	P5	P4	P3	P2	P1	P0

- **P0-P31: Edge /Level Interrupt Source Selection**

0: The interrupt source is a falling edge detection (if PIO\_ELSR = 0) or low-level detection event (if PIO\_ELSR = 1).

1: The interrupt source is a rising edge detection (if PIO\_ELSR = 0) or high-level detection event (if PIO\_ELSR = 1).

### 32.7.45 PIO Write Protection Mode Register

**Name:** PIO\_WPMR

**Address:** 0x400E0EE4 (PIOA), 0x400E10E4 (PIOB), 0x4800C0E4 (PIOC)

**Access:** Read/Write

**Reset:** See [Table 32-2](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

For more information on write-protecting registers, refer to [Section 32.5.13 "Register Write Protection"](#).

- **WPEN: Write Protection Enable**

0: Disables the write protection if WPKEY corresponds to 0x50494F ("PIO" in ASCII).

1: Enables the write protection if WPKEY corresponds to 0x50494F ("PIO" in ASCII).

See [Section 32.5.13 "Register Write Protection"](#) for the list of registers that can be protected.

- **WPKEY: Write Protection Key.**

Value	Name	Description
0x50494F	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.



### 32.7.46 PIO Write Protection Status Register

**Name:** PIO\_WPSR

**Address:** 0x400E0EE8 (PIOA), 0x400E10E8 (PIOB), 0x4800C0E8 (PIOC)

**Access:** Read-only

**Reset:** See [Table 32-2](#)

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

- **WPVS: Write Protection Violation Status**

0: No write protection violation has occurred since the last read of the PIO\_WPSR register.

1: A write protection violation has occurred since the last read of the PIO\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protection Violation Source**

When WPVS = 1, WPVSR indicates the register address offset at which a write access has been attempted.

### 32.7.47 PIO Schmitt Trigger Register

**Name:** PIO\_SCHMITT

**Address:** 0x400E0F00 (PIOA), 0x400E1100 (PIOB), 0x4800C100 (PIOC)

**Access:** Read/Write

**Reset:** See [Table 32-2](#)

31	30	29	28	27	26	25	24
SCHMITT31	SCHMITT30	SCHMITT29	SCHMITT28	SCHMITT27	SCHMITT26	SCHMITT25	SCHMITT24
23	22	21	20	19	18	17	16
SCHMITT23	SCHMITT22	SCHMITT21	SCHMITT20	SCHMITT19	SCHMITT18	SCHMITT17	SCHMITT16
15	14	13	12	11	10	9	8
SCHMITT15	SCHMITT14	SCHMITT13	SCHMITT12	SCHMITT11	SCHMITT10	SCHMITT9	SCHMITT8
7	6	5	4	3	2	1	0
SCHMITT7	SCHMITT6	SCHMITT5	SCHMITT4	SCHMITT3	SCHMITT2	SCHMITT1	SCHMITT0

- **SCHMITTx [x=0..31]: Schmitt Trigger Control**

0: Schmitt trigger is enabled.

1: Schmitt trigger is disabled.

### 32.7.48 PIO I/O Drive Register 1

**Name:** PIO\_DRIVER1

**Address:** 0x400E0F18 (PIOA), 0x400E1118 (PIOB), 0x4800C118 (PIOC)

**Access:** Read/Write

**Reset:** See [Table 32-2](#)

31	30	29	28	27	26	25	24
LINE15				LINE14			
				LINE13			
				LINE12			
23	22	21	20	19	18	17	16
LINE11				LINE10			
				LINE9			
				LINE8			
15	14	13	12	11	10	9	8
LINE7				LINE6			
				LINE5			
				LINE4			
7	6	5	4	3	2	1	0
LINE3				LINE2			
				LINE1			
				LINE0			

- **LINE<sub>x</sub> [x=0..15]: Drive of PIO Line x**

Value	Name	Description
0	HI_DRIVE	High drive
1	ME_DRIVE	Medium drive
2	LO_DRIVE	Low drive
3		Reserved

### 32.7.49 PIO I/O Drive Register 2

**Name:** PIO\_DRIVER2

**Address:** 0x400E0F1C (PIOA), 0x400E111C (PIOB), 0x4800C11C (PIOC)

**Access:** Read/Write

**Reset:** See [Table 32-2](#)

31	30	29	28	27	26	25	24
LINE31		LINE30		LINE29		LINE28	
23	22	21	20	19	18	17	16
LINE27		LINE26		LINE25		LINE24	
15	14	13	12	11	10	9	8
LINE23		LINE22		LINE21		LINE20	
7	6	5	4	3	2	1	0
LINE19		LINE18		LINE17		LINE16	

- **LINE<sub>x</sub> [x=16..31]:** Drive of PIO line x

Value	Name	Description
0	HI_DRIVE	High drive
1	ME_DRIVE	Medium drive
2	LO_DRIVE	Low drive
3		Reserved

## 33. Serial Peripheral Interface (SPI)

### 33.1 Description

The Serial Peripheral Interface (SPI) circuit is a synchronous serial data link that provides communication with external devices in Master or Slave Mode. It also enables communication between processors if an external processor is connected to the system.

The Serial Peripheral Interface is essentially a shift register that serially transmits data bits to other SPIs. During a data transfer, one SPI system acts as the “master” which controls the data flow, while the other devices act as “slaves” which have data shifted into and out by the master. Different CPUs can take turn being masters (Multiple Master Protocol opposite to Single Master Protocol where one CPU is always the master while all of the others are always slaves) and one master may simultaneously shift data into multiple slaves. However, only one slave may drive its output to write data back to the master at any given time.

A slave device is selected when the master asserts its NSS signal. If multiple slave devices exist, the master generates a separate slave select signal for each slave (NPCS).

The SPI system consists of two data lines and two control lines:

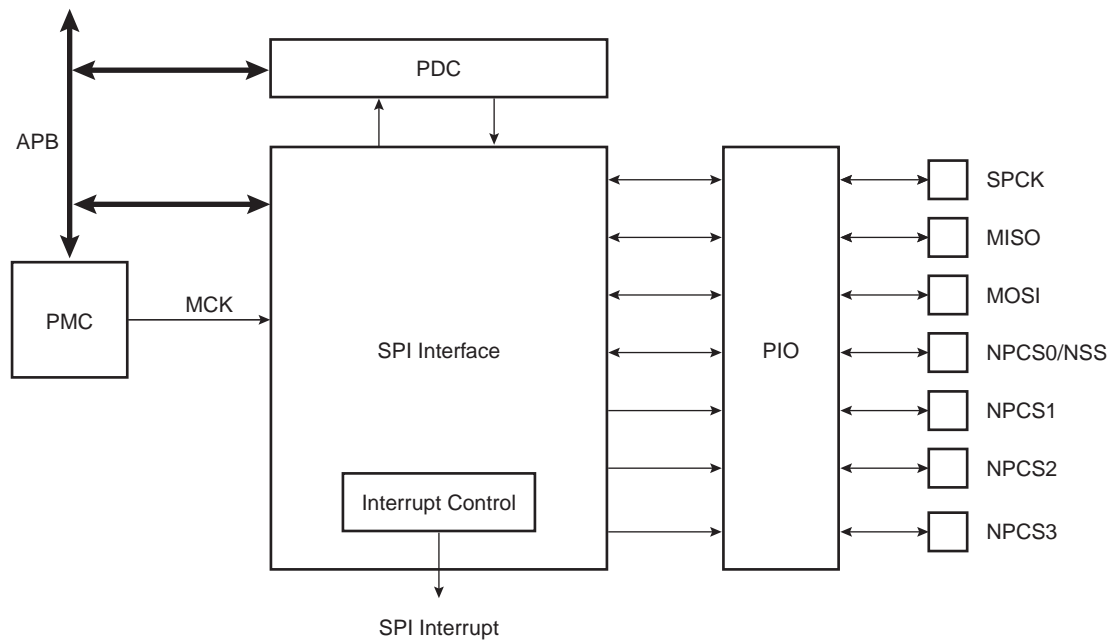
- Master Out Slave In (MOSI): This data line supplies the output data from the master shifted into the input(s) of the slave(s).
- Master In Slave Out (MISO): This data line supplies the output data from a slave to the input of the master. There may be no more than one slave transmitting data during any particular transfer.
- Serial Clock (SPCK): This control line is driven by the master and regulates the flow of the data bits. The master may transmit data at a variety of baud rates; the SPCK line cycles once for each bit that is transmitted.
- Slave Select (NSS): This control line allows slaves to be turned on and off by hardware.

## 33.2 Embedded Characteristics

- Supports Communication with Serial External Devices
  - Master Mode can drive SPCK up to peripheral clock (bounded by maximum bus clock divided by 2)
  - Slave Mode operates on SPCK, asynchronously to Core and Bus Clock
  - Four Chip Selects with External Decoder Support Allow Communication with Up to 15 Peripherals
  - Serial Memories, such as DataFlash and 3-wire EEPROMs
  - Serial Peripherals, such as ADCs, DACs, LCD Controllers, CAN Controllers and Sensors
  - External Coprocessors
- Master or Slave Serial Peripheral Bus Interface
  - 8-bit to 16-bit Programmable Data Length Per Chip Select
  - Programmable Phase and Polarity Per Chip Select
  - Programmable Transfer Delay Between Consecutive Transfers and Delay before SPI Clock per Chip Select
  - Programmable Delay Between Chip Selects
  - Selectable Mode Fault Detection
- Connection to PDC Channel Capabilities Optimizes Data Transfers
  - One Channel for the Receiver, One Channel for the Transmitter

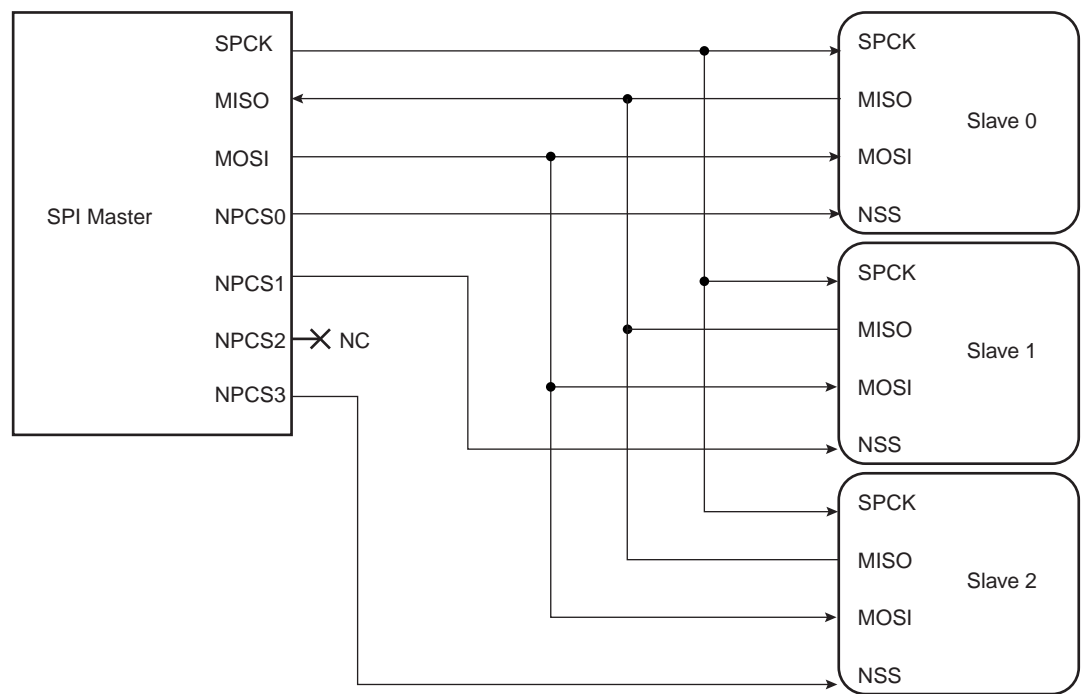
### 33.3 Block Diagram

Figure 33-1. Block Diagram



### 33.4 Application Block Diagram

Figure 33-2. Application Block Diagram: Single Master/Multiple Slave Implementation



## 33.5 Signal Description

Table 33-1. Signal Description

Pin Name	Pin Description	Type	
		Master	Slave
MISO	Master In Slave Out	Input	Output
MOSI	Master Out Slave In	Output	Input
SPCK	Serial Clock	Output	Input
NPCS1-NPCS3	Peripheral Chip Selects	Output	Unused
NPCS0/NSS	Peripheral Chip Select/Slave Select	Output	Input

## 33.6 Product Dependencies

### 33.6.1 I/O Lines

The pins used for interfacing the compliant external devices may be multiplexed with PIO lines. The programmer must first program the PIO controllers to assign the SPI pins to their peripheral functions.

Table 33-2. I/O Lines

Instance	Signal	I/O Line	Peripheral
SPI0	SPI0_MISO	PA6	A
SPI0	SPI0_MOSI	PA7	A
SPI0	SPI0_NPCS0	PA5	A
SPI0	SPI0_NPCS1	PA21	A
SPI0	SPI0_NPCS2	PA22	A
SPI0	SPI0_NPCS3	PA23	A
SPI0	SPI0_SPCK	PA8	A
SPI1	SPI1_MISO	PC3	A
SPI1	SPI1_MOSI	PC4	A
SPI1	SPI1_NPCS0	PC2	A
SPI1	SPI1_NPCS1	PC6	B
SPI1	SPI1_NPCS2	PC7	B
SPI1	SPI1_NPCS3	PC8	B
SPI1	SPI1_SPCK	PC5	A

### 33.6.2 Power Management

The SPI may be clocked through the Power Management Controller (PMC), thus the programmer must first configure the PMC to enable the SPI clock.



### 33.6.3 Interrupt

The SPI interface has an interrupt line connected to the Interrupt Controller. Handling the SPI interrupt requires programming the interrupt controller before configuring the SPI.

**Table 33-3. Peripheral IDs**

Instance	ID
SPI0	21
SPI1	40

### 33.6.4 Peripheral DMA Controller (PDC)

The SPI interface can be used in conjunction with the PDC in order to reduce processor overhead. For a full description of the PDC, refer to the corresponding section in the full datasheet.

## 33.7 Functional Description

### 33.7.1 Modes of Operation

The SPI operates in Master Mode or in Slave Mode.

Operation in Master Mode is programmed by writing at 1 the MSTR bit in the Mode Register. The pins NPCS0 to NPCS3 are all configured as outputs, the SPCK pin is driven, the MISO line is wired on the receiver input and the MOSI line driven as an output by the transmitter.

If the MSTR bit is written at 0, the SPI operates in Slave Mode. The MISO line is driven by the transmitter output, the MOSI line is wired on the receiver input, the SPCK pin is driven by the transmitter to synchronize the receiver. The NPCS0 pin becomes an input, and is used as a Slave Select signal (NSS). The pins NPCS1 to NPCS3 are not driven and can be used for other purposes.

The data transfers are identically programmable for both modes of operations. The baud rate generator is activated only in Master Mode.

### 33.7.2 Data Transfer

Four combinations of polarity and phase are available for data transfers. The clock polarity is programmed with the CPOL bit in the Chip Select Register. The clock phase is programmed with the NCPHA bit. These two parameters determine the edges of the clock signal on which data is driven and sampled. Each of the two parameters has two possible states, resulting in four possible combinations that are incompatible with one another. Thus, a master/slave pair must use the same parameter pair values to communicate. If multiple slaves are used and fixed in different configurations, the master must reconfigure itself each time it needs to communicate with a different slave.

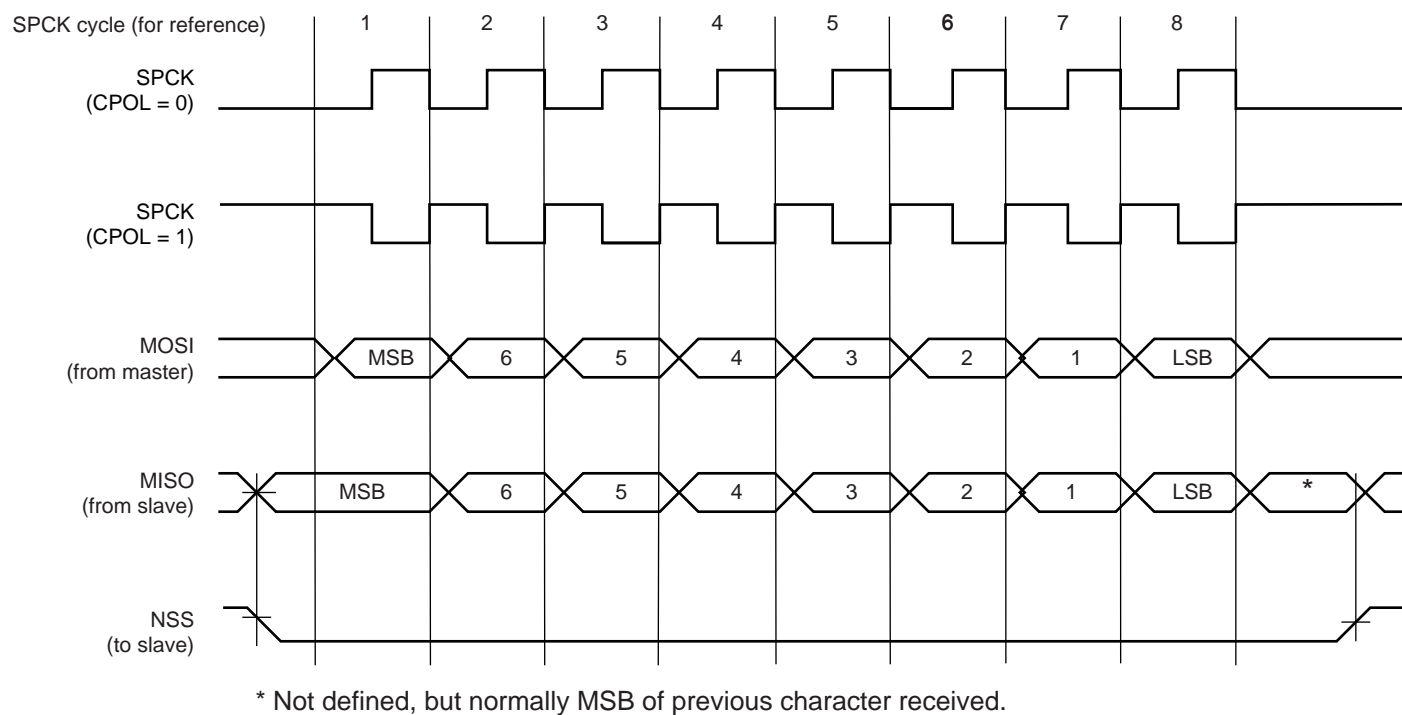
Table 33-4 shows the four modes and corresponding parameter settings.

**Table 33-4. SPI Bus Protocol Mode**

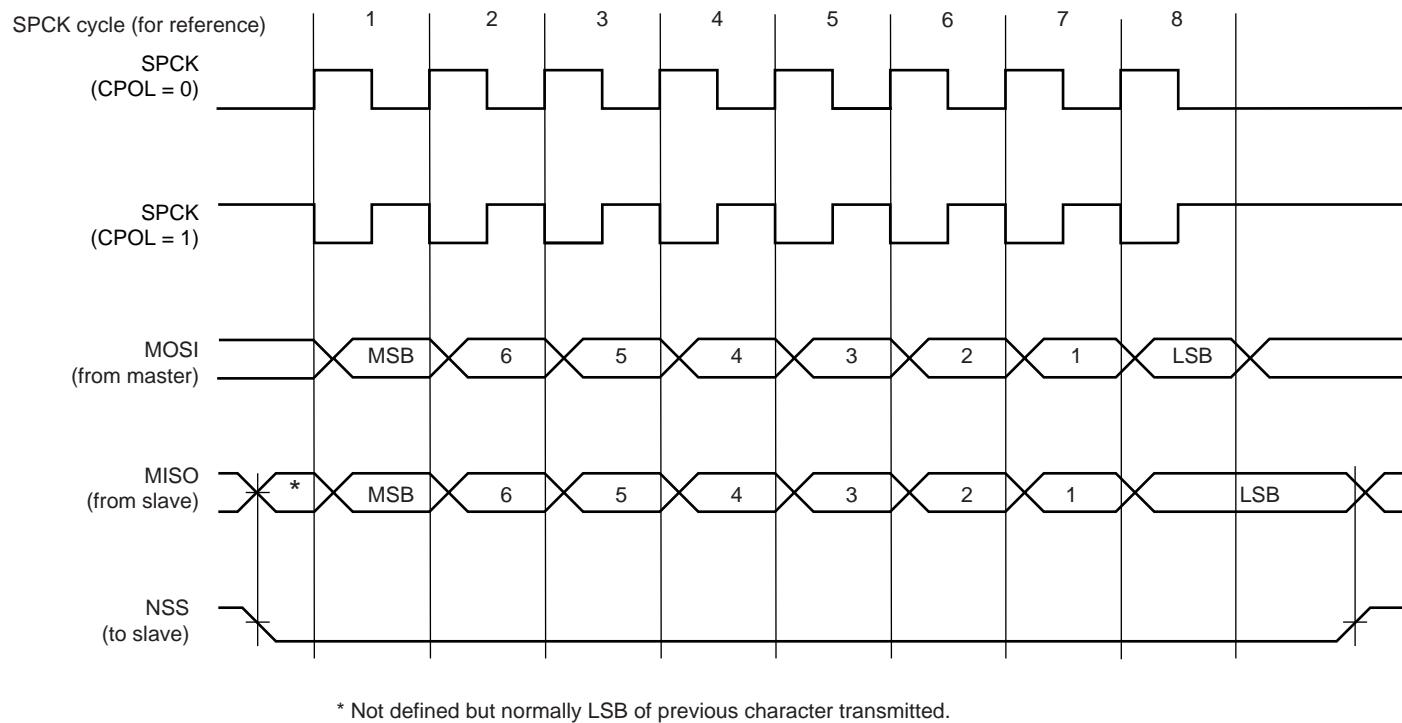
SPI Mode	CPOL	NCPHA	Shift SPCK Edge	Capture SPCK Edge	SPCK Inactive Level
0	0	1	Falling	Rising	Low
1	0	0	Rising	Falling	Low
2	1	1	Rising	Falling	High
3	1	0	Falling	Rising	High

Figure 33-3 and Figure 33-4 show examples of data transfers.

**Figure 33-3. SPI Transfer Format (NCPHA = 1, 8 bits per transfer)**



**Figure 33-4. SPI Transfer Format (NCPHA = 0, 8 bits per transfer)**



### 33.7.3 Master Mode Operations

When configured in Master Mode, the SPI operates on the clock generated by the internal programmable baud rate generator. It fully controls the data transfers to and from the slave(s) connected to the SPI bus. The SPI drives the chip select line to the slave and the serial clock signal (SPCK).

The SPI features two holding registers, the Transmit Data Register and the Receive Data Register, and a single Shift Register. The holding registers maintain the data flow at a constant rate.

After enabling the SPI, a data transfer begins when the processor writes to the SPI\_TDR (Transmit Data Register). The written data is immediately transferred in the Shift Register and transfer on the SPI bus starts. While the data in the Shift Register is shifted on the MOSI line, the MISO line is sampled and shifted in the Shift Register. Receiving data cannot occur without transmitting data. If receiving mode is not needed, for example when communicating with a slave receiver only (such as an LCD), the receive status flags in the status register can be discarded.

Before writing the TDR, the PCS field in the SPI\_MR register must be set in order to select a slave.

After enabling the SPI, a data transfer begins when the processor writes to the SPI\_TDR (Transmit Data Register). The written data is immediately transferred in the Shift Register and transfer on the SPI bus starts. While the data in the Shift Register is shifted on the MOSI line, the MISO line is sampled and shifted in the Shift Register. Transmission cannot occur without reception.

Before writing the TDR, the PCS field must be set in order to select a slave.

If new data is written in SPI\_TDR during the transfer, it stays in it until the current transfer is completed. Then, the received data is transferred from the Shift Register to SPI\_RDR, the data in SPI\_TDR is loaded in the Shift Register and a new transfer starts.

The transfer of a data written in SPI\_TDR in the Shift Register is indicated by the TDRE bit (Transmit Data Register Empty) in the Status Register (SPI\_SR). When new data is written in SPI\_TDR, this bit is cleared. The TDRE bit is used to trigger the TransmitPDC channel.

The end of transfer is indicated by the TXEMPTY flag in the SPI\_SR register. If a transfer delay (DLYBCT) is greater than 0 for the last transfer, TXEMPTY is set after the completion of said delay. The master clock (MCK) can be switched off at this time.

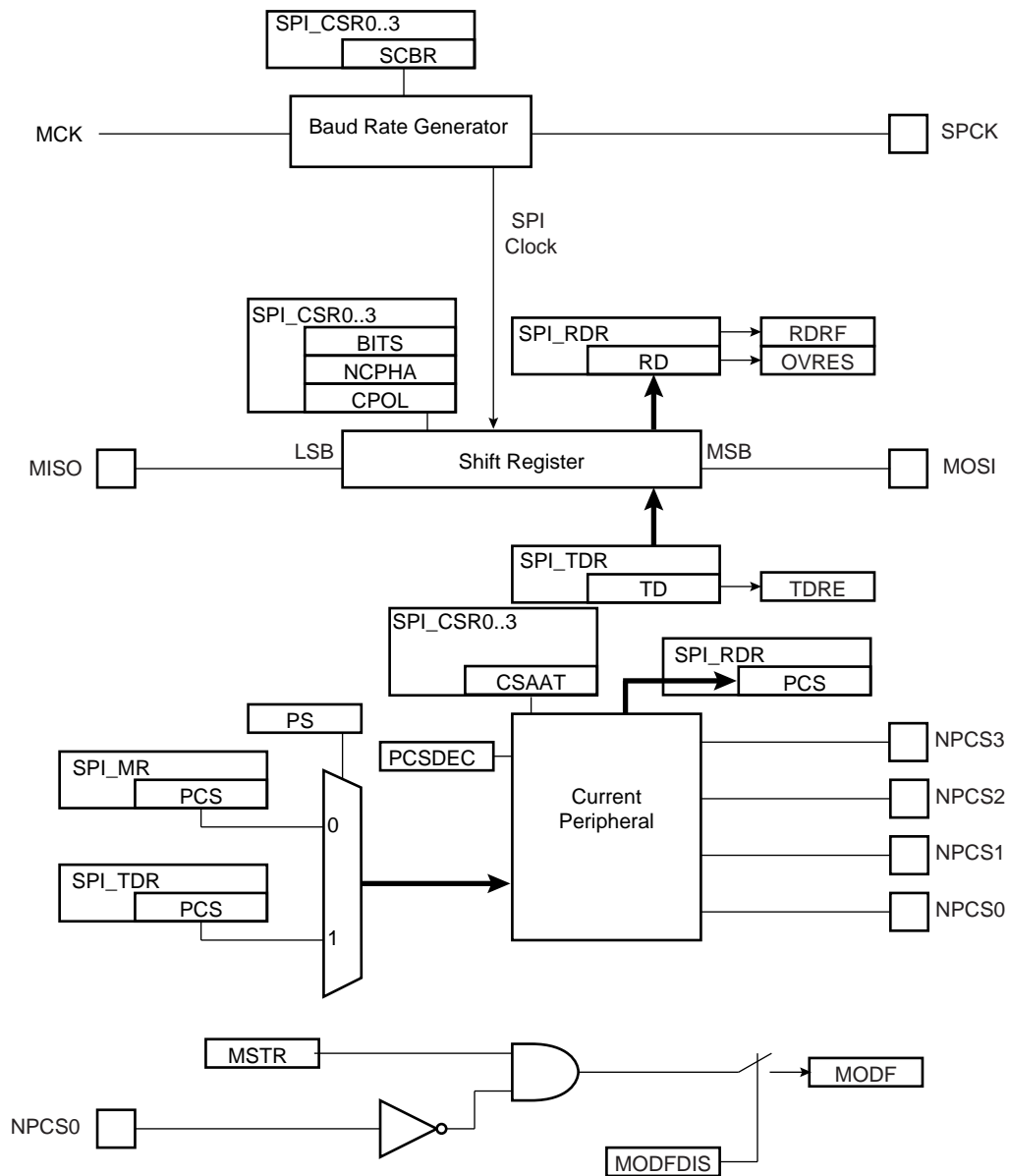
The transfer of received data from the Shift Register in SPI\_RDR is indicated by the RDRF bit (Receive Data Register Full) in the Status Register (SPI\_SR). When the received data is read, the RDRF bit is cleared.

If the SPI\_RDR (Receive Data Register) has not been read before new data is received, the Overrun Error bit (OVRES) in SPI\_SR is set. As long as this flag is set, data is loaded in SPI\_RDR. The user has to read the status register to clear the OVRES bit.

[Figure 33-5](#), shows a block diagram of the SPI when operating in Master Mode. [Figure 33-6 on page 621](#) shows a flow chart describing how transfers are handled.

### 33.7.3.1 Master Mode Block Diagram

Figure 33-5. Master Mode Block Diagram



### 33.7.3.2 Master Mode Flow Diagram

Figure 33-6. Master Mode Flow Diagram

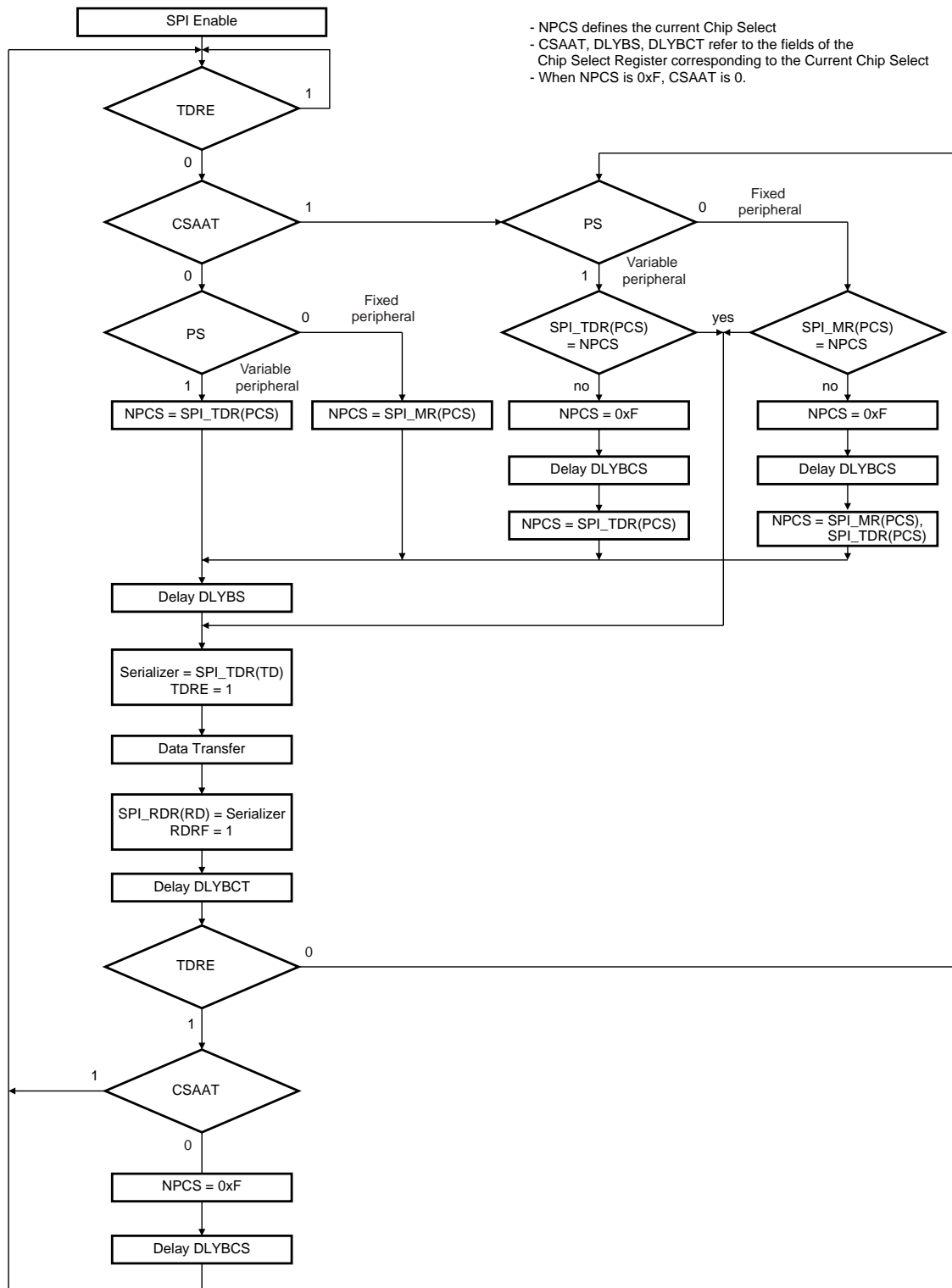


Figure 33-7 shows Transmit Data Register Empty (TDRE), Receive Data Register (RDRF) and Transmission Register Empty (TXEMPTY) status flags behavior within the SPI\_SR (Status Register) during an 8-bit data transfer in fixed mode and no Peripheral Data Controller involved.

**Figure 33-7. Status Register Flags Behavior**

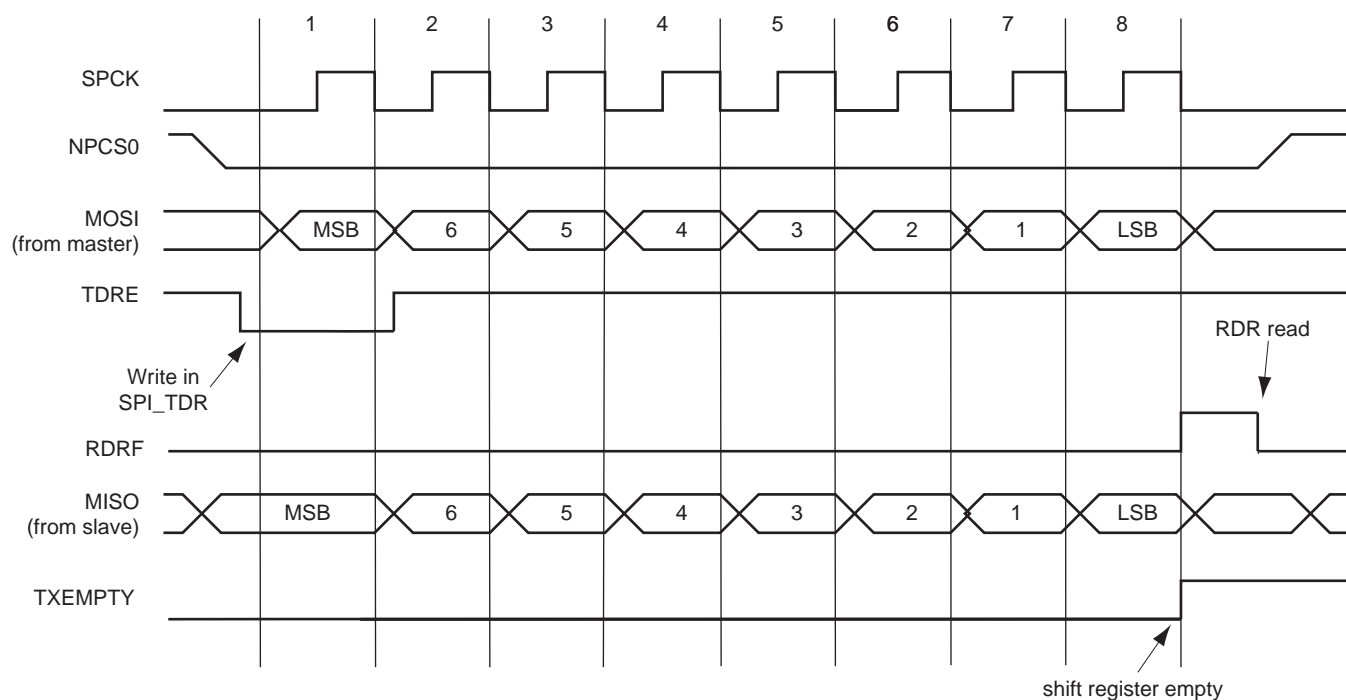
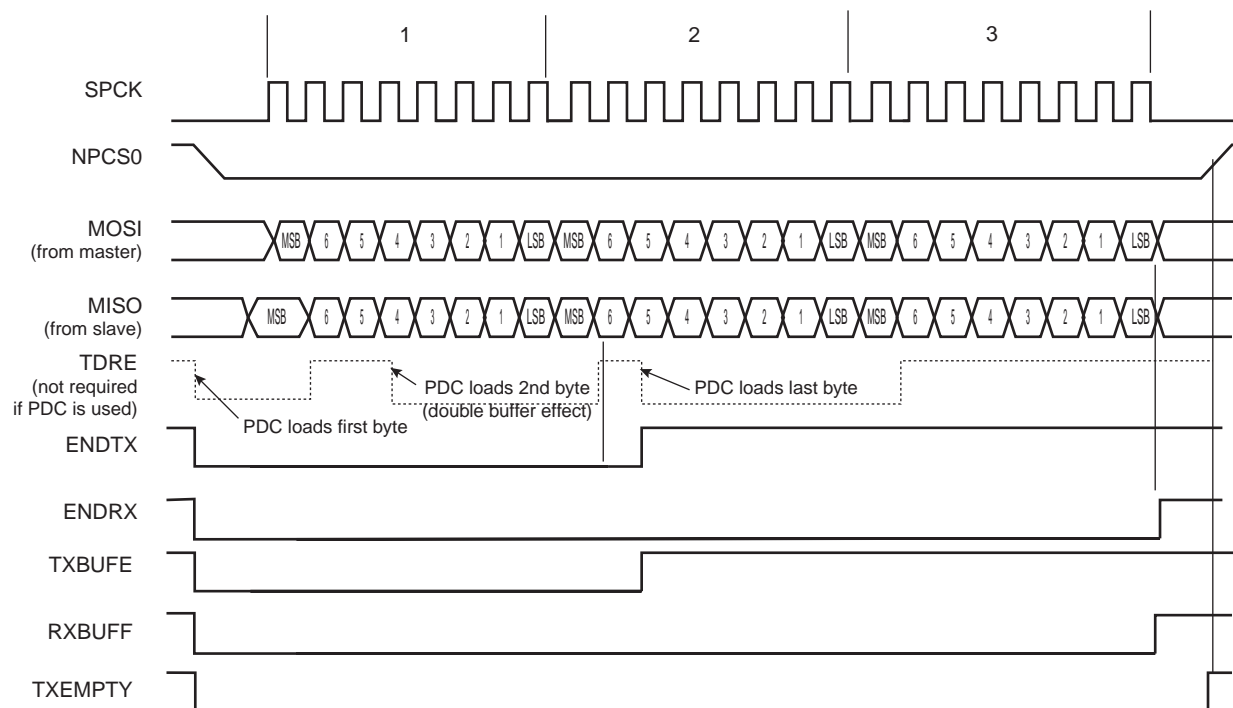


Figure 33-8 shows Transmission Register Empty (TXEMPTY), End of RX buffer (ENDRX), End of TX buffer (ENDTX), RX Buffer Full (RXBUFF) and TX Buffer Empty (TXBUFE) status flags behavior within the SPI\_SR (Status Register) during an 8-bit data transfer in fixed mode with the Peripheral Data Controller involved. The PDC is programmed to transfer and receive three data. The next pointer and counter are not used. The RDRF and TDRE are not shown because these flags are managed by the PDC when using the PDC.

**Figure 33-8. PDC Status Register Flags Behavior**



### 33.7.3.3 Clock Generation

The SPI Baud rate clock is generated by dividing the Master Clock (MCK), by a value between 1 and 255.

This allows a maximum operating baud rate at up to Master Clock and a minimum operating baud rate of MCK divided by 255.

Programming the SCBR field at 0 is forbidden. Triggering a transfer while SCBR is at 0 can lead to unpredictable results.

At reset, SCBR is 0 and the user has to program it at a valid value before performing the first transfer.

The divisor can be defined independently for each chip select, as it has to be programmed in the SCBR field of the Chip Select Registers. This allows the SPI to automatically adapt the baud rate for each interfaced peripheral without reprogramming.

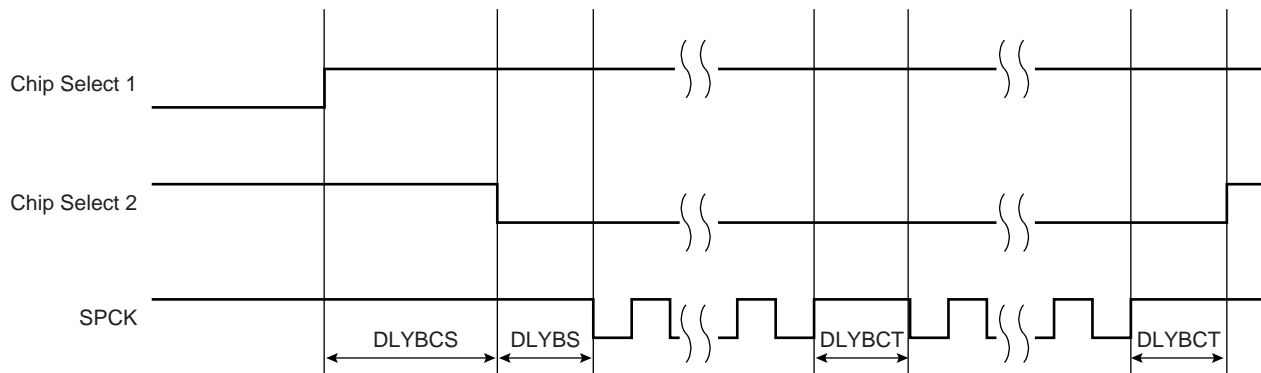
### 33.7.3.4 Transfer Delays

Figure 33-9 shows a chip select transfer change and consecutive transfers on the same chip select. Three delays can be programmed to modify the transfer waveforms:

- The delay between chip selects, programmable only once for all the chip selects by writing the DLYBCS field in the Mode Register. Allows insertion of a delay between release of one chip select and before assertion of a new one.
- The delay before SPCK, independently programmable for each chip select by writing the field DLYBS. Allows the start of SPCK to be delayed after the chip select has been asserted.
- The delay between consecutive transfers, independently programmable for each chip select by writing the DLYBCT field. Allows insertion of a delay between two transfers occurring on the same chip select

These delays allow the SPI to be adapted to the interfaced peripherals and their speed and bus release time.

**Figure 33-9. Programmable Delays**



### 33.7.3.5 Peripheral Selection

The serial peripherals are selected through the assertion of the NPCS0 to NPCS3 signals. By default, all the NPCS signals are high before and after each transfer.

- Fixed Peripheral Select: SPI exchanges data with only one peripheral

Fixed Peripheral Select is activated by writing the PS bit to zero in SPI\_MR (Mode Register). In this case, the current peripheral is defined by the PCS field in SPI\_MR and the PCS field in the SPI\_TDR has no effect.

- Variable Peripheral Select: Data can be exchanged with more than one peripheral without having to reprogram the NPCS field in the SPI\_MR register.

Variable Peripheral Select is activated by setting PS bit to one. The PCS field in SPI\_TDR is used to select the current peripheral. This means that the peripheral selection can be defined for each new data. The value to write in the SPI\_TDR register as the following format.

[xxxxxxx(7-bit) + LASTXFER(1-bit)<sup>(1)</sup> + xxxx(4-bit) + PCS (4-bit) + DATA (8 to 16-bit)] with PCS equals to the chip select to assert as defined in [Section 33.8.4](#) (SPI Transmit Data Register) and LASTXFER bit at 0 or 1 depending on CSAAT bit.

Note: 1. Optional.

CSAAT, LASTXFER and CSNAAT bits are discussed in [Section 33.7.3.9 "Peripheral Deselection with PDC"](#).

If LASTXFER is used, the command must be issued before writing the last character. Instead of LASTXFER, the user can use the SPIDIS command. After the end of the PDC transfer, wait for the TXEMPTY flag, then write SPIDIS into the SPI\_CR register (this will not change the configuration register values); the NPCS will be deactivated after the last character transfer. Then, another PDC transfer can be started if the SPIEN was previously written in the SPI\_CR register.

### 33.7.3.6 SPI Peripheral DMA Controller (PDC)

In both fixed and variable mode the Peripheral DMA Controller (PDC) can be used to reduce processor overhead.

The Fixed Peripheral Selection allows buffer transfers with a single peripheral. Using the PDC is an optimal means, as the size of the data transfer between the memory and the SPI is either 8 bits or 16 bits. However, changing the peripheral selection requires the Mode Register to be reprogrammed.

The Variable Peripheral Selection allows buffer transfers with multiple peripherals without reprogramming the Mode Register. Data written in SPI\_TDR is 32 bits wide and defines the real data to be transmitted and the peripheral it is destined to. Using the PDC in this mode requires 32-bit wide buffers, with the data in the LSBs and the PCS and LASTXFER fields in the MSBs, however the SPI still controls the number of bits (8 to 16) to be transferred through MISO and MOSI lines with the chip select configuration registers. This is not the optimal means in term of memory size for the buffers, but it provides a very effective means to exchange data with several peripherals without any intervention of the processor.



## Transfer Size

Depending on the data size to transmit, from 8 to 16 bits, the PDC manages automatically the type of pointer's size it has to point to. The PDC will perform the following transfer size depending on the mode and number of bits per data.

Fixed Mode:

- 8-bit Data:  
Byte transfer, PDC Pointer Address = Address + 1 byte,  
PDC Counter = Counter - 1
- 8-bit to 16-bit Data:  
2 bytes transfer. n-bit data transfer with don't care data (MSB) filled with 0's,  
PDC Pointer Address = Address + 2 bytes,  
PDC Counter = Counter - 1

Variable Mode:

In variable Mode, PDC Pointer Address = Address + 4 bytes and PDC Counter = Counter - 1 for 8 to 16-bit transfer size. When using the PDC, the TDRE and RDRF flags are handled by the PDC, thus the user's application does not have to check those bits. Only End of RX Buffer (ENDRX), End of TX Buffer (ENDTX), Buffer Full (RXBUFF), TX Buffer Empty (TXBUFE) are significant. For further details about the Peripheral DMA Controller and user interface, refer to the PDC section of the product datasheet.

### 33.7.3.7 Peripheral Chip Select Decoding

The user can program the SPI to operate with up to 15 peripherals by decoding the four Chip Select lines, NPCS0 to NPCS3 with 1 of up to 16 decoder/demultiplexer. This can be enabled by writing the PCSDEC bit at 1 in the Mode Register (SPI\_MR).

When operating without decoding, the SPI makes sure that in any case only one chip select line is activated, i.e., one NPCS line driven low at a time. If two bits are defined low in a PCS field, only the lowest numbered chip select is driven low.

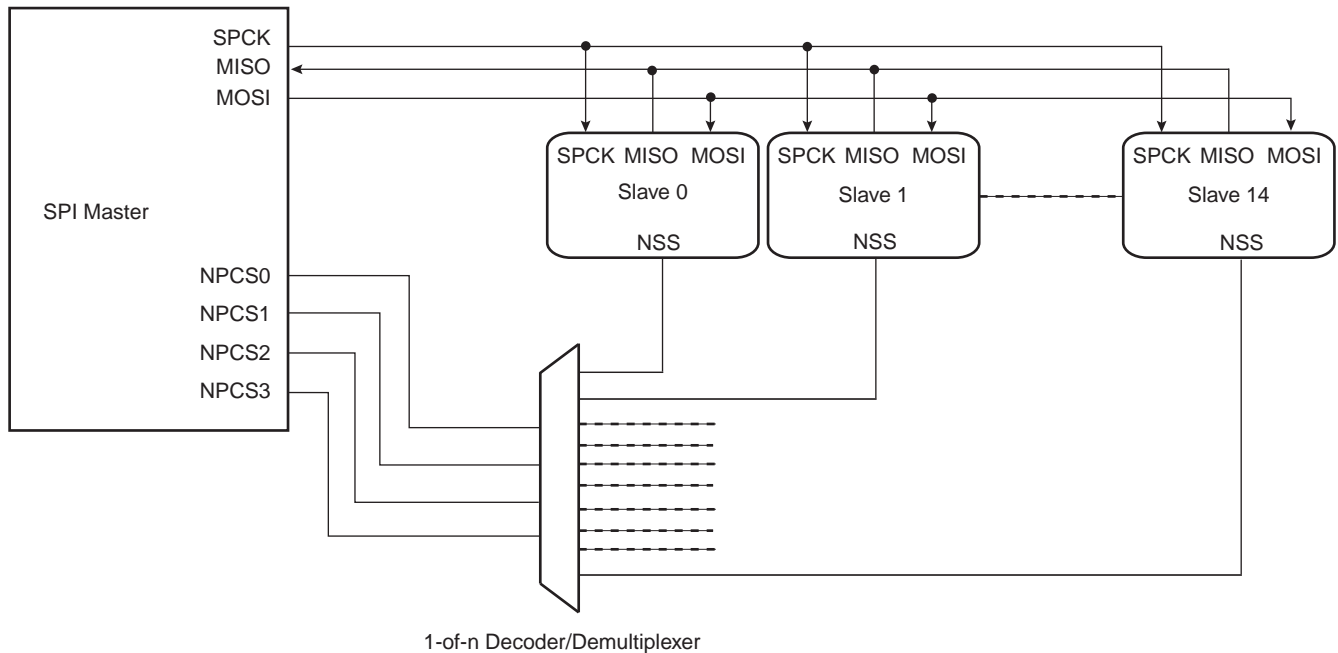
When operating with decoding, the SPI directly outputs the value defined by the PCS field on NPCS lines of either the Mode Register or the Transmit Data Register (depending on PS).

As the SPI sets a default value of 0xF on the chip select lines (i.e. all chip select lines at 1) when not processing any transfer, only 15 peripherals can be decoded.

The SPI has only four Chip Select Registers, not 15. As a result, when decoding is activated, each chip select defines the characteristics of up to four peripherals. As an example, SPI\_CR0 defines the characteristics of the externally decoded peripherals 0 to 3, corresponding to the PCS values 0x0 to 0x3. Thus, the user has to make sure to connect compatible peripherals on the decoded chip select lines 0 to 3, 4 to 7, 8 to 11 and 12 to 14. [Figure 33-10](#) below shows such an implementation.

If the CSAAT bit is used, with or without the PDC, the Mode Fault detection for NPCS0 line must be disabled. This is not needed for all other chip select lines since Mode Fault Detection is only on NPCS0.

**Figure 33-10. Chip Select Decoding Application Block Diagram: Single Master/Multiple Slave Implementation**



### 33.7.3.8 Peripheral Deselection without PDC

During a transfer of more than one data on a Chip Select without the PDC, the `SPI_TDR` is loaded by the processor, the flag `TDRE` rises as soon as the content of the `SPI_TDR` is transferred into the internal shift register. When this flag is detected high, the `SPI_TDR` can be reloaded. If this reload by the processor occurs before the end of the current transfer and if the next transfer is performed on the same chip select as the current transfer, the Chip Select is not de-asserted between the two transfers. But depending on the application software handling the SPI status register flags (by interrupt or polling method) or servicing other interrupts or other tasks, the processor may not reload the `SPI_TDR` in time to keep the chip select active (low). A null Delay Between Consecutive Transfer (`DLYBCT`) value in the `SPI_CSR` register, will give even less time for the processor to reload the `SPI_TDR`. With some SPI slave peripherals, requiring the chip select line to remain active (low) during a full set of transfers might lead to communication errors.

To facilitate interfacing with such devices, the Chip Select Register [`CSR0...CSR3`] can be programmed with the `CSAAT` bit (Chip Select Active After Transfer) at 1. This allows the chip select lines to remain in their current state (low = active) until transfer to another chip select is required. Even if the `SPI_TDR` is not reloaded the chip select will remain active. To have the chip select line to raise at the end of the transfer the Last transfer Bit (`LASTXFER`) in the `SPI_MR` register must be set at 1 before writing the last data to transmit into the `SPI_TDR`.

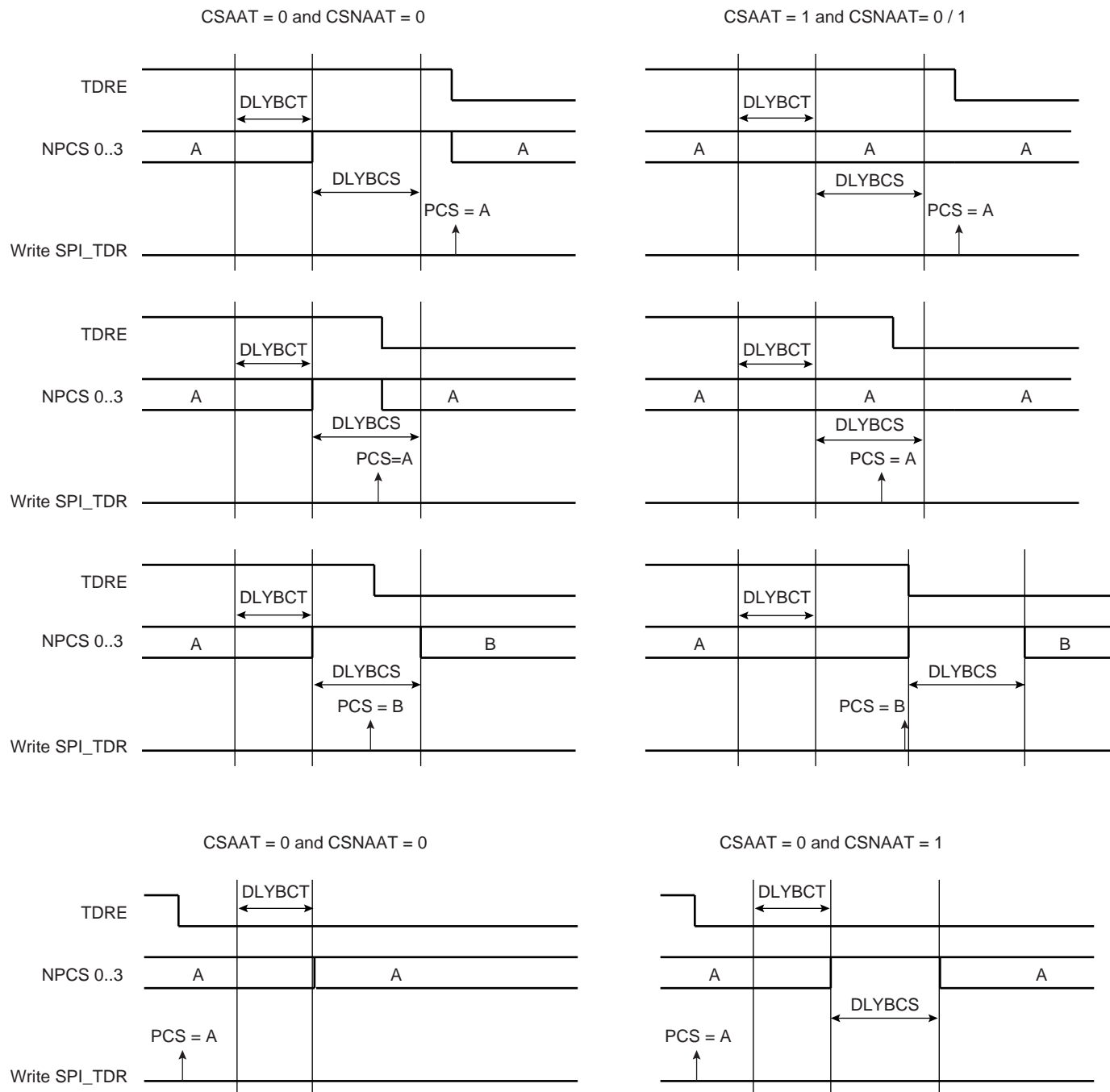
### 33.7.3.9 Peripheral Deselection with PDC

When the Peripheral DMA Controller is used, the chip select line will remain low during the whole transfer since the `TDRE` flag is managed by the PDC itself. The reloading of the `SPI_TDR` by the PDC is done as soon as `TDRE` flag is set to one. In this case the use of `CSAAT` bit might not be needed. However, it may happen that when other PDC channels connected to other peripherals are in use as well, the SPI PDC might be delayed by another (PDC with a higher priority on the bus). Having PDC buffers in slower memories like flash memory or SDRAM compared to fast internal SRAM, may lengthen the reload time of the `SPI_TDR` by the PDC as well. This means that the `SPI_TDR` might not be reloaded in time to keep the chip select line low. In this case the chip select line may toggle between data transfer and according to some SPI Slave devices, the communication might get lost. The use of the `CSAAT` bit might be needed. When the `CSAAT` bit is set at 0, the `NPCS` does not rise in all cases between two transfers on the same peripheral. During a transfer on a Chip Select, the flag `TDRE` rises as soon as the content of the `SPI_TDR` is transferred into the internal shifter. When this flag is detected the `SPI_TDR` can be reloaded. If this reload occurs before the end of the current transfer and if the next transfer is performed on the same chip select as the current transfer, the Chip Select is not de-asserted between the two transfers. This might lead to difficulties for interfacing with some serial peripherals requiring

the chip select to be de-asserted after each transfer. To facilitate interfacing with such devices, the Chip Select Register can be programmed with the CSNAAT bit (Chip Select Not Active After Transfer) at 1. This allows to de-assert systematically the chip select lines during a time DLYBCS. (The value of the CSAAT bit is taken into account only if the CSAAT bit is set at 0 for the same Chip Select).

Figure 33-11 shows different peripheral deselection cases and the effect of the CSAAT and CSNAAT bits.

**Figure 33-11. Peripheral Deselection**



### 33.7.3.10 Mode Fault Detection

A mode fault is detected when the SPI is programmed in Master Mode and a low level is driven by an external master on the NPCSS0/NSS signal. In this case, multi-master configuration, NPCSS0, MOSI, MISO and SPCK pins must be configured in open drain (through the PIO controller). When a mode fault is detected, the MODF bit in the SPI\_SR is set until the SPI\_SR is read and the SPI is automatically disabled until re-enabled by writing the SPIEN bit in the SPI\_CR (Control Register) at 1.

By default, the Mode Fault detection circuitry is enabled. The user can disable Mode Fault detection by setting the MODFDIS bit in the SPI Mode Register (SPI\_MR).

### 33.7.4 SPI Slave Mode

When operating in Slave Mode, the SPI processes data bits on the clock provided on the SPI clock pin (SPCK).

The SPI waits for NSS to go active before receiving the serial clock from an external master. When NSS falls, the clock is validated on the serializer, which processes the number of bits defined by the BITS field of the Chip Select Register 0 (SPI\_CSR0). These bits are processed following a phase and a polarity defined respectively by the NCPHA and CPOL bits of the SPI\_CSR0. Note that BITS, CPOL and NCPHA of the other Chip Select Registers have no effect when the SPI is programmed in Slave Mode.

The bits are shifted out on the MISO line and sampled on the MOSI line.

(For more information on BITS field, see also, the [\(Note:\)](#) below the register table; [Section 33.8.9 “SPI Chip Select Register” on page 641.](#))

When all the bits are processed, the received data is transferred in the Receive Data Register and the RDRF bit rises. If the SPI\_RDR (Receive Data Register) has not been read before new data is received, the Overrun Error bit (OVRES) in SPI\_SR is set. As long as this flag is set, data is loaded in SPI\_RDR. The user has to read the status register to clear the OVRES bit.

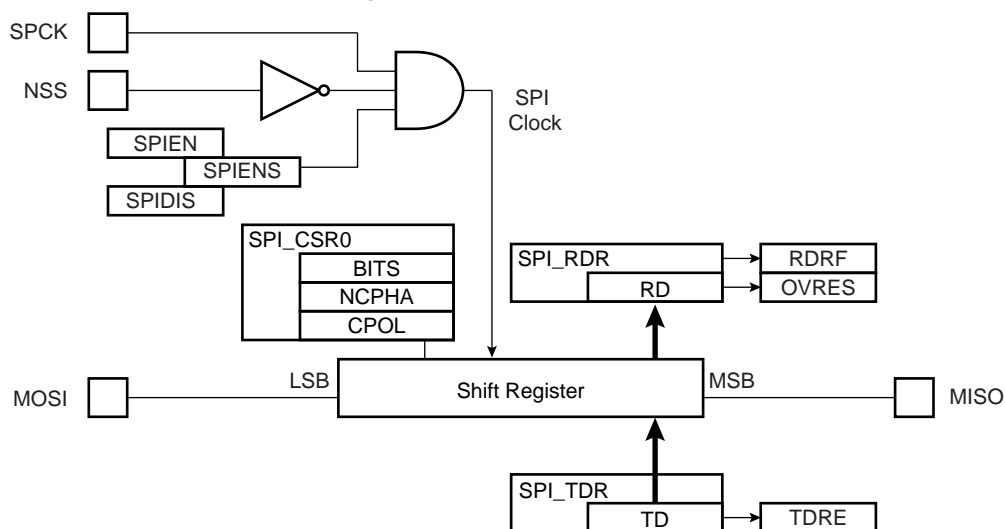
When a transfer starts, the data shifted out is the data present in the Shift Register. If no data has been written in the Transmit Data Register (SPI\_TDR), the last data received is transferred. If no data has been received since the last reset, all bits are transmitted low, as the Shift Register resets at 0.

When a first data is written in SPI\_TDR, it is transferred immediately in the Shift Register and the TDRE bit rises. If new data is written, it remains in SPI\_TDR until a transfer occurs, i.e. NSS falls and there is a valid clock on the SPCK pin. When the transfer occurs, the last data written in SPI\_TDR is transferred in the Shift Register and the TDRE bit rises. This enables frequent updates of critical variables with single transfers.

Then, a new data is loaded in the Shift Register from the Transmit Data Register. In case no character is ready to be transmitted, i.e. no character has been written in SPI\_TDR since the last load from SPI\_TDR to the Shift Register, the SPI\_TDR is retransmitted. In this case the Underrun Error Status Flag (UNDES) is set in the SPI\_SR.

Figure 33-12 shows a block diagram of the SPI when operating in Slave Mode.

**Figure 33-12. Slave Mode Functional Block Diagram**



### 33.7.5 Write Protected Registers

To prevent any single software error that may corrupt SPI behavior, the registers listed below can be write-protected by setting the WPEN bit in the SPI Write Protection Mode Register (SPI\_WPMR).

If a write access in a write-protected register is detected, then the WPVS flag in the SPI Write Protection Status Register (SPI\_WPSR) is set and the field WPVSRC indicates in which register the write access has been attempted.

The WPVS flag is automatically reset after reading the SPI Write Protection Status Register (SPI\_WPSR).

List of the write-protected registers:

- [Section 33.8.2 "SPI Mode Register"](#)
- [Section 33.8.9 "SPI Chip Select Register"](#)

## 33.8 Serial Peripheral Interface (SPI) User Interface

Table 33-5. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	SPI_CR	Write-only	---
0x04	Mode Register	SPI_MR	Read-write	0x0
0x08	Receive Data Register	SPI_RDR	Read-only	0x0
0x0C	Transmit Data Register	SPI_TDR	Write-only	---
0x10	Status Register	SPI_SR	Read-only	0x000000F0
0x14	Interrupt Enable Register	SPI_IER	Write-only	---
0x18	Interrupt Disable Register	SPI_IDR	Write-only	---
0x1C	Interrupt Mask Register	SPI_IMR	Read-only	0x0
0x20 - 0x2C	Reserved			
0x30	Chip Select Register 0	SPI_CSR0	Read-write	0x0
0x34	Chip Select Register 1	SPI_CSR1	Read-write	0x0
0x38	Chip Select Register 2	SPI_CSR2	Read-write	0x0
0x3C	Chip Select Register 3	SPI_CSR3	Read-write	0x0
0x40 - 0xE0	Reserved	–	–	–
0xE4	Write Protection Control Register	SPI_WPMR	Read-write	0x0
0xE8	Write Protection Status Register	SPI_WPSR	Read-only	0x0
0x00EC - 0x00F8	Reserved	–	–	–
0x00FC	Reserved	–	–	–
0x100 - 0x124	Reserved for PDC Registers	–	–	–

### 33.8.1 SPI Control Register

**Name:** SPI\_CR

**Address:** 0x40008000 (0), 0x48000000 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	LASTXFER
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
SWRST	–	–	–	–	–	SPIDIS	SPIEN

- **SPIEN: SPI Enable**

0 = No effect.

1 = Enables the SPI to transfer and receive data.

- **SPIDIS: SPI Disable**

0 = No effect.

1 = Disables the SPI.

As soon as SPIDIS is set, SPI finishes its transfer.

All pins are set in input mode and no data is received or transmitted.

If a transfer is in progress, the transfer is finished before the SPI is disabled.

If both SPIEN and SPIDIS are equal to one when the control register is written, the SPI is disabled.

- **SWRST: SPI Software Reset**

0 = No effect.

1 = Reset the SPI. A software-triggered hardware reset of the SPI interface is performed.

The SPI is in slave mode after software reset.

PDC channels are not affected by software reset.

- **LASTXFER: Last Transfer**

0 = No effect.

1 = The current NPCS will be deasserted after the character written in TD has been transferred. When CSAAT is set, this allows to close the communication with the current serial peripheral by raising the corresponding NPCS line as soon as TD transfer has completed.

Refer to [Section 33.7.3.5 "Peripheral Selection"](#) for more details.

### 33.8.2 SPI Mode Register

**Name:** SPI\_MR

**Address:** 0x40008004 (0), 0x48000004 (1)

**Access:** Read-write

31	30	29	28	27	26	25	24
DLYBCS							
23	22	21	20	19	18	17	16
–	–	–	–	PCS			
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
LLB	–	WDRBT	MODFDIS	–	PCSDEC	PS	MSTR

This register can only be written if the WPEN bit is cleared in "SPI Write Protection Mode Register".

- **MSTR: Master/Slave Mode**

0 = SPI is in Slave mode.

1 = SPI is in Master mode.

- **PS: Peripheral Select**

0 = Fixed Peripheral Select.

1 = Variable Peripheral Select.

- **PCSDEC: Chip Select Decode**

0 = The chip selects are directly connected to a peripheral device.

1 = The four chip select lines are connected to a 4- to 16-bit decoder.

When PCSDEC equals one, up to 15 Chip Select signals can be generated with the four lines using an external 4- to 16-bit decoder. The Chip Select Registers define the characteristics of the 15 chip selects according to the following rules:

SPI\_CSR0 defines peripheral chip select signals 0 to 3.

SPI\_CSR1 defines peripheral chip select signals 4 to 7.

SPI\_CSR2 defines peripheral chip select signals 8 to 11.

SPI\_CSR3 defines peripheral chip select signals 12 to 14.

- **MODFDIS: Mode Fault Detection**

0 = Mode fault detection is enabled.

1 = Mode fault detection is disabled.

- **WDRBT: Wait Data Read Before Transfer**

0 = No Effect. In master mode, a transfer can be initiated whatever the state of the Receive Data Register is.

1 = In Master Mode, a transfer can start only if the Receive Data Register is empty, i.e. does not contain any unread data. This mode prevents overrun error in reception.

- **LLB: Local Loopback Enable**

0 = Local loopback path disabled.

1 = Local loopback path enabled

LLB controls the local loopback on the data serializer for testing in Master Mode only. (MISO is internally connected on MOSI.)



- **PCS: Peripheral Chip Select**

This field is only used if Fixed Peripheral Select is active (PS = 0).

If PCSDEC = 0:

PCS = xxx0	NPCS[3:0] = 1110
PCS = xx01	NPCS[3:0] = 1101
PCS = x011	NPCS[3:0] = 1011
PCS = 0111	NPCS[3:0] = 0111
PCS = 1111	forbidden (no peripheral is selected)
(x = don't care)	

If PCSDEC = 1:

NPCS[3:0] output signals = PCS.

- **DLYBCS: Delay Between Chip Selects**

This field defines the delay from NPCS inactive to the activation of another NPCS. The DLYBCS time guarantees non-overlapping chip selects and solves bus contentions in case of peripherals having long data float times.

If DLYBCS is less than or equal to six, six MCK periods will be inserted by default.

Otherwise, the following equation determines the delay:

$$\text{Delay Between Chip Selects} = \frac{DLYBCS}{MCK}$$

### 33.8.3 SPI Receive Data Register

**Name:** SPI\_RDR

**Address:** 0x40008008 (0), 0x48000008 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	PCS			
15	14	13	12	11	10	9	8
RD							
7	6	5	4	3	2	1	0
RD							

- **RD: Receive Data**

Data received by the SPI Interface is stored in this register right-justified. Unused bits read zero.

- **PCS: Peripheral Chip Select**

In Master Mode only, these bits indicate the value on the NPCS pins at the end of a transfer. Otherwise, these bits read zero.

**Note:** When using variable peripheral select mode (PS = 1 in SPI\_MR) it is mandatory to also set the WDRBT field to 1 if the SPI\_RDR PCS field is to be processed.

### 33.8.4 SPI Transmit Data Register

**Name:** SPI\_TDR

**Address:** 0x4000800C (0), 0x4800000C (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	LASTXFER
23	22	21	20	19	18	17	16
–	–	–	–	PCS			
15	14	13	12	11	10	9	8
TD							
7	6	5	4	3	2	1	0
TD							

- **TD: Transmit Data**

Data to be transmitted by the SPI Interface is stored in this register. Information to be transmitted must be written to the transmit data register in a right-justified format.

- **PCS: Peripheral Chip Select**

This field is only used if Variable Peripheral Select is active (PS = 1).

If PCSDEC = 0:

PCS = xxx0	NPCS[3:0] = 1110
PCS = xx01	NPCS[3:0] = 1101
PCS = x011	NPCS[3:0] = 1011
PCS = 0111	NPCS[3:0] = 0111
PCS = 1111	forbidden (no peripheral is selected)
(x = don't care)	

If PCSDEC = 1:

NPCS[3:0] output signals = PCS

- **LASTXFER: Last Transfer**

0 = No effect.

1 = The current NPCS will be deasserted after the character written in TD has been transferred. When CSAAT is set, this allows to close the communication with the current serial peripheral by raising the corresponding NPCS line as soon as TD transfer has completed.

This field is only used if Variable Peripheral Select is active (PS = 1).

### 33.8.5 SPI Status Register

**Name:** SPI\_SR

**Address:** 0x40008010 (0), 0x48000010 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	SPIENS
15	14	13	12	11	10	9	8
—	—	—	—	—	UNDES	TXEMPTY	NSSR
7	6	5	4	3	2	1	0
TXBUFE	RXBUFF	ENDTX	ENDRX	OVRES	MODF	TDRE	RDRF

- **RDRF: Receive Data Register Full**

0 = No data has been received since the last read of SPI\_RDR

1 = Data has been received and the received data has been transferred from the serializer to SPI\_RDR since the last read of SPI\_RDR.

- **TDRE: Transmit Data Register Empty**

0 = Data has been written to SPI\_TDR and not yet transferred to the serializer.

1 = The last data written in the Transmit Data Register has been transferred to the serializer.

TDRE equals zero when the SPI is disabled or at reset. The SPI enable command sets this bit to one.

- **MODF: Mode Fault Error**

0 = No Mode Fault has been detected since the last read of SPI\_SR.

1 = A Mode Fault occurred since the last read of the SPI\_SR.

- **OVRES: Overrun Error Status**

0 = No overrun has been detected since the last read of SPI\_SR.

1 = An overrun has occurred since the last read of SPI\_SR.

An overrun occurs when SPI\_RDR is loaded at least twice from the serializer since the last read of the SPI\_RDR.

- **ENDRX: End of RX buffer**

0 = The Receive Counter Register has not reached 0 since the last write in SPI\_RCR<sup>(1)</sup> or SPI\_RNCR<sup>(1)</sup>.

1 = The Receive Counter Register has reached 0 since the last write in SPI\_RCR<sup>(1)</sup> or SPI\_RNCR<sup>(1)</sup>.

- **ENDTX: End of TX buffer**

0 = The Transmit Counter Register has not reached 0 since the last write in SPI\_TCR<sup>(1)</sup> or SPI\_TNCR<sup>(1)</sup>.

1 = The Transmit Counter Register has reached 0 since the last write in SPI\_TCR<sup>(1)</sup> or SPI\_TNCR<sup>(1)</sup>.

- **RXBUFF: RX Buffer Full**

0 = SPI\_RCR<sup>(1)</sup> or SPI\_RNCR<sup>(1)</sup> has a value other than 0.

1 = Both SPI\_RCR<sup>(1)</sup> and SPI\_RNCR<sup>(1)</sup> have a value of 0.

- **TXBUFE: TX Buffer Empty**

0 = SPI\_TCR<sup>(1)</sup> or SPI\_TNCR<sup>(1)</sup> has a value other than 0.

1 = Both SPI\_TCR<sup>(1)</sup> and SPI\_TNCR<sup>(1)</sup> have a value of 0.

- **NSSR: NSS Rising**

0 = No rising edge detected on NSS pin since last read.

1 = A rising edge occurred on NSS pin since last read.

- **TXEMPTY: Transmission Registers Empty**

0 = As soon as data is written in SPI\_TDR.

1 = SPI\_TDR and internal shifter are empty. If a transfer delay has been defined, TXEMPTY is set after the completion of such delay.

- **UNDES: Underrun Error Status (Slave Mode Only)**

0 = No underrun has been detected since the last read of SPI\_SR.

1 = A transfer begins whereas no data has been loaded in the Transmit Data Register.

- **SPIENS: SPI Enable Status**

0 = SPI is disabled.

1 = SPI is enabled.

Note: 1. SPI\_RCR, SPI\_RNCR, SPI\_TCR, SPI\_TNCR are physically located in the PDC.

### 33.8.6 SPI Interrupt Enable Register

**Name:** SPI\_IER

**Address:** 0x40008014 (0), 0x48000014 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	UNDES	TXEMPTY	NSSR
7	6	5	4	3	2	1	0
TXBUFE	RXBUFF	ENDTX	ENDRX	OVRES	MODF	TDRE	RDRF

0 = No effect.

1 = Enables the corresponding interrupt.

- **RDRF: Receive Data Register Full Interrupt Enable**
- **TDRE: SPI Transmit Data Register Empty Interrupt Enable**
- **MODF: Mode Fault Error Interrupt Enable**
- **OVRES: Overrun Error Interrupt Enable**
- **ENDRX: End of Receive Buffer Interrupt Enable**
- **ENDTX: End of Transmit Buffer Interrupt Enable**
- **RXBUFF: Receive Buffer Full Interrupt Enable**
- **TXBUFE: Transmit Buffer Empty Interrupt Enable**
- **NSSR: NSS Rising Interrupt Enable**
- **TXEMPTY: Transmission Registers Empty Enable**
- **UNDES: Underrun Error Interrupt Enable**

### 33.8.7 SPI Interrupt Disable Register

**Name:** SPI\_IDR

**Address:** 0x40008018 (0), 0x48000018 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	UNDES	TXEMPTY	NSSR
7	6	5	4	3	2	1	0
TXBUFE	RXBUFF	ENDTX	ENDRX	OVRES	MODF	TDRE	RDRF

0 = No effect.

1 = Disables the corresponding interrupt.

- **RDRF: Receive Data Register Full Interrupt Disable**
- **TDRE: SPI Transmit Data Register Empty Interrupt Disable**
- **MODF: Mode Fault Error Interrupt Disable**
- **OVRES: Overrun Error Interrupt Disable**
- **ENDRX: End of Receive Buffer Interrupt Disable**
- **ENDTX: End of Transmit Buffer Interrupt Disable**
- **RXBUFF: Receive Buffer Full Interrupt Disable**
- **TXBUFE: Transmit Buffer Empty Interrupt Disable**
- **NSSR: NSS Rising Interrupt Disable**
- **TXEMPTY: Transmission Registers Empty Disable**
- **UNDES: Underrun Error Interrupt Disable**

### 33.8.8 SPI Interrupt Mask Register

**Name:** SPI\_IMR

**Address:** 0x4000801C (0), 0x4800001C (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	UNDES	TXEMPTY	NSSR
7	6	5	4	3	2	1	0
TXBUFE	RXBUFF	ENDTX	ENDRX	OVRES	MODF	TDRE	RDRF

0 = The corresponding interrupt is not enabled.

1 = The corresponding interrupt is enabled.

- **RDRF: Receive Data Register Full Interrupt Mask**
- **TDRE: SPI Transmit Data Register Empty Interrupt Mask**
- **MODF: Mode Fault Error Interrupt Mask**
- **OVRES: Overrun Error Interrupt Mask**
- **ENDRX: End of Receive Buffer Interrupt Mask**
- **ENDTX: End of Transmit Buffer Interrupt Mask**
- **RXBUFF: Receive Buffer Full Interrupt Mask**
- **TXBUFE: Transmit Buffer Empty Interrupt Mask**
- **NSSR: NSS Rising Interrupt Mask**
- **TXEMPTY: Transmission Registers Empty Mask**
- **UNDES: Underrun Error Interrupt Mask**



### 33.8.9 SPI Chip Select Register

**Name:** SPI\_CSRx[x=0..3]  
**Address:** 0x40008030 (0), 0x48000030 (1)  
**Access:** Read/Write

31	30	29	28	27	26	25	24
DLYBCT							
23	22	21	20	19	18	17	16
DLYBS							
15	14	13	12	11	10	9	8
SCBR							
7	6	5	4	3	2	1	0
BITS				CSAAT	CSNAAT	NCPHA	CPOL

This register can only be written if the WPEN bit is cleared in "SPI Write Protection Mode Register".

**Note:** SPI\_CSRx registers must be written even if the user wants to use the defaults. The BITS field will not be updated with the translated value unless the register is written.

- **CPOL: Clock Polarity**

0 = The inactive state value of SPCK is logic level zero.

1 = The inactive state value of SPCK is logic level one.

CPOL is used to determine the inactive state value of the serial clock (SPCK). It is used with NCPHA to produce the required clock/data relationship between master and slave devices.

- **NCPHA: Clock Phase**

0 = Data is changed on the leading edge of SPCK and captured on the following edge of SPCK.

1 = Data is captured on the leading edge of SPCK and changed on the following edge of SPCK.

NCPHA determines which edge of SPCK causes data to change and which edge causes data to be captured. NCPHA is used with CPOL to produce the required clock/data relationship between master and slave devices.

- **CSNAAT: Chip Select Not Active After Transfer (Ignored if CSAAT = 1)**

0 = The Peripheral Chip Select does not rise between two transfers if the SPI\_TDR is reloaded before the end of the first transfer and if the two transfers occur on the same Chip Select.

1 = The Peripheral Chip Select rises systematically after each transfer performed on the same slave. It remains active after the end of transfer for a minimal duration of:

$$- \frac{DLYBCT}{MCK} \text{ (if DLYBCT field is different from 0)}$$

$$- \frac{DLYBCT + 1}{MCK} \text{ (if DLYBCT field equals 0)}$$

- **CSAAT: Chip Select Active After Transfer**

0 = The Peripheral Chip Select Line rises as soon as the last transfer is achieved.

1 = The Peripheral Chip Select does not rise after the last transfer is achieved. It remains active until a new transfer is requested on a different chip select.

- **BITS: Bits Per Transfer**

(See the (Note:) below the register table; [Section 33.8.9 "SPI Chip Select Register" on page 641.](#))

The BITS field determines the number of data bits transferred. Reserved values should not be used.

Value	Name	Description
0	8_BIT	8 bits for transfer
1	9_BIT	9 bits for transfer
2	10_BIT	10 bits for transfer
3	11_BIT	11 bits for transfer
4	12_BIT	12 bits for transfer
5	13_BIT	13 bits for transfer
6	14_BIT	14 bits for transfer
7	15_BIT	15 bits for transfer
8	16_BIT	16 bits for transfer
9	–	Reserved
10	–	Reserved
11	–	Reserved
12	–	Reserved
13	–	Reserved
14	–	Reserved
15	–	Reserved

#### • SCBR: Serial Clock Baud Rate

In Master Mode, the SPI Interface uses a modulus counter to derive the SPCK baud rate from the Master Clock MCK. The Baud rate is selected by writing a value from 1 to 255 in the SCBR field. The following equations determine the SPCK baud rate:

$$\text{SPCK Baudrate} = \frac{MCK}{SCBR}$$

Programming the SCBR field at 0 is forbidden. Triggering a transfer while SCBR is at 0 can lead to unpredictable results.

At reset, SCBR is 0 and the user has to program it at a valid value before performing the first transfer.

Note: If one of the SCBR fields in SPI\_CSRx is set to 1, the other SCBR fields in SPI\_CSRx must be set to 1 as well, if they are required to process transfers. If they are not used to transfer data, they can be set at any value.

#### • DLYBS: Delay Before SPCK

This field defines the delay from NPCS valid to the first valid SPCK transition.

When DLYBS equals zero, the NPCS valid to SPCK transition is 1/2 the SPCK clock period.

Otherwise, the following equations determine the delay:

$$\text{Delay Before SPCK} = \frac{DLYBS}{MCK}$$

#### • DLYBCT: Delay Between Consecutive Transfers

This field defines the delay between two consecutive transfers with the same peripheral without removing the chip select. The delay is always inserted after each transfer and before removing the chip select if needed.

When DLYBCT equals zero, no delay between consecutive transfers is inserted and the clock keeps its duty cycle over the character transfers.

Otherwise, the following equation determines the delay:

$$\text{Delay Between Consecutive Transfers} = \frac{32 \times DLYBCT}{MCK}$$

### 33.8.10 SPI Write Protection Mode Register

**Name:** SPI\_WPMR

**Address:** 0x400080E4 (0), 0x480000E4 (1)

**Access:** Read-write

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

- **WPEN: Write Protect Enable**

0: Disables the Write Protect if WPKEY corresponds to 0x535049 ("SPI" in ASCII).

1: Enables the Write Protect if WPKEY corresponds to 0x535049 ("SPI" in ASCII).

Protects the registers:

- [Section 33.8.2 "SPI Mode Register"](#)
- [Section 33.8.9 "SPI Chip Select Register"](#)

- **WPKEY: Write Protect Key**

Value	Name	Description
0x535049	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

### 33.8.11 SPI Write Protection Status Register

**Name:** SPI\_WPSR

**Address:** 0x400080E8 (0), 0x480000E8 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

- **WPVS: Write Protection Violation Status**

0 = No Write Protect Violation has occurred since the last read of the SPI\_WPSR register.

1 = A Write Protect Violation has occurred since the last read of the SPI\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protection Violation Source**

This Field indicates the APB Offset of the register concerned by the violation (SPI\_MR or SPI\_CSRx)

## 34. Two-wire Interface (TWI)

### 34.1 Description

The Atmel Two-wire Interface (TWI) interconnects components on a unique two-wire bus, made up of one clock line and one data line with speeds of up to 400 Kbits per second, based on a byte-oriented transfer format. It can be used with any Atmel Two-wire Interface bus Serial EEPROM and I<sup>2</sup>C compatible device such as Real Time Clock (RTC), Dot Matrix/Graphic LCD Controllers and Temperature Sensor, to name but a few. The TWI is programmable as a master or a slave with sequential or single-byte access. Multiple master capability is supported.

Arbitration of the bus is performed internally and puts the TWI in slave mode automatically if the bus arbitration is lost.

A configurable baud rate generator permits the output data rate to be adapted to a wide range of core clock frequencies.

Table 34-1 lists the compatibility level of the Atmel Two-wire Interface in Master Mode and a full I<sup>2</sup>C compatible device.

**Table 34-1. Atmel TWI compatibility with I<sup>2</sup>C Standard**

I <sup>2</sup> C Standard	Atmel TWI
Standard Mode Speed (100 kHz)	Supported
Fast Mode Speed (400 kHz)	Supported
7 or 10 bits Slave Addressing	Supported
START BYTE <sup>(1)</sup>	Not Supported
Repeated Start (Sr) Condition	Supported
ACK and NACK Management	Supported
Slope control and input filtering (Fast mode)	Not Supported
Clock stretching	Supported
Multi Master Capability	Supported

Note: 1. START + b000000001 + Ack + Sr

### 34.2 Embedded Characteristics

- 2 TWIs
- Compatible with Atmel Two-wire Interface Serial Memory and I<sup>2</sup>C Compatible Devices<sup>(1)</sup>
- One, Two or Three Bytes for Slave Address
- Sequential Read-write Operations
- Master, Multi-master and Slave Mode Operation
- Bit Rate: Up to 400 Kbit/s
- General Call Supported in Slave mode
- SMBUS Quick Command Supported in Master Mode
- Connection to Peripheral DMA Controller (PDC) Channel Capabilities Optimizes Data Transfers
  - One Channel for the Receiver, One Channel for the Transmitter

Note: 1. See Table 34-1 for details on compatibility with I<sup>2</sup>C Standard.

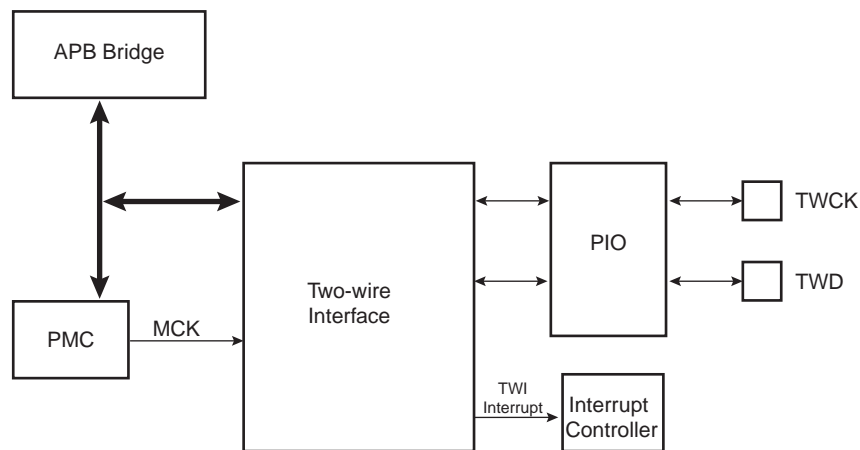
### 34.3 List of Abbreviations

Table 34-2. Abbreviations

Abbreviation	Description
TWI	Two-wire Interface
A	Acknowledge
NA	Non Acknowledge
P	Stop
S	Start
Sr	Repeated Start
SADR	Slave Address
ADR	Any address except SADR
R	Read
W	Write

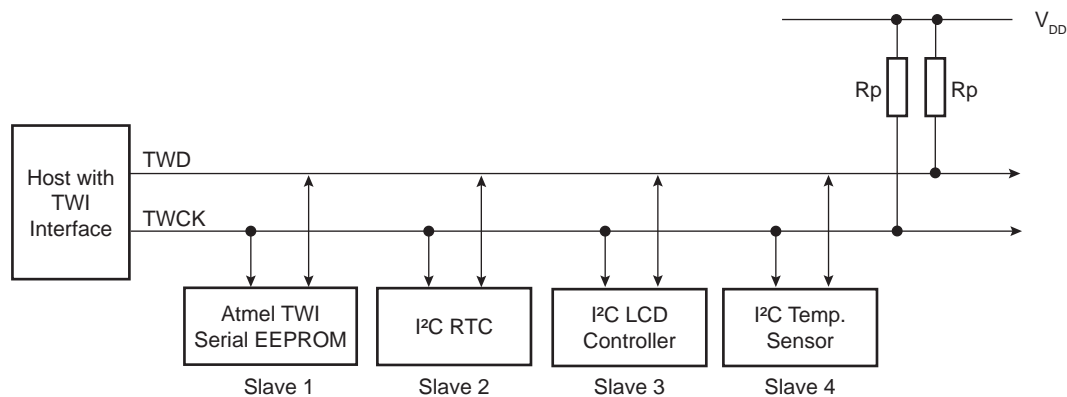
### 34.4 Block Diagram

Figure 34-1. Block Diagram



### 34.5 Application Block Diagram

Figure 34-2. Application Block Diagram



Rp: Pull-up value as given by the I²C Standard

#### 34.5.1 I/O Lines Description

Table 34-3. I/O Lines Description

Pin Name	Pin Description	Type
TWD	Two-wire Serial Data	Input/Output
TWCK	Two-wire Serial Clock	Input/Output

### 34.6 Product Dependencies

#### 34.6.1 I/O Lines

Both TWD and TWCK are bidirectional lines, connected to a positive supply voltage via a current source or pull-up resistor (see Figure 34-2). When the bus is free, both lines are high. The output stages of devices connected to the bus must have an open-drain or open-collector to perform the wired-AND function.

TWD and TWCK pins may be multiplexed with PIO lines. To enable the TWI, the programmer must perform the following step:

- Program the PIO controller to dedicate TWD and TWCK as peripheral lines.

The user must not program TWD and TWCK as open-drain. It is already done by the hardware.

Table 34-4. I/O Lines

Instance	Signal	I/O Line	Peripheral
TWI0	TWCK0	PA25	A
TWI0	TWD0	PA24	A
TWI1	TWCK1	PB1	A
TWI1	TWD1	PB0	A

#### 34.6.2 Power Management

The TWI interface may be clocked through the Power Management Controller (PMC), thus the programmer must first configure the PMC to enable the TWI clock.

### 34.6.3 Interrupt

The TWI interface has an interrupt line connected to the Interrupt Controller. In order to handle interrupts, the Interrupt Controller must be programmed before configuring the TWI.

**Table 34-5. Peripheral IDs**

Instance	ID
TWI0	19
TWI1	20

## 34.7 Functional Description

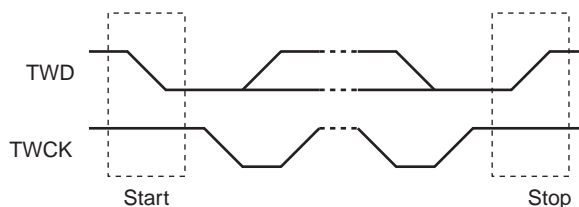
### 34.7.1 Transfer Format

The data put on the TWD line must be 8 bits long. Data is transferred MSB first; each byte must be followed by an acknowledgement. The number of bytes per transfer is unlimited (see [Figure 34-4](#)).

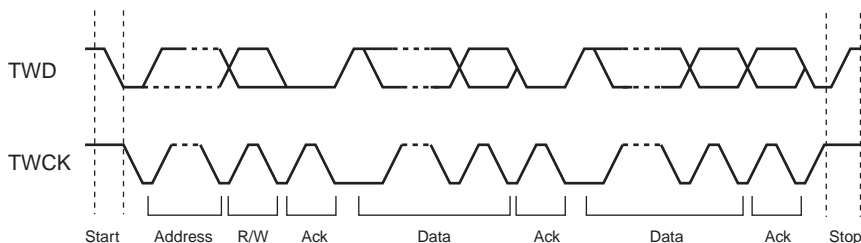
Each transfer begins with a START condition and terminates with a STOP condition (see [Figure 34-3](#)).

- A high-to-low transition on the TWD line while TWCK is high defines the START condition.
- A low-to-high transition on the TWD line while TWCK is high defines a STOP condition.

**Figure 34-3. START and STOP Conditions**



**Figure 34-4. Transfer Format**



### 34.7.2 Modes of Operation

The TWI has different modes of operations:

- Master transmitter mode
- Master receiver mode
- Multi-master transmitter mode
- Multi-master receiver mode
- Slave transmitter mode
- Slave receiver mode

These modes are described in the following sections.



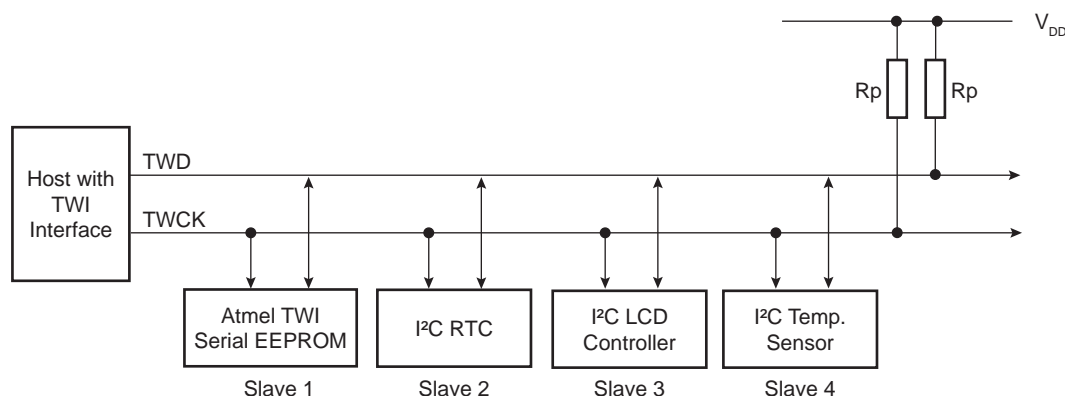
## 34.8 Master Mode

### 34.8.1 Definition

The Master is the device that starts a transfer, generates a clock and stops it.

### 34.8.2 Application Block Diagram

Figure 34-5. Master Mode Typical Application Block Diagram



Rp: Pull-up value as given by the I²C Standard

### 34.8.3 Programming Master Mode

The following registers have to be programmed before entering Master mode:

1. DADR (+ IADRSZ + IADR if a 10 bit device is addressed): The device address is used to access slave devices in read or write mode.
2. CKDIV + CHDIV + CLDIV: Clock Waveform.
3. SVDIS: Disable the slave mode.
4. MSEN: Enable the master mode.

### 34.8.4 Master Transmitter Mode

After the master initiates a Start condition when writing into the Transmit Holding Register, TWI\_THR, it sends a 7-bit slave address, configured in the Master Mode register (DADR in TWI\_MMR), to notify the slave device. The bit following the slave address indicates the transfer direction, 0 in this case (MREAD = 0 in TWI\_MMR).

The TWI transfers require the slave to acknowledge each received byte. During the acknowledge clock pulse (9th pulse), the master releases the data line (HIGH), enabling the slave to pull it down in order to generate the acknowledge. The master polls the data line during this clock pulse and sets the Not Acknowledge bit (NACK) in the status register if the slave does not acknowledge the byte. As with the other status bits, an interrupt can be generated if enabled in the interrupt enable register (TWI\_IER). If the slave acknowledges the byte, the data written in the TWI\_THR, is then shifted in the internal shifter and transferred. When an acknowledge is detected, the TXRDY bit is set until a new write in the TWI\_THR.

TXRDY is used as Transmit Ready for the PDC transmit channel.

While no new data is written in the TWI\_THR, the Serial Clock Line is tied low. When new data is written in the TWI\_THR, the SCL is released and the data is sent. To generate a STOP event, the STOP command must be performed by writing in the STOP field of TWI\_CR.

After a Master Write transfer, the Serial Clock line is stretched (tied low) while no new data is written in the TWI\_THR or until a STOP command is performed.

See [Figure 34-6](#), [Figure 34-7](#), and [Figure 34-8](#).

Figure 34-6. Master Write with One Data Byte

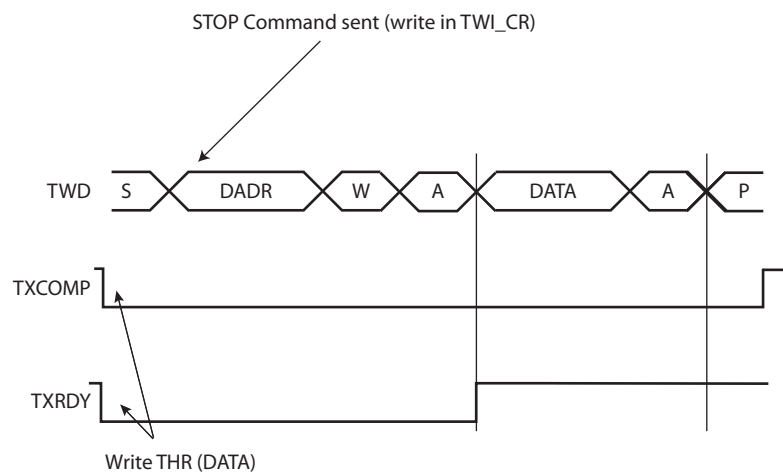
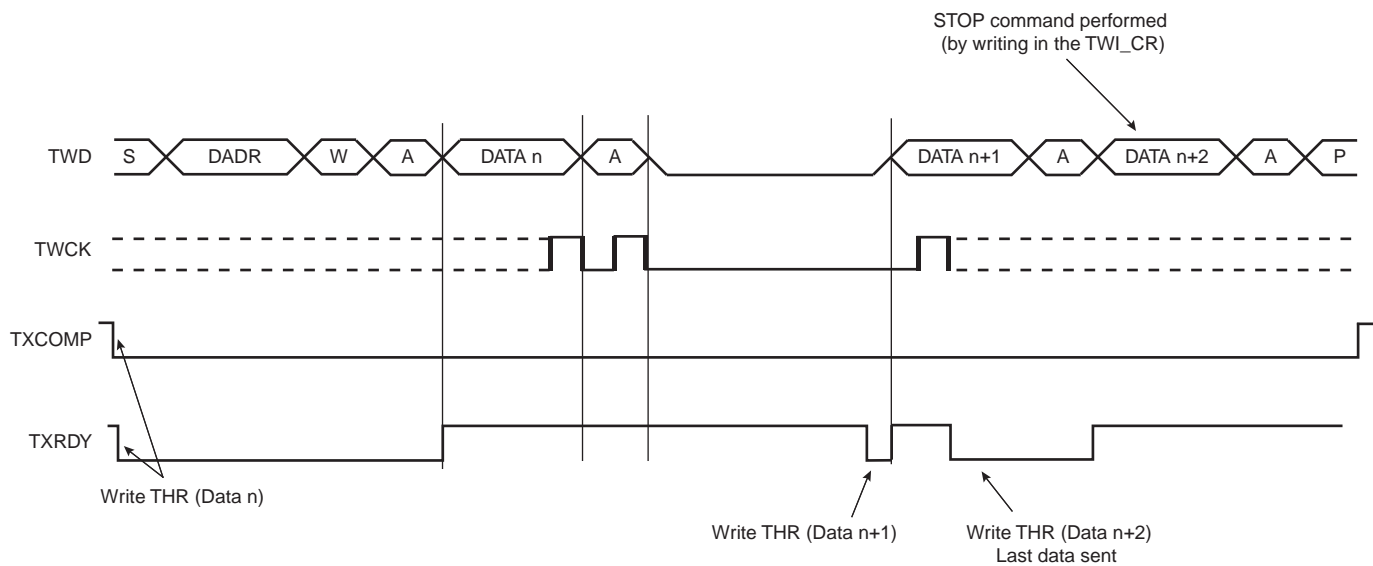
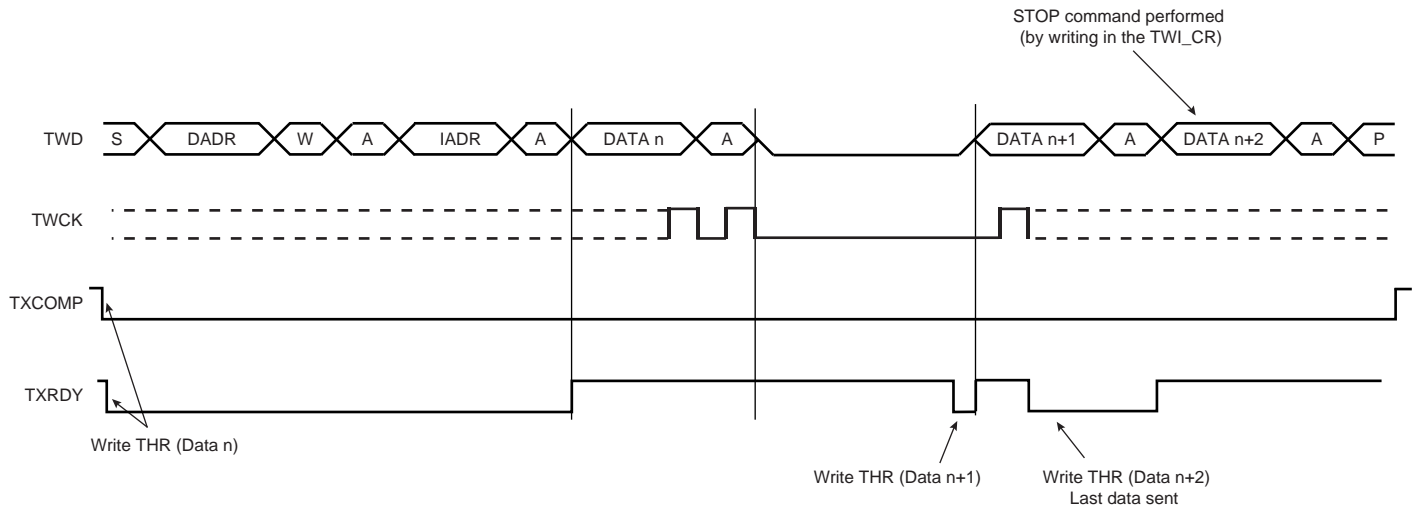


Figure 34-7. Master Write with Multiple Data Bytes



**Figure 34-8. Master Write with One Byte Internal Address and Multiple Data Bytes**



### 34.8.5 Master Receiver Mode

The read sequence begins by setting the START bit. After the start condition has been sent, the master sends a 7-bit slave address to notify the slave device. The bit following the slave address indicates the transfer direction, 1 in this case (MREAD = 1 in TWI\_MMR). During the acknowledge clock pulse (9th pulse), the master releases the data line (HIGH), enabling the slave to pull it down in order to generate the acknowledge. The master polls the data line during this clock pulse and sets the NACK bit in the status register if the slave does not acknowledge the byte.

If an acknowledge is received, the master is then ready to receive data from the slave. After data has been received, the master sends an acknowledge condition to notify the slave that the data has been received except for the last data, after the stop condition. See [Figure 34-9](#). When the RXRDY bit is set in the status register, a character has been received in the receive-holding register (TWI\_RHR). The RXRDY bit is reset when reading the TWI\_RHR.

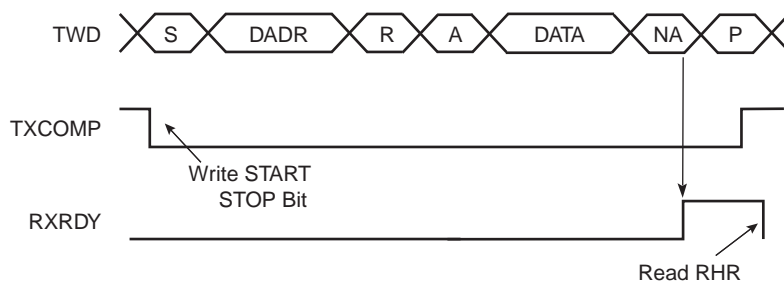
When a single data byte read is performed, with or without internal address (IADR), the START and STOP bits must be set at the same time. See [Figure 34-9](#). When a multiple data byte read is performed, with or without internal address (IADR), the STOP bit must be set after the next-to-last data received. See [Figure 34-10](#). For Internal Address usage see [Section 34.8.6](#).

If the receive holding register (TWI\_RHR) is full (RXRDY high) and the master is receiving data, the Serial Clock Line will be tied low before receiving the last bit of the data and until the TWI\_RHR is read. Once the TWI\_RHR is read, the master will stop stretching the Serial Clock Line and end the data reception. See [Figure 34-11](#).

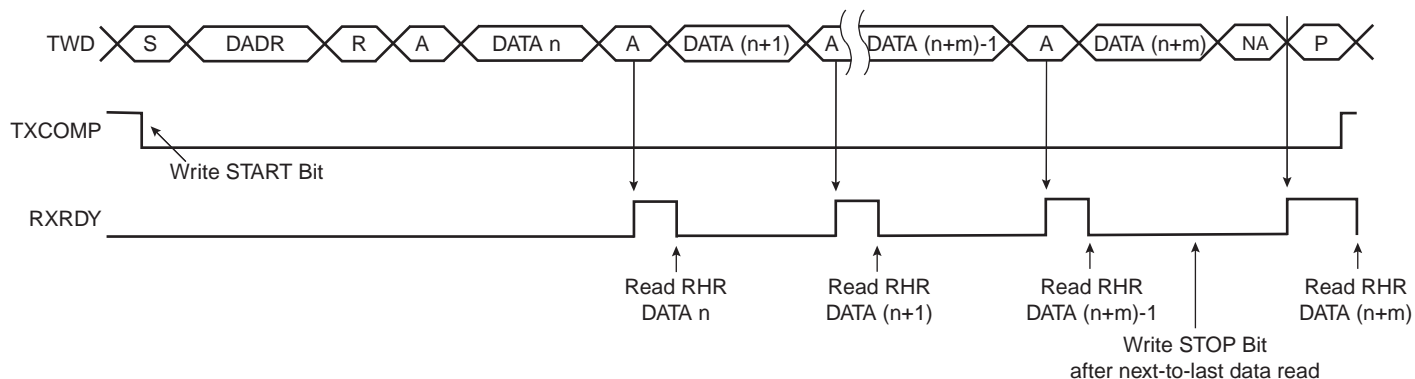
**Warning:** When receiving multiple bytes in master read mode, if the next-to-last access is not read (the RXRDY flag remains high), the last access will not be completed until TWI\_RHR is read. The last access stops on the next-to-last bit (clock stretching). When the TWI\_RHR is read there is only half a bit period to send the stop bit command, else another read access might occur (spurious access).

A possible workaround is to raise the STOP BIT command before reading the TWI\_RHR on the next-to-last access (within IT handler).

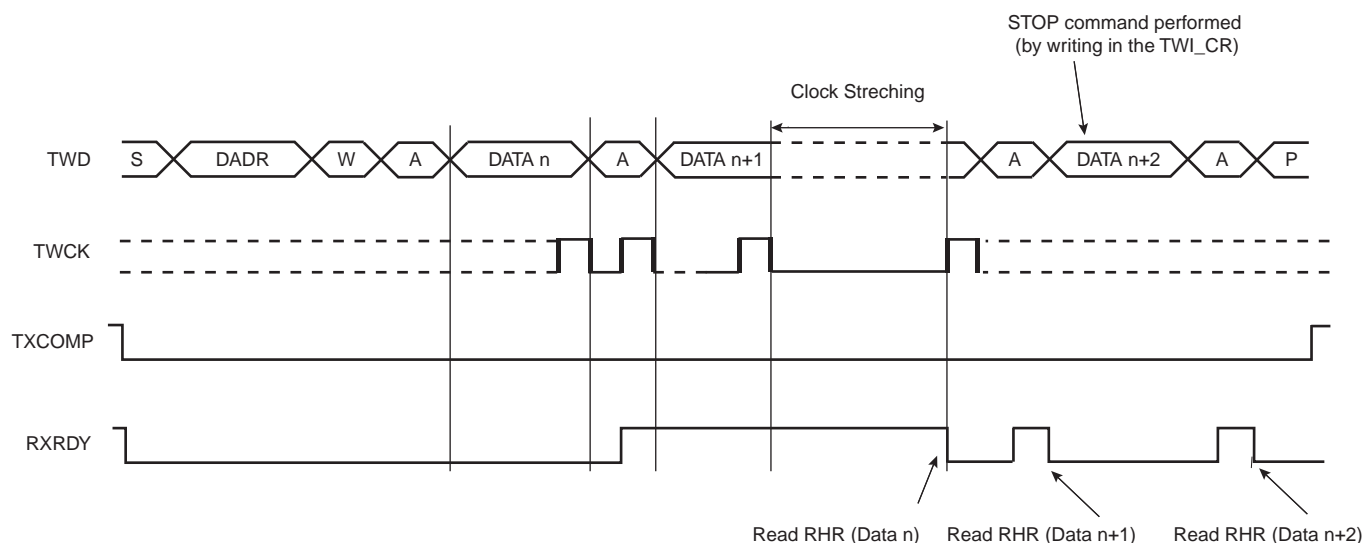
**Figure 34-9. Master Read with One Data Byte**



**Figure 34-10. Master Read with Multiple Data Bytes**



**Figure 34-11. Master Read Clock Stretching with Multiple Data Bytes**



RXRDY is used as Receive Ready for the PDC receive channel.

### 34.8.6 Internal Address

The TWI interface can perform various transfer formats: Transfers with 7-bit slave address devices and 10-bit slave address devices.

### 34.8.6.1 7-bit Slave Addressing

When Addressing 7-bit slave devices, the internal address bytes are used to perform random address (read or write) accesses to reach one or more data bytes, within a memory page location in a serial memory, for example. When performing read operations with an internal address, the TWI performs a write operation to set the internal address into the slave device, and then switch to Master Receiver mode. Note that the second start condition (after sending the IADR) is sometimes called “repeated start” (Sr) in I<sup>2</sup>C fully-compatible devices. See [Figure 34-13](#). See [Figure 34-12](#) and [Figure 34-14](#) for Master Write operation with internal address.

The three internal address bytes are configurable through the Master Mode register (TWI\_MMR).

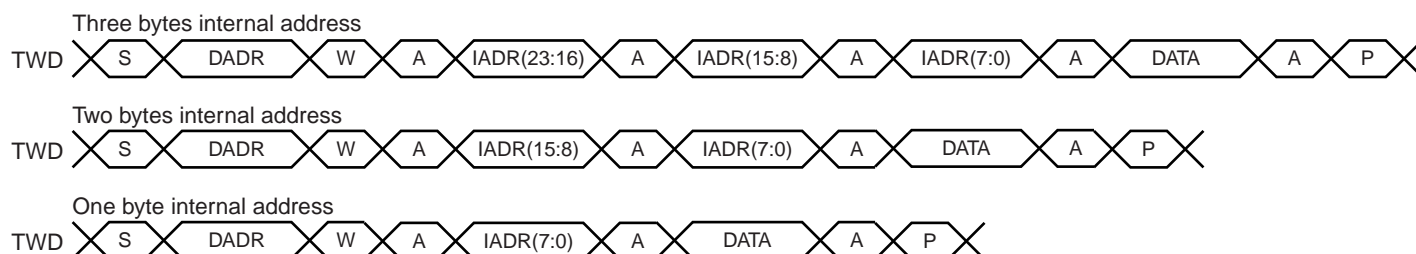
If the slave device supports only a 7-bit address, i.e., no internal address, IADRSZ must be set to 0.

[Table 34-6](#) shows the abbreviations used in [Figure 34-12](#) and [Figure 34-13](#).

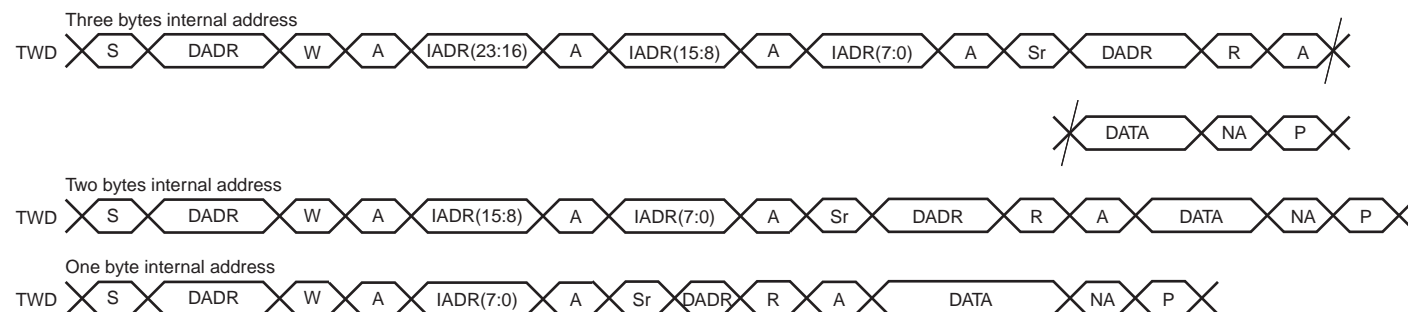
**Table 34-6. Abbreviations**

Abbreviation	Definition
S	Start
Sr	Repeated Start
P	Stop
W	Write
R	Read
A	Acknowledge
NA	Not Acknowledge
DADR	Device Address
IADR	Internal Address

**Figure 34-12. Master Write with One, Two or Three Bytes Internal Address and One Data Byte**



**Figure 34-13. Master Read with One, Two or Three Bytes Internal Address and One Data Byte**



### 34.8.6.2 10-bit Slave Addressing

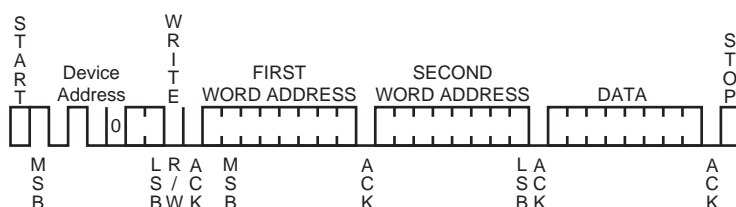
For a slave address higher than 7 bits, the user must configure the address size (IADRSZ) and set the other slave address bits in the internal address register (TWI\_IADR). The two remaining Internal address bytes, IADR[15:8] and IADR[23:16] can be used the same as in 7-bit Slave Addressing.

**Example:** Address a 10-bit device (10-bit device address is b1 b2 b3 b4 b5 b6 b7 b8 b9 b10)

1. Program IADRSZ = 1,
2. Program DADR with 1 1 1 1 0 b1 b2 (b1 is the MSB of the 10-bit address, b2, etc.)
3. Program TWI\_IADR with b3 b4 b5 b6 b7 b8 b9 b10 (b10 is the LSB of the 10-bit address)

Figure 34-14 below shows a byte write to an Atmel AT24LC512 EEPROM. This demonstrates the use of internal addresses to access the device.

Figure 34-14. Internal Address Usage



### 34.8.7 Using the Peripheral DMA Controller (PDC)

The use of the PDC significantly reduces the CPU load.

To assure correct implementation, respect the following programming sequences:

#### 34.8.7.1 Data Transmit with the PDC

1. Initialize the transmit PDC (memory pointers, transfer size - 1).
2. Configure the master (DADR, CKDIV, etc.) or slave mode.
3. Start the transfer by setting the PDC TXTEN bit.
4. Wait for the PDC ENDTX Flag either by using the polling method or ENDTX interrupt.
5. Disable the PDC by setting the PDC TXTDIS bit.
6. Wait for the TXRDY flag in TWI\_SR.
7. Set the STOP command in TWI\_CR.
8. Write the last character in TWI\_THR.
9. (Optional) Wait for the TXCOMP flag in TWI\_SR before disabling the peripheral clock if required.

#### 34.8.7.2 Data Receive with the PDC

The PDC transfer size must be defined with the buffer size minus 2. The two remaining characters must be managed without PDC to ensure that the exact number of bytes are received whatever the system bus latency conditions encountered during the end of buffer transfer period.

In slave mode, the number of characters to receive must be known in order to configure the PDC.

1. Initialize the receive PDC (memory pointers, transfer size - 2).
2. Configure the master (DADR, CKDIV, etc.) or slave mode.
3. Set the PDC RXTEN bit.
4. (Master Only) Write the START bit in the TWI\_CR to start the transfer.
5. Wait for the PDC ENDRX Flag either by using polling method or ENDRX interrupt.
6. Disable the PDC by setting the PDC RXTDIS bit.
7. Wait for the RXRDY flag in TWI\_SR.
8. Set the STOP command in TWI\_CR.

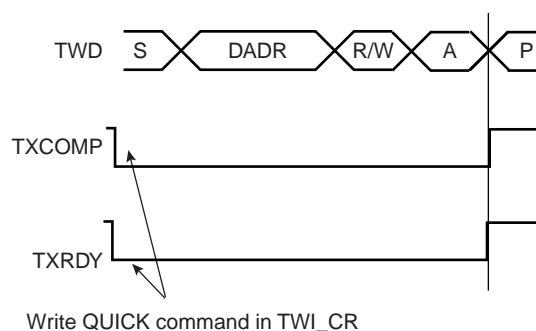
9. Read the penultimate character in TWI\_RHR.
10. Wait for the RXRDY flag in TWI\_SR.
11. Read the last character in TWI\_RHR.
12. (Optional) Wait for the TXCOMP flag in TWI\_SR before disabling the peripheral clock if required.

### 34.8.8 SMBUS Quick Command (Master Mode Only)

The TWI interface can perform a Quick Command:

1. Configure the master mode (DADR, CKDIV, etc.).
2. Write the MREAD bit in the TWI\_MMR at the value of the one-bit command to be sent.
3. Start the transfer by setting the QUICK bit in the TWI\_CR.

**Figure 34-15. SMBUS Quick Command**



### 34.8.9 Read-write Flowcharts

The following flowcharts shown in [Figure 34-17 on page 657](#), [Figure 34-18 on page 658](#), [Figure 34-19 on page 659](#), [Figure 34-20 on page 660](#) and [Figure 34-21 on page 661](#) give examples for read and write operations. A polling or interrupt method can be used to check the status bits. The interrupt method requires that the interrupt enable register (TWI\_IER) be configured first.

Figure 34-16. TWI Write Operation with Single Data Byte without Internal Address

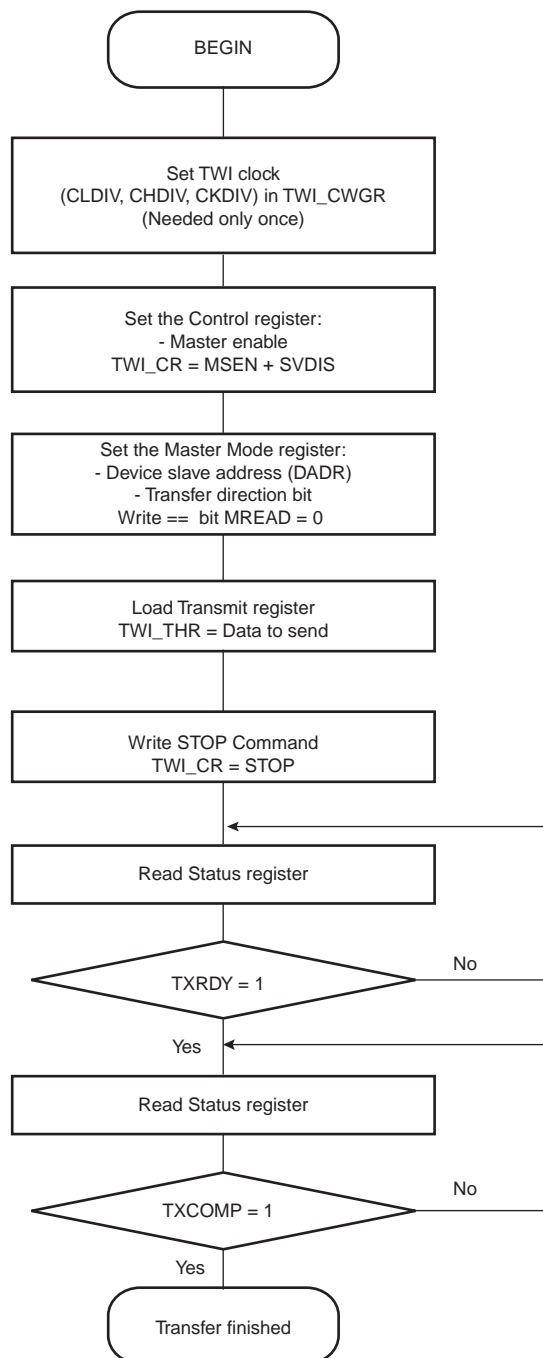




Figure 34-17. TWI Write Operation with Single Data Byte and Internal Address

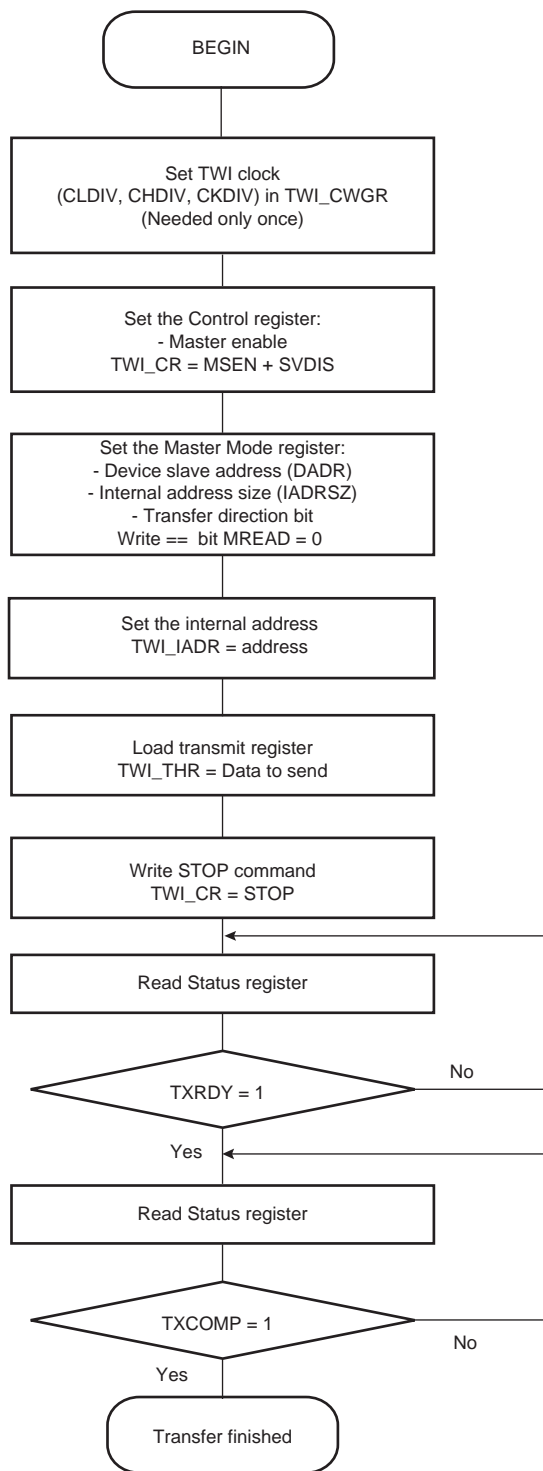


Figure 34-18. TWI Write Operation with Multiple Data Bytes with or without Internal Address

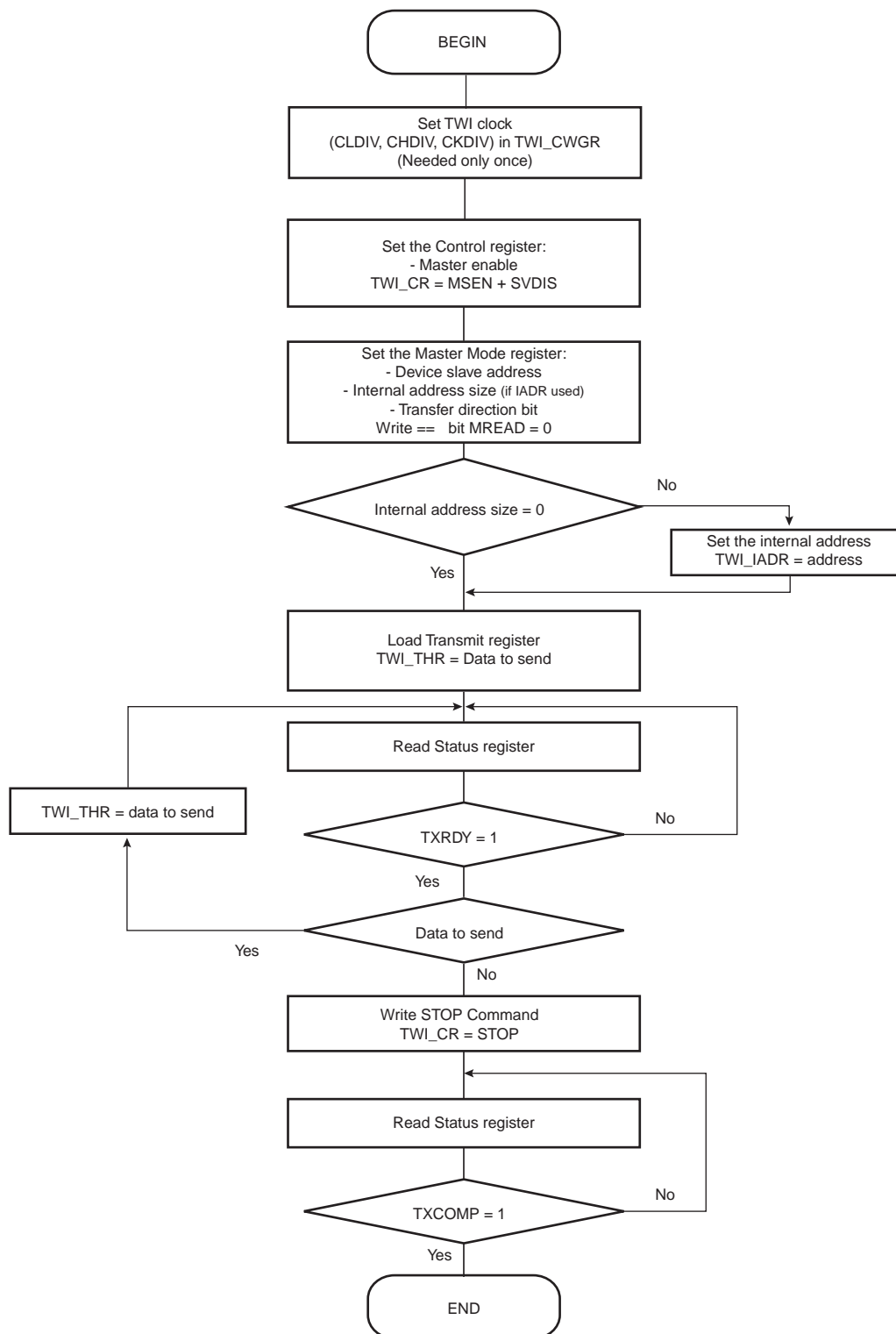


Figure 34-19. TWI Read Operation with Single Data Byte without Internal Address

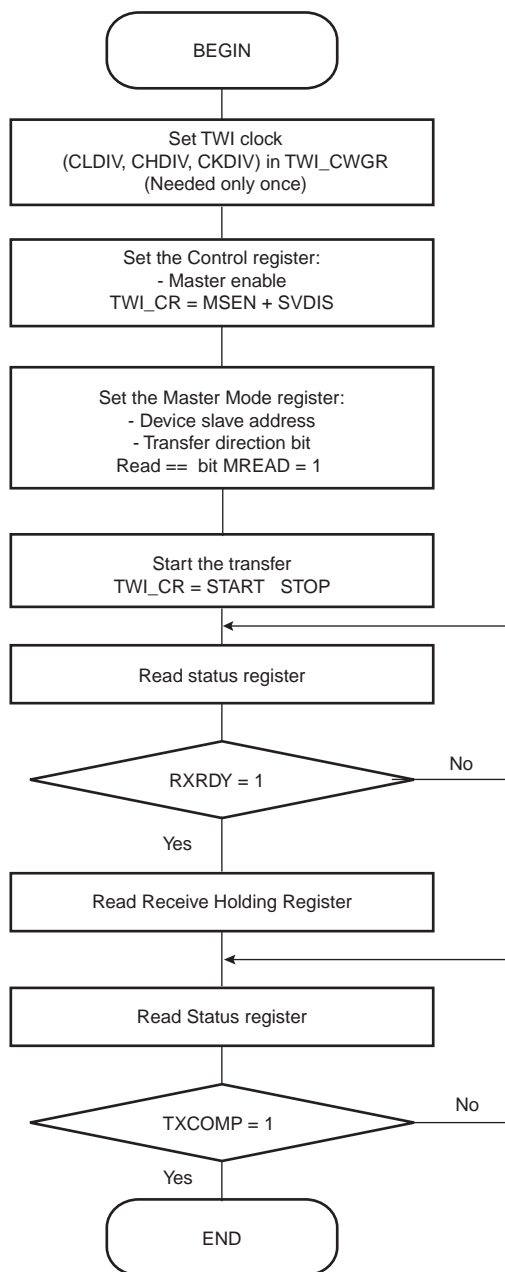


Figure 34-20. TWI Read Operation with Single Data Byte and Internal Address

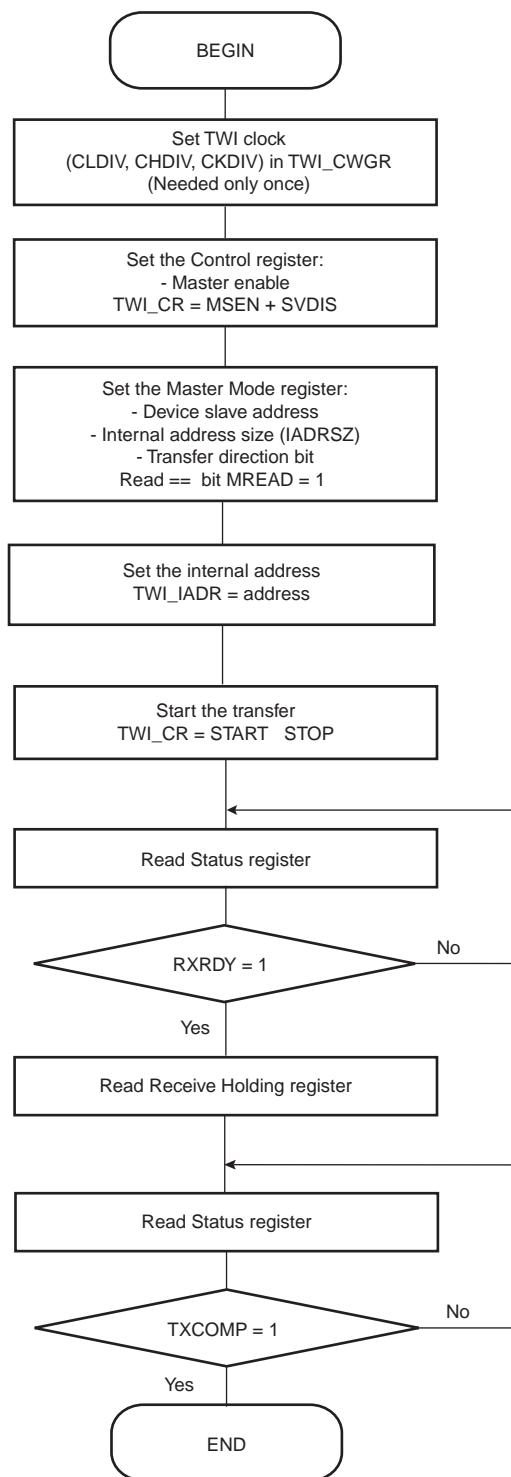
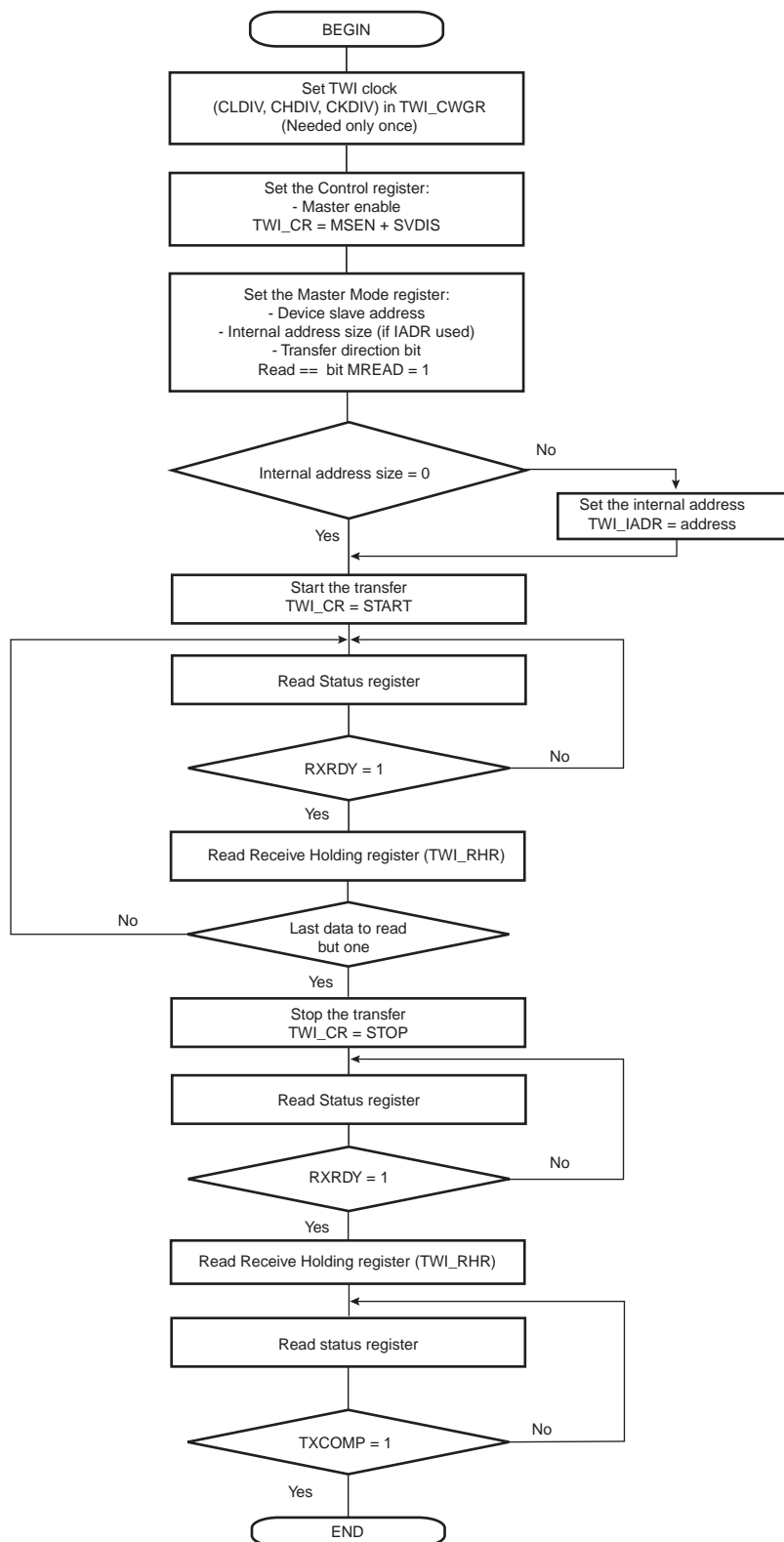


Figure 34-21. TWI Read Operation with Multiple Data Bytes with or without Internal Address



## 34.9 Multi-master Mode

### 34.9.1 Definition

More than one master may handle the bus at the same time without data corruption by using arbitration.

Arbitration starts as soon as two or more masters place information on the bus at the same time, and stops (arbitration is lost) for the master that intends to send a logical one while the other master sends a logical zero.

As soon as arbitration is lost by a master, it stops sending data and listens to the bus in order to detect a stop. When the stop is detected, the master who has lost arbitration may put its data on the bus by respecting arbitration.

Arbitration is illustrated in [Figure 34-23 on page 663](#).

### 34.9.2 Different Multi-master Modes

Two multi-master modes may be distinguished:

1. TWI is considered as a Master only and will never be addressed.
2. TWI may be either a Master or a Slave and may be addressed.

Note: In both Multi-master modes arbitration is supported.

#### 34.9.2.1 TWI as Master Only

In this mode, TWI is considered as a Master only (MSEN is always at one) and must be driven like a Master with the ARBLST (ARBitration Lost) flag in addition.

If arbitration is lost (ARBLST = 1), the programmer must reinitiate the data transfer.

If the user starts a transfer (ex.: DADR + START + W + Write in THR) and if the bus is busy, the TWI automatically waits for a STOP condition on the bus to initiate the transfer (see [Figure 34-22 on page 663](#)).

Note: The state of the bus (busy or free) is not indicated in the user interface.

#### 34.9.2.2 TWI as Master or Slave

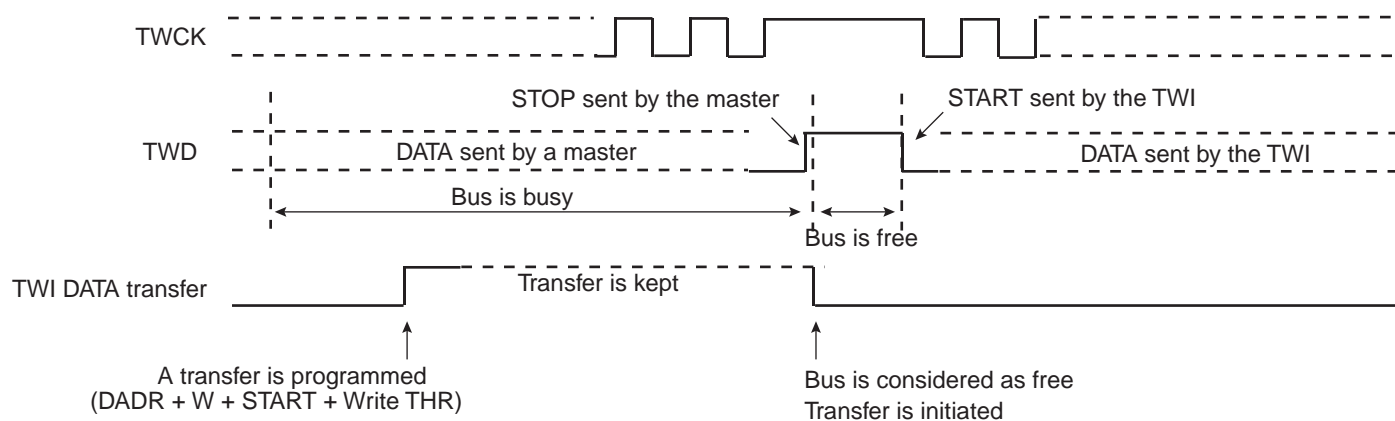
The automatic reversal from Master to Slave is not supported in case of a lost arbitration.

Then, in the case where TWI may be either a Master or a Slave, the programmer must manage the pseudo Multi-master mode described in the steps below.

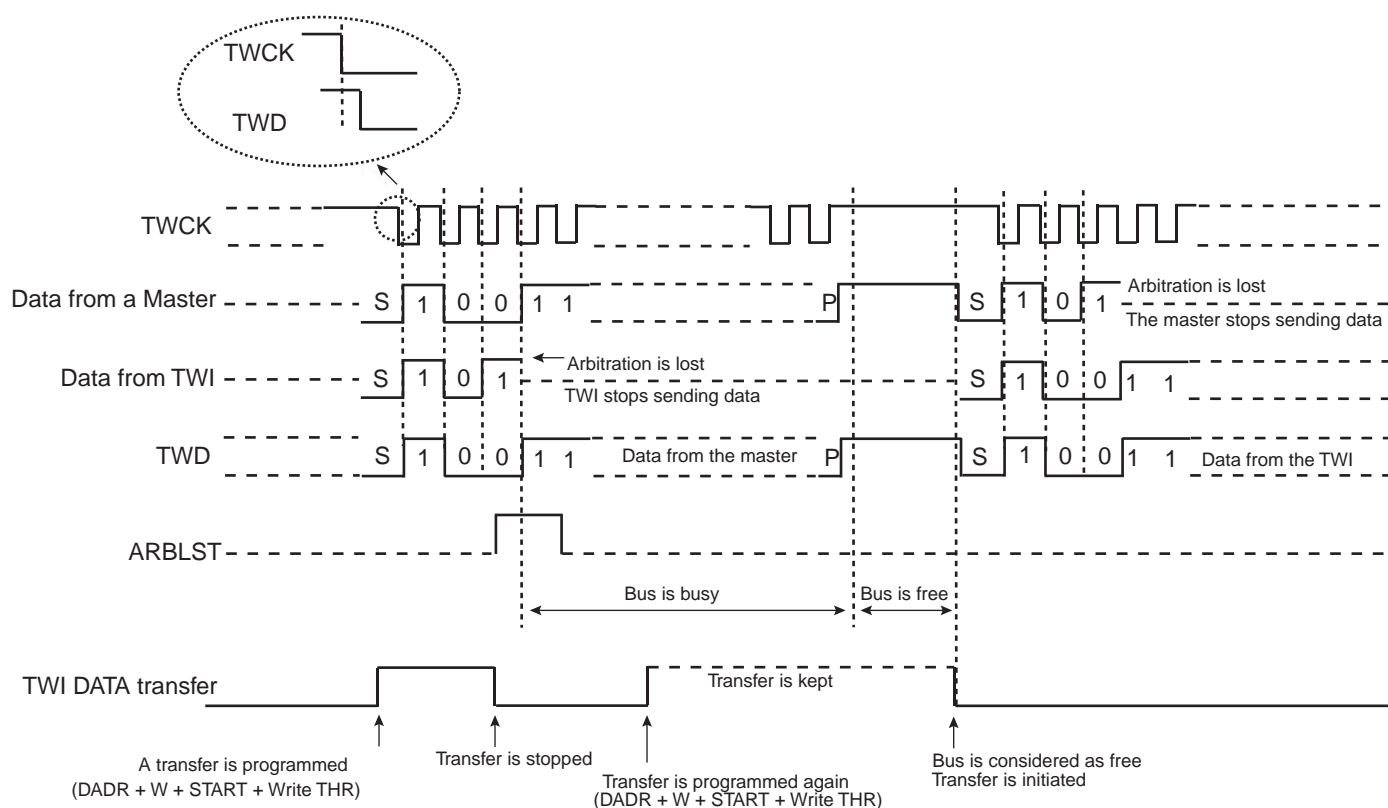
1. Program TWI in Slave mode (SADR + MSDIS + SVEN) and perform Slave Access (if TWI is addressed).
2. If TWI has to be set in Master mode, wait until TXCOMP flag is at 1.
3. Program Master mode (DADR + SVDIS + MSEN) and start the transfer (ex: START + Write in THR).
4. As soon as the Master mode is enabled, TWI scans the bus in order to detect if it is busy or free. When the bus is considered as free, TWI initiates the transfer.
5. As soon as the transfer is initiated and until a STOP condition is sent, the arbitration becomes relevant and the user must monitor the ARBLST flag.
6. If the arbitration is lost (ARBLST is set to 1), the user must program the TWI in Slave mode in the case where the Master that won the arbitration wanted to access the TWI.
7. If TWI has to be set in Slave mode, wait until TXCOMP flag is at 1 and then program the Slave mode.

Note: In the case where the arbitration is lost and TWI is addressed, TWI will not acknowledge even if it is programmed in Slave mode as soon as ARBLST is set to 1. Then, the Master must repeat SADR.

**Figure 34-22. Programmer Sends Data While the Bus is Busy**

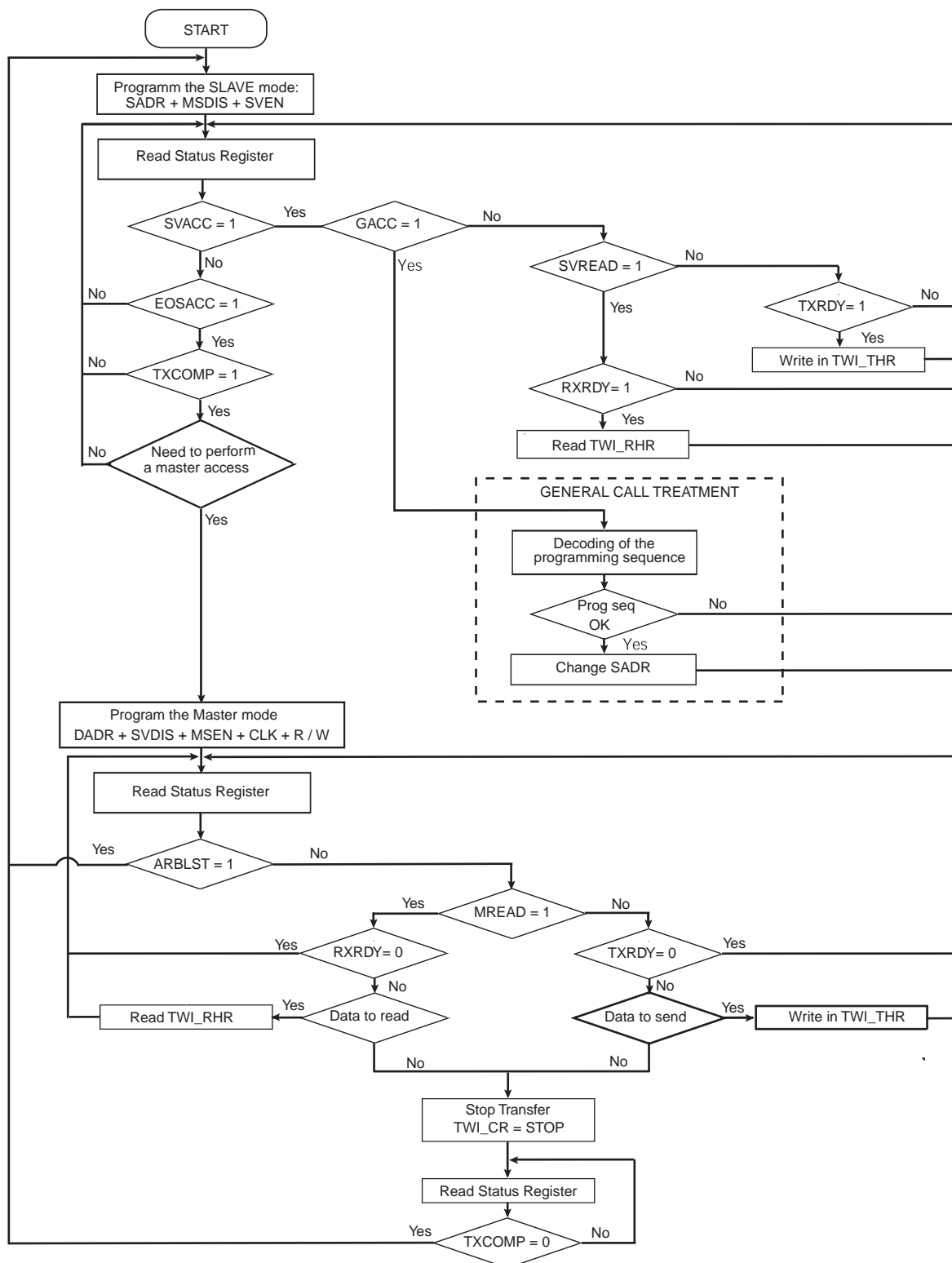


**Figure 34-23. Arbitration Cases**



The flowchart shown in [Figure 34-24 on page 664](#) gives an example of read and write operations in Multi-master mode.

Figure 34-24. Multi-master Flowchart





## 34.10 Slave Mode

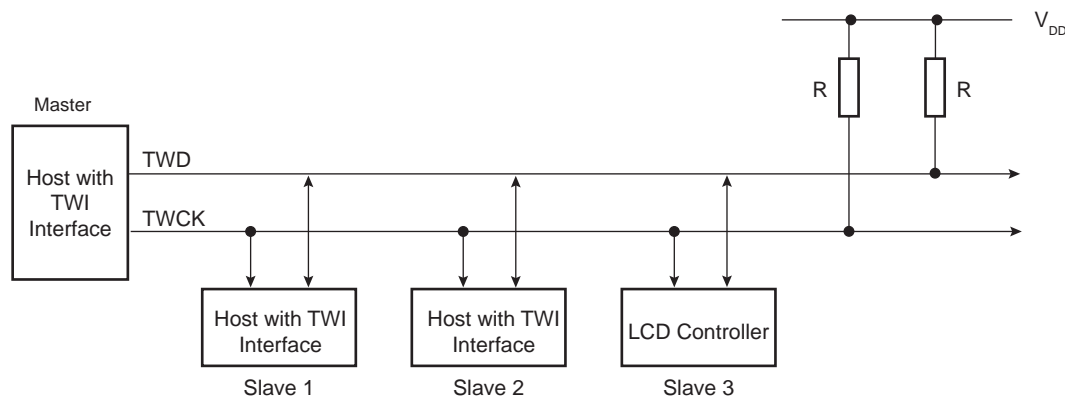
### 34.10.1 Definition

The Slave Mode is defined as a mode where the device receives the clock and the address from another device called the master.

In this mode, the device never initiates and never completes the transmission (START, REPEATED\_START and STOP conditions are always provided by the master).

### 34.10.2 Application Block Diagram

Figure 34-25. Slave Mode Typical Application Block Diagram



### 34.10.3 Programming Slave Mode

The following fields must be programmed before entering Slave mode:

1. SADR (TWI\_SMR): The slave device address is used in order to be accessed by master devices in read or write mode.
2. MSDIS (TWI\_CR): Disable the master mode.
3. SVEN (TWI\_CR): Enable the slave mode.

As the device receives the clock, values written in TWI\_CWGR are not taken into account.

### 34.10.4 Receiving Data

After a Start or Repeated Start condition is detected and if the address sent by the Master matches with the Slave address programmed in the SADR (Slave Address) field, SVACC (Slave ACCESS) flag is set and SVREAD (Slave READ) indicates the direction of the transfer.

SVACC remains high until a STOP condition or a repeated START is detected. When such a condition is detected, EOSACC (End Of Slave ACCESS) flag is set.

#### 34.10.4.1 Read Sequence

In the case of a Read sequence (SVREAD is high), TWI transfers data written in the TWI\_THR (TWI Transmit Holding Register) until a STOP condition or a REPEATED\_START + an address different from SADR is detected. Note that at the end of the read sequence TXCOMP (Transmission Complete) flag is set and SVACC reset.

As soon as data is written in the TWI\_THR, TXRDY (Transmit Holding Register Ready) flag is reset, and it is set when the shift register is empty and the sent data acknowledged or not. If the data is not acknowledged, the NACK flag is set.

Note that a STOP or a repeated START always follows a NACK.

See [Figure 34-26 on page 666](#).

### 34.10.4.2 Write Sequence

In the case of a Write sequence (SVREAD is low), the RXRDY (Receive Holding Register Ready) flag is set as soon as a character has been received in the TWI\_RHR (TWI Receive Holding Register). RXRDY is reset when reading the TWI\_RHR.

TWI continues receiving data until a STOP condition or a REPEATED\_START + an address different from SADR is detected. Note that at the end of the write sequence TXCOMP flag is set and SVACC reset.

See [Figure 34-27 on page 667](#).

### 34.10.4.3 Clock Synchronization Sequence

In the case where TWI\_THR or TWI\_RHR is not written/read in time, TWI performs a clock synchronization.

Clock stretching information is given by the SCLWS (Clock Wait state) bit.

See [Figure 34-29 on page 668](#) and [Figure 34-30 on page 669](#).

### 34.10.4.4 General Call

In the case where a GENERAL CALL is performed, GACC (General Call ACCess) flag is set.

After GACC is set, it is up to the programmer to interpret the meaning of the GENERAL CALL and to decode the new address programming sequence.

See [Figure 34-28 on page 667](#).

## 34.10.5 Data Transfer

### 34.10.5.1 Read Operation

The read mode is defined as a data requirement from the master.

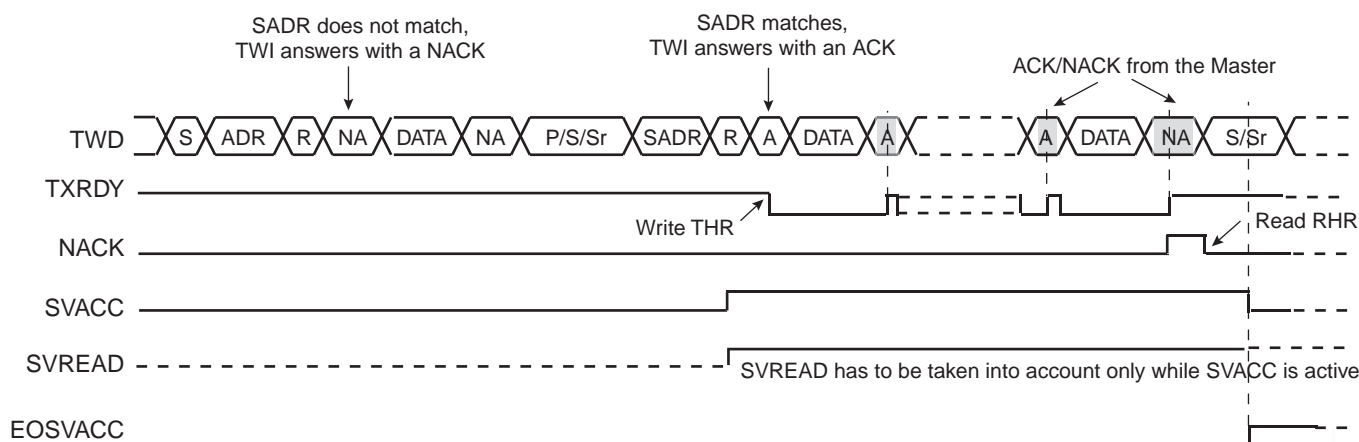
After a START or a REPEATED START condition is detected, the decoding of the address starts. If the slave address (SADR) is decoded, SVACC is set and SVREAD indicates the direction of the transfer.

Until a STOP or REPEATED START condition is detected, TWI continues sending data loaded in the TWI\_THR.

If a STOP condition or a REPEATED START + an address different from SADR is detected, SVACC is reset.

[Figure 34-26 on page 666](#) describes the write operation.

**Figure 34-26. Read Access Ordered by a MASTER**



- Notes:
1. When SVACC is low, the state of SVREAD becomes irrelevant.
  2. TXRDY is reset when data has been transmitted from TWI\_THR to the shift register and set when this data has been acknowledged or non acknowledged.

### 34.10.5.2 Write Operation

The write mode is defined as a data transmission from the master.

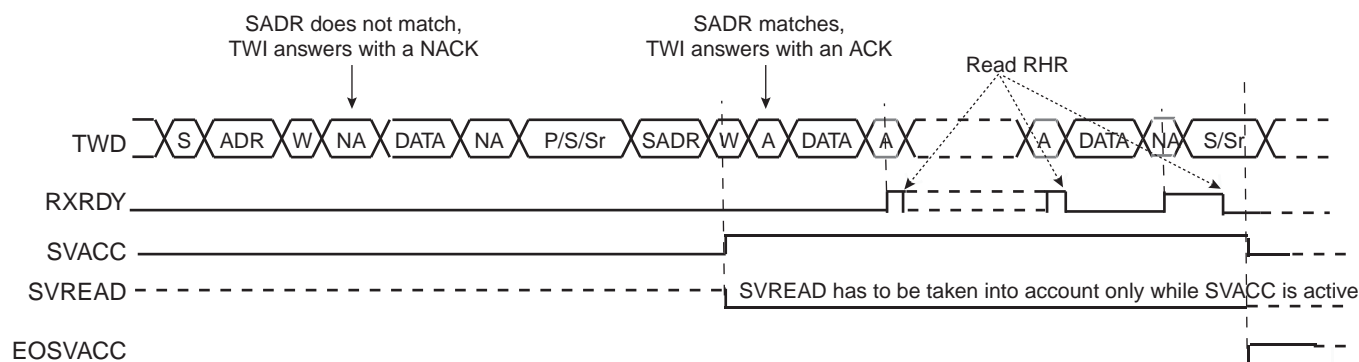
After a START or a REPEATED START, the decoding of the address starts. If the slave address is decoded, SVACC is set and SVREAD indicates the direction of the transfer (SVREAD is low in this case).

Until a STOP or REPEATED START condition is detected, TWI stores the received data in the TWI\_RHR.

If a STOP condition or a REPEATED START + an address different from SADR is detected, SVACC is reset.

Figure 34-27 describes the Write operation.

**Figure 34-27. Write Access Ordered by a Master**



- Notes:
1. When SVACC is low, the state of SVREAD becomes irrelevant.
  2. RXRDY is set when data has been transmitted from the shift register to the TWI\_RHR and reset when this data is read.

### 34.10.5.3 General Call

The general call is performed in order to change the address of the slave.

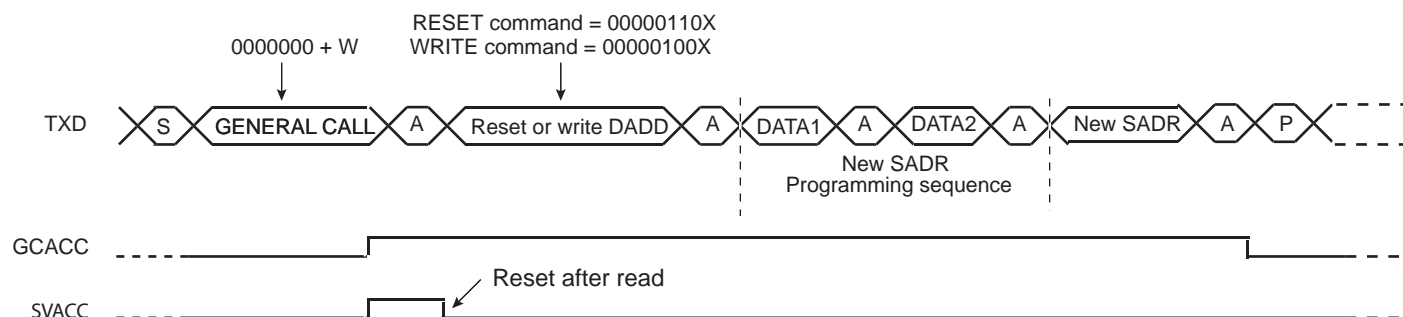
If a GENERAL CALL is detected, GACC is set.

After the detection of General Call, it is up to the programmer to decode the commands which come afterwards.

In case of a WRITE command, the programmer has to decode the programming sequence and program a new SADR if the programming sequence matches.

Figure 34-28 describes the General Call access.

**Figure 34-28. Master Performs a General Call**



- Note:
- This method allows the user to create an own programming sequence by choosing the programming bytes and the number of them. The programming sequence has to be provided to the master.

### 34.10.5.4 Clock Synchronization

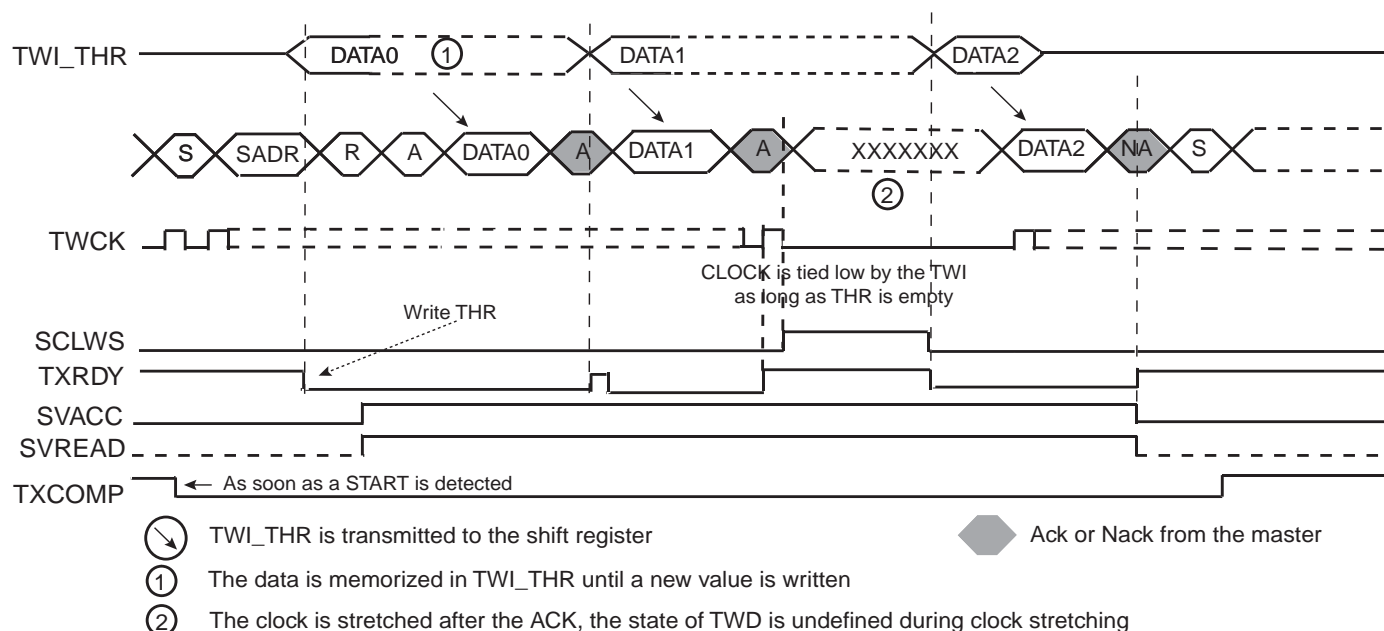
In both read and write modes, it may happen that TWI\_THR/TWI\_RHR buffer is not filled /emptied before the emission/reception of a new character. In this case, to avoid sending/receiving undesired data, a clock stretching mechanism is implemented.

#### Clock Synchronization in Read Mode

The clock is tied low if the shift register is empty and if a STOP or REPEATED START condition was not detected. It is tied low until the shift register is loaded.

Figure 34-29 describes the clock synchronization in Read mode.

**Figure 34-29. Clock Synchronization in Read Mode**



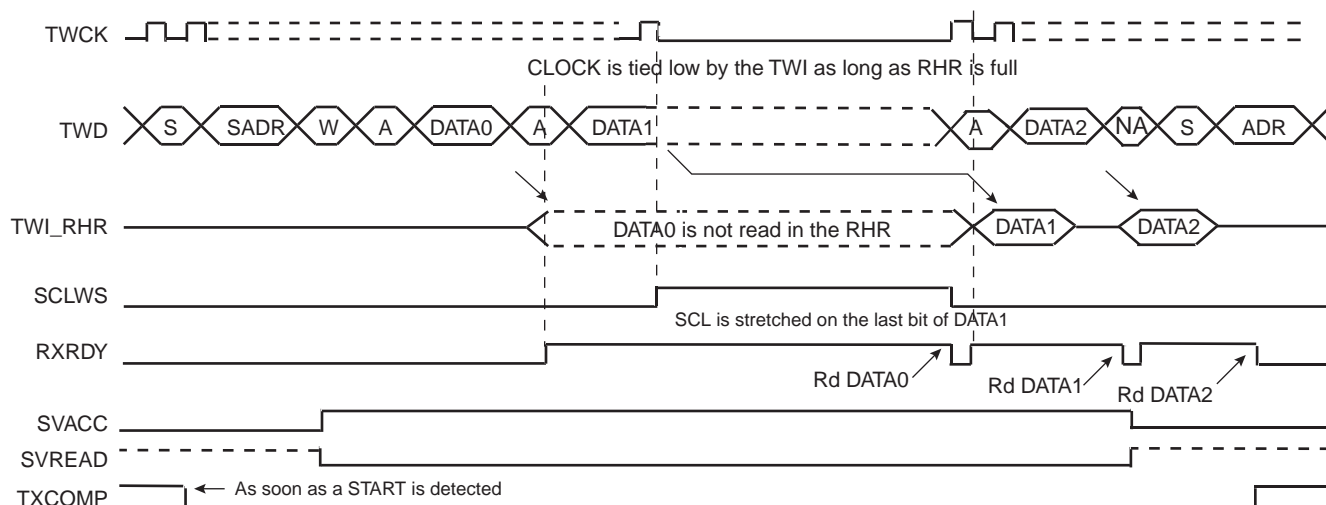
- Notes:
1. TXRDY is reset when data has been written in the TWI\_THR to the shift register and set when this data has been acknowledged or non acknowledged.
  2. At the end of the read sequence, TXCOMP is set after a STOP or after a REPEATED\_START + an address different from SADR.
  3. SCLWS is automatically set when the clock synchronization mechanism is started.

#### Clock Synchronization in Write Mode

The clock is tied low if the shift register and the TWI\_RHR is full. If a STOP or REPEATED\_START condition was not detected, it is tied low until TWI\_RHR is read.

Figure 34-30 describes the clock synchronization in Read mode.

**Figure 34-30. Clock Synchronization in Write Mode**



- Notes:
1. At the end of the read sequence, TXCOMP is set after a STOP or after a REPEATED\_START + an address different from SADR.
  2. SCLWS is automatically set when the clock synchronization mechanism is started and automatically reset when the mechanism is finished.

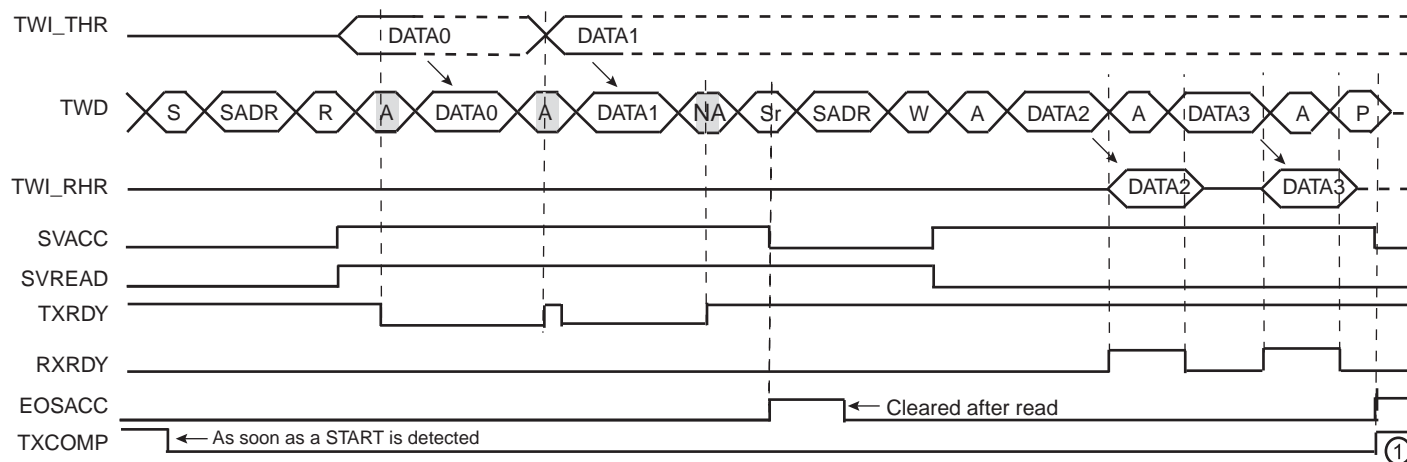
#### 34.10.5.5 Reversal after a Repeated Start

##### Reversal of Read to Write

The master initiates the communication by a read command and finishes it by a write command.

Figure 34-31 describes the repeated start + reversal from Read to Write mode.

**Figure 34-31. Repeated Start + Reversal from Read to Write Mode**

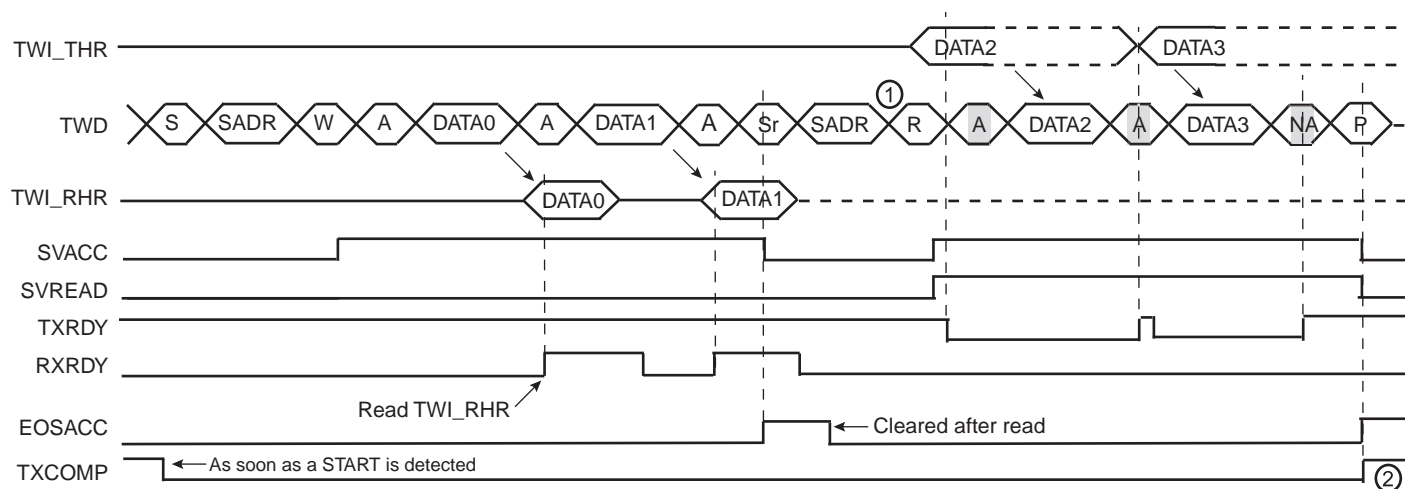


- Note:
1. TXCOMP is only set at the end of the transmission because after the repeated start, SADR is detected again.

## Reversal of Write to Read

The master initiates the communication by a write command and finishes it by a read command. Figure 34-32 describes the repeated start + reversal from Write to Read mode.

**Figure 34-32. Repeated Start + Reversal from Write to Read Mode**



- Notes:
1. In this case, if TWI\_THR has not been written at the end of the read command, the clock is automatically stretched before the ACK.
  2. TXCOMP is only set at the end of the transmission because after the repeated start, SADR is detected again.

## 34.10.6 Using the Peripheral DMA Controller (PDC) in Slave Mode

The use of the PDC significantly reduces the CPU load.

### 34.10.6.1 Data Transmit with the PDC in Slave Mode

The following procedure shows an example to transmit data with PDC.

1. Initialize the transmit PDC (memory pointers, transfer size).
2. Start the transfer by setting the PDC TXTEN bit.
3. Wait for the PDC ENDTX Flag either by using the polling method or ENDTX interrupt.
4. Disable the PDC by setting the PDC TXTDIS bit.
5. (Optional) Wait for the TXCOMP flag in TWI\_SR before disabling the peripheral clock if required.

### 34.10.6.2 Data Receive with the PDC in Slave Mode

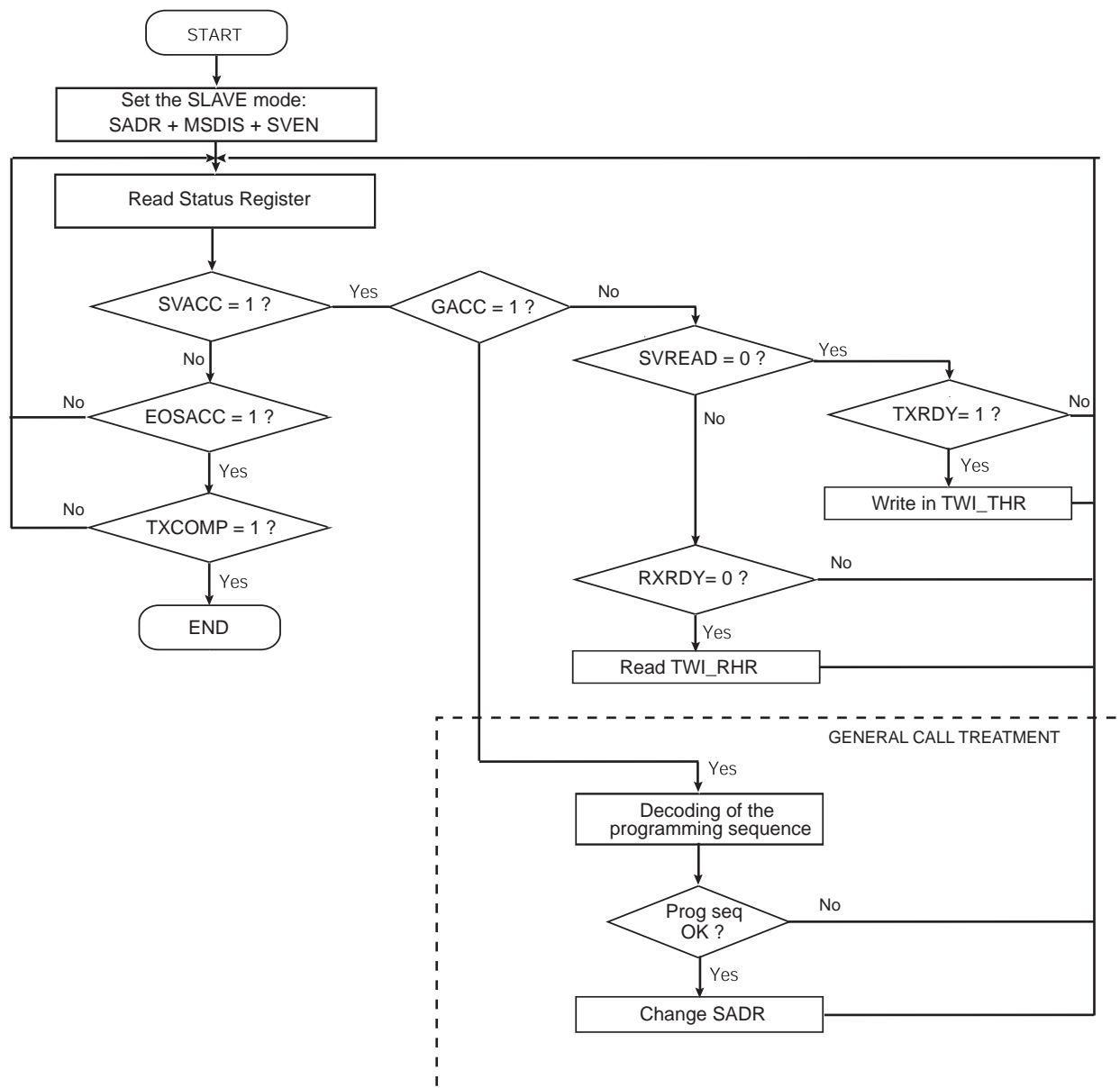
The following procedure shows an example to transmit data with PDC where the number of characters to receive is known.

1. Initialize the receive PDC (memory pointers, transfer size).
2. Set the PDC RXTEN bit.
3. Wait for the PDC ENDRX flag either by using polling method or ENDRX interrupt.
4. Disable the PDC by setting the PDC RXTDIS bit.
5. (Optional) Wait for the TXCOMP flag in TWI\_SR before disabling the peripheral clock if required.

### 34.10.7 Read Write Flowcharts

The flowchart shown in Figure 34-33 gives an example of read and write operations in Slave mode. A polling or interrupt method can be used to check the status bits. The interrupt method requires that the interrupt enable register (TWI\_IER) be configured first.

Figure 34-33. Read Write Flowchart in Slave Mode



## 34.11 Write Protection System

In order to bring security to the TWI, a write protection system has been implemented.

The write protection mode prevents the write of the “TWI Clock Waveform Generator Register” and the “TWI Slave Mode Register”. When this mode is enabled and one of the protected registers is written, an error is generated in the “TWI Write Protection Status Register” and the register write request is canceled. When a write protection error occurs the WPVS flag is set and the address of the corresponding canceled register write is available in the WPVSRC field of the “TWI Write Protection Status Register”.

Due to the nature of the write protection feature, enabling and disabling the write protection mode requires the use of a security code. Thus when enabling or disabling the write protection mode the WPKEY field of the “TWI Write Protection Mode Register” must be filled with the “TWI” ASCII code (0x545749) otherwise the register write will be canceled.



## 34.12 Two-wire Interface (TWI) User Interface

**Table 34-7. Register Mapping**

Offset	Register	Name	Access	Reset
0x00	Control Register	TWI_CR	Write-only	–
0x04	Master Mode Register	TWI_MMR	Read-write	0x00000000
0x08	Slave Mode Register	TWI_SMR	Read-write	0x00000000
0x0C	Internal Address Register	TWI_IADR	Read-write	0x00000000
0x10	Clock Waveform Generator Register	TWI_CWGR	Read-write	0x00000000
0x14–0x1C	Reserved	–	–	–
0x20	Status Register	TWI_SR	Read-only	0x0000F009
0x24	Interrupt Enable Register	TWI_IER	Write-only	–
0x28	Interrupt Disable Register	TWI_IDR	Write-only	–
0x2C	Interrupt Mask Register	TWI_IMR	Read-only	0x00000000
0x30	Receive Holding Register	TWI_RHR	Read-only	0x00000000
0x34	Transmit Holding Register	TWI_THR	Write-only	0x00000000
0x38–0xE0	Reserved	–	–	–
0xE4	Protection Mode Register	TWI_WPMR	Read-write	0x00000000
0xE8	Protection Status Register	TWI_WPSR	Read-only	0x00000000
0xEC–0xFC	Reserved	–	–	–
0x100–0x128	Reserved for PDC registers	–	–	–

Note: All unlisted offset values are considered as “reserved”.

### 34.12.1 TWI Control Register

Name: TWI\_CR

Address: 0x40018000 (0), 0x4001C000 (1)

Access: Write-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
SWRST	QUICK	SVDIS	SVEN	MSDIS	MSEN	STOP	START

- **START: Send a START Condition**

0: No effect.

1: A frame beginning with a START bit is transmitted according to the features defined in the mode register.

This action is necessary when the TWI peripheral wants to read data from a slave. When configured in Master Mode with a write operation, a frame is sent as soon as the user writes a character in the Transmit Holding Register (TWI\_THR).

- **STOP: Send a STOP Condition**

0: No effect.

1: STOP Condition is sent just after completing the current byte transmission in master read mode.

- In single data byte master read, the START and STOP must both be set.
- In multiple data bytes master read, the STOP must be set after the last data received but one.
- In master read mode, if a NACK bit is received, the STOP is automatically performed.
- In master data write operation, a STOP condition will be sent after the transmission of the current data is finished.

- **MSEN: TWI Master Mode Enabled**

0: No effect.

1: If MSDIS = 0, the master mode is enabled.

Note: Switching from Slave to Master mode is only permitted when TXCOMP = 1.

- **MSDIS: TWI Master Mode Disabled**

0: No effect.

1: The master mode is disabled, all pending data is transmitted. The shifter and holding characters (if it contains data) are transmitted in case of write operation. In read operation, the character being transferred must be completely received before disabling.

- **SVEN: TWI Slave Mode Enabled**

0: No effect.

1: If SVDIS = 0, the slave mode is enabled.

Note: Switching from Master to Slave mode is only permitted when TXCOMP = 1.

- **SVDIS: TWI Slave Mode Disabled**

0: No effect.

1: The slave mode is disabled. The shifter and holding characters (if it contains data) are transmitted in case of read operation. In write operation, the character being transferred must be completely received before disabling.

- **QUICK: SMBUS Quick Command**

0: No effect.

1: If Master mode is enabled, a SMBUS Quick Command is sent.

- **SWRST: Software Reset**

0: No effect.

1: Equivalent to a system reset.

### 34.12.2 TWI Master Mode Register

**Name:** TWI\_MMR

**Address:** 0x40018004 (0), 0x4001C004 (1)

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	DADR						
15	14	13	12	11	10	9	8
–	–	–	MREAD	–	–	IADRSZ	
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

- **IADRSZ: Internal Device Address Size**

Value	Name	Description
0	NONE	No internal device address
1	1_BYTE	One-byte internal device address
2	2_BYTE	Two-byte internal device address
3	3_BYTE	Three-byte internal device address

- **MREAD: Master Read Direction**

0: Master write direction.

1: Master read direction.

- **DADR: Device Address**

The device address is used to access slave devices in read or write mode. Those bits are only used in Master mode.

### 34.12.3 TWI Slave Mode Register

Name: TWI\_SMR

Address: 0x40018008 (0), 0x4001C008 (1)

Access: Read-write

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	SADR						
15	14	13	12	11	10	9	8
–	–	–	–	–	–		
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	–

This register can only be written if the WPEN bit is cleared in the [“TWI Write Protection Mode Register”](#).

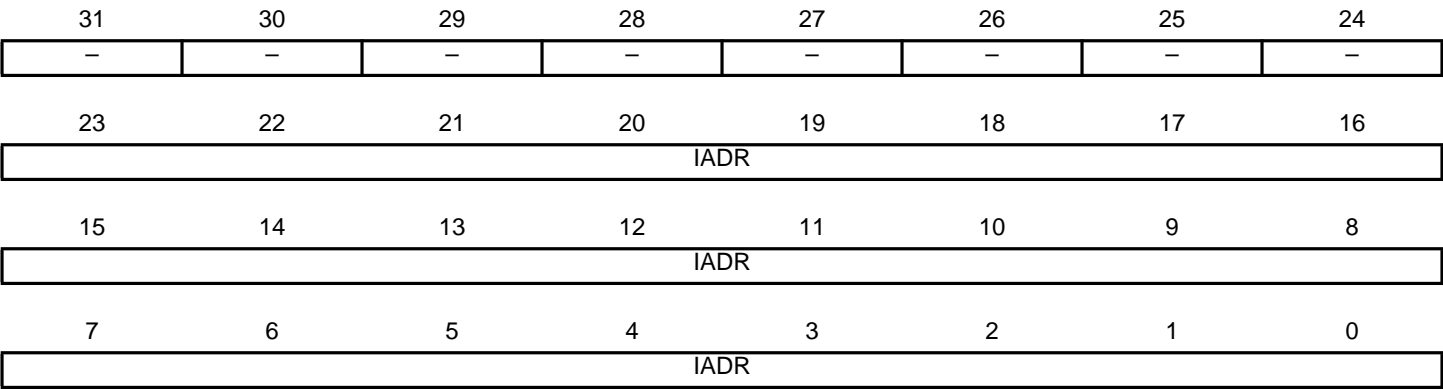
- **SADR: Slave Address**

The slave device address is used in Slave mode in order to be accessed by master devices in read or write mode.

SADR must be programmed before enabling the Slave mode or after a general call. Writes at other times have no effect.

34.12.4 TWI Internal Address Register

Name: TWI\_IADR  
Address: 0x4001800C (0), 0x4001C00C (1)  
Access: Read-write  
Reset: 0x00000000



- IADR: Internal Address  
0, 1, 2 or 3 bytes depending on IADRSZ.

### 34.12.5 TWI Clock Waveform Generator Register

Name: TWI\_CWGR

Address: 0x40018010 (0), 0x4001C010 (1)

Access: Read-write

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	CKDIV		
15	14	13	12	11	10	9	8
CHDIV							
7	6	5	4	3	2	1	0
CLDIV							

This register can only be written if the WPEN bit is cleared in the [“TWI Write Protection Mode Register”](#).

TWI\_CWGR is only used in Master mode.

- **CLDIV: Clock Low Divider**

The SCL low period is defined as follows:

$$t_{low} = ((CLDIV \times 2^{CKDIV}) + 4) \times t_{MCK}$$

- **CHDIV: Clock High Divider**

The SCL high period is defined as follows:

$$t_{high} = ((CHDIV \times 2^{CKDIV}) + 4) \times t_{MCK}$$

- **CKDIV: Clock Divider**

The CKDIV is used to increase both SCL high and low periods.

### 34.12.6 TWI Status Register

Name: TWI\_SR

Address: 0x40018020 (0), 0x4001C020 (1)

Access: Read-only

Reset: 0x0000F009

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXBUFE	RXBUFF	ENDTX	ENDRX	EOSACC	SCLWS	ARBLST	NACK
7	6	5	4	3	2	1	0
–	OVRE	GACC	SVACC	SVREAD	TXRDY	RXRDY	TXCOMP

- **TXCOMP: Transmission Completed (automatically set / reset)**

TXCOMP used in Master mode:

0: During the length of the current frame.

1: When both holding and shifter registers are empty and STOP condition has been sent.

*TXCOMP behavior in Master mode* can be seen in [Figure 34-8 on page 651](#) and in [Figure 34-10 on page 652](#).

TXCOMP used in Slave mode:

0: As soon as a Start is detected.

1: After a Stop or a Repeated Start + an address different from SADR is detected.

*TXCOMP behavior in Slave mode* can be seen in [Figure 34-29 on page 668](#), [Figure 34-30 on page 669](#), [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#).

- **RXRDY: Receive Holding Register Ready (automatically set / reset)**

0: No character has been received since the last TWI\_RHR read operation.

1: A byte has been received in the TWI\_RHR since the last read.

*RXRDY behavior in Master mode* can be seen in [Figure 34-10 on page 652](#).

*RXRDY behavior in Slave mode* can be seen in [Figure 34-27 on page 667](#), [Figure 34-30 on page 669](#), [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#).

- **TXRDY: Transmit Holding Register Ready (automatically set / reset)**

TXRDY used in Master mode:

0: The transmit holding register has not been transferred into shift register. Set to 0 when writing into TWI\_THR.

1: As soon as a data byte is transferred from TWI\_THR to internal shifter or if a NACK error is detected, TXRDY is set at the same time as TXCOMP and NACK. TXRDY is also set when MSEN is set (enable TWI).

*TXRDY behavior in Master mode* can be seen in [Figure 34.8.4 on page 649](#).

TXRDY used in Slave mode:

0: As soon as data is written in the TWI\_THR, until this data has been transmitted and acknowledged (ACK or NACK).



1: It indicates that the TWI\_THR is empty and that data has been transmitted and acknowledged.

If TXRDY is high and if a NACK has been detected, the transmission will be stopped. Thus when TRDY = NACK = 1, the programmer must not fill TWI\_THR to avoid losing it.

*TXRDY behavior in Slave mode* can be seen in [Figure 34-26 on page 666](#), [Figure 34-29 on page 668](#), [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#).

- **SVREAD: Slave Read (automatically set / reset)**

This bit is only used in Slave mode. When SVACC is low (no Slave access has been detected) SVREAD is irrelevant.

0: Indicates that a write access is performed by a Master.

1: Indicates that a read access is performed by a Master.

*SVREAD behavior* can be seen in [Figure 34-26 on page 666](#), [Figure 34-27 on page 667](#), [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#).

- **SVACC: Slave Access (automatically set / reset)**

This bit is only used in Slave mode.

0: TWI is not addressed. SVACC is automatically cleared after a NACK or a STOP condition is detected.

1: Indicates that the address decoding sequence has matched (A Master has sent SADR). SVACC remains high until a NACK or a STOP condition is detected.

*SVACC behavior* can be seen in [Figure 34-26 on page 666](#), [Figure 34-27 on page 667](#), [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#).

- **GACC: General Call Access (clear on read)**

This bit is only used in Slave mode.

0: No General Call has been detected.

1: A General Call has been detected. After the detection of General Call, if need be, the programmer may acknowledge this access and decode the following bytes and respond according to the value of the bytes.

*GACC behavior* can be seen in [Figure 34-28 on page 667](#).

- **OVRE: Overrun Error (clear on read)**

This bit is only used in Master mode.

0: TWI\_RHR has not been loaded while RXRDY was set

1: TWI\_RHR has been loaded while RXRDY was set. Reset by read in TWI\_SR when TXCOMP is set.

- **NACK: Not Acknowledged (clear on read)**

NACK used in Master mode:

0: Each data byte has been correctly received by the far-end side TWI slave component.

1: A data byte or an address byte has not been acknowledged by the slave component. Set at the same time as TXCOMP.

NACK used in Slave Read mode:

0: Each data byte has been correctly received by the Master.

1: In read mode, a data byte has not been acknowledged by the Master. When NACK is set the programmer must not fill TWI\_THR even if TXRDY is set, because it means that the Master will stop the data transfer or re initiate it.

Note that in Slave Write mode all data are acknowledged by the TWI.

- **ARBLST: Arbitration Lost (clear on read)**

This bit is only used in Master mode.

0: Arbitration won.

1: Arbitration lost. Another master of the TWI bus has won the multi-master arbitration. TXCOMP is set at the same time.

- **SCLWS: Clock Wait State (automatically set / reset)**

This bit is only used in Slave mode.

0: The clock is not stretched.

1: The clock is stretched. TWI\_THR / TWI\_RHR buffer is not filled / emptied before the emission / reception of a new character.

*SCLWS behavior* can be seen in [Figure 34-29 on page 668](#) and [Figure 34-30 on page 669](#).

- **EOSACC: End Of Slave Access (clear on read)**

This bit is only used in Slave mode.

0: A slave access is being performing.

1: The Slave Access is finished. End Of Slave Access is automatically set as soon as SVACC is reset.

*EOSACC behavior* can be seen in [Figure 34-31 on page 669](#) and [Figure 34-32 on page 670](#)

- **ENDRX: End of RX buffer**

0: The Receive Counter Register has not reached 0 since the last write in TWI\_RCR or TWI\_RNCR.

1: The Receive Counter Register has reached 0 since the last write in TWI\_RCR or TWI\_RNCR.

- **ENDTX: End of TX buffer**

0: The Transmit Counter Register has not reached 0 since the last write in TWI\_TCR or TWI\_TNCR.

1: The Transmit Counter Register has reached 0 since the last write in TWI\_TCR or TWI\_TNCR.

- **RXBUFF: RX Buffer Full**

0: TWI\_RCR or TWI\_RNCR have a value other than 0.

1: Both TWI\_RCR and TWI\_RNCR have a value of 0.

- **TXBUFE: TX Buffer Empty**

0: TWI\_TCR or TWI\_TNCR have a value other than 0.

1: Both TWI\_TCR and TWI\_TNCR have a value of 0.

### 34.12.7 TWI Interrupt Enable Register

Name: TWI\_IER

Address: 0x40018024 (0), 0x4001C024 (1)

Access: Write-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXBUFE	RXBUFF	ENDTX	ENDRX	EOSACC	SCL_WS	ARBLST	NACK
7	6	5	4	3	2	1	0
–	OVRE	GACC	SVACC	–	TXRDY	RXRDY	TXCOMP

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Enables the corresponding interrupt.

- **TXCOMP:** Transmission Completed Interrupt Enable
- **RXRDY:** Receive Holding Register Ready Interrupt Enable
- **TXRDY:** Transmit Holding Register Ready Interrupt Enable
- **SVACC:** Slave Access Interrupt Enable
- **GACC:** General Call Access Interrupt Enable
- **OVRE:** Overrun Error Interrupt Enable
- **NACK:** Not Acknowledge Interrupt Enable
- **ARBLST:** Arbitration Lost Interrupt Enable
- **SCL\_WS:** Clock Wait State Interrupt Enable
- **EOSACC:** End Of Slave Access Interrupt Enable
- **ENDRX:** End of Receive Buffer Interrupt Enable
- **ENDTX:** End of Transmit Buffer Interrupt Enable
- **RXBUFF:** Receive Buffer Full Interrupt Enable
- **TXBUFE:** Transmit Buffer Empty Interrupt Enable

### 34.12.8 TWI Interrupt Disable Register

Name: TWI\_IDR

Address: 0x40018028 (0), 0x4001C028 (1)

Access: Write-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXBUFE	RXBUFF	ENDTX	ENDRX	EOSACC	SCL_WS	ARBLST	NACK
7	6	5	4	3	2	1	0
–	OVRE	GACC	SVACC	–	TXRDY	RXRDY	TXCOMP

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Disables the corresponding interrupt.

- **TXCOMP:** Transmission Completed Interrupt Disable
- **RXRDY:** Receive Holding Register Ready Interrupt Disable
- **TXRDY:** Transmit Holding Register Ready Interrupt Disable
- **SVACC:** Slave Access Interrupt Disable
- **GACC:** General Call Access Interrupt Disable
- **OVRE:** Overrun Error Interrupt Disable
- **NACK:** Not Acknowledge Interrupt Disable
- **ARBLST:** Arbitration Lost Interrupt Disable
- **SCL\_WS:** Clock Wait State Interrupt Disable
- **EOSACC:** End Of Slave Access Interrupt Disable
- **ENDRX:** End of Receive Buffer Interrupt Disable
- **ENDTX:** End of Transmit Buffer Interrupt Disable
- **RXBUFF:** Receive Buffer Full Interrupt Disable
- **TXBUFE:** Transmit Buffer Empty Interrupt Disable

### 34.12.9 TWI Interrupt Mask Register

Name: TWI\_IMR

Address: 0x4001802C (0), 0x4001C02C (1)

Access: Read-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXBUFE	RXBUFF	ENDTX	ENDRX	EOSACC	SCL_WS	ARBLST	NACK
7	6	5	4	3	2	1	0
–	OVRE	GACC	SVACC	–	TXRDY	RXRDY	TXCOMP

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is disabled.

1: The corresponding interrupt is enabled.

- **TXCOMP: Transmission Completed Interrupt Mask**
- **RXRDY: Receive Holding Register Ready Interrupt Mask**
- **TXRDY: Transmit Holding Register Ready Interrupt Mask**
- **SVACC: Slave Access Interrupt Mask**
- **GACC: General Call Access Interrupt Mask**
- **OVRE: Overrun Error Interrupt Mask**
- **NACK: Not Acknowledge Interrupt Mask**
- **ARBLST: Arbitration Lost Interrupt Mask**
- **SCL\_WS: Clock Wait State Interrupt Mask**
- **EOSACC: End Of Slave Access Interrupt Mask**
- **ENDRX: End of Receive Buffer Interrupt Mask**
- **ENDTX: End of Transmit Buffer Interrupt Mask**
- **RXBUFF: Receive Buffer Full Interrupt Mask**
- **TXBUFE: Transmit Buffer Empty Interrupt Mask**

### 34.12.10 TWI Receive Holding Register

Name: TWI\_RHR

Address: 0x40018030 (0), 0x4001C030 (1)

Access: Read-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
RXDATA							

- RXDATA: Master or Slave Receive Holding Data

### 34.12.11 TWI Transmit Holding Register

Name: TWI\_THR

Address: 0x40018034 (0), 0x4001C034 (1)

Access: Write-only

Reset: 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
TXDATA							

- TXDATA: Master or Slave Transmit Holding Data

### 34.12.12 TWI Write Protection Mode Register

**Name:** TWI\_WPMR

**Address:** 0x400180E4 (0), 0x4001C0E4 (1)

**Access:** Read-write

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

- **WPEN: Write Protect Enable**

0: Disables the Write Protect if WPKEY corresponds to 0x545749 ("TWI" in ASCII).

1: Enables the Write Protect if WPKEY corresponds to 0x545749 ("TWI" in ASCII).

The write protected registers are:

- ["TWI Clock Waveform Generator Register"](#)
- ["TWI Slave Mode Register"](#)

- **WPKEY: Write Protect Key**

Value	Name	Description
0x545749	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0



### 34.12.13 TWI Write Protection Status Register

**Name:** TWI\_WPSR

**Address:** 0x400180E8 (0), 0x4001C0E8 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
WPVSR							
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

- **WPVS: Write Protect Violation Status**

0: No Write Protect Violation has occurred since the last read of the TWI\_WPSR.

1: A Write Protect Violation has occurred since the last read of the TWI\_WPSR. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protect Violation Source**

When WPVS is active, this field indicates the write-protected register (through address offset or code) in which a write access has been attempted.

**Note:** Reading TWI\_WPSR automatically clears all fields.

# 35. Universal Asynchronous Receiver Transmitter (UART)

## 35.1 Description

The Universal Asynchronous Receiver Transmitter (UART) features a two-pin UART that can be used for communication and trace purposes and offers an ideal medium for in-situ programming solutions.

Moreover, the association with a peripheral DMA controller (PDC) permits packet handling for these tasks with processor time reduced to a minimum.

The optical link transceiver establishes electrically isolated serial communication with hand-held equipment, such as calibrators compliant with ANSI-C12.18 or IEC62056-21 norms.

## 35.2 Embedded Characteristics

- Two-pin UART
  - Independent Receiver and Transmitter with a Common Programmable Baud Rate Generator
  - Even, Odd, Mark or Space Parity Generation
  - Parity, Framing and Overrun Error Detection
  - Automatic Echo, Local Loopback and Remote Loopback Channel Modes
  - Digital Filter on Receive Line
  - Interrupt Generation
  - Support for Two PDC Channels with Connection to Receiver and Transmitter
  - Optical Link Transceiver for Communication Compliant with ANSI-C12.18 or IEC62056-21 Norms

## 35.3 Block Diagram

Figure 35-1. UART Functional Block Diagram

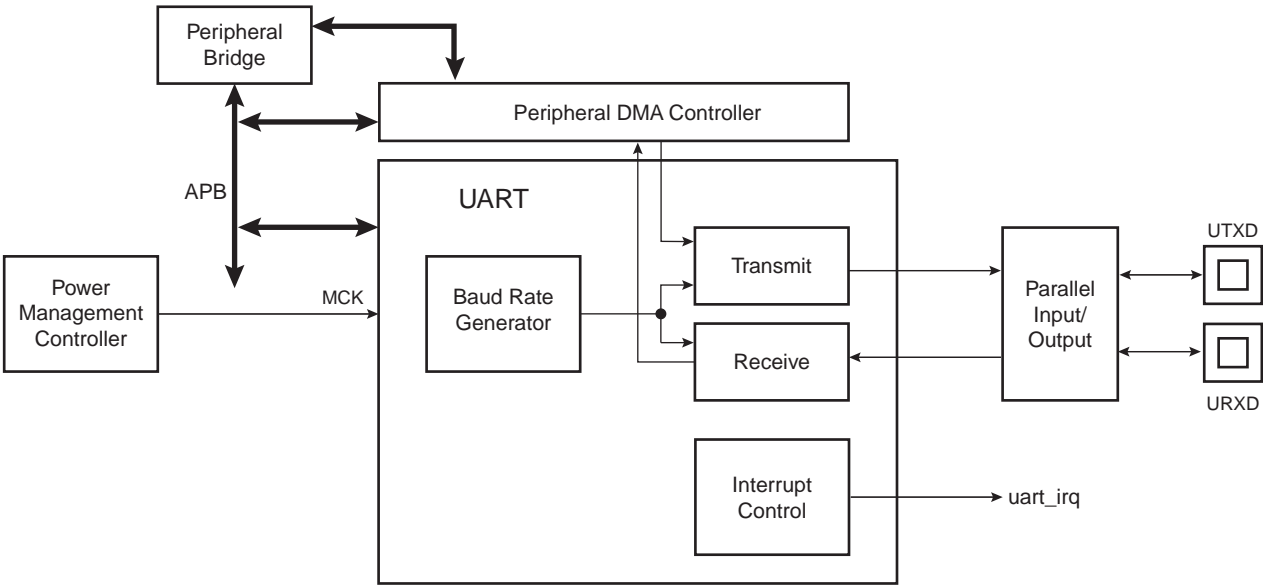


Table 35-1. UART Pin Description

Pin Name	Description	Type
URXD	UART Receive Data	Input
UTXD	UART Transmit Data	Output

## 35.4 Product Dependencies

### 35.4.1 I/O Lines

The UART pins are multiplexed with PIO lines. The user must first configure the corresponding PIO Controller to enable I/O line operations of the UART.

**Table 35-2. I/O Lines**

Instance	Signal	I/O Line	Peripheral
UART0	URXD0	PB4	A
UART0	UTXD0	PB5	A
UART1	URXD1	PC1	A
UART1	UTXD1	PC0	A

### 35.4.2 Power Management

The UART clock can be controlled through the Power Management Controller (PMC). In this case, the user must first configure the PMC to enable the UART clock. Usually, the peripheral identifier used for this purpose is 1.

### 35.4.3 Interrupt Source

The UART interrupt line is connected to one of the interrupt sources of the Interrupt Controller. Interrupt handling requires programming of the Interrupt Controller before configuring the UART.

### 35.4.4 Optical Interface

The UART optical interface requires configuration of the PMC to generate 4096 kHz or 8192 kHz on the PLLA prior to any transfer.

## 35.5 UART Operations

The UART operates in asynchronous mode only and supports only 8-bit character handling (with parity). It has no clock pin.

The UART is made up of a receiver and a transmitter that operate independently, and a common baud rate generator. Receiver timeout and transmitter time guard are not implemented. However, all the implemented features are compatible with those of a standard USART.

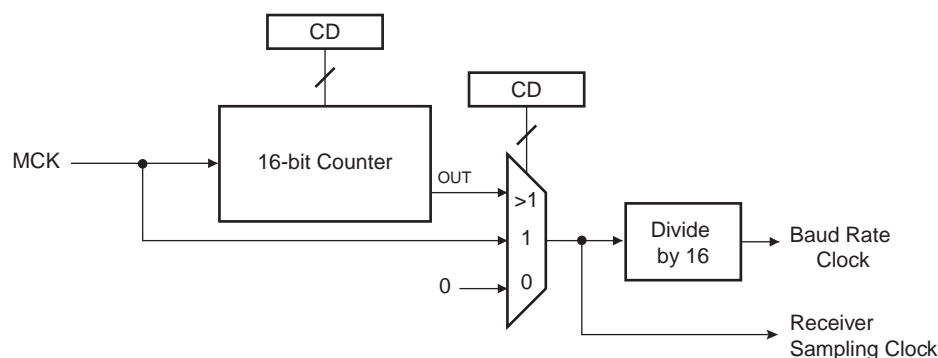
### 35.5.1 Baud Rate Generator

The baud rate generator provides the bit period clock named baud rate clock to both the receiver and the transmitter.

The baud rate clock is the master clock divided by 16 times the value (CD) written in UART\_BRGR (Baud Rate Generator Register). If UART\_BRGR is set to 0, the baud rate clock is disabled and the UART remains inactive. The maximum allowable baud rate is Master Clock divided by 16. The minimum allowable baud rate is Master Clock divided by (16 x 65536).

$$\text{Baud Rate} = \frac{\text{MCK}}{16 \times \text{CD}}$$

**Figure 35-2. Baud Rate Generator**



## 35.5.2 Receiver

### 35.5.2.1 Receiver Reset, Enable and Disable

After device reset, the UART receiver is disabled and must be enabled before being used. The receiver can be enabled by writing the Control register (UART\_CR) with the bit RXEN at 1. At this command, the receiver starts looking for a start bit.

The programmer can disable the receiver by writing UART\_CR with the bit RXDIS at 1. If the receiver is waiting for a start bit, it is immediately stopped. However, if the receiver has already detected a start bit and is receiving the data, it waits for the stop bit before actually stopping its operation.

The receiver can be put in reset state by writing UART\_CR with the bit RSTRX at 1. In this case, the receiver immediately stops its current operations and is disabled, whatever its current state. If RSTRX is applied when data is being processed, this data is lost.

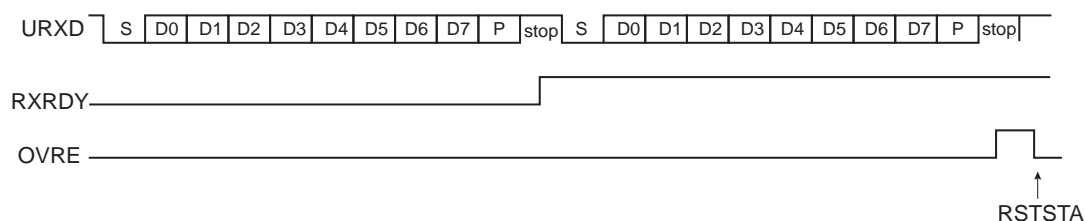
### 35.5.2.2 Start Detection and Data Sampling

The UART only supports asynchronous operations, and this affects only its receiver. The UART receiver detects the start of a received character by sampling the URXD signal until it detects a valid start bit. A low level (space) on URXD is interpreted as a valid start bit if it is detected for more than seven cycles of the sampling clock, which is 16 times the baud rate. Hence, a space that is longer than 7/16 of the bit period is detected as a valid start bit. A space which is 7/16 of a bit period or shorter is ignored and the receiver continues to wait for a valid start bit.

When a valid start bit has been detected, the receiver samples the URXD at the theoretical midpoint of each bit. It is assumed that each bit lasts 16 cycles of the sampling clock (1-bit period) so the bit sampling point is eight cycles (0.5-bit period) after the start of the bit. The first sampling point is therefore 24 cycles (1.5-bit periods) after detecting the falling edge of the start bit.

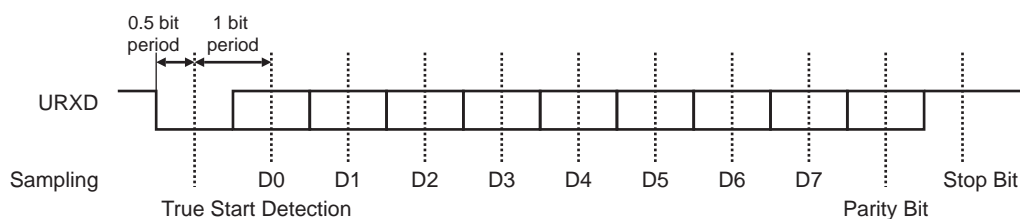
Each subsequent bit is sampled 16 cycles (1-bit period) after the previous one.

**Figure 35-3. Start Bit Detection**



**Figure 35-4. Character Reception**

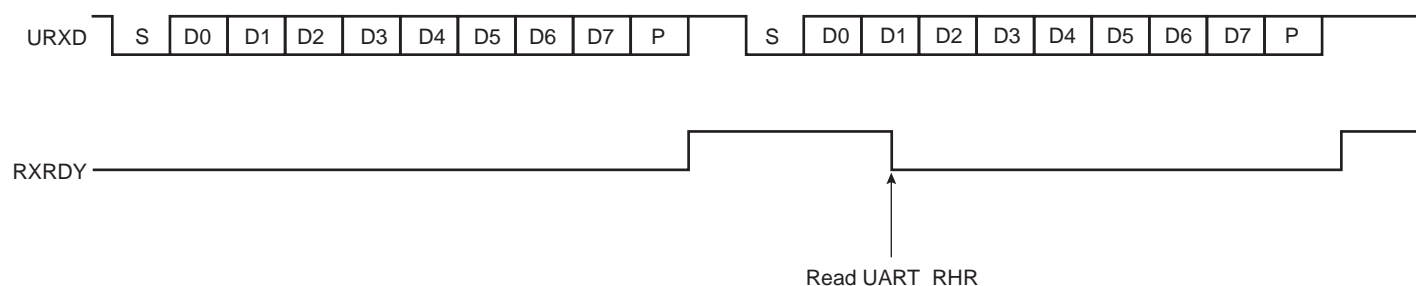
Example: 8-bit, parity enabled 1 stop



### 35.5.2.3 Receiver Ready

When a complete character is received, it is transferred to the Receive Holding register (UART\_RHR) and the RXRDY status bit in the Status register (UART\_SR) is set. The bit RXRDY is automatically cleared when UART\_RHR is read.

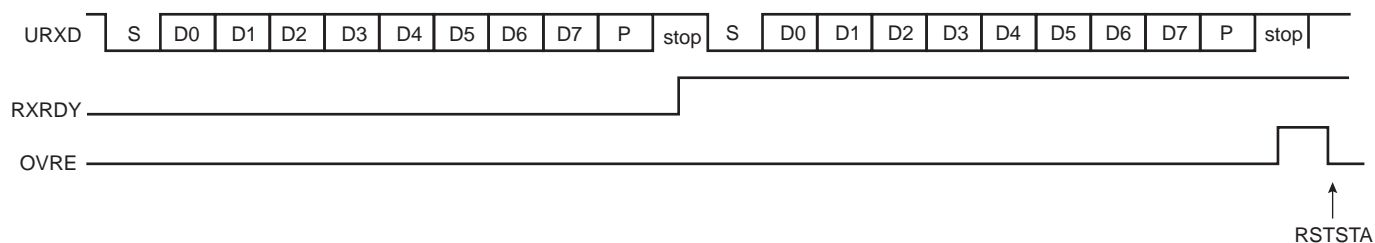
**Figure 35-5. Receiver Ready**



### 35.5.2.4 Receiver Overrun

The OVRE status bit in UART\_SR is set if UART\_RHR has not been read by the software (or the Peripheral Data Controller or DMA Controller) since the last transfer, the RXRDY bit is still set and a new character is received. OVRE is cleared when the software writes a 1 to the bit RSTSTA (Reset Status) in UART\_CR.

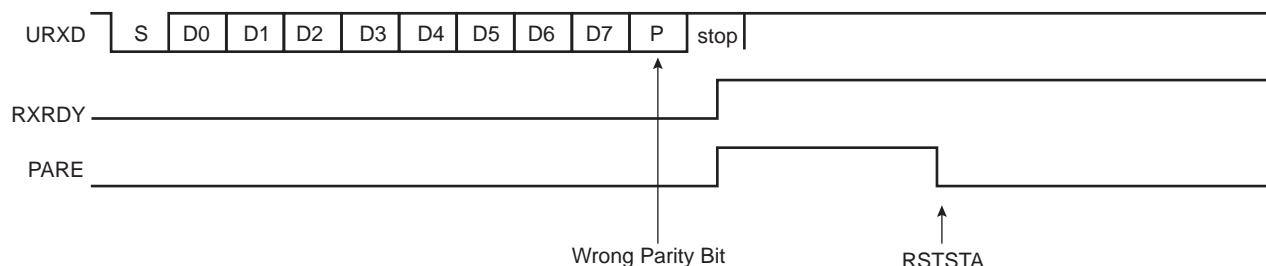
**Figure 35-6. Receiver Overrun**



### 35.5.2.5 Parity Error

Each time a character is received, the receiver calculates the parity of the received data bits, in accordance with the field PAR in the Mode register (UART\_MR). It then compares the result with the received parity bit. If different, the parity error bit PARE in UART\_SR is set at the same time RXRDY is set. The parity bit is cleared when UART\_CR is written with the bit RSTSTA (Reset Status) at 1. If a new character is received before the reset status command is written, the PARE bit remains at 1.

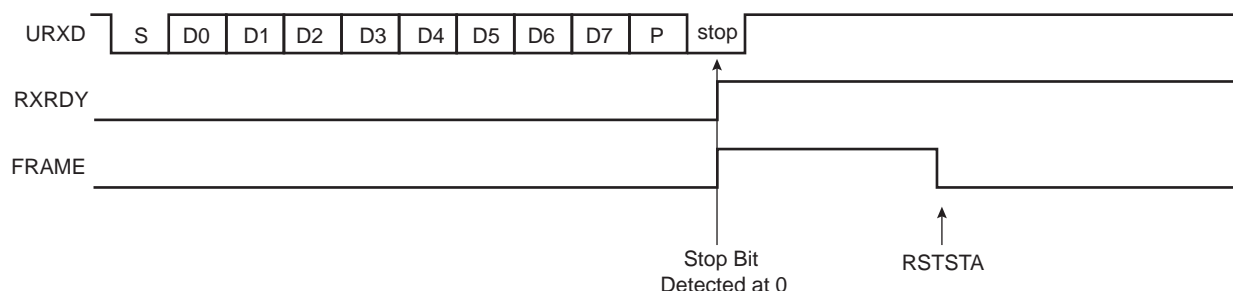
**Figure 35-7. Parity Error**



### 35.5.2.6 Receiver Framing Error

When a start bit is detected, it generates a character reception when all the data bits have been sampled. The stop bit is also sampled and when it is detected at 0, the FRAME (Framing Error) bit in UART\_SR is set at the same time the RXRDY bit is set. The FRAME bit remains high until the control register UART\_CR is written with the bit RSTSTA at 1.

**Figure 35-8. Receiver Framing Error**



### 35.5.2.7 Receiver Digital Filter

The UART embeds a digital filter on the receive line. It is disabled by default and can be enabled by writing a logical 1 in the FILTER bit of UART\_MR. When enabled, the receive line is sampled using the 16x bit clock and a three-sample filter (majority 2 over 3) determines the value of the line.

## 35.5.3 Transmitter

### 35.5.3.1 Transmitter Reset, Enable and Disable

After device reset, the UART transmitter is disabled and must be enabled before being used. The transmitter is enabled by writing UART\_CR with the bit TXEN at 1. From this command, the transmitter waits for a character to be written in the Transmit Holding Register (UART\_THR) before actually starting the transmission.

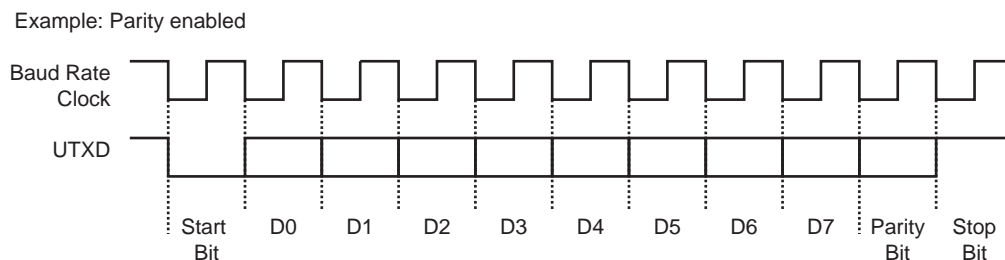
The programmer can disable the transmitter by writing UART\_CR with the bit TXDIS at 1. If the transmitter is not operating, it is immediately stopped. However, if a character is being processed into the Shift Register and/or a character has been written in the Transmit Holding Register, the characters are completed before the transmitter is actually stopped.

The programmer can also put the transmitter in its reset state by writing the UART\_CR with the bit RSTTX at 1. This immediately stops the transmitter, whether or not it is processing characters.

### 35.5.3.2 Transmit Format

The UART transmitter drives the pin UTXD at the baud rate clock speed. The line is driven depending on the format defined in UART\_MR and the data stored in the Shift Register. One start bit at level 0, then the 8 data bits, from the lowest to the highest bit, one optional parity bit and one stop bit at 1 are consecutively shifted out as shown in the following figure. The field PARE in UART\_MR defines whether or not a parity bit is shifted out. When a parity bit is enabled, it can be selected between an odd parity, an even parity, or a fixed space or mark bit.

**Figure 35-9. Character Transmission**

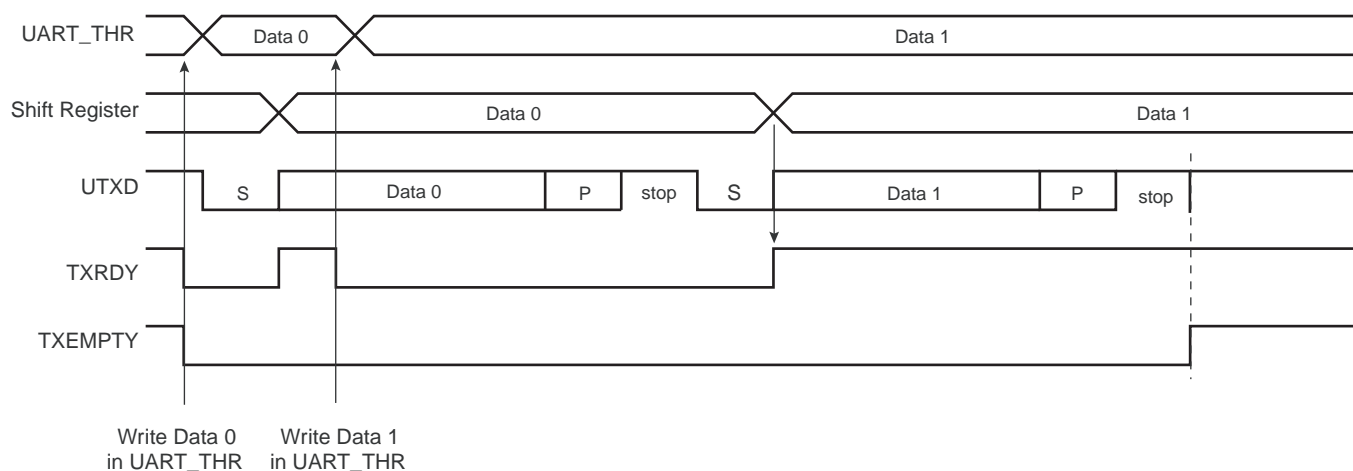


### 35.5.3.3 Transmitter Control

When the transmitter is enabled, the bit TXRDY (Transmitter Ready) is set in UART\_SR. The transmission starts when the programmer writes in the Transmit Holding register (UART\_THR), and after the written character is transferred from UART\_THR to the Shift Register. The TXRDY bit remains high until a second character is written in UART\_THR. As soon as the first character is completed, the last character written in UART\_THR is transferred into the shift register and TXRDY rises again, showing that the holding register is empty.

When both the Shift Register and UART\_THR are empty, i.e., all the characters written in UART\_THR have been processed, the TXEMPTY bit rises after the last stop bit has been completed.

**Figure 35-10. Transmitter Control**



### 35.5.4 Optical Interface

To use the optical interface circuitry, the PLLA clock must be ready and programmed to generate a frequency within the range of 4096 up to 8192 kHz. This range allows a modulation by a clock with an adjustable frequency from 30 up to 60 kHz.

The optical interface is enabled by writing a 1 to OPT\_EN in US\_MR (see [“UART Mode Register” on page 702](#)).

When OPT\_EN=1, the URXD pad is automatically configured in analog mode and the analog comparator is enabled (see [Figure 35-11 on page 696](#)).

To match the characteristics of the off-chip optical receiver circuitry, the voltage reference threshold of the embedded comparator can be adjusted from VDDIO/10 up to VDD/2 by programming the OPT\_CMPH field in US\_MR.

The NRZ output of the UART transmitter sub-module is modulated with the 30 up to 60 kHz modulation clock prior to driving the PIO controller.

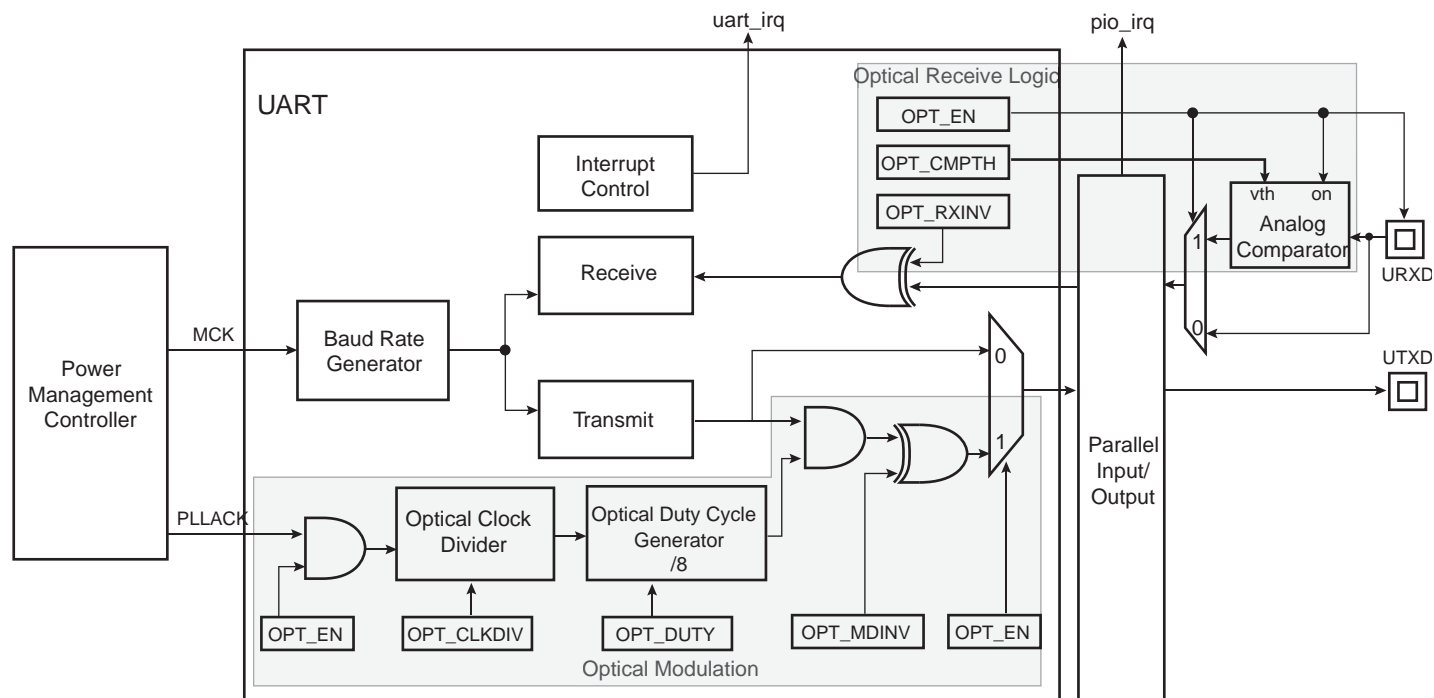
A logical 0 on the UART transmitter sub-module output generates the said modulated signal (see [Figure 35-12 on page 697](#)) having a frequency programmable from 30 kHz up to 60 kHz (38 kHz is the default value assuming the PLLA clock

frequency is 8192 kHz). A logical 1 on the UART transmitter sub-module output generates a stuck-at 1 output signal (no modulation). The idle polarity of the modulated signal is 1 (OPT\_MDIV=0 in US\_MR).

The idle polarity of the modulated signal can be inverted by programming 1 into the OPT\_MDIV field in US\_MR.

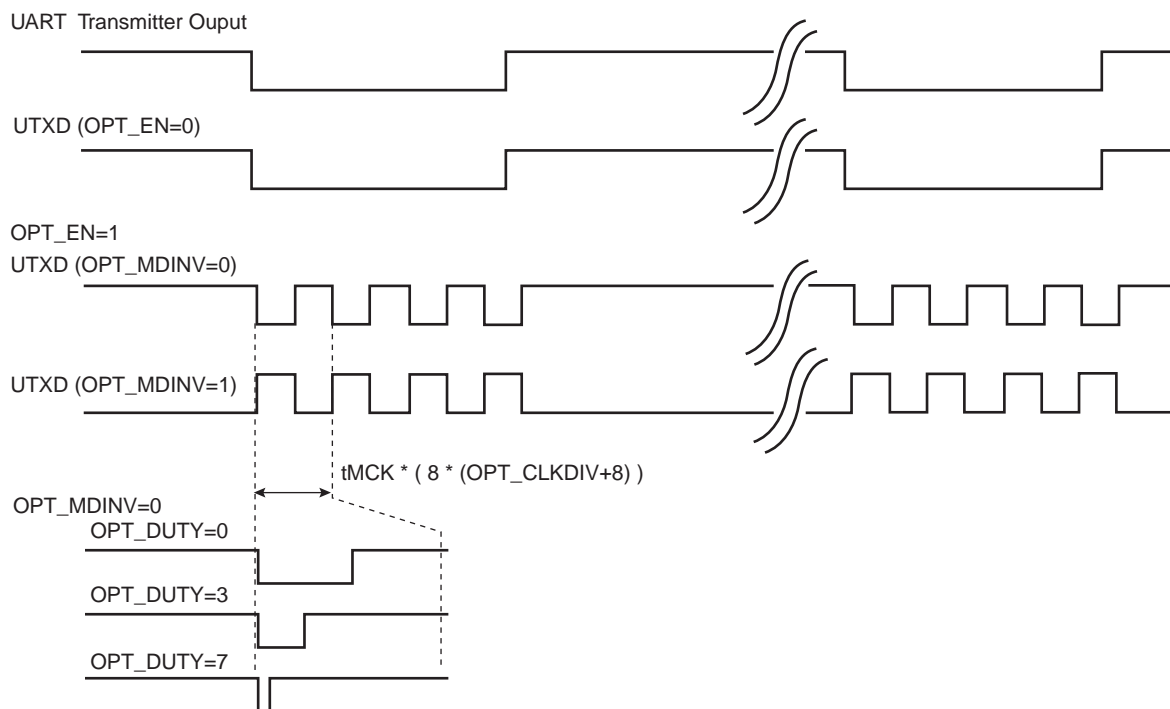
The duty cycle of the modulated signal can be adjusted from 6.25% up to 50% (default value) by steps of 6.25% by programming the OPT\_DUTY field in US\_MR.

**Figure 35-11. Optical Interface Block Diagram**





**Figure 35-12. Optical Interface Waveforms**



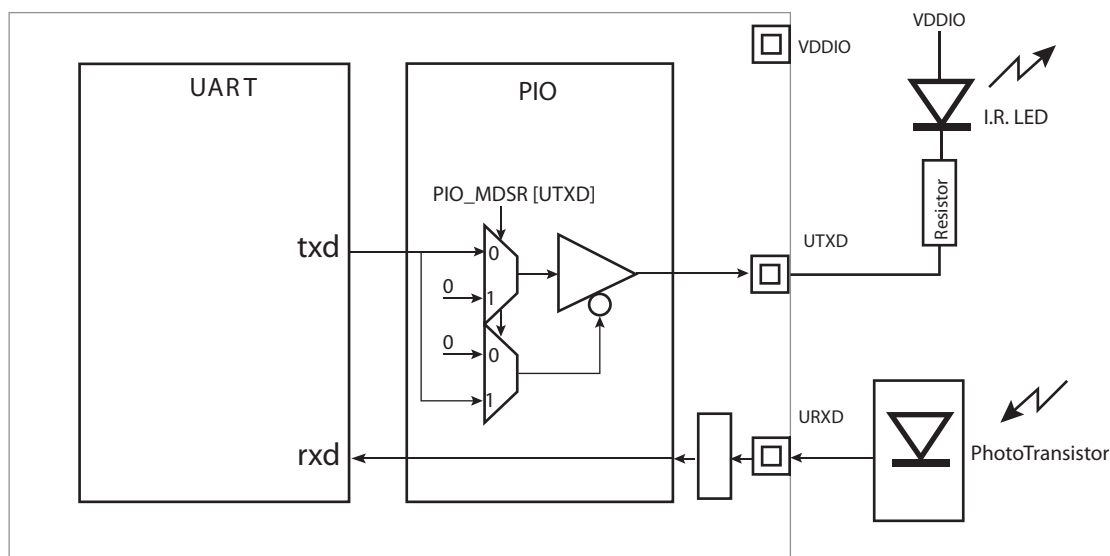
The default configuration values of the optical link circuitries allow the 38 kHz modulation, a 50% duty cycle and an idle polarity allowing a direct drive of an IR LED through a resistor (see [Figure 35-13 on page 698](#)).

Refer to the Electrical Characteristics section for drive capability of the buffer associated with the UTXD output.

In case of direct drive of the IR LED as shown in [Figure 35-13 on page 698](#), the PIO must be programmed in multi-driver mode (open-drain). To do so, the adequate index and values must be programmed into the PIO Multi-driver Enable (PIO\_MDER) register (status reported on the PIO Multi-driver Status (PIO\_MDSR) register). Refer to the PIO section for details.

If an off-chip current amplifier is used to drive the transmitting of the IR LED, the PIO may be programmed in default drive mode (non open-drain) for the line index driving the UTXD output, or in open-drain mode depending on the type of external circuitry.

**Figure 35-13. Optical Interface Connected to IR Components**



### 35.5.5 Peripheral DMA Controller (PDC)

Both the receiver and the transmitter of the UART are connected to a PDC.

The peripheral data controller channels are programmed via registers that are mapped within the UART user interface from the offset 0x100. The status bits are reported in UART\_SR and generate an interrupt.

The RXRDY bit triggers the PDC channel data transfer of the receiver. This results in a read of the data in UART\_RHR. The TXRDY bit triggers the PDC channel data transfer of the transmitter. This results in a write of data in UART\_THR.

### 35.5.6 Test Modes

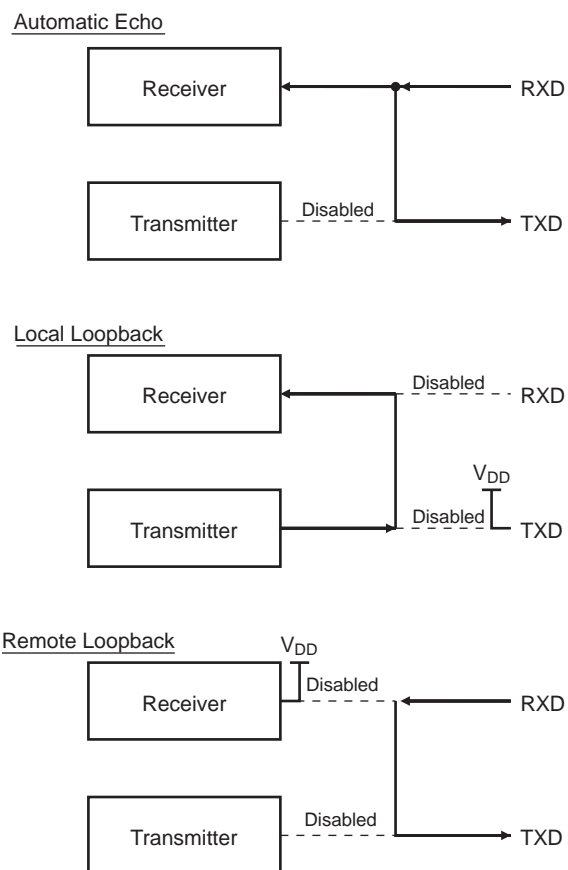
The UART supports three test modes. These modes of operation are programmed by using the CHMODE field in UART\_MR.

The automatic echo mode allows bit-by-bit retransmission. When a bit is received on the URXD line, it is sent to the UTXD line. The transmitter operates normally, but has no effect on the UTXD line.

The local loopback mode allows the transmitted characters to be received. UTXD and URXD pins are not used and the output of the transmitter is internally connected to the input of the receiver. The URXD pin level has no effect and the UTXD line is held high, as in idle state.

The remote loopback mode directly connects the URXD pin to the UTXD line. The transmitter and the receiver are disabled and have no effect. This mode allows a bit-by-bit retransmission.

Figure 35-14. Test Modes



## 35.6 Universal Asynchronous Receiver Transmitter (UART) User Interface

Table 35-3. Register Mapping

Offset	Register	Name	Access	Reset
0x0000	Control Register	UART_CR	Write-only	–
0x0004	Mode Register	UART_MR	Read/Write	0x0013_0000
0x0008	Interrupt Enable Register	UART_IER	Write-only	–
0x000C	Interrupt Disable Register	UART_IDR	Write-only	–
0x0010	Interrupt Mask Register	UART_IMR	Read-only	0x0
0x0014	Status Register	UART_SR	Read-only	–
0x0018	Receive Holding Register	UART_RHR	Read-only	0x0
0x001C	Transmit Holding Register	UART_THR	Write-only	–
0x0020	Baud Rate Generator Register	UART_BRGR	Read/Write	0x0
0x0024 - 0x003C	Reserved	–	–	–
0x0040 - 0x00E8	Reserved	–	–	–
0x00EC - 0x00FC	Reserved	–	–	–
0x0100 - 0x0128	Reserved for PDC registers	–	–	–

### 35.6.1 UART Control Register

**Name:** UART\_CR

**Address:** 0x400E0600 (0), 0x48004000 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	RSTSTA
7	6	5	4	3	2	1	0
TXDIS	TXEN	RXDIS	RXEN	RSTTX	RSTRX	–	–

- **RSTRX: Reset Receiver**

0: No effect.

1: The receiver logic is reset and disabled. If a character is being received, the reception is aborted.

- **RSTTX: Reset Transmitter**

0: No effect.

1: The transmitter logic is reset and disabled. If a character is being transmitted, the transmission is aborted.

- **RXEN: Receiver Enable**

0: No effect.

1: The receiver is enabled if RXDIS is 0.

- **RXDIS: Receiver Disable**

0: No effect.

1: The receiver is disabled. If a character is being processed and RSTRX is not set, the character is completed before the receiver is stopped.

- **TXEN: Transmitter Enable**

0: No effect.

1: The transmitter is enabled if TXDIS is 0.

- **TXDIS: Transmitter Disable**

0: No effect.

1: The transmitter is disabled. If a character is being processed and a character has been written in the UART\_THR and RSTTX is not set, both characters are completed before the transmitter is stopped.

- **RSTSTA: Reset Status Bits**

0: No effect.

1: Resets the status bits PARE, FRAME and OVRE in the UART\_SR.

### 35.6.2 UART Mode Register

**Name:** UART\_MR

**Address:** 0x400E0604 (0), 0x48004004 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
—	OPT_CMPTH			—	OPT_DUTY		
23	22	21	20	19	18	17	16
—	—	—	OPT_CLKDIV				
15	14	13	12	11	10	9	8
CHMODE		—	—	PAR			—
7	6	5	4	3	2	1	0
—	—	—	FILTER	—	OPT_MDINV	OPT_RXINV	OPT_EN

- **OPT\_EN: UART Optical Interface Enable**

Value	Name	Description
0	DISABLED	The UART TX data is not inverted before modulation.
1	ENABLED	The UART TX data is inverted before modulation.

- **OPT\_RXINV: UART Receive Data Inverted**

Value	Name	Description
0	DISABLED	The Comparator Data Output is not inverted before entering UART.
1	ENABLED	The Comparator Data Output is inverted before entering UART.

- **OPT\_MDINV: UART Modulated Data Inverted**

Value	Name	Description
0	DISABLED	The Output of the Modulator is not inverted.
1	ENABLED	The Output of the Modulator is inverted.

- **FILTER: Receiver Digital Filter**

0 (DISABLED): UART does not filter the receive line.

1 (ENABLED): UART filters the receive line using a three-sample filter (16x-bit clock) (2 over 3 majority).

- **PAR: Parity Type**

Value	Name	Description
0	EVEN	Even Parity
1	ODD	Odd Parity
2	SPACE	Space: parity forced to 0
3	MARK	Mark: parity forced to 1
4	NO	No parity

- **CHMODE: Channel Mode**

Value	Name	Description
0	NORMAL	Normal mode
1	AUTOMATIC	Automatic echo
2	LOCAL_LOOPBACK	Local loopback
3	REMOTE_LOOPBACK	Remote loopback

- **OPT\_CLKDIV: Optical Link Clock Divider**

0 to 31: The optical modulation clock frequency is defined by  $PLLACK / (8 * (OPT\_CLKDIV + 8))$ .

- **OPT\_DUTY: Optical Link Modulation Clock Duty Cycle**

Value	Name	Description
0	DUTY_50	Modulation clock duty cycle is 50%.
1	DUTY_43P75	Modulation clock duty cycle is 43.75%.
2	DUTY_37P5	Modulation clock duty cycle is 37.5%.
3	DUTY_31P25	Modulation clock duty cycle is 31.75%.
4	DUTY_25	Modulation clock duty cycle is 25%.
5	DUTY_18P75	Modulation clock duty cycle is 18.75%.
6	DUTY_12P5	Modulation clock duty cycle is 12.5%.
7	DUTY_6P25	Modulation clock duty cycle is 6.25%.

- **OPT\_CMPH: Receive Path Comparator Threshold**

Value	Name	Description
0	VDDIO_DIV2	Comparator threshold is VDDIO/2 Volts.
1	VDDIO_DIV2P5	Comparator threshold is VDDIO/2.5 Volts.
2	VDDIO_DIV3P3	Comparator threshold is VDDIO/3.3 Volts.
3	VDDIO_DIV5	Comparator threshold is VDDIO/5 Volts.
4	VDDIO_DIV10	Comparator threshold is VDDIO/10 Volts.

### 35.6.3 UART Interrupt Enable Register

**Name:** UART\_IER

**Address:** 0x400E0608 (0), 0x48004008 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	–	TXEMPTY	–
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Enables the corresponding interrupt.

- **RXRDY: Enable RXRDY Interrupt**
- **TXRDY: Enable TXRDY Interrupt**
- **ENDRX: Enable End of Receive Transfer Interrupt**
- **ENDTX: Enable End of Transmit Interrupt**
- **OVRE: Enable Overrun Error Interrupt**
- **FRAME: Enable Framing Error Interrupt**
- **PARE: Enable Parity Error Interrupt**
- **TXEMPTY: Enable TXEMPTY Interrupt**
- **TXBUFE: Enable Buffer Empty Interrupt**
- **RXBUFF: Enable Buffer Full Interrupt**



### 35.6.4 UART Interrupt Disable Register

**Name:** UART\_IDR

**Address:** 0x400E060C (0), 0x4800400C (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	RXBUFF	TXBUFE	—	TXEMPTY	—
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	—	TXRDY	RXRDY

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Disables the corresponding interrupt.

- **RXRDY: Disable RXRDY Interrupt**
- **TXRDY: Disable TXRDY Interrupt**
- **ENDRX: Disable End of Receive Transfer Interrupt**
- **ENDTX: Disable End of Transmit Interrupt**
- **OVRE: Disable Overrun Error Interrupt**
- **FRAME: Disable Framing Error Interrupt**
- **PARE: Disable Parity Error Interrupt**
- **TXEMPTY: Disable TXEMPTY Interrupt**
- **TXBUFE: Disable Buffer Empty Interrupt**
- **RXBUFF: Disable Buffer Full Interrupt**

### 35.6.5 UART Interrupt Mask Register

**Name:** UART\_IMR

**Address:** 0x400E0610 (0), 0x48004010 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	–	TXEMPTY	–
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is disabled.

1: The corresponding interrupt is enabled.

- **RXRDY: Mask RXRDY Interrupt**
- **TXRDY: Disable TXRDY Interrupt**
- **ENDRX: Mask End of Receive Transfer Interrupt**
- **ENDTX: Mask End of Transmit Interrupt**
- **OVRE: Mask Overrun Error Interrupt**
- **FRAME: Mask Framing Error Interrupt**
- **PARE: Mask Parity Error Interrupt**
- **TXEMPTY: Mask TXEMPTY Interrupt**
- **TXBUFE: Mask TXBUFE Interrupt**
- **RXBUFF: Mask RXBUFF Interrupt**

### 35.6.6 UART Status Register

**Name:** UART\_SR

**Address:** 0x400E0614 (0), 0x48004014 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	RXBUFF	TXBUFE	—	TXEMPTY	—
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	—	TXRDY	RXRDY

- **RXRDY: Receiver Ready**

0: No character has been received since the last read of the UART\_RHR, or the receiver is disabled.

1: At least one complete character has been received, transferred to UART\_RHR and not yet read.

- **TXRDY: Transmitter Ready**

0: A character has been written to UART\_THR and not yet transferred to the Shift Register, or the transmitter is disabled.

1: There is no character written to UART\_THR not yet transferred to the Shift Register.

- **ENDRX: End of Receiver Transfer**

0: The end of transfer signal from the receiver Peripheral Data Controller channel is inactive.

1: The end of transfer signal from the receiver Peripheral Data Controller channel is active.

- **ENDTX: End of Transmitter Transfer**

0: The end of transfer signal from the transmitter Peripheral Data Controller channel is inactive.

1: The end of transfer signal from the transmitter Peripheral Data Controller channel is active.

- **OVRE: Overrun Error**

0: No overrun error has occurred since the last RSTSTA.

1: At least one overrun error has occurred since the last RSTSTA.

- **FRAME: Framing Error**

0: No framing error has occurred since the last RSTSTA.

1: At least one framing error has occurred since the last RSTSTA.

- **PARE: Parity Error**

0: No parity error has occurred since the last RSTSTA.

1: At least one parity error has occurred since the last RSTSTA.

- **TXEMPTY: Transmitter Empty**

0: There are characters in UART\_THR, or characters being processed by the transmitter, or the transmitter is disabled.

1: There are no characters in UART\_THR and there are no characters being processed by the transmitter.

- **TXBUFE: Transmission Buffer Empty**

0: The buffer empty signal from the transmitter PDC channel is inactive.

1: The buffer empty signal from the transmitter PDC channel is active.

- **RXBUFF: Receive Buffer Full**

0: The buffer full signal from the receiver PDC channel is inactive.

1: The buffer full signal from the receiver PDC channel is active.

35.6.7 UART Receiver Holding Register

**Name:** UART\_RHR  
**Address:** 0x400E0618 (0), 0x48004018 (1)  
**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
RXCHR							

- **RXCHR: Received Character**  
Last received character if RXRDY is set.

### 35.6.8 UART Transmit Holding Register

**Name:** UART\_THR

**Address:** 0x400E061C (0), 0x4800401C (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
TXCHR							

- **TXCHR: Character to be Transmitted**

Next character to be transmitted after the current character if TXRDY is not set.

35.6.9 UART Baud Rate Generator Register

**Name:** UART\_BRGR  
**Address:** 0x400E0620 (0), 0x48004020 (1)  
**Access:** Read/Write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
CD							
7	6	5	4	3	2	1	0
CD							

- **CD: Clock Divisor**  
0: Baud Rate Clock is disabled  
1 to 65,535:  $MCK / (CD \times 16)$

## 36. Universal Synchronous Asynchronous Receiver Transceiver (USART)

### 36.1 TDescription

The Universal Synchronous Asynchronous Receiver Transceiver (USART) provides one full duplex universal synchronous asynchronous serial link. Data frame format is widely programmable (data length, parity, number of stop bits) to support a maximum of standards. The receiver implements parity error, framing error and overrun error detection. The receiver time-out enables handling variable-length frames and the transmitter timeguard facilitates communications with slow remote devices. Multidrop communications are also supported through address bit handling in reception and transmission.

The USART features three test modes: remote loopback, local loopback and automatic echo.

The USART supports specific operating modes providing interfaces on RS485 and SPI buses, with ISO7816 T = 0 or T = 1 smart card slots and infrared transceivers. The hardware handshaking feature enables an out-of-band flow control by automatic management of the pins RTS and CTS.

The USART supports the connection to the Peripheral DMA Controller, which enables data transfers to the transmitter and from the receiver. The PDC provides chained buffer management without any intervention of the processor.

### 36.2 Embedded Characteristics

- Programmable Baud Rate Generator
- 5- to 9-bit Full-duplex Synchronous or Asynchronous Serial Communications
  - 1, 1.5 or 2 Stop Bits in Asynchronous Mode or 1 or 2 Stop Bits in Synchronous Mode
  - Parity Generation and Error Detection
  - Framing Error Detection, Overrun Error Detection
  - MSB- or LSB-first
  - Optional Break Generation and Detection
  - By 8 or by 16 Over-sampling Receiver Frequency
  - Optional Hardware Handshaking RTS-CTS
  - Receiver Time-out and Transmitter Timeguard
  - Optional Multidrop Mode with Address Generation and Detection
- RS485 with Driver Control Signal
- ISO7816, T = 0 or T = 1 Protocols for Interfacing with Smart Cards
  - NACK Handling, Error Counter with Repetition and Iteration Limit
- IrDA Modulation and Demodulation
  - Communication at up to 115.2 Kbps
- SPI Mode
  - Master or Slave
  - Serial Clock Programmable Phase and Polarity
  - SPI Serial Clock (SCK) Frequency up to Internal Clock Frequency MCK/6
- Test Modes
  - Remote Loopback, Local Loopback, Automatic Echo
- Supports Connection of:
  - Two Peripheral DMA Controller Channels (PDC)
- Offers Buffer Transfer without Processor Intervention



36.3 Block Diagram

Figure 36-1. USART Block Diagram

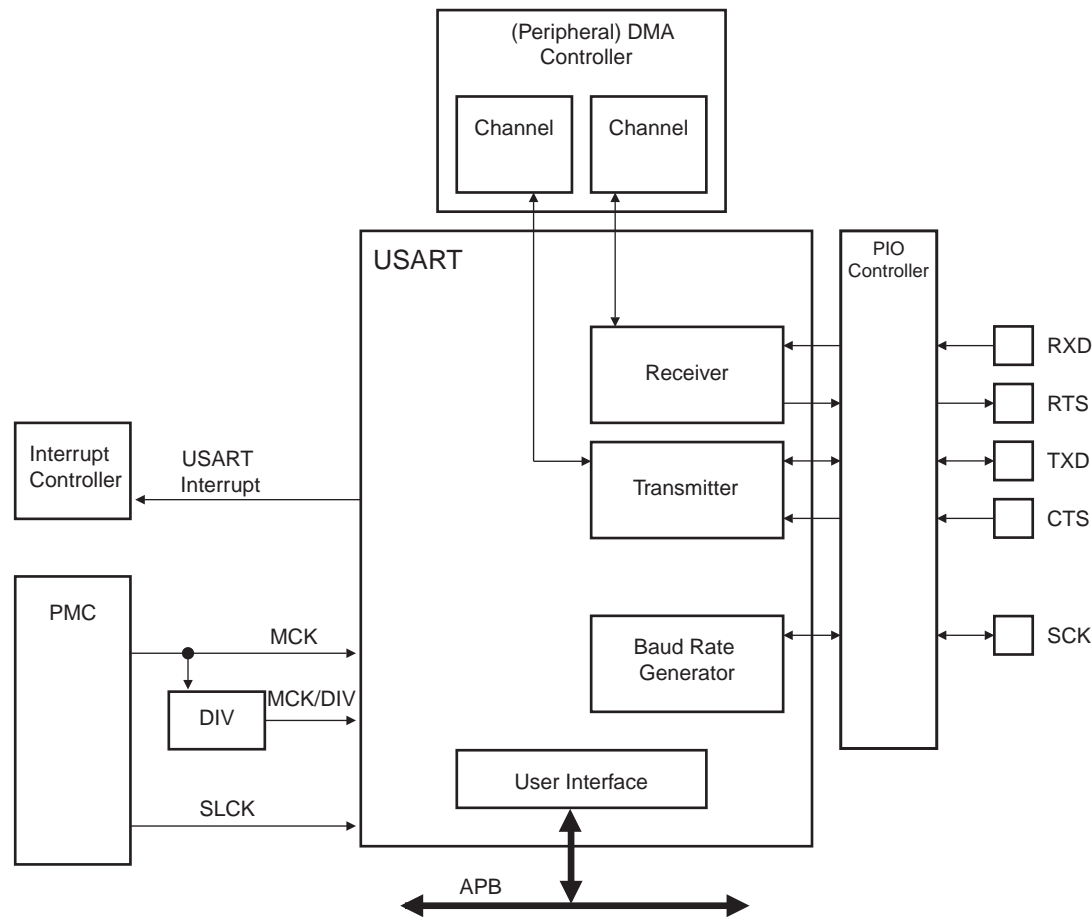
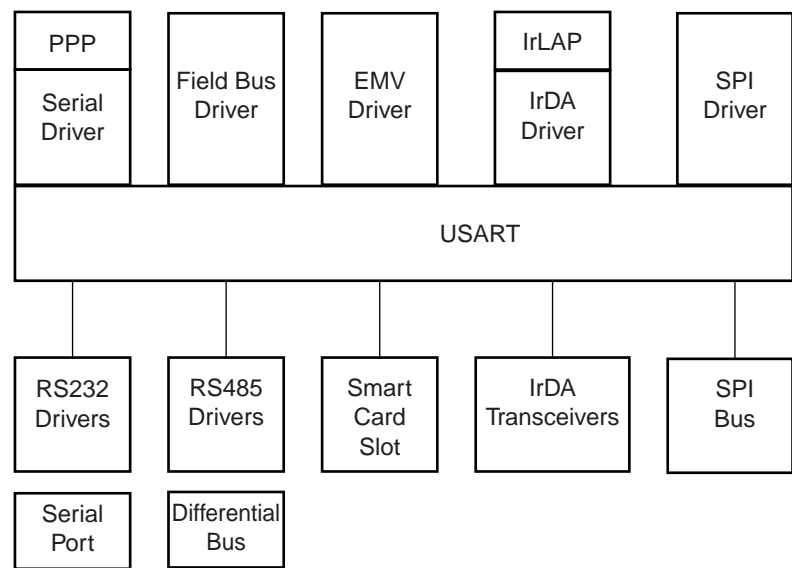


Table 36-1. SPI Operating Mode

PIN	USART	SPI Slave	SPI Master
RXD	RXD	MOSI	MISO
TXD	TXD	MISO	MOSI
RTS	RTS	–	CS
CTS	CTS	CS	–

# 36.4 Application Block Diagram

Figure 36-2. Application Block Diagram



## 36.5 I/O Lines Description

Table 36-2. I/O Line Description

Name	Description	Type	Active Level
SCK	Serial Clock	I/O	—
TXD	Transmit Serial Data or Master Out Slave In (MOSI) in SPI Master Mode or Master In Slave Out (MISO) in SPI Slave Mode	I/O	—
RXD	Receive Serial Data or Master In Slave Out (MISO) in SPI Master Mode or Master Out Slave In (MOSI) in SPI Slave Mode	Input	—
CTS	Clear to Send or Slave Select (NSS) in SPI Slave Mode	Input	Low
RTS	Request to Send or Slave Select (NSS) in SPI Master Mode	Output	Low

## 36.6 Product Dependencies

### 36.6.1 I/O Lines

The pins used for interfacing the USART may be multiplexed with the PIO lines. The programmer must first program the PIO controller to assign the desired USART pins to their peripheral function. If I/O lines of the USART are not used by the application, they can be used for other purposes by the PIO Controller.

To prevent the TXD line from falling when the USART is disabled, the use of an internal pull up is mandatory. If the hardware handshaking feature is used, the internal pull up on TXD must also be enabled.

**Table 36-3. I/O Lines**

Instance	Signal	I/O Line	Peripheral
USART0	CTS0	PA20	A
USART0	RTS0	PA19	A
USART0	RXD0	PB16	A
USART0	SCK0	PB18	A
USART0	TXD0	PB17	A
USART1	CTS1	PA18	A
USART1	RTS1	PA17	A
USART1	RXD1	PA11	A
USART1	SCK1	PA16	A
USART1	TXD1	PA12	A
USART2	CTS2	PA15	A
USART2	RTS2	PA14	A
USART2	RXD2	PA9	A
USART2	SCK2	PA13	A
USART2	TXD2	PA10	A
USART3	CTS3	PA1	A
USART3	RTS3	PA0	A
USART3	RXD3	PA3	A
USART3	SCK3	PA2	A
USART3	TXD3	PA4	A
USART4	CTS4	PA26	A
USART4	RTS4	PB22	A
USART4	RXD4	PB19	A
USART4	SCK4	PB21	A
USART4	TXD4	PB20	A

### 36.6.2 Power Management

The USART is not continuously clocked. The programmer must first enable the USART Clock in the Power Management Controller (PMC) before using the USART. However, if the application does not require USART operations, the USART

clock can be stopped when not needed and be restarted later. In this case, the USART will resume its operations where it left off.

Configuring the USART does not require the USART clock to be enabled.

### 36.6.3 Interrupt

The USART interrupt line is connected on one of the internal sources of the Interrupt Controller. Using the USART

**Table 36-4. Peripheral IDs**

Instance	ID
USART0	14
USART1	15
USART2	16
USART3	17
USART4	18

interrupt requires the Interrupt Controller to be programmed first. Note that it is not recommended to use the USART interrupt line in edge sensitive mode.

## 36.7 Functional Description

The USART is capable of managing several types of serial synchronous or asynchronous communications.

It supports the following communication modes:

- 5- to 9-bit full-duplex asynchronous serial communication
  - MSB- or LSB-first
  - 1, 1.5 or 2 stop bits
  - Parity even, odd, marked, space or none
  - By 8 or by 16 over-sampling receiver frequency
  - Optional hardware handshaking
  - Optional break management
  - Optional multidrop serial communication
- High-speed 5- to 9-bit full-duplex synchronous serial communication
  - MSB- or LSB-first
  - 1 or 2 stop bits
  - Parity even, odd, marked, space or none
  - By 8 or by 16 over-sampling frequency
  - Optional hardware handshaking
  - Optional break management
  - Optional multidrop serial communication
- RS485 with driver control signal
- ISO7816, T0 or T1 protocols for interfacing with smart cards
  - NACK handling, error counter with repetition and iteration limit, inverted data.
- InfraRed IrDA Modulation and Demodulation
- SPI Mode
  - Master or Slave
  - Serial Clock Programmable Phase and Polarity
  - SPI Serial Clock (SCK) Frequency up to Internal Clock Frequency MCK/6
- Test modes
  - Remote loopback, local loopback, automatic echo

### 36.7.1 Baud Rate Generator

The Baud Rate Generator provides the bit period clock named the Baud Rate Clock to both the receiver and the transmitter.

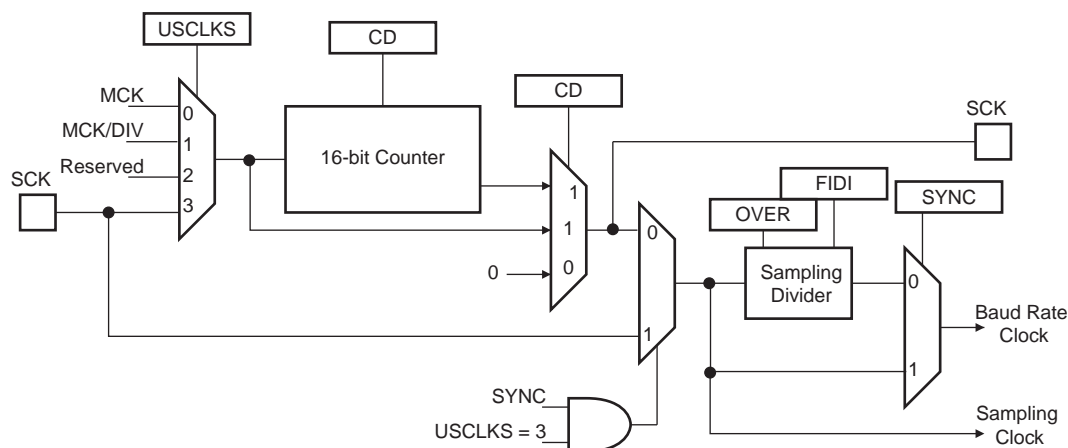
The Baud Rate Generator clock source can be selected by setting the USCLKS field in the Mode Register (US\_MR) between:

- The Master Clock MCK
- A division of the Master Clock, the divider being product dependent, but generally set to 8
- The external clock, available on the SCK pin

The Baud Rate Generator is based upon a 16-bit divider, which is programmed with the CD field of the Baud Rate Generator Register (US\_BRGR). If a zero is written to CD, the Baud Rate Generator does not generate any clock. If a one is written to CD, the divider is bypassed and becomes inactive.

If the external SCK clock is selected, the duration of the low and high levels of the signal provided on the SCK pin must be longer than a Master Clock (MCK) period. The frequency of the signal provided on SCK must be at least 3 times lower than MCK in USART mode, or 6 times lower in SPI mode.

**Figure 36-3. Baud Rate Generator**



#### 36.7.1.1 Baud Rate in Asynchronous Mode

If the USART is programmed to operate in asynchronous mode, the selected clock is first divided by CD, which is field programmed in the US\_BRGR. The resulting clock is provided to the receiver as a sampling clock and then divided by 16 or 8, depending on the programming of the OVER bit in the US\_MR.

If OVER is set, the receiver sampling is eight times higher than the baud rate clock. If OVER is cleared, the sampling is performed at 16 times the baud rate clock.

The baud rate is calculated as per the following formula:

$$Baudrate = \frac{SelectedClock}{(8(2 - Over)CD)}$$

This gives a maximum baud rate of MCK divided by 8, assuming that MCK is the highest possible clock and that the OVER bit is set.

*Baud Rate Calculation Example*

Table 36-5 shows calculations of CD to obtain a baud rate at 38,400 bit/s for different source clock frequencies. This table also shows the actual resulting baud rate and the error.

**Table 36-5. Baud Rate Example (OVER = 0)**

Source Clock (MHz)	Expected Baud Rate (Bit/s)	Calculation Result	CD	Actual Baud Rate (Bit/s)	Error
3,686,400	38,400	6.00	6	38,400.00	0.00%
4,915,200	38,400	8.00	8	38,400.00	0.00%
5,000,000	38,400	8.14	8	39,062.50	1.70%
7,372,800	38,400	12.00	12	38,400.00	0.00%
8,000,000	38,400	13.02	13	38,461.54	0.16%
12,000,000	38,400	19.53	20	37,500.00	2.40%
12,288,000	38,400	20.00	20	38,400.00	0.00%
14,318,180	38,400	23.30	23	38,908.10	1.31%
14,745,600	38,400	24.00	24	38,400.00	0.00%
18,432,000	38,400	30.00	30	38,400.00	0.00%
24,000,000	38,400	39.06	39	38,461.54	0.16%
24,576,000	38,400	40.00	40	38,400.00	0.00%
25,000,000	38,400	40.69	40	38,109.76	0.76%
32,000,000	38,400	52.08	52	38,461.54	0.16%
32,768,000	38,400	53.33	53	38,641.51	0.63%
33,000,000	38,400	53.71	54	38,194.44	0.54%
40,000,000	38,400	65.10	65	38,461.54	0.16%
50,000,000	38,400	81.38	81	38,580.25	0.47%

The baud rate is calculated with the following formula:

$$BaudRate = MCK / CD \times 16$$

The baud rate error is calculated with the following formula. It is not recommended to work with an error higher than 5%.

$$Error = 1 - \left( \frac{ExpectedBaudRate}{ActualBaudRate} \right)$$

### 36.7.1.2 Fractional Baud Rate in Asynchronous Mode

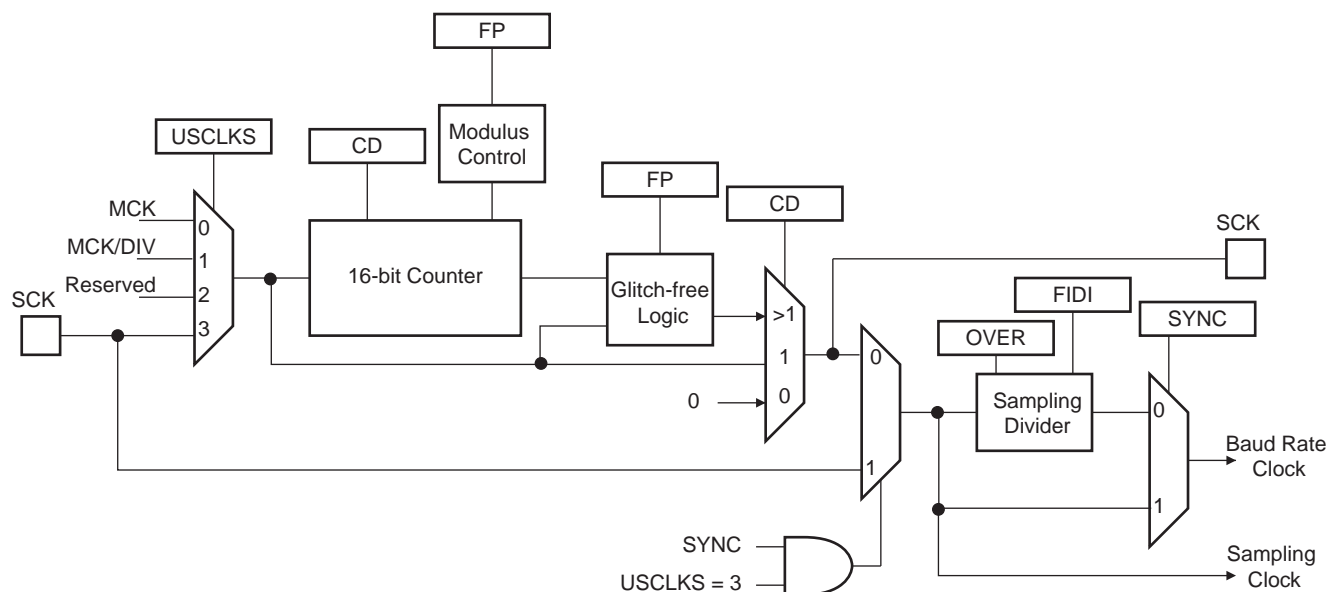
The Baud Rate generator previously defined is subject to the following limitation: the output frequency changes by only integer multiples of the reference frequency. An approach to this problem is to integrate a fractional N clock generator that has a high resolution. The generator architecture is modified to obtain baud rate changes by a fraction of the reference source clock. This fractional part is programmed with the FP field in the US\_BRGR. If FP is not 0, the fractional part is activated. The resolution is one eighth of the clock divider. This feature is only available when using USART normal mode. The fractional baud rate is calculated using the following formula:

$$Baudrate = \frac{SelectedClock}{\left( 8(2 - Over) \left( CD + \frac{FP}{8} \right) \right)}$$

The modified architecture is presented below:



**Figure 36-4. Fractional Baud Rate Generator**



### 36.7.1.3 Baud Rate in Synchronous Mode or SPI Mode

If the USART is programmed to operate in synchronous mode, the selected clock is simply divided by the field CD in the US\_BRGR.

$$\text{BaudRate} = \frac{\text{SelectedClock}}{CD}$$

In synchronous mode, if the external clock is selected (USCLKS = 3), the clock is provided directly by the signal on the USART SCK pin. No division is active. The value written in US\_BRGR has no effect. The external clock frequency must be at least 3 times lower than the system clock. In synchronous mode master (USCLKS = 0 or 1, CLK0 set to 1), the receive part limits the SCK maximum frequency to MCK/3 in USART mode, or MCK/6 in SPI mode.

When either the external clock SCK or the internal clock divided (MCK/DIV) is selected, the value programmed in CD must be even if the user has to ensure a 50:50 mark/space ratio on the SCK pin. If the internal clock MCK is selected, the Baud Rate Generator ensures a 50:50 duty cycle on the SCK pin, even if the value programmed in CD is odd.

### 36.7.1.4 Baud Rate in ISO 7816 Mode

The ISO7816 specification defines the bit rate with the following formula:

$$B = \frac{Di}{Fi} \times f$$

where:

- B is the bit rate
- Di is the bit-rate adjustment factor
- Fi is the clock frequency division factor
- f is the ISO7816 clock frequency (Hz)

Di is a binary value encoded on a 4-bit field, named DI, as represented in [Table 36-6](#).

**Table 36-6. Binary and Decimal Values for Di**

DI field	0001	0010	0011	0100	0101	0110	1000	1001
Di (decimal)	1	2	4	8	16	32	12	20

Fi is a binary value encoded on a 4-bit field, named FI, as represented in [Table 36-7](#).

**Table 36-7. Binary and Decimal Values for Fi**

Fi field	0000	0001	0010	0011	0100	0101	0110	1001	1010	1011	1100	1101
Fi (decimal)	372	372	558	744	1116	1488	1860	512	768	1024	1536	2048

[Table 36-8](#) shows the resulting Fi/Di Ratio, which is the ratio between the ISO7816 clock and the baud rate clock.

**Table 36-8. Possible Values for the Fi/Di Ratio**

Fi/Di	372	558	774	1116	1488	1806	512	768	1024	1536	2048
1	372	558	744	1116	1488	1860	512	768	1024	1536	2048
2	186	279	372	558	744	930	256	384	512	768	1024
4	93	139.5	186	279	372	465	128	192	256	384	512
8	46.5	69.75	93	139.5	186	232.5	64	96	128	192	256
16	23.25	34.87	46.5	69.75	93	116.2	32	48	64	96	128
32	11.62	17.43	23.25	34.87	46.5	58.13	16	24	32	48	64
12	31	46.5	62	93	124	155	42.66	64	85.33	128	170.6
20	18.6	27.9	37.2	55.8	74.4	93	25.6	38.4	51.2	76.8	102.4

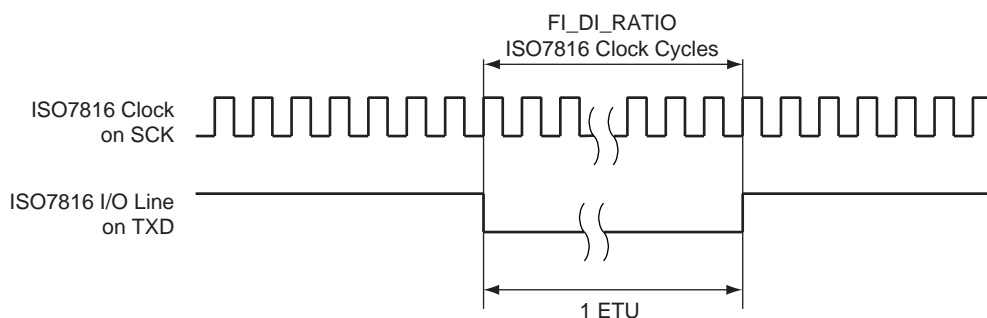
If the USART is configured in ISO7816 Mode, the clock selected by the USCLKS field in the US\_MR is first divided by the value programmed in the field CD in the US\_BRGR. The resulting clock can be provided to the SCK pin to feed the smart card clock inputs. This means that the CLKO bit can be set in US\_MR.

This clock is then divided by the value programmed in the FI\_DI\_RATIO field in the FI\_DI\_Ratio register (US\_FIDI). This is performed by the Sampling Divider, which performs a division by up to **2047** in ISO7816 Mode. The non-integer values of the Fi/Di Ratio are not supported and the user must program the FI\_DI\_RATIO field to a value as close as possible to the expected value.

The FI\_DI\_RATIO field resets to the value 0x174 (372 in decimal) and is the most common divider between the ISO7816 clock and the bit rate (Fi = 372, Di = 1).

[Figure 36-5](#) shows the relation between the Elementary Time Unit, corresponding to a bit time, and the ISO 7816 clock.

**Figure 36-5. Elementary Time Unit (ETU)**



### 36.7.2 Receiver and Transmitter Control

After reset, the receiver is disabled. The user must enable the receiver by setting the RXEN bit in the Control Register (US\_CR). However, the receiver registers can be programmed before the receiver clock is enabled.

After reset, the transmitter is disabled. The user must enable it by setting the TXEN bit in the US\_CR. However, the transmitter registers can be programmed before being enabled.

The Receiver and the Transmitter can be enabled together or independently.

At any time, the software can perform a reset on the receiver or the transmitter of the USART by setting the corresponding bit, RSTRX and RSTTX respectively, in the US\_CR. The software resets clear the status flag and reset internal state machines but the user interface configuration registers hold the value configured prior to software reset. Regardless of what the receiver or the transmitter is performing, the communication is immediately stopped.

The user can also independently disable the receiver or the transmitter by setting RXDIS and TXDIS respectively in the US\_CR. If the receiver is disabled during a character reception, the USART waits until the end of reception of the current character, then the reception is stopped. If the transmitter is disabled while it is operating, the USART waits the end of transmission of both the current character and character being stored in the Transmit Holding Register (US\_THR). If a timeguard is programmed, it is handled normally.

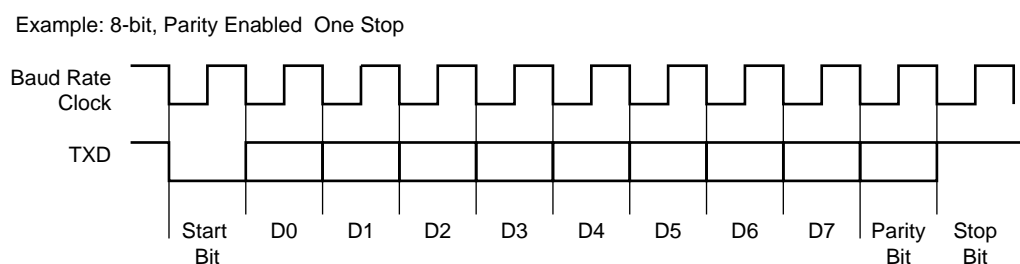
### 36.7.3 Synchronous and Asynchronous Modes

#### 36.7.3.1 Transmitter Operations

The transmitter performs the same in both synchronous and asynchronous operating modes (SYNC = 0 or SYNC = 1). One start bit, up to 9 data bits, one optional parity bit and up to two stop bits are successively shifted out on the TXD pin at each falling edge of the programmed serial clock.

The number of data bits is selected by the CHRL field and the MODE 9 bit in the Mode Register (US\_MR). Nine bits are selected by setting the MODE 9 bit regardless of the CHRL field. The parity bit is set according to the PAR field in US\_MR. The even, odd, space, marked or none parity bit can be configured. The MSBF field in the US\_MR configures which data bit is sent first. If written to 1, the most significant bit is sent first. If written to 0, the less significant bit is sent first. The number of stop bits is selected by the NBSTOP field in the US\_MR. The 1.5 stop bit is supported in asynchronous mode only.

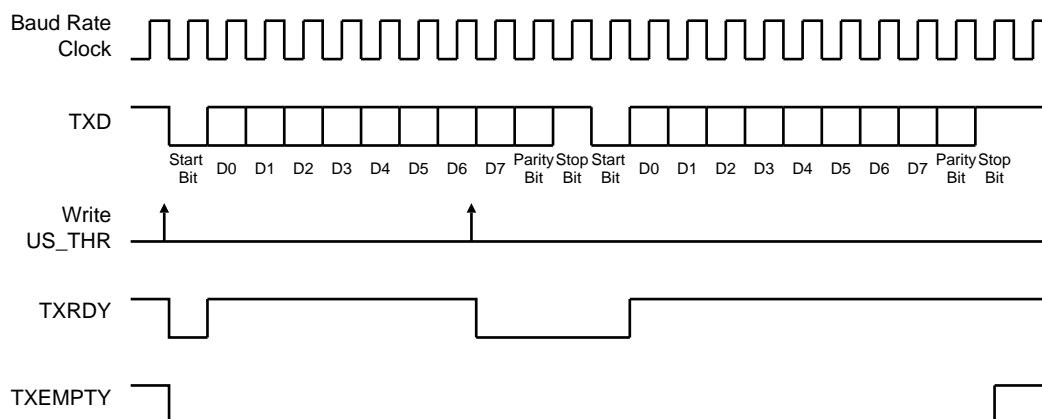
**Figure 36-6. Character Transmit**



The characters are sent by writing in the Transmit Holding Register (US\_THR). The transmitter reports two status bits in the Channel Status Register (US\_CSR): TXRDY (Transmitter Ready), which indicates that US\_THR is empty and TXEMPTY, which indicates that all the characters written in US\_THR have been processed. When the current character processing is completed, the last character written in US\_THR is transferred into the Shift Register of the transmitter and US\_THR becomes empty, thus TXRDY rises.

Both TXRDY and TXEMPTY bits are low when the transmitter is disabled. Writing a character in US\_THR while TXRDY is low has no effect and the written character is lost.

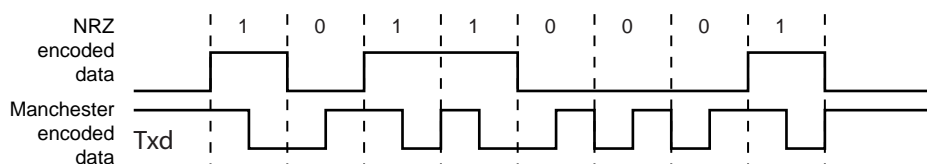
**Figure 36-7. Transmitter Status**



### 36.7.3.2 Manchester Encoder

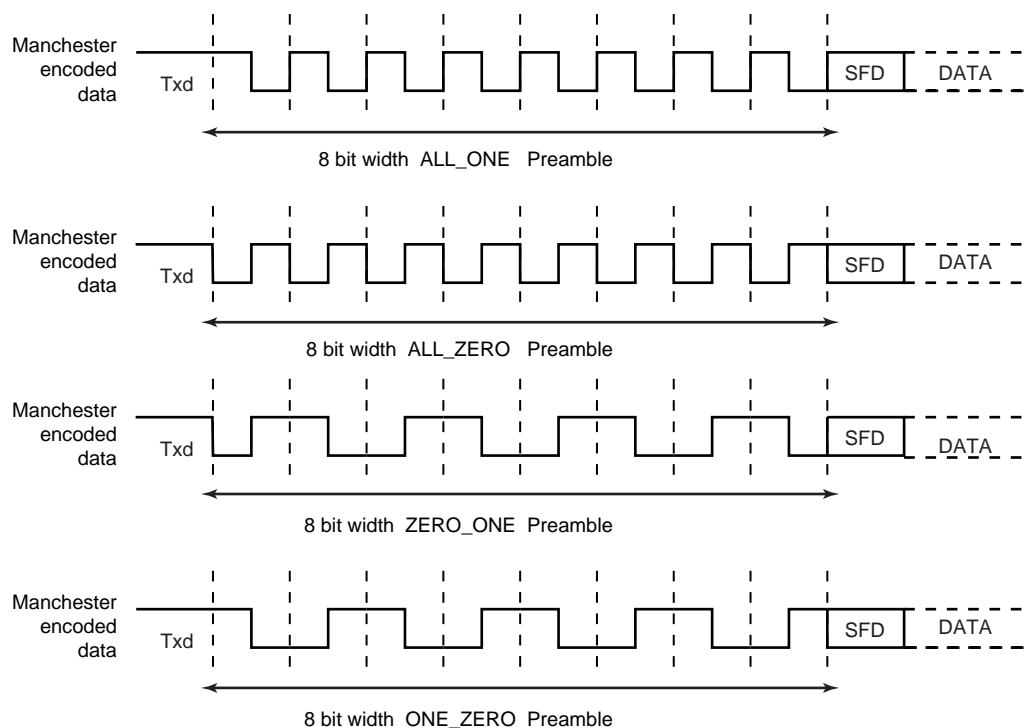
When the Manchester encoder is in use, characters transmitted through the USART are encoded based on biphase Manchester II format. To enable this mode, set the MAN field in the US\_MR register to 1. Depending on polarity configuration, a logic level (zero or one), is transmitted as a coded signal one-to-zero or zero-to-one. Thus, a transition always occurs at the midpoint of each bit time. It consumes more bandwidth than the original NRZ signal (2x) but the receiver has more error control since the expected input must show a change at the center of a bit cell. An example of Manchester encoded sequence is: the byte 0xB1 or 10110001 encodes to 10 01 10 10 01 01 01 10, assuming the default polarity of the encoder. [Figure 36-8](#) illustrates this coding scheme.

**Figure 36-8. NRZ to Manchester Encoding**



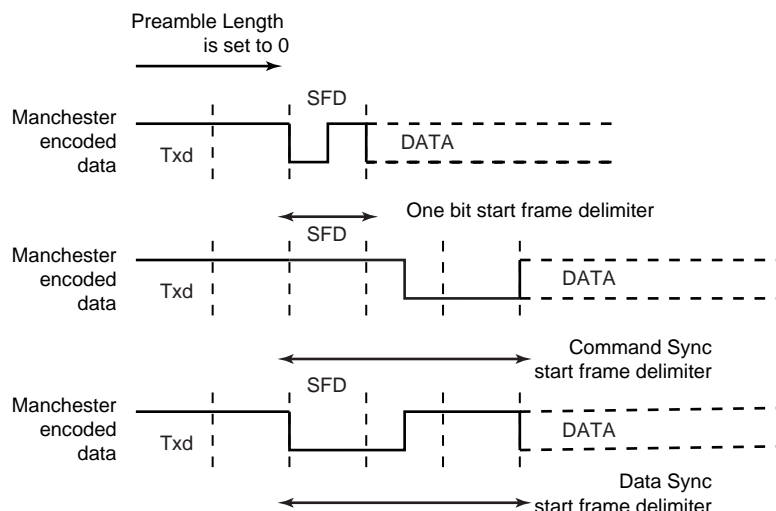
The Manchester encoded character can also be encapsulated by adding both a configurable preamble and a start frame delimiter pattern. Depending on the configuration, the preamble is a training sequence, composed of a predefined pattern with a programmable length from 1 to 15 bit times. If the preamble length is set to 0, the preamble waveform is not generated prior to any character. The preamble pattern is chosen among the following sequences: ALL\_ONE, ALL\_ZERO, ONE\_ZERO or ZERO\_ONE, writing the field TX\_PP in the US\_MAN register, the field TX\_PL is used to configure the preamble length. [Figure 36-9](#) illustrates and defines the valid patterns. To improve flexibility, the encoding scheme can be configured using the TX\_MPOL field in the US\_MAN register. If the TX\_MPOL field is set to zero (default), a logic zero is encoded with a zero-to-one transition and a logic one is encoded with a one-to-zero transition. If the TX\_MPOL field is set to one, a logic one is encoded with a one-to-zero transition and a logic zero is encoded with a zero-to-one transition.

**Figure 36-9. Preamble Patterns, Default Polarity Assumed**



A start frame delimiter is to be configured using the ONEBIT field in the US\_MR register. It consists of a user-defined pattern that indicates the beginning of a valid data. [Figure 36-10](#) illustrates these patterns. If the start frame delimiter, also known as the start bit, is one bit, (ONEBIT to 1), a logic zero is Manchester encoded and indicates that a new character is being sent serially on the line. If the start frame delimiter is a synchronization pattern also referred to as sync (ONEBIT to 0), a sequence of three bit times is sent serially on the line to indicate the start of a new character. The sync waveform is in itself an invalid Manchester waveform as the transition occurs at the middle of the second bit time. Two distinct sync patterns are used: the command sync and the data sync. The command sync has a logic one level for one and a half bit times, then a transition to logic zero for the second one and a half bit times. If the MODSYNC field in the US\_MR is set to 1, the next character is a command. If it is set to 0, the next character is a data. When direct memory access is used, the MODSYNC field can be immediately updated with a modified character located in memory. To enable this mode, VAR\_SYNC field in US\_MR register must be set to 1. In this case, the MODSYNC field in the US\_MR is bypassed and the sync configuration is held in the TXSYNH in the US\_THR. The USART character format is modified and includes sync information.

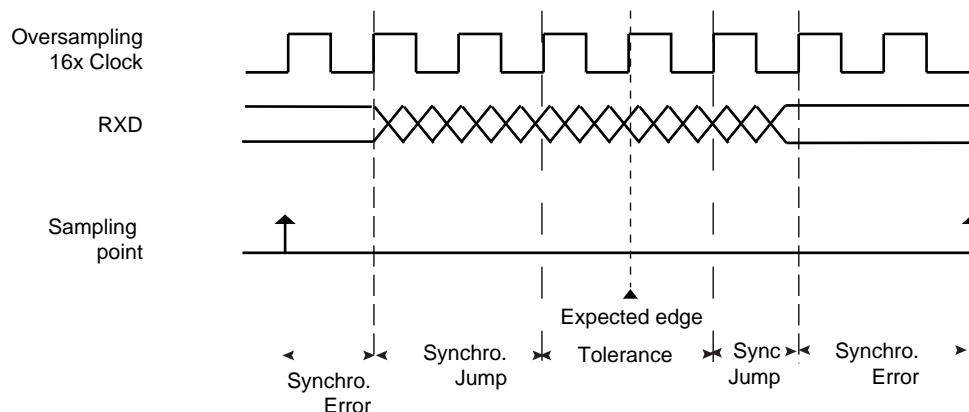
**Figure 36-10. Start Frame Delimiter**



### Drift Compensation

Drift compensation is available only in 16X oversampling mode. An hardware recovery system allows a larger clock drift. To enable the hardware system, the bit in the USART\_MAN register must be set. If the RXD edge is one 16X clock cycle from the expected edge, this is considered as normal jitter and no corrective actions is taken. If the RXD event is between 4 and 2 clock cycles before the expected edge, then the current period is shortened by one clock cycle. If the RXD event is between 2 and 3 clock cycles after the expected edge, then the current period is lengthened by one clock cycle. These intervals are considered to be drift and so corrective actions are automatically taken.

**Figure 36-11. Bit Resynchronization**



### 36.7.3.3 Asynchronous Receiver

If the USART is programmed in asynchronous operating mode (SYNC = 0), the receiver oversamples the RXD input line. The oversampling is either 16 or 8 times the baud rate clock, depending on the OVER bit in the US\_MR.

The receiver samples the RXD line. If the line is sampled during one half of a bit time to 0, a start bit is detected and data, parity and stop bits are successively sampled on the bit rate clock.

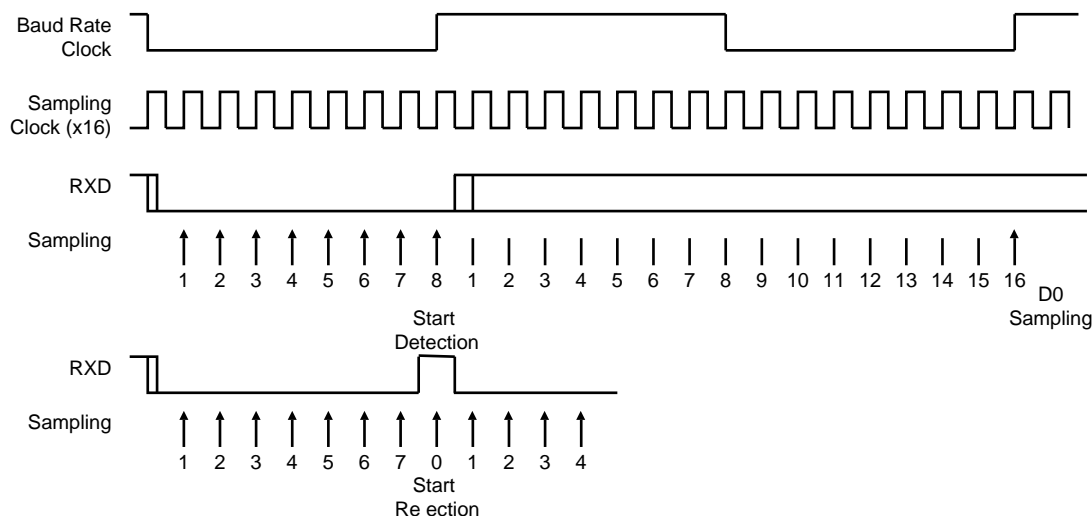
If the oversampling is 16, (OVER to 0), a start is detected at the eighth sample to 0. Then, data bits, parity bit and stop bit are sampled on each 16 sampling clock cycle. If the oversampling is 8 (OVER to 1), a start bit is detected at the fourth sample to 0. Then, data bits, parity bit and stop bit are sampled on each 8 sampling clock cycle.

The number of data bits, first bit sent and parity mode are selected by the same fields and bits as the transmitter, i.e., respectively CHRL, MODE9, MSBF and PAR. For the synchronization mechanism **only**, the number of stop bits has no

effect on the receiver as it considers only one stop bit, regardless of the field NBSTOP, so that resynchronization between the receiver and the transmitter can occur. Moreover, as soon as the stop bit is sampled, the receiver starts looking for a new start bit so that resynchronization can also be accomplished when the transmitter is operating with one stop bit.

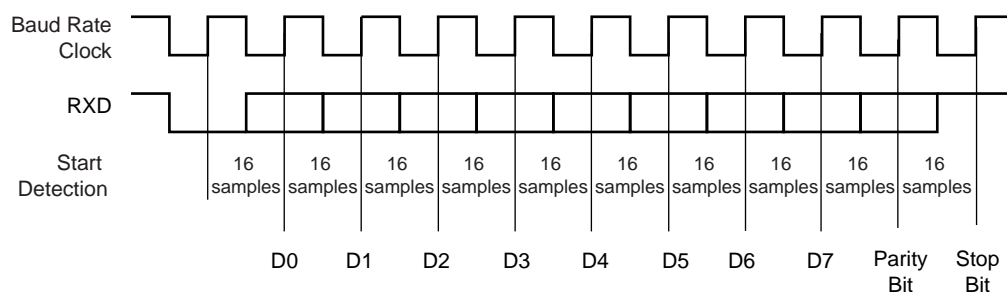
Figure 36-12 and Figure 36-13 illustrate start detection and character reception when USART operates in asynchronous mode.

**Figure 36-12. Asynchronous Start Detection**



**Figure 36-13. Asynchronous Character Reception**

Example: 8-bit, Parity Enabled



### 36.7.3.4 Manchester Decoder

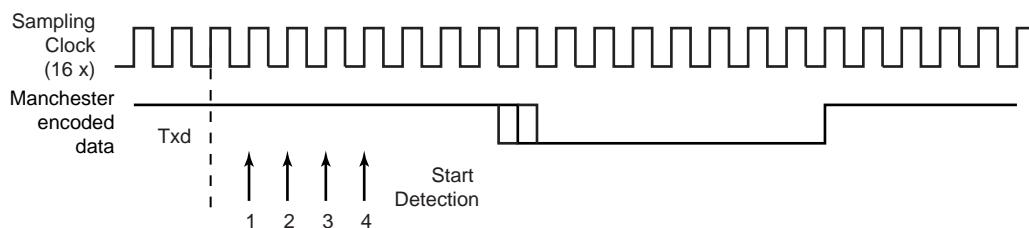
When the MAN field in the US\_MR is set to 1, the Manchester decoder is enabled. The decoder performs both preamble and start frame delimiter detection. One input line is dedicated to Manchester encoded input data.

An optional preamble sequence can be defined, its length is user-defined and totally independent of the emitter side. Use RX\_PL in US\_MAN register to configure the length of the preamble sequence. If the length is set to 0, no preamble is detected and the function is disabled. In addition, the polarity of the input stream is programmable with RX\_MPOL field in US\_MAN register. Depending on the desired application the preamble pattern matching is to be defined via the RX\_PP field in US\_MAN. See Figure 36-9 for available preamble patterns.

Unlike preamble, the start frame delimiter is shared between Manchester Encoder and Decoder. So, if ONEBIT field is set to 1, only a zero encoded Manchester can be detected as a valid start frame delimiter. If ONEBIT is set to 0, only a sync pattern is detected as a valid start frame delimiter. Decoder operates by detecting transition on incoming stream. If RXD is sampled during one quarter of a bit time to zero, a start bit is detected. See Figure 36-14. The sample pulse rejection mechanism applies.

In order to increase the compatibility the RXIDLVL bit in the US\_MAN register allows to inform the USART block of the Rx line idle state value (Rx line undriven), it can be either level one (pull-up) or level zero (pull-down). By default this bit is set to one (Rx line is at level 1 if undriven).

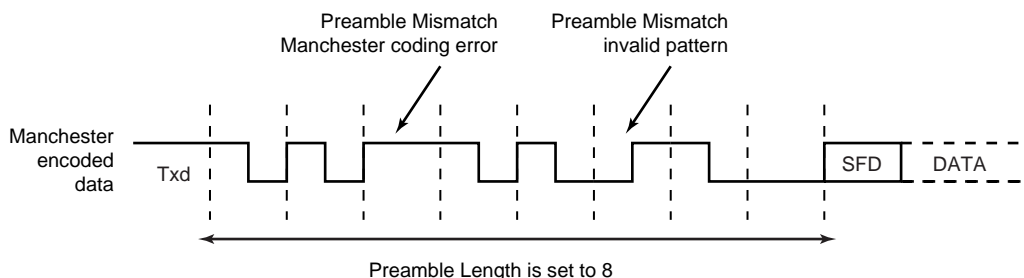
**Figure 36-14. Asynchronous Start Bit Detection**



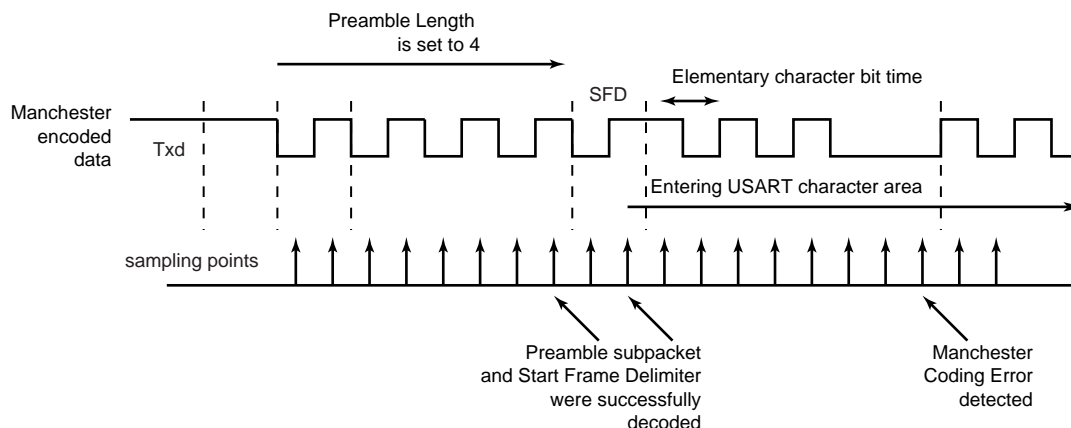
The receiver is activated and starts Preamble and Frame Delimiter detection, sampling the data at one quarter and then three quarters. If a valid preamble pattern or start frame delimiter is detected, the receiver continues decoding with the same synchronization. If the stream does not match a valid pattern or a valid start frame delimiter, the receiver resynchronizes on the next valid edge. The minimum time threshold to estimate the bit value is three quarters of a bit time.

If a valid preamble (if used) followed with a valid start frame delimiter is detected, the incoming stream is decoded into NRZ data and passed to USART for processing. Figure 36-15 illustrates Manchester pattern mismatch. When incoming data stream is passed to the USART, the receiver is also able to detect Manchester code violation. A code violation is a lack of transition in the middle of a bit cell. In this case, MANE flag in the US\_CSR is raised. It is cleared by writing the US\_CR with the RSTSTA bit to 1. See Figure 36-16 for an example of Manchester error detection during data phase.

**Figure 36-15. Preamble Pattern Mismatch**



**Figure 36-16. Manchester Error Flag**





When the start frame delimiter is a sync pattern (ONEBIT field to 0), both command and data delimiter are supported. If a valid sync is detected, the received character is written as RXCHR field in the US\_RHR register and the RXSYNH is updated. RXCHR is set to 1 when the received character is a command, and it is set to 0 if the received character is a data. This mechanism alleviates and simplifies the direct memory access as the character contains its own sync field in the same register.

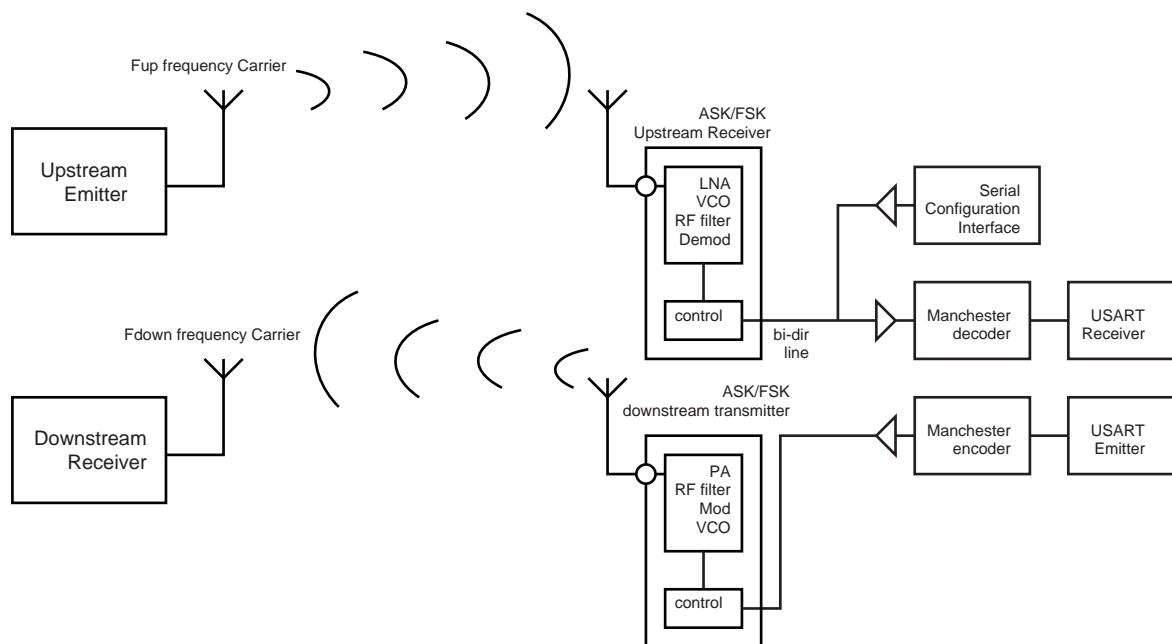
As the decoder is setup to be used in unipolar mode, the first bit of the frame has to be a zero-to-one transition.

### 36.7.3.5 Radio Interface: Manchester Encoded USART Application

This section describes low data rate RF transmission systems and their integration with a Manchester encoded USART. These systems are based on transmitter and receiver ICs that support ASK and FSK modulation schemes.

The goal is to perform full duplex radio transmission of characters using two different frequency carriers. See the configuration in [Figure 36-17](#).

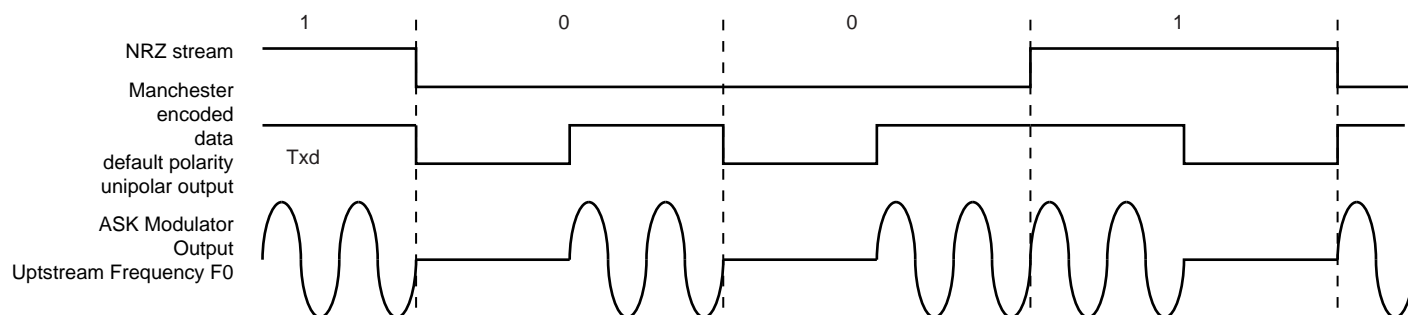
**Figure 36-17. Manchester Encoded Characters RF Transmission**



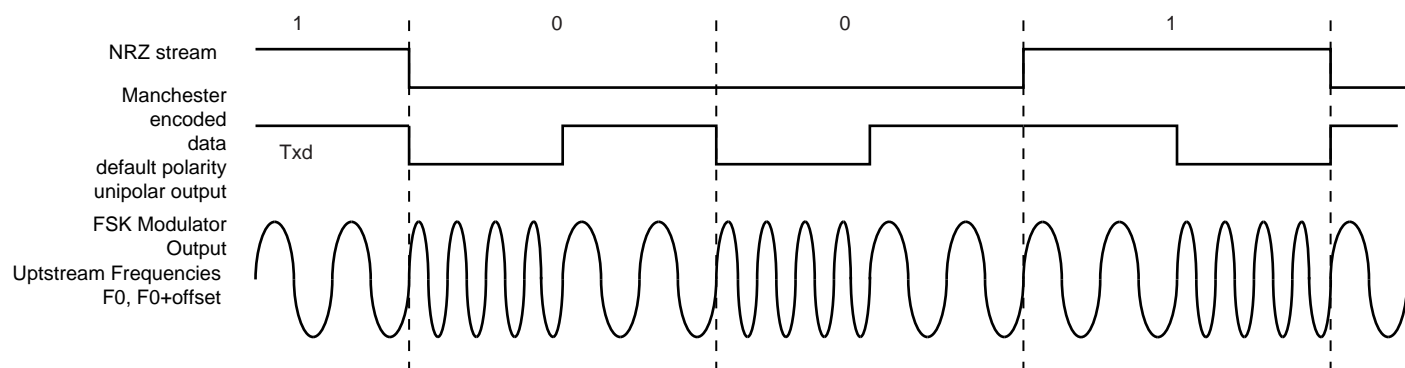
The USART module is configured as a Manchester encoder/decoder. Looking at the downstream communication channel, Manchester encoded characters are serially sent to the RF emitter. This may also include a user defined preamble and a start frame delimiter. Mostly, preamble is used in the RF receiver to distinguish between a valid data from a transmitter and signals due to noise. The Manchester stream is then modulated. See [Figure 36-18](#) for an example of ASK modulation scheme. When a logic one is sent to the ASK modulator, the power amplifier, referred to as PA, is enabled and transmits an RF signal at downstream frequency. When a logic zero is transmitted, the RF signal is turned off. If the FSK modulator is activated, two different frequencies are used to transmit data. When a logic 1 is sent, the modulator outputs an RF signal at frequency F0 and switches to F1 if the data sent is a 0. See [Figure 36-19](#).

From the receiver side, another carrier frequency is used. The RF receiver performs a bit check operation examining demodulated data stream. If a valid pattern is detected, the receiver switches to receiving mode. The demodulated stream is sent to the Manchester decoder. Because of bit checking inside RF IC, the data transferred to the microcontroller is reduced by a user-defined number of bits. The Manchester preamble length is to be defined in accordance with the RF IC configuration.

**Figure 36-18. ASK Modulator Output**



**Figure 36-19. FSK Modulator Output**



### 36.7.3.6 Synchronous Receiver

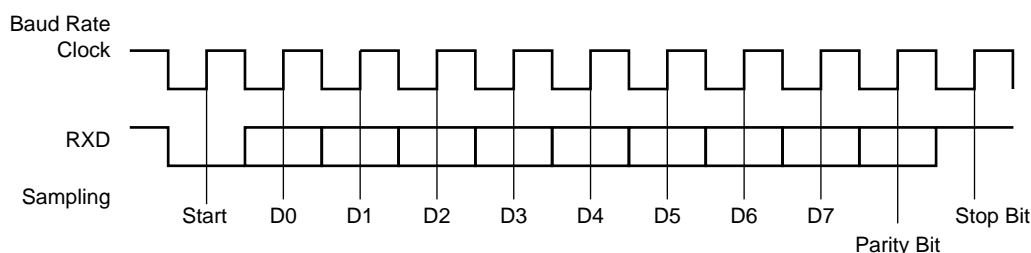
In synchronous mode ( $\text{SYNC} = 1$ ), the receiver samples the RXD signal on each rising edge of the baud rate clock. If a low level is detected, it is considered as a start. All data bits, the parity bit and the stop bits are sampled and the receiver waits for the next start bit. Synchronous mode operations provide a high speed transfer capability.

Configuration fields and bits are the same as in asynchronous mode.

Figure 36-20 illustrates a character reception in synchronous mode.

**Figure 36-20. Synchronous Mode Character Reception**

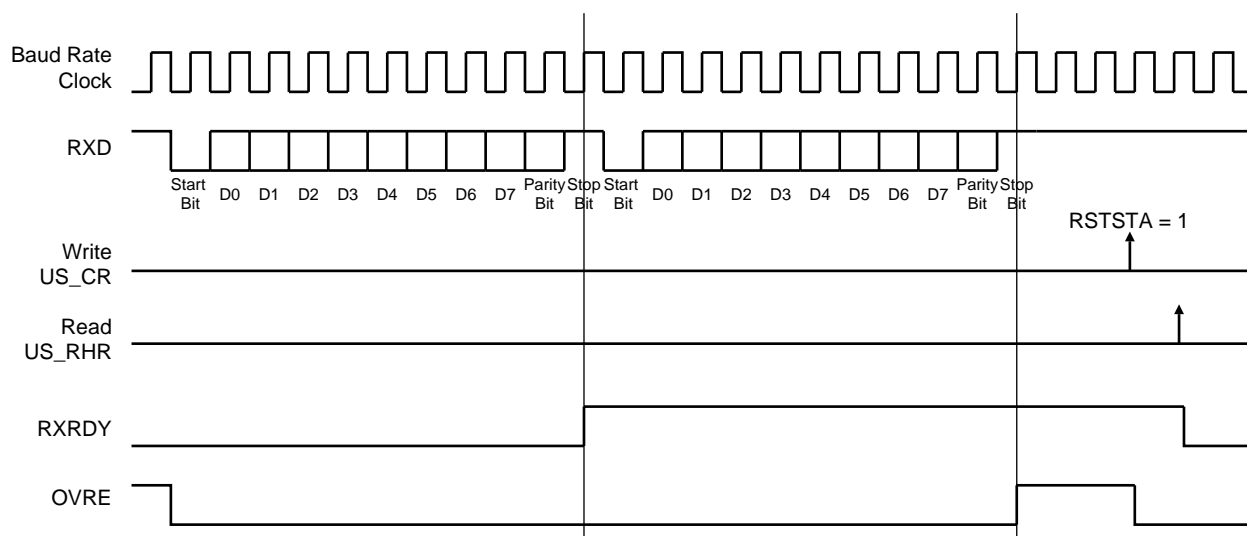
Example: 8-bit, Parity Enabled 1 Stop



### 36.7.3.7 Receiver Operations

When a character reception is completed, it is transferred to the Receive Holding Register (US\_RHR) and the RXRDY bit in the US\_CSR rises. If a character is completed while the RXRDY is set, the OVRE (Overrun Error) bit is set. The last character is transferred into US\_RHR and overwrites the previous one. The OVRE bit is cleared by writing the US\_CR with the RSTSTA (Reset Status) bit to 1.

**Figure 36-21. Receiver Status**



### 36.7.3.8 Parity

The USART supports five parity modes selected by writing to the PAR field in the US\_MR. The PAR field also enables the Multidrop mode, see [“Multidrop Mode” on page 732](#). Even and odd parity bit generation and error detection are supported.

If even parity is selected, the parity generator of the transmitter drives the parity bit to 0 if a number of 1s in the character data bit is even, and to 1 if the number of 1s is odd. Accordingly, the receiver parity checker counts the number of received 1s and reports a parity error if the sampled parity bit does not correspond. If odd parity is selected, the parity generator of the transmitter drives the parity bit to 1 if a number of 1s in the character data bit is even, and to 0 if the number of 1s is odd. Accordingly, the receiver parity checker counts the number of received 1s and reports a parity error if the sampled parity bit does not correspond. If the mark parity is used, the parity generator of the transmitter drives the parity bit to 1 for all characters. The receiver parity checker reports an error if the parity bit is sampled to 0. If the space parity is used, the parity generator of the transmitter drives the parity bit to 0 for all characters. The receiver parity checker reports an error if the parity bit is sampled to 1. If parity is disabled, the transmitter does not generate any parity bit and the receiver does not report any parity error.

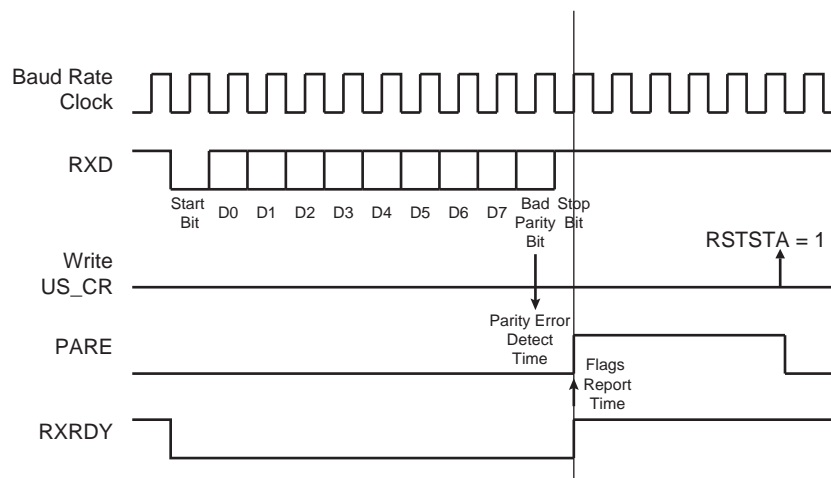
[Table 36-9](#) shows an example of the parity bit for the character 0x41 (character ASCII “A”) depending on the configuration of the USART. Because there are two bits to 1, 1 bit is added when a parity is odd, or 0 is added when a parity is even.

**Table 36-9. Parity Bit Examples**

Character	Hexadecimal	Binary	Parity Bit	Parity Mode
A	0x41	0100 0001	1	Odd
A	0x41	0100 0001	0	Even
A	0x41	0100 0001	1	Mark
A	0x41	0100 0001	0	Space
A	0x41	0100 0001	None	None

When the receiver detects a parity error, it sets the PARE (Parity Error) bit in the US\_CSR. The PARE bit can be cleared by writing the US\_CR with the RSTSTA bit to 1. [Figure 36-22](#) illustrates the parity bit status setting and clearing.

**Figure 36-22. Parity Error**



### 36.7.3.9 Multidrop Mode

If the value 0x6 or 0x07 is written to the PAR field in the US\_MR, the USART runs in Multidrop Mode. This mode differentiates the data characters and the address characters. Data is transmitted with the parity bit to 0 and addresses are transmitted with the parity bit to 1.

If the USART is configured in multidrop mode, the receiver sets the PARE parity error bit when the parity bit is high and the transmitter is able to send a character with the parity bit high when a one is written to the SENDA bit in the US\_CR.

To handle parity error, the PARE bit is cleared when a one is written to the RSTSTA bit in the US\_CR.

The transmitter sends an address byte (parity bit set) when SENDA is written to in the US\_CR. In this case, the next byte written to the US\_THR is transmitted as an address. Any character written in the US\_THR without having written the command SENDA is transmitted normally with the parity to 0.

### 36.7.3.10 Transmitter Timeguard

The timeguard feature enables the USART interface with slow remote devices.

The timeguard function enables the transmitter to insert an idle state on the TXD line between two characters. This idle state actually acts as a long stop bit.

The duration of the idle state is programmed in the TG field of the Transmitter Timeguard Register (US\_TTGR). When this field is written to zero no timeguard is generated. Otherwise, the transmitter holds a high level on TXD after each transmitted byte during the number of bit periods programmed in TG in addition to the number of stop bits.

As illustrated in [Figure 36-23](#), the behavior of TXRDY and TXEMPTY status bits is modified by the programming of a timeguard. TXRDY rises only when the start bit of the next character is sent, and thus remains to 0 during the timeguard transmission if a character has been written in US\_THR. TXEMPTY remains low until the timeguard transmission is completed as the timeguard is part of the current character being transmitted.

**Figure 36-23. Timeguard Operations**

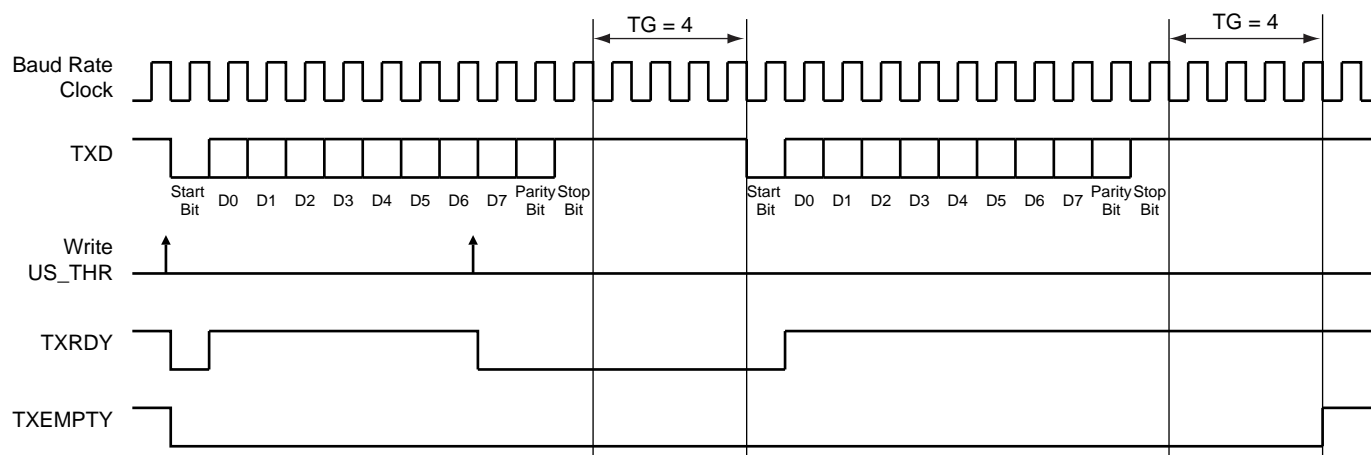


Table 36-10 indicates the maximum length of a timeguard period that the transmitter can handle in relation to the function of the baud rate.

**Table 36-10. Maximum Timeguard Length Depending on Baud Rate**

Baud Rate (Bit/s)	Bit Time (μs)	Timeguard (ms)
1,200	833	212.50
9,600	104	26.56
14,400	69.4	17.71
19,200	52.1	13.28
28,800	34.7	8.85
38,400	26	6.63
56,000	17.9	4.55
57,600	17.4	4.43
115,200	8.7	2.21

### 36.7.3.11 Receiver Time-out

The Receiver Time-out provides support in handling variable-length frames. This feature detects an idle condition on the RXD line. When a time-out is detected, the bit TIMEOUT in the US\_CSR rises and can generate an interrupt, thus indicating to the driver an end of frame.

The time-out delay period (during which the receiver waits for a new character) is programmed in the TO field of the Receiver Time-out Register (US\_RTOR). If the TO field is written to 0, the Receiver Time-out is disabled and no time-out is detected. The TIMEOUT bit in the US\_CSR remains at 0. Otherwise, the receiver loads a 16-bit counter with the value programmed in TO. This counter is decremented at each bit period and reloaded each time a new character is received. If the counter reaches 0, the TIMEOUT bit in the US\_CSR rises. Then, the user can either:

- Stop the counter clock until a new character is received. This is performed by writing a one to the STTTO (Start Time-out) bit in the US\_CR. In this case, the idle state on RXD before a new character is received will not provide a time-out. This prevents having to handle an interrupt before a character is received and allows waiting for the next idle state on RXD after a frame is received.
- Obtain an interrupt while no character is received. This is performed by writing a one to the RETTO (Reload and Start Time-out) bit in the US\_CR. If RETTO is performed, the counter starts counting down immediately from the value TO. This enables generation of a periodic interrupt so that a user time-out can be handled, for example when no key is pressed on a keyboard.

If STTTO is performed, the counter clock is stopped until a first character is received. The idle state on RXD before the start of the frame does not provide a time-out. This prevents having to obtain a periodic interrupt and enables a wait of the end of frame when the idle state on RXD is detected.

If RETTO is performed, the counter starts counting down immediately from the value TO. This enables generation of a periodic interrupt so that a user time-out can be handled, for example when no key is pressed on a keyboard.

Figure 36-24 shows the block diagram of the Receiver Time-out feature.

**Figure 36-24. Receiver Time-out Block Diagram**

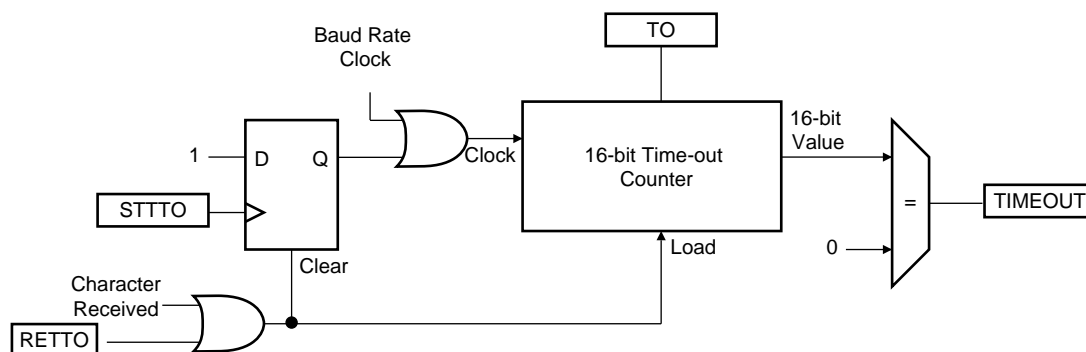


Table 36-11 gives the maximum time-out period for some standard baud rates.

**Table 36-11. Maximum Time-out Period**

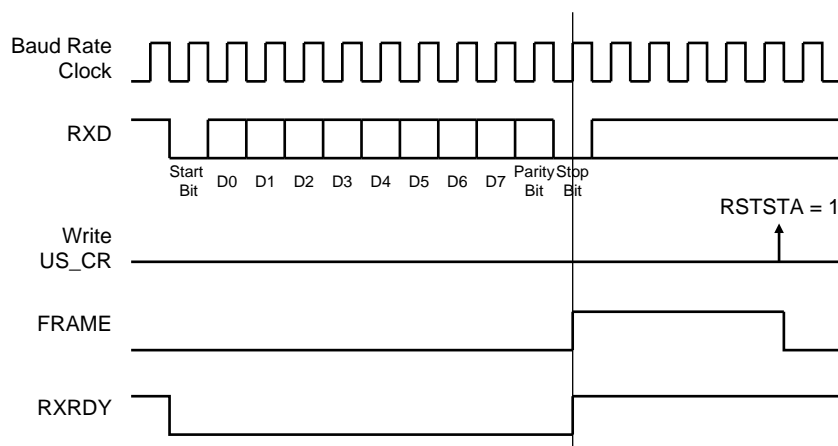
Baud Rate (Bit/s)	Bit Time ( $\mu$ s)	Time-out (ms)
600	1,667	109,225
1,200	833	54,613
2,400	417	27,306
4,800	208	13,653
9,600	104	6,827
14,400	69	4,551
19,200	52	3,413
28,800	35	2,276
38,400	26	1,704
56,000	18	1,170
57,600	17	1,138
200,000	5	328

### 36.7.3.12 Framing Error

The receiver is capable of detecting framing errors. A framing error happens when the stop bit of a received character is detected at level 0. This can occur if the receiver and the transmitter are fully desynchronized.

A framing error is reported on the FRAME bit of the Channel Status Register (US\_CSR). The FRAME bit is asserted in the middle of the stop bit as soon as the framing error is detected. It is cleared by writing the Control Register (US\_CR) with the RSTSTA bit to 1.

**Figure 36-25. Framing Error Status**



### 36.7.3.13 Transmit Break

The user can request the transmitter to generate a break condition on the TXD line. A break condition drives the TXD line low during at least one complete character. It appears the same as a 0x00 character sent with the parity and the stop bits to 0. However, the transmitter holds the TXD line at least during one character until the user requests the break condition to be removed.

A break is transmitted by writing the Control Register (US\_CR) with the STTBRK bit to 1. This can be performed at any time, either while the transmitter is empty (no character in either the Shift Register or in US\_THR) or when a character is being transmitted. If a break is requested while a character is being shifted out, the character is first completed before the TXD line is held low.

Once STTBRK command is requested further STTBRK commands are ignored until the end of the break is completed.

The break condition is removed by writing US\_CR with the STPBRK bit to 1. If the STPBRK is requested before the end of the minimum break duration (one character, including start, data, parity and stop bits), the transmitter ensures that the break condition completes.

The transmitter considers the break as though it is a character, i.e., the STTBRK and STPBRK commands are taken into account only if the TXRDY bit in US\_CSR is to 1 and the start of the break condition clears the TXRDY and TXEMPTY bits as if a character is processed.

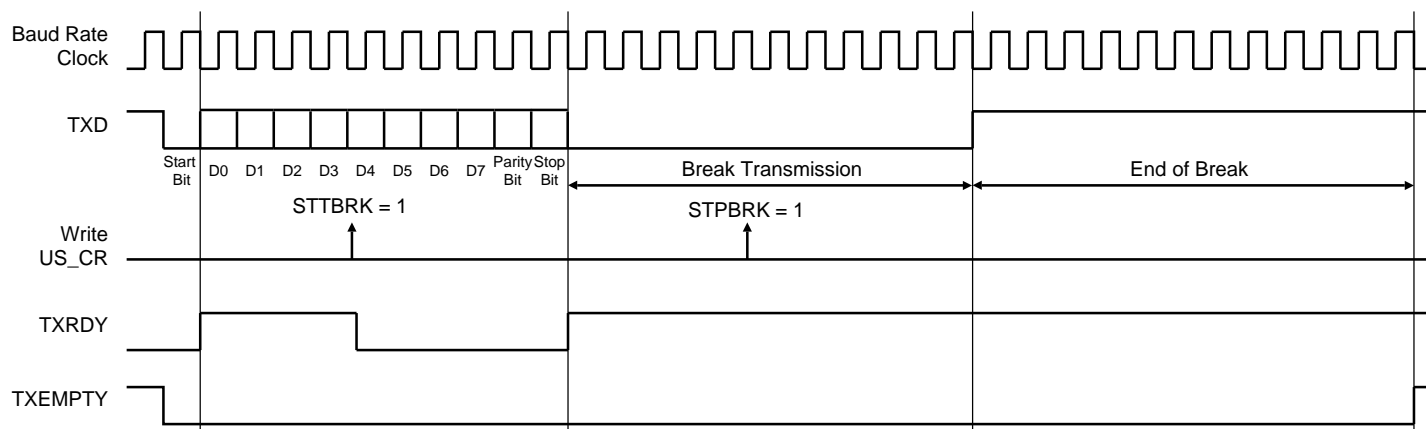
Writing US\_CR with both STTBRK and STPBRK bits to 1 can lead to an unpredictable result. All STPBRK commands requested without a previous STTBRK command are ignored. A byte written into the Transmit Holding Register while a break is pending, but not started, is ignored.

After the break condition, the transmitter returns the TXD line to 1 for a minimum of 12 bit times. Thus, the transmitter ensures that the remote receiver detects correctly the end of break and the start of the next character. If the timeguard is programmed with a value higher than 12, the TXD line is held high for the timeguard period.

After holding the TXD line for this period, the transmitter resumes normal operations.

Figure 36-26 illustrates the effect of both the Start Break (STTBRK) and Stop Break (STPBRK) commands on the TXD line.

**Figure 36-26. Break Transmission**



### 36.7.3.14 Receive Break

The receiver detects a break condition when all data, parity and stop bits are low. This corresponds to detecting a framing error with data to 0x00, but FRAME remains low.

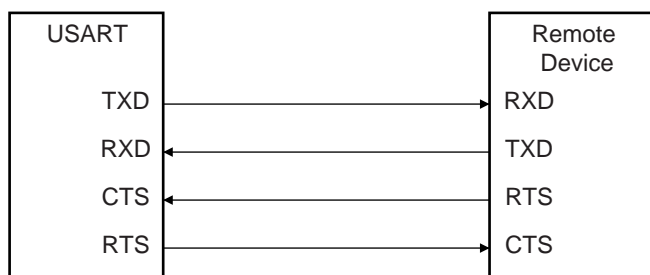
When the low stop bit is detected, the receiver asserts the RXBRK bit in US\_CSR. This bit may be cleared by writing the Control Register (US\_CR) with the bit RSTSTA to 1.

An end of receive break is detected by a high level for at least 2/16 of a bit period in asynchronous operating mode or one sample at high level in synchronous operating mode. The end of break detection also asserts the RXBRK bit.

### 36.7.3.15 Hardware Handshaking

The USART features a hardware handshaking out-of-band flow control. The RTS and CTS pins are used to connect with the remote device, as shown in [Figure 36-27](#).

**Figure 36-27. Connection with a Remote Device for Hardware Handshaking**



Setting the USART to operate with hardware handshaking is performed by writing the USART\_MODE field in the Mode Register (US\_MR) to the value 0x2.

The USART behavior when hardware handshaking is enabled is the same as the behavior in standard synchronous or asynchronous mode, except that the receiver drives the RTS pin as described below and the level on the CTS pin modifies the behavior of the transmitter as described below. Using this mode requires using the PDC channel for reception. The transmitter can handle hardware handshaking in any case.

[Figure 36-28](#) shows how the receiver operates if hardware handshaking is enabled. The RTS pin is driven high if the receiver is disabled and if the status RXBUFF (Receive Buffer Full) coming from the PDC channel is high. Normally, the remote device does not start transmitting while its CTS pin (driven by RTS) is high. As soon as the Receiver is enabled, the RTS falls, indicating to the remote device that it can start transmitting. Defining a new buffer to the PDC clears the status bit RXBUFF and, as a result, asserts the pin RTS low.



**Figure 36-28. Receiver Behavior when Operating with Hardware Handshaking**

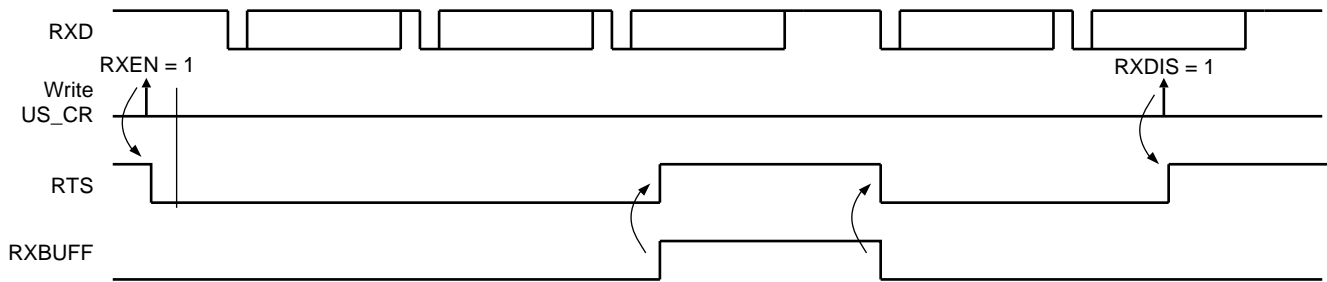
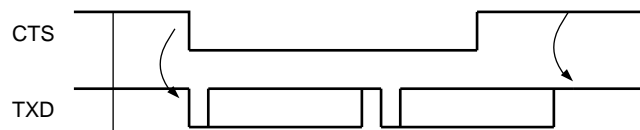


Figure 36-29 shows how the transmitter operates if hardware handshaking is enabled. The CTS pin disables the transmitter. If a character is being processing, the transmitter is disabled only after the completion of the current character and transmission of the next character happens as soon as the pin CTS falls.

**Figure 36-29. Transmitter Behavior when Operating with Hardware Handshaking**



### 36.7.4 ISO7816 Mode

The USART features an ISO7816-compatible operating mode. This mode permits interfacing with smart cards and Security Access Modules (SAM) communicating through an ISO7816 link. Both T = 0 and T = 1 protocols defined by the ISO7816 specification are supported.

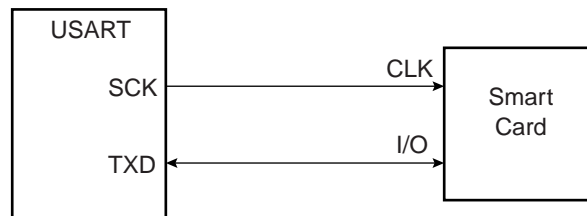
Setting the USART in ISO7816 mode is performed by writing the USART\_MODE field in the Mode Register (US\_MR) to the value 0x4 for protocol T = 0 and to the value 0x5 for protocol T = 1.

#### 36.7.4.1 ISO7816 Mode Overview

The ISO7816 is a half duplex communication on only one bidirectional line. The baud rate is determined by a division of the clock provided to the remote device (see “[Baud Rate Generator](#)” on page 719).

The USART connects to a smart card as shown in Figure 36-30. The TXD line becomes bidirectional and the Baud Rate Generator feeds the ISO7816 clock on the SCK pin. As the TXD pin becomes bidirectional, its output remains driven by the output of the transmitter but only when the transmitter is active while its input is directed to the input of the receiver. The USART is considered as the master of the communication as it generates the clock.

**Figure 36-30. Connection of a Smart Card to the USART**



When operating in ISO7816, either in T = 0 or T = 1 modes, the character format is fixed. The configuration is 8 data bits, even parity and 1 or 2 stop bits, regardless of the values programmed in the CHRL, MODE9, PAR and CHMODE fields. MSBF can be used to transmit LSB or MSB first. Parity Bit (PAR) can be used to transmit in normal or inverse mode. Refer to “[USART Mode Register](#)” on page 754 and “[PAR: Parity Type](#)” on page 755.

The USART cannot operate concurrently in both receiver and transmitter modes as the communication is unidirectional at a time. It has to be configured according to the required mode by enabling or disabling either the receiver or the

transmitter as desired. Enabling both the receiver and the transmitter at the same time in ISO7816 mode may lead to unpredictable results.

The ISO7816 specification defines an inverse transmission format. Data bits of the character must be transmitted on the I/O line at their negative value.

#### 36.7.4.2 Protocol T = 0

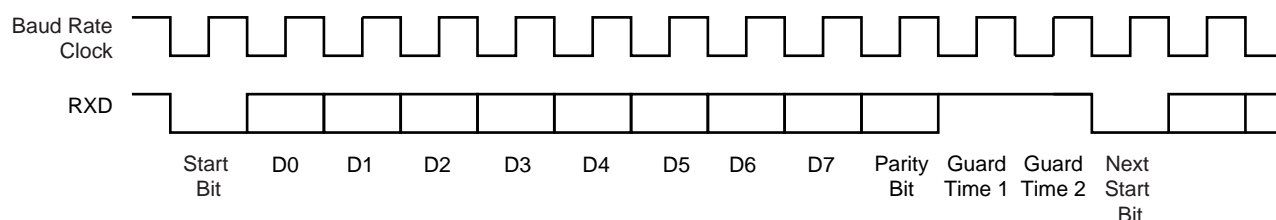
In T = 0 protocol, a character is made up of one start bit, eight data bits, one parity bit and one guard time, which lasts two bit times. The transmitter shifts out the bits and does not drive the I/O line during the guard time.

If no parity error is detected, the I/O line remains to 1 during the guard time and the transmitter can continue with the transmission of the next character, as shown in Figure 36-31.

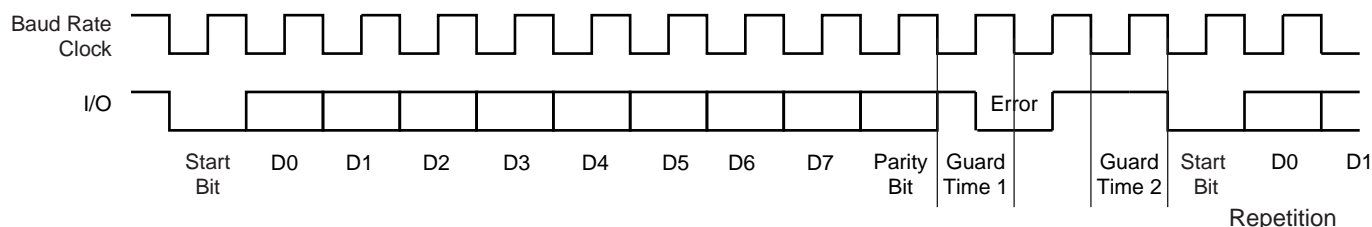
If a parity error is detected by the receiver, it drives the I/O line to 0 during the guard time, as shown in Figure 36-32. This error bit is also named NACK, for Non Acknowledge. In this case, the character lasts 1 bit time more, as the guard time length is the same and is added to the error bit time which lasts 1 bit time.

When the USART is the receiver and it detects an error, it does not load the erroneous character in the Receive Holding Register (US\_RHR). It appropriately sets the PARE bit in the Status Register (US\_SR) so that the software can handle the error.

**Figure 36-31. T = 0 Protocol without Parity Error**



**Figure 36-32. T = 0 Protocol with Parity Error**



#### Receive Error Counter

The USART receiver also records the total number of errors. This can be read in the Number of Error (US\_NER) register. The NB\_ERRORS field can record up to 255 errors. Reading US\_NER automatically clears the NB\_ERRORS field.

#### Receive NACK Inhibit

The USART can also be configured to inhibit an error. This can be achieved by setting the INACK bit in the Mode Register (US\_MR). If INACK is to 1, no error signal is driven on the I/O line even if a parity bit is detected.

Moreover, if INACK is set, the erroneous received character is stored in the Receive Holding Register, as if no error occurred and the RXRDY bit does rise.

#### Transmit Character Repetition

When the USART is transmitting a character and gets a NACK, it can automatically repeat the character before moving on to the next one. Repetition is enabled by writing the MAX\_ITERATION field in the US\_MR at a value higher than 0. Each character can be transmitted up to eight times; the first transmission plus seven repetitions.

If MAX\_ITERATION does not equal zero, the USART repeats the character as many times as the value loaded in MAX\_ITERATION.

When the USART repetition number reaches MAX\_ITERATION, the ITERATION bit is set in the Channel Status Register (US\_CSR). If the repetition of the character is acknowledged by the receiver, the repetitions are stopped and the iteration counter is cleared.

The ITERATION bit in US\_CSR can be cleared by writing the Control Register with the RSTIT bit to 1.

#### Disable Successive Receive NACK

The receiver can limit the number of successive NACKs sent back to the remote transmitter. This is programmed by setting the bit DSNACK in the US\_MR. The maximum number of NACKs transmitted is programmed in the MAX\_ITERATION field. As soon as MAX\_ITERATION is reached, the character is considered as correct, an acknowledge is sent on the line and the ITERATION bit in the US\_CSR is set.

#### 36.7.4.3 Protocol T = 1

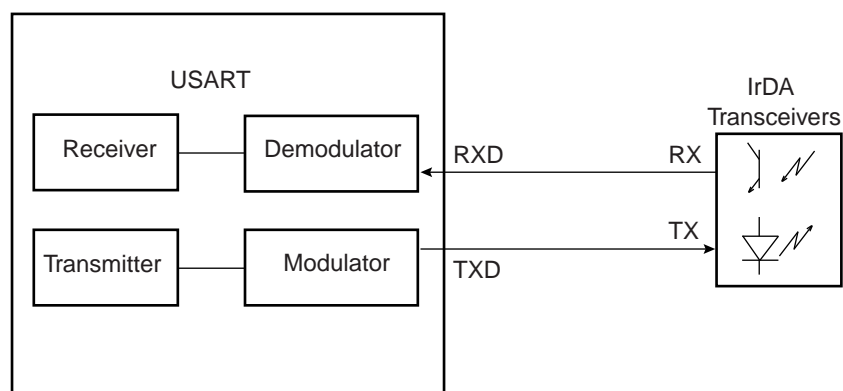
When operating in ISO7816 protocol T = 1, the transmission is similar to an asynchronous format with only one stop bit. The parity is generated when transmitting and checked when receiving. Parity error detection sets the PARE bit in the US\_CSR.

#### 36.7.5 IrDA Mode

The USART features an IrDA mode supplying half-duplex point-to-point wireless communication. It embeds the modulator and demodulator which allows a glueless connection to the infrared transceivers, as shown in [Figure 36-33](#). The modulator and demodulator are compliant with the IrDA specification version 1.1 and support data transfer speeds ranging from 2.4 Kb/s to 115.2 Kb/s.

The USART IrDA mode is enabled by setting the USART\_MODE field in the US\_MR to the value 0x8. The IrDA Filter Register (US\_IF) allows configuring the demodulator filter. The USART transmitter and receiver operate in a normal asynchronous mode and all parameters are accessible. Note that the modulator and the demodulator are activated.

**Figure 36-33. Connection to IrDA Transceivers**



The receiver and the transmitter must be enabled or disabled according to the direction of the transmission to be managed.

To receive IrDA signals, the following needs to be done:

- Disable TX and Enable RX
- Configure the TXD pin as PIO and set it as an output to 0 (to avoid LED emission). Disable the internal pull-up (better for power consumption).
- Receive data

### 36.7.5.1 IrDA Modulation

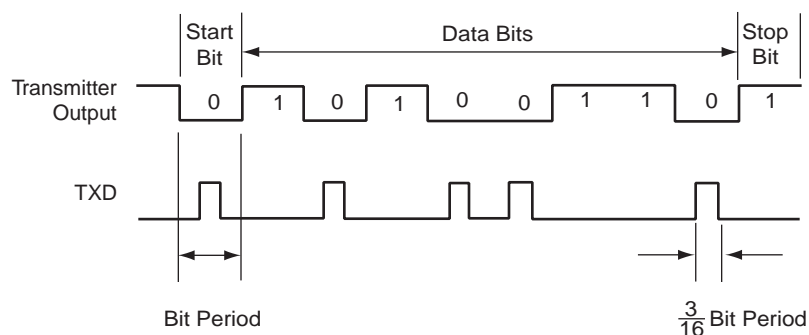
For baud rates up to and including 115.2 Kb/s, the RZI modulation scheme is used. “0” is represented by a light pulse of 3/16th of a bit time. Some examples of signal pulse duration are shown in [Table 36-12](#).

**Table 36-12. IrDA Pulse Duration**

Baud Rate	Pulse Duration (3/16)
2.4 Kb/s	78.13 $\mu$ s
9.6 Kb/s	19.53 $\mu$ s
19.2 Kb/s	9.77 $\mu$ s
38.4 Kb/s	4.88 $\mu$ s
57.6 Kb/s	3.26 $\mu$ s
115.2 Kb/s	1.63 $\mu$ s

[Figure 36-34](#) shows an example of character transmission.

**Figure 36-34. IrDA Modulation**



### 36.7.5.2 IrDA Baud Rate

[Table 36-13](#) gives some examples of CD values, baud rate error and pulse duration. Note that the requirement on the maximum acceptable error of  $\pm 1.87\%$  must be met.

**Table 36-13. IrDA Baud Rate Error**

Peripheral Clock	Baud Rate (Bit/s)	CD	Baud Rate Error	Pulse Time ( $\mu$ s)
3,686,400	115,200	2	0.00%	1.63
20,000,000	115,200	11	1.38%	1.63
32,768,000	115,200	18	1.25%	1.63
40,000,000	115,200	22	1.38%	1.63
3,686,400	57,600	4	0.00%	3.26
20,000,000	57,600	22	1.38%	3.26
32,768,000	57,600	36	1.25%	3.26
40,000,000	57,600	43	0.93%	3.26
3,686,400	38,400	6	0.00%	4.88
20,000,000	38,400	33	1.38%	4.88
32,768,000	38,400	53	0.63%	4.88
40,000,000	38,400	65	0.16%	4.88
3,686,400	19,200	12	0.00%	9.77
20,000,000	19,200	65	0.16%	9.77

**Table 36-13. IrDA Baud Rate Error (Continued)**

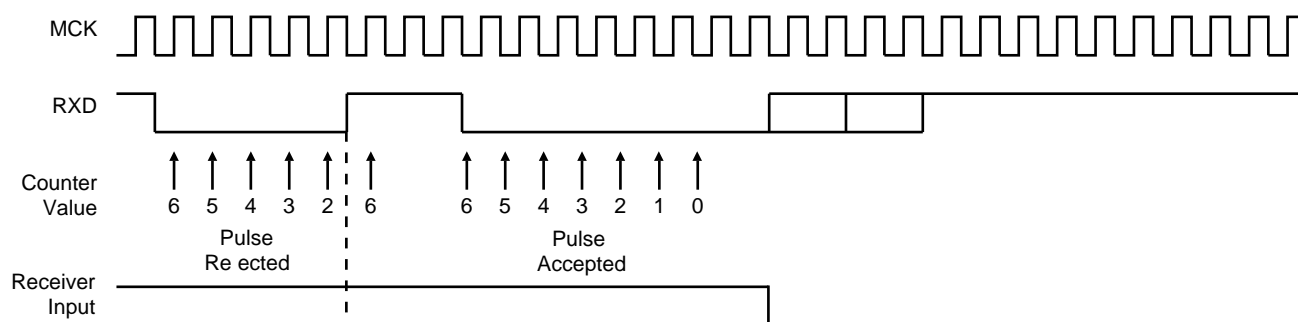
Peripheral Clock	Baud Rate (Bit/s)	CD	Baud Rate Error	Pulse Time (μs)
32,768,000	19,200	107	0.31%	9.77
40,000,000	19,200	130	0.16%	9.77
3,686,400	9,600	24	0.00%	19.53
20,000,000	9,600	130	0.16%	19.53
32,768,000	9,600	213	0.16%	19.53
40,000,000	9,600	260	0.16%	19.53
3,686,400	2,400	96	0.00%	78.13
20,000,000	2,400	521	0.03%	78.13
32,768,000	2,400	853	0.04%	78.13

### 36.7.5.3 IrDA Demodulator

The demodulator is based on the IrDA Receive filter comprised of an 8-bit down counter which is loaded with the value programmed in US\_IF. When a falling edge is detected on the RXD pin, the Filter Counter starts counting down at the Master Clock (MCK) speed. If a rising edge is detected on the RXD pin, the counter stops and is reloaded with US\_IF. If no rising edge is detected when the counter reaches 0, the input of the receiver is driven low during one bit time.

Figure 36-35 illustrates the operations of the IrDA demodulator.

**Figure 36-35. IrDA Demodulator Operations**



The programmed value in the US\_IF register must always meet the following criteria:

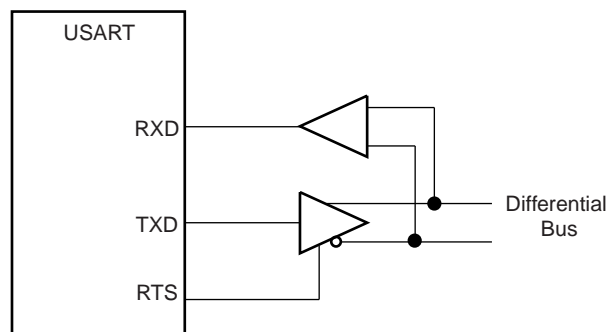
$$t_{MCK} * (IRDA\_FILTER + 3) < 1.41 \mu s$$

As the IrDA mode uses the same logic as the ISO7816, note that the FI\_DI\_RATIO field in US\_FIDI must be set to a value higher than 0 in order to assure IrDA communications operate correctly.

### 36.7.6 RS485 Mode

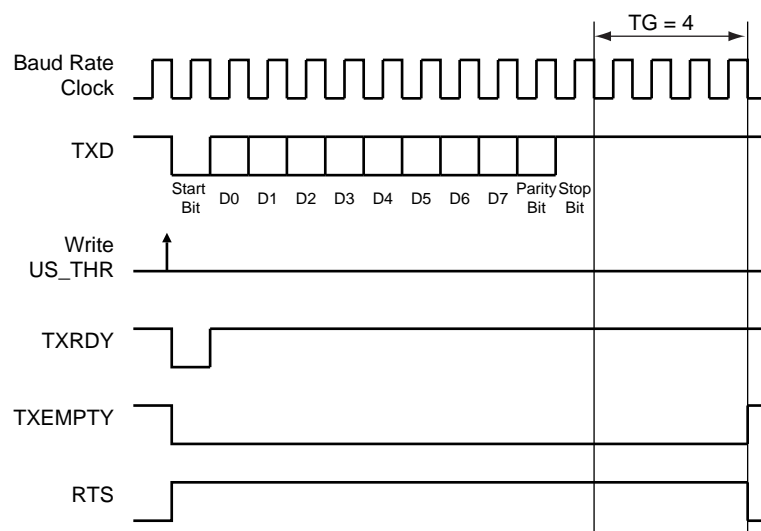
The USART features the RS485 mode to enable line driver control. While operating in RS485 mode, the USART behaves as though in asynchronous or synchronous mode and configuration of all the parameters is possible. The difference is that the RTS pin is driven high when the transmitter is operating. The behavior of the RTS pin is controlled by the TXEMPTY bit. A typical connection of the USART to a RS485 bus is shown in Figure 36-36.

**Figure 36-36. Typical Connection to a RS485 Bus**



The USART is set in RS485 mode by writing the value 0x1 to the USART\_MODE field in the Mode Register (US\_MR). The RTS pin is at a level inverse to the TXEMPTY bit. Significantly, the RTS pin remains high when a timeguard is programmed so that the line can remain driven after the last character completion. [Figure 36-37](#) gives an example of the RTS waveform during a character transmission when the timeguard is enabled.

**Figure 36-37. Example of RTS Drive with Timeguard**



### 36.7.7 SPI Mode

The Serial Peripheral Interface (SPI) Mode is a synchronous serial data link that provides communication with external devices in Master or Slave Mode. It also enables communication between processors if an external processor is connected to the system.

The Serial Peripheral Interface is essentially a shift register that serially transmits data bits to other SPIs. During a data transfer, one SPI system acts as the “master” which controls the data flow, while the other devices act as “slaves” which have data shifted into and out by the master. Different CPUs can take turns being masters and one master may simultaneously shift data into multiple slaves. (Multiple Master Protocol is the opposite of Single Master Protocol, where one CPU is always the master while all of the others are always slaves.) However, only one slave may drive its output to write data back to the master at any given time.

A slave device is selected when its NSS signal is asserted by the master. The USART in SPI Master mode can address only one SPI Slave because it can generate only one NSS signal.

The SPI system consists of two data lines and two control lines:

- Master Out Slave In (MOSI): This data line supplies the output data from the master shifted into the input of the slave.
- Master In Slave Out (MISO): This data line supplies the output data from a slave to the input of the master.
- Serial Clock (SCK): This control line is driven by the master and regulates the flow of the data bits. The master may transmit data at a variety of baud rates. The SCK line cycles once for each bit that is transmitted.
- Slave Select (NSS): This control line allows the master to select or deselect the slave.

#### 36.7.7.1 Modes of Operation

The USART can operate in SPI Master Mode or in SPI Slave Mode.

Operation in SPI Master Mode is programmed by writing 0xE to the USART\_MODE field in the Mode Register (US\_MR). In this case the SPI lines must be connected as described below:

- The MOSI line is driven by the output pin TXD
- The MISO line drives the input pin RXD
- The SCK line is driven by the output pin SCK
- The NSS line is driven by the output pin RTS

Operation in SPI Slave Mode is programmed by writing 0xF to the USART\_MODE field in the Mode Register. In this case the SPI lines must be connected as described below:

- The MOSI line drives the input pin RXD
- The MISO line is driven by the output pin TXD
- The SCK line drives the input pin SCK
- The NSS line drives the input pin CTS

In order to avoid unpredicted behavior, any change of the SPI Mode must be followed by a software reset of the transmitter and of the receiver (except the initial configuration after a hardware reset). (See [Section 36.7.2 “Receiver and Transmitter Control”](#)).

#### 36.7.7.2 Baud Rate

In SPI Mode, the baud rate generator operates in the same way as in USART synchronous mode: [See “Baud Rate in Synchronous Mode or SPI Mode” on page 721](#). However, there are some restrictions:

In SPI Master Mode:

- The external clock SCK must not be selected (USCLKS  $\neq$  0x3), and the bit CLKO must be set to ‘1’ in the US\_MR, in order to generate correctly the serial clock on the SCK pin.
- To obtain correct behavior of the receiver and the transmitter, the value programmed in CD must be superior or equal to 6.
- If the internal clock divided (MCK/DIV) is selected, the value programmed in CD must be even to ensure a 50:50 mark/space ratio on the SCK pin, this value can be odd if the internal clock is selected (MCK).

In SPI Slave Mode:

- The external clock (SCK) selection is forced regardless of the value of the USCLKS field in the US\_MR. Likewise, the value written in US\_BRGR has no effect, because the clock is provided directly by the signal on the USART SCK pin.
- To obtain correct behavior of the receiver and the transmitter, the external clock (SCK) frequency must be at least 6 times lower than the system clock.

### 36.7.7.3 Data Transfer

Up to nine data bits are successively shifted out on the TXD pin at each rising or falling edge (depending of CPOL and CPHA) of the programmed serial clock. There is no Start bit, no Parity bit and no Stop bit.

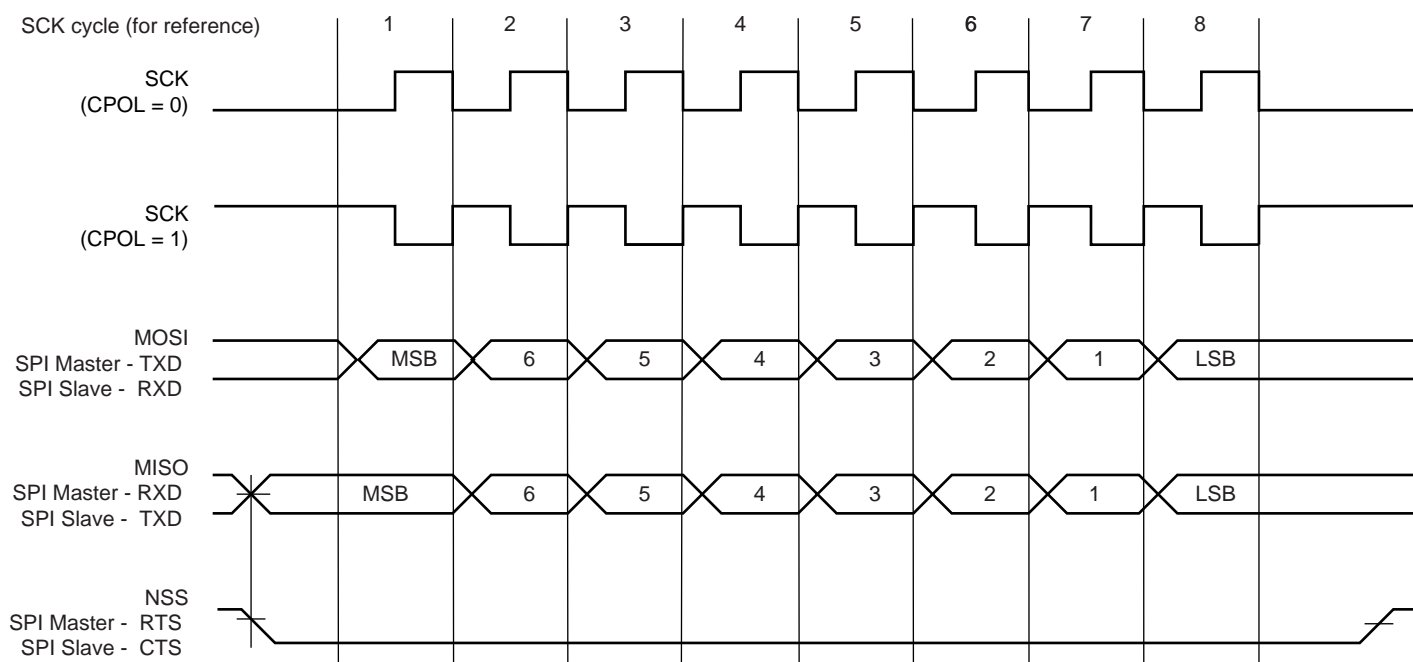
The number of data bits is selected by the CHRL field and the MODE 9 bit in the US\_MR. The nine bits are selected by setting the MODE 9 bit regardless of the CHRL field. The MSB data bit is always sent first in SPI Mode (Master or Slave).

Four combinations of polarity and phase are available for data transfers. The clock polarity is programmed with the CPOL bit in the US\_MR. The clock phase is programmed with the CPHA bit. These two parameters determine the edges of the clock signal upon which data is driven and sampled. Each of the two parameters has two possible states, resulting in four possible combinations that are incompatible with one another. Thus, a master/slave pair must use the same parameter pair values to communicate. If multiple slaves are used and fixed in different configurations, the master must reconfigure itself each time it needs to communicate with a different slave.

**Table 36-14. SPI Bus Protocol Mode**

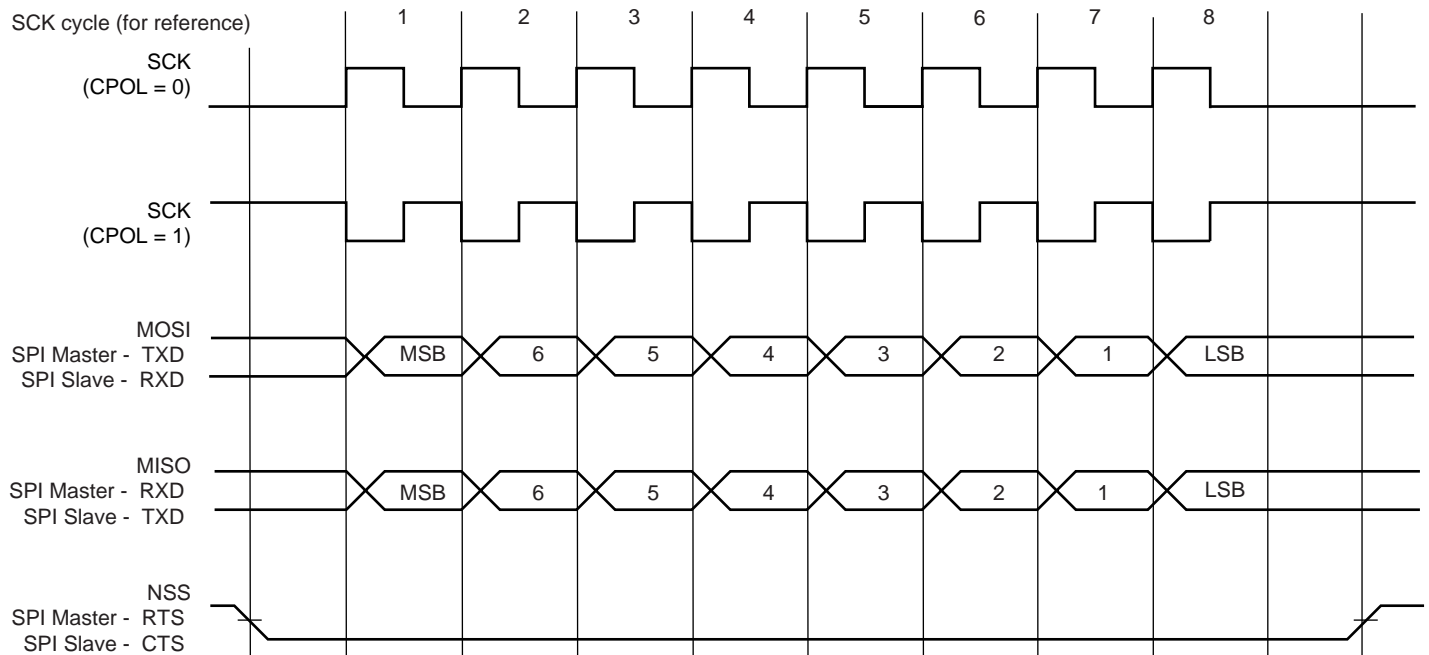
SPI Bus Protocol Mode	CPOL	CPHA
0	0	1
1	0	0
2	1	1
3	1	0

**Figure 36-38. SPI Transfer Format (CPHA = 1, 8 bits per transfer)**





**Figure 36-39. SPI Transfer Format (CPHA = 0, 8 bits per transfer)**



#### 36.7.7.4 Receiver and Transmitter Control

See “Receiver and Transmitter Control” on page 722.

#### 36.7.7.5 Character Transmission

The characters are sent by writing in the Transmit Holding Register (US\_THR). An additional condition for transmitting a character can be added when the USART is configured in SPI master mode. In the USART\_MR, the value configured on INACK field can prevent any character transmission (even if US\_THR has been written) while the receiver side is not ready (character not read). When WRDBT equals 0, the character is transmitted whatever the receiver status. If WRDBT is set to 1, the transmitter waits for the receiver holding register to be read before transmitting the character (RXRDY flag cleared), thus preventing any overflow (character loss) on the receiver side.

The transmitter reports two status bits in the Channel Status Register (US\_CSR): TXRDY (Transmitter Ready), which indicates that US\_THR is empty and TXEMPTY, which indicates that all the characters written in US\_THR have been processed. When the current character processing is completed, the last character written in US\_THR is transferred into the Shift Register of the transmitter and US\_THR becomes empty, thus TXRDY rises.

Both TXRDY and TXEMPTY bits are low when the transmitter is disabled. Writing a character in US\_THR while TXRDY is low has no effect and the written character is lost.

If the USART is in SPI Slave Mode and if a character must be sent while the US\_THR is empty, the UNRE (Underrun Error) bit is set. The TXD transmission line stays at high level during all this time. The UNRE bit is cleared by writing a one to the RSTSTA (Reset Status) bit in the Control Register (US\_CR).

In SPI Master Mode, the slave select line (NSS) is asserted at low level 1 Tbit (Time bit) before the transmission of the MSB bit and released at high level 1 Tbit after the transmission of the LSB bit. So, the slave select line (NSS) is always released between each character transmission and a minimum delay of 3 Tbits always inserted. However, in order to address slave devices supporting the CSAAT mode (Chip Select Active After Transfer), the slave select line (NSS) can be forced at low level by writing a one to the RTSEN bit in the US\_CR. The slave select line (NSS) can be released at high level only by writing a one to the RTSDIS bit in the US\_CR (for example, when all data have been transferred to the slave device).

In SPI Slave Mode, the transmitter does not require a falling edge of the slave select line (NSS) to initiate a character transmission but only a low level. However, this low level must be present on the slave select line (NSS) at least 1 Tbit before the first serial clock cycle corresponding to the MSB bit.

### 36.7.7.6 Character Reception

When a character reception is completed, it is transferred to the Receive Holding Register (US\_RHR) and the RXRDY bit in the Status Register (US\_CSR) rises. If a character is completed while RXRDY is set, the OVRE (Overrun Error) bit is set. The last character is transferred into US\_RHR and overwrites the previous one. The OVRE bit is cleared by writing a one to the RSTSTA (Reset Status) bit the US\_CR.

To ensure correct behavior of the receiver in SPI Slave Mode, the master device sending the frame must ensure a minimum delay of 1 Tbit between each character transmission. The receiver does not require a falling edge of the slave select line (NSS) to initiate a character reception but only a low level. However, this low level must be present on the slave select line (NSS) at least 1 Tbit before the first serial clock cycle corresponding to the MSB bit.

### 36.7.7.7 Receiver Timeout

Because the receiver baud rate clock is active only during data transfers in SPI Mode, a receiver timeout is impossible in this mode, whatever the Time-out value is (field TO) in the Time-out Register (US\_RTOR).

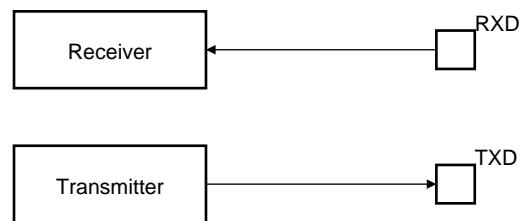
## 36.7.8 Test Modes

The USART can be programmed to operate in three different test modes. The internal loopback capability allows on-board diagnostics. In the loopback mode the USART interface pins are disconnected or not and reconfigured for loopback internally or externally.

### 36.7.8.1 Normal Mode

Normal mode connects the RXD pin on the receiver input and the transmitter output on the TXD pin.

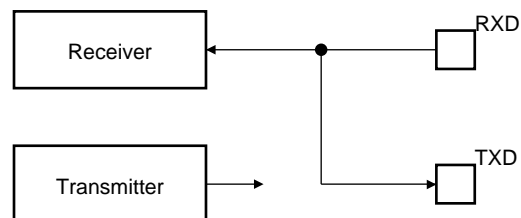
Figure 36-40. Normal Mode Configuration



### 36.7.8.2 Automatic Echo Mode

Automatic echo mode allows bit-by-bit retransmission. When a bit is received on the RXD pin, it is sent to the TXD pin, as shown in Figure 36-41. Programming the transmitter has no effect on the TXD pin. The RXD pin is still connected to the receiver input, thus the receiver remains active.

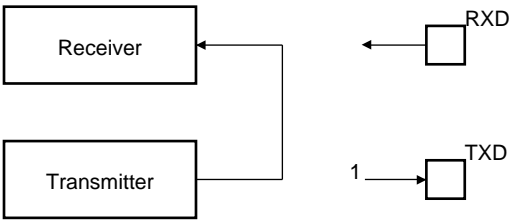
Figure 36-41. Automatic Echo Mode Configuration



### 36.7.8.3 Local Loopback Mode

Local loopback mode connects the output of the transmitter directly to the input of the receiver, as shown in Figure 36-42. The TXD and RXD pins are not used. The RXD pin has no effect on the receiver and the TXD pin is continuously driven high, as in idle state.

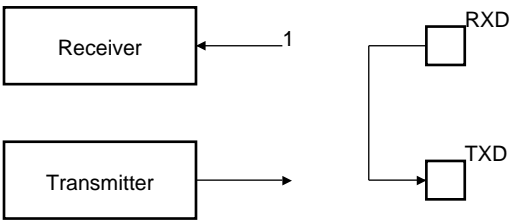
Figure 36-42. Local Loopback Mode Configuration



36.7.8.4 Remote Loopback Mode

Remote loopback mode directly connects the RXD pin to the TXD pin, as shown in [Figure 36-43](#). The transmitter and the receiver are disabled and have no effect. This mode allows bit-by-bit retransmission.

Figure 36-43. Remote Loopback Mode Configuration



### 36.7.9 Write Protection Registers

To prevent any single software error that may corrupt USART behavior, certain address spaces can be write-protected by setting the WPEN bit in the USART Write Protect Mode Register (US\_WPMR).

If a write access to the protected registers is detected, then the WPVS flag in the USART Write Protect Status Register (US\_WPSR) is set and the field WPVSR indicates in which register the write access has been attempted.

The WPVS flag is automatically reset by reading the US\_WPMR with the appropriate access key (WPKEY).

The protected registers are:

- “USART Mode Register”
- “USART Baud Rate Generator Register”
- “USART Receiver Time-out Register”
- “USART Transmitter Timeguard Register”
- “USART FI DI RATIO Register”
- “USART IrDA FILTER Register”
- “USART Manchester Configuration Register”

## 36.8 Universal Synchronous Asynchronous Receiver Transmitter (USART) User Interface

Table 36-15. Register Mapping

Offset	Register	Name	Access	Reset
0x0000	Control Register	US_CR	Write-only	–
0x0004	Mode Register	US_MR	Read-write	–
0x0008	Interrupt Enable Register	US_IER	Write-only	–
0x000C	Interrupt Disable Register	US_IDR	Write-only	–
0x0010	Interrupt Mask Register	US_IMR	Read-only	0x0
0x0014	Channel Status Register	US_CSR	Read-only	–
0x0018	Receiver Holding Register	US_RHR	Read-only	0x0
0x001C	Transmitter Holding Register	US_THR	Write-only	–
0x0020	Baud Rate Generator Register	US_BRGR	Read-write	0x0
0x0024	Receiver Time-out Register	US_RTOR	Read-write	0x0
0x0028	Transmitter Timeguard Register	US_TTGR	Read-write	0x0
0x2C–0x3C	Reserved	–	–	–
0x0040	FI DI Ratio Register	US_FIDI	Read-write	0x174
0x0044	Number of Errors Register	US_NER	Read-only	–
0x0048	Reserved	–	–	–
0x004C	IrDA Filter Register	US_IF	Read-write	0x0
0x0050	Manchester Encoder Decoder Register	US_MAN	Read-write	0xB0011004
0x0054–0x005C	Reserved	–	–	–
0x0060–0x00E0	Reserved	–	–	–
0x00E4	Write Protect Mode Register	US_WPMR	Read-write	0x0
0x00E8	Write Protect Status Register	US_WPSR	Read-only	0x0
0x00EC–0x00FC	Reserved	–	–	–
0x100–0x128	Reserved for PDC Registers	–	–	–

### 36.8.1 USART Control Register

**Name:** US\_CR

**Address:** 0x40024000 (0), 0x40028000 (1), 0x4002C000 (2), 0x40030000 (3), 0x40034000 (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	RTSDIS	RTSEN	–	–
15	14	13	12	11	10	9	8
RETTO	RSTNACK	RSTIT	SENA	STTTO	STPBRK	STTBRK	RSTSTA
7	6	5	4	3	2	1	0
TXDIS	TXEN	RXDIS	RXEN	RSTTX	RSTRX	–	–

For SPI control, see [“USART Control Register \(SPI\\_MODE\)”](#) on page 752.

- **RSTRX: Reset Receiver**

0: No effect.

1: Resets the receiver.

- **RSTTX: Reset Transmitter**

0: No effect.

1: Resets the transmitter.

- **RXEN: Receiver Enable**

0: No effect.

1: Enables the receiver, if RXDIS is 0.

- **RXDIS: Receiver Disable**

0: No effect.

1: Disables the receiver.

- **TXEN: Transmitter Enable**

0: No effect.

1: Enables the transmitter if TXDIS is 0.

- **TXDIS: Transmitter Disable**

0: No effect.

1: Disables the transmitter.

- **RSTSTA: Reset Status Bits**

0: No effect.

1: Resets the status bits PARE, FRAME, OVRE, MANERR and RXBRK in US\_CSR.

- **STTBK: Start Break**

0: No effect.

1: Starts transmission of a break after the characters present in US\_THR and the Transmit Shift Register have been transmitted. No effect if a break is already being transmitted.

- **STPBK: Stop Break**

0: No effect.

1: Stops transmission of the break after a minimum of one character length and transmits a high level during 12-bit periods. No effect if no break is being transmitted.

- **STTTO: Start Time-out**

0: No effect.

1: Starts waiting for a character before clocking the time-out counter. Resets the status bit TIMEOUT in US\_CSR.

- **SENDA: Send Address**

0: No effect.

1: In Multidrop Mode only, the next character written to the US\_THR is sent with the address bit set.

- **RSTIT: Reset Iterations**

0: No effect.

1: Resets ITERATION in US\_CSR. No effect if the ISO7816 is not enabled.

- **RSTNACK: Reset Non Acknowledge**

0: No effect

1: Resets NACK in US\_CSR.

- **RETTO: Rearm Time-out**

0: No effect

1: Restart Time-out

- **RTSEN: Request to Send Enable**

0: No effect.

1: Drives the pin RTS to 0.

- **RTSDIS: Request to Send Disable**

0: No effect.

1: Drives the pin RTS to 1.

### 36.8.2 USART Control Register (SPI\_MODE)

**Name:** US\_CR (SPI\_MODE)

**Address:** 0x40024000 (0), 0x40028000 (1), 0x4002C000 (2), 0x40030000 (3), 0x40034000 (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	RCS	FCS	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	RSTSTA
7	6	5	4	3	2	1	0
TXDIS	TXEN	RXDIS	RXEN	RSTTX	RSTRX	–	–

This configuration is relevant only if USART\_MODE=0xE or 0xF in “[USART Mode Register](#)” on page 754.

- **RSTRX: Reset Receiver**

0: No effect.

1: Resets the receiver.

- **RSTTX: Reset Transmitter**

0: No effect.

1: Resets the transmitter.

- **RXEN: Receiver Enable**

0: No effect.

1: Enables the receiver, if RXDIS is 0.

- **RXDIS: Receiver Disable**

0: No effect.

1: Disables the receiver.

- **TXEN: Transmitter Enable**

0: No effect.

1: Enables the transmitter if TXDIS is 0.

- **TXDIS: Transmitter Disable**

0: No effect.

1: Disables the transmitter.



- **RSTSTA: Reset Status Bits**

0: No effect.

1: Resets the status bits OVRE, UNRE in US\_CSR.

- **FCS: Force SPI Chip Select**

Applicable if USART operates in SPI Master Mode (USART\_MODE = 0xE):

0: No effect.

1: Forces the Slave Select Line NSS (RTS pin) to 0, even if USART is not transmitting, in order to address SPI slave devices supporting the CSAAT Mode (Chip Select Active After Transfer).

- **RCS: Release SPI Chip Select**

Applicable if USART operates in SPI Master Mode (USART\_MODE = 0xE):

0: No effect.

1: Releases the Slave Select Line NSS (RTS pin).

### 36.8.3 USART Mode Register

**Name:** US\_MR

**Address:** 0x40024004 (0), 0x40028004 (1), 0x4002C004 (2), 0x40030004 (3), 0x40034004 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
ONEBIT	MODSYNC	MAN	FILTER	–	MAX_ITERATION		
23	22	21	20	19	18	17	16
INVDATA	VAR_SYNC	DSNACK	INACK	OVER	CLKO	MODE9	MSBF
15	14	13	12	11	10	9	8
CHMODE		NBSTOP		PAR		SYNC	
7	6	5	4	3	2	1	0
CHRL		USCLKS		USART_MODE			

This register can only be written if the WPEN bit is cleared in “USART Write Protect Mode Register” on page 778.

For SPI configuration, see “USART Mode Register (SPI\_MODE)” on page 757.

#### • USART\_MODE: USART Mode of Operation

Value	Name	Description
0x0	NORMAL	Normal mode
0x1	RS485	RS485
0x2	HW_HANDSHAKING	Hardware Handshaking
0x4	IS07816_T_0	IS07816 Protocol: T = 0
0x6	IS07816_T_1	IS07816 Protocol: T = 1
0x8	IRDA	IrDA
0xE	SPI_MASTER	SPI Master
0xF	SPI_SLAVE	SPI Slave

The PDC transfers are supported in all USART Mode of Operation.

#### • USCLKS: Clock Selection

Value	Name	Description
0	MCK	Master Clock MCK is selected
1	DIV	Internal Clock Divided MCK/DIV (DIV=8) is selected
3	SCK	Serial Clock SLK is selected

#### • CHRL: Character Length

Value	Name	Description
0	5_BIT	Character length is 5 bits

1	6_BIT	Character length is 6 bits
2	7_BIT	Character length is 7 bits
3	8_BIT	Character length is 8 bits

- **SYNC: Synchronous Mode Select**

0: USART operates in Asynchronous Mode.

1: USART operates in Synchronous Mode.

- **PAR: Parity Type**

Value	Name	Description
0	EVEN	Even parity
1	ODD	Odd parity
2	SPACE	Parity forced to 0 (Space)
3	MARK	Parity forced to 1 (Mark)
4	NO	No parity
6	MULTIDROP	Multidrop mode

- **NBSTOP: Number of Stop Bits**

Value	Name	Description
0	1_BIT	1 stop bit
1	1_5_BIT	1.5 stop bit (SYNC = 0) or reserved (SYNC = 1)
2	2_BIT	2 stop bits

- **CHMODE: Channel Mode**

Value	Name	Description
0	NORMAL	Normal Mode
1	AUTOMATIC	Automatic Echo. Receiver input is connected to the TXD pin.
2	LOCAL_LOOPBACK	Local Loopback. Transmitter output is connected to the Receiver Input.
3	REMOTE_LOOPBACK	Remote Loopback. RXD pin is internally connected to the TXD pin.

- **MSBF: Bit Order**

0: Least Significant Bit is sent/received first.

1: Most Significant Bit is sent/received first.

- **MODE9: 9-bit Character Length**

0: CHRL defines character length.

1: 9-bit character length.

- **CLKO: Clock Output Select**

0: The USART does not drive the SCK pin.

1: The USART drives the SCK pin if USCLKS does not select the external clock SCK.

- **OVER: Oversampling Mode**

0: 16x Oversampling.

1: 8x Oversampling.

- **INACK: Inhibit Non Acknowledge**

0: The NACK is generated.

1: The NACK is not generated.

- **DSNACK: Disable Successive NACK**

0: NACK is sent on the ISO line as soon as a parity error occurs in the received character (unless INACK is set).

1: Successive parity errors are counted up to the value specified in the MAX\_ITERATION field. These parity errors generate a NACK on the ISO line. As soon as this value is reached, no additional NACK is sent on the ISO line. The flag ITERATION is asserted.

- **INVDATA: Inverted Data**

0: The data field transmitted on TXD line is the same as the one written in US\_THR register or the content read in US\_RHR is the same as RXD line. Normal mode of operation.

1: The data field transmitted on TXD line is inverted (voltage polarity only) compared to the value written on US\_THR register or the content read in US\_RHR is inverted compared to what is received on RXD line (or ISO7816 IO line). Inverted Mode of operation, useful for contactless card application. To be used with configuration bit MSBF.

- **VAR\_SYNC: Variable Synchronization of Command/Data Sync Start Frame Delimiter**

0: User defined configuration of command or data sync field depending on MODSYNC value.

1: The sync field is updated when a character is written into US\_THR register.

- **MAX\_ITERATION: Maximum Number of Automatic Iteration**

0–7: Defines the maximum number of iterations in mode ISO7816, protocol T = 0.

- **FILTER: Infrared Receive Line Filter**

0: The USART does not filter the receive line.

1: The USART filters the receive line using a three-sample filter (1/16-bit clock) (2 over 3 majority).

- **MAN: Manchester Encoder/Decoder Enable**

0: Manchester Encoder/Decoder are disabled.

1: Manchester Encoder/Decoder are enabled.

- **MODSYNC: Manchester Synchronization Mode**

0: The Manchester Start bit is a 0 to 1 transition

1: The Manchester Start bit is a 1 to 0 transition.

- **ONEBIT: Start Frame Delimiter Selector**

0: Start Frame delimiter is COMMAND or DATA SYNC.

1: Start Frame delimiter is One Bit.

### 36.8.4 USART Mode Register (SPI\_MODE)

**Name:** US\_MR (SPI\_MODE)

**Address:** 0x40024004 (0), 0x40028004 (1), 0x4002C004 (2), 0x40030004 (3), 0x40034004 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	WRDBT	–	–	–	CPOL
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	CPHA
7	6	5	4	3	2	1	0
CHRL		USCLKS		USART_MODE			

This configuration is relevant only if USART\_MODE = 0xE or 0xF in “[USART Mode Register](#)” on page 754.

This register can only be written if the WPEN bit is cleared in “[USART Write Protect Mode Register](#)” on page 778.

#### • USART\_MODE: USART Mode of Operation

Value	Name	Description
0xE	SPI_MASTER	SPI Master
0xF	SPI_SLAVE	SPI Slave

#### • USCLKS: Clock Selection

Value	Name	Description
0	MCK	Master Clock MCK is selected
1	DIV	Internal Clock Divided MCK/DIV (DIV=8) is selected
3	SCK	Serial Clock SLK is selected

#### • CHRL: Character Length

Value	Name	Description
3	8_BIT	Character length is 8 bits

#### • CPHA: SPI Clock Phase

- Applicable if USART operates in SPI Mode (USART\_MODE = 0xE or 0xF):

0: Data is changed on the leading edge of SPCK and captured on the following edge of SPCK.

1: Data is captured on the leading edge of SPCK and changed on the following edge of SPCK.

CPHA determines which edge of SPCK causes data to change and which edge causes data to be captured. CPHA is used with CPOL to produce the required clock/data relationship between master and slave devices.

- **CHMODE: Channel Mode**

Value	Name	Description
0	NORMAL	Normal Mode
1	AUTOMATIC	Automatic Echo. Receiver input is connected to the TXD pin.
2	LOCAL_LOOPBACK	Local Loopback. Transmitter output is connected to the Receiver Input.
3	REMOTE_LOOPBACK	Remote Loopback. RXD pin is internally connected to the TXD pin.

- **CPOL: SPI Clock Polarity**

Applicable if USART operates in SPI Mode (Slave or Master, USART\_MODE = 0xE or 0xF):

0: The inactive state value of SPCK is logic level zero.

1: The inactive state value of SPCK is logic level one.

CPOL is used to determine the inactive state value of the serial clock (SPCK). It is used with CPHA to produce the required clock/data relationship between master and slave devices.

- **WRDBT: Wait Read Data Before Transfer**

0: The character transmission starts as soon as a character is written into US\_THR register (assuming TXRDY was set).

1: The character transmission starts when a character is written and only if RXRDY flag is cleared (Receiver Holding Register has been read).

### 36.8.5 USART Interrupt Enable Register

**Name:** US\_IER

**Address:** 0x40024008 (0), 0x40028008 (1), 0x4002C008 (2), 0x40030008 (3), 0x40034008 (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	MANE
23	22	21	20	19	18	17	16
–	–	–	–	CTSIC	–	–	–
15	14	13	12	11	10	9	8
–	–	NACK	RXBUFF	TXBUFE	ITER	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

For SPI specific configuration, see [“USART Interrupt Enable Register \(SPI\\_MODE\)” on page 760](#).

The following configuration values are valid for all listed bit names of this register:

0: No effect

1: Enables the corresponding interrupt.

- **RXRDY: RXRDY Interrupt Enable**
- **TXRDY: TXRDY Interrupt Enable**
- **RXBRK: Receiver Break Interrupt Enable**
- **ENDRX: End of Receive Transfer Interrupt Enable (available in all USART modes of operation)**
- **ENDTX: End of Transmit Interrupt Enable (available in all USART modes of operation)**
- **OVRE: Overrun Error Interrupt Enable**
- **FRAME: Framing Error Interrupt Enable**
- **PARE: Parity Error Interrupt Enable**
- **TIMEOUT: Time-out Interrupt Enable**
- **TXEMPTY: TXEMPTY Interrupt Enable**
- **ITER: Max number of Repetitions Reached Interrupt Enable**
- **TXBUFE: Buffer Empty Interrupt Enable (available in all USART modes of operation)**
- **RXBUFF: Buffer Full Interrupt Enable (available in all USART modes of operation)**
- **NACK: Non Acknowledge Interrupt Enable**
- **CTSIC: Clear to Send Input Change Interrupt Enable**
- **MANE: Manchester Error Interrupt Enable**

### 36.8.6 USART Interrupt Enable Register (SPI\_MODE)

**Name:** US\_IER (SPI\_MODE)

**Address:** 0x40024008 (0), 0x40028008 (1), 0x4002C008 (2), 0x40030008 (3), 0x40034008 (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	UNRE	TXEMPTY	–
7	6	5	4	3	2	1	0
–	–	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

This configuration is relevant only if USART\_MODE = 0xE or 0xF in “[USART Mode Register](#)” on page 754.

The following configuration values are valid for all listed bit names of this register:

0: No effect

1: Enables the corresponding interrupt.

- **RXRDY: RXRDY Interrupt Enable**
- **TXRDY: TXRDY Interrupt Enable**
- **OVRE: Overrun Error Interrupt Enable**
- **TXEMPTY: TXEMPTY Interrupt Enable**
- **UNRE: SPI Underrun Error Interrupt Enable**



### 36.8.7 USART Interrupt Disable Register

**Name:** US\_IDR

**Address:** 0x4002400C (0), 0x4002800C (1), 0x4002C00C (2), 0x4003000C (3), 0x4003400C (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	MANE
23	22	21	20	19	18	17	16
–	–	–	–	CTSIC	–	–	–
15	14	13	12	11	10	9	8
–	–	NACK	RXBUFF	TXBUFE	ITER	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

For SPI specific configuration, see [“USART Interrupt Disable Register \(SPI\\_MODE\)” on page 762](#).

The following configuration values are valid for all listed bit names of this register:

0: No effect

1: Disables the corresponding interrupt.

- **RXRDY: RXRDY Interrupt Disable**
- **TXRDY: TXRDY Interrupt Disable**
- **RXBRK: Receiver Break Interrupt Disable**
- **ENDRX: End of Receive Transfer Interrupt Disable (available in all USART modes of operation)**
- **ENDTX: End of Transmit Interrupt Disable (available in all USART modes of operation)**
- **OVRE: Overrun Error Interrupt Enable**
- **FRAME: Framing Error Interrupt Disable**
- **PARE: Parity Error Interrupt Disable**
- **TIMEOUT: Time-out Interrupt Disable**
- **TXEMPTY: TXEMPTY Interrupt Disable**
- **ITER: Max Number of Repetitions Reached Interrupt Disable**
- **TXBUFE: Buffer Empty Interrupt Disable (available in all USART modes of operation)**
- **RXBUFF: Buffer Full Interrupt Disable (available in all USART modes of operation)**
- **NACK: Non Acknowledge Interrupt Disable**
- **CTSIC: Clear to Send Input Change Interrupt Disable**
- **MANE: Manchester Error Interrupt Disable**

### 36.8.8 USART Interrupt Disable Register (SPI\_MODE)

**Name:** US\_IDR (SPI\_MODE)

**Address:** 0x4002400C (0), 0x4002800C (1), 0x4002C00C (2), 0x4003000C (3), 0x4003400C (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	UNRE	TXEMPTY	–
7	6	5	4	3	2	1	0
–	–	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

This configuration is relevant only if USART\_MODE = 0xE or 0xF in “[USART Mode Register](#)” on page 754.

The following configuration values are valid for all listed bit names of this register:

0: No effect

1: Disables the corresponding interrupt.

- **RXRDY: RXRDY Interrupt Disable**
- **TXRDY: TXRDY Interrupt Disable**
- **OVRE: Overrun Error Interrupt Disable**
- **TXEMPTY: TXEMPTY Interrupt Disable**
- **UNRE: SPI Underrun Error Interrupt Disable**

### 36.8.9 USART Interrupt Mask Register

**Name:** US\_IMR

**Address:** 0x40024010 (0), 0x40028010 (1), 0x4002C010 (2), 0x40030010 (3), 0x40034010 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	MANE
23	22	21	20	19	18	17	16
–	–	–	–	CTSIC	–	–	–
15	14	13	12	11	10	9	8
–	–	NACK	RXBUFF	TXBUFE	ITER	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

For SPI specific configuration, see [“USART Interrupt Mask Register \(SPI\\_MODE\)” on page 764](#).

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is not enabled.

1: The corresponding interrupt is enabled.

- **RXRDY: RXRDY Interrupt Mask**
- **TXRDY: TXRDY Interrupt Mask**
- **RXBRK: Receiver Break Interrupt Mask**
- **ENDRX: End of Receive Transfer Interrupt Mask (available in all USART modes of operation)**
- **ENDTX: End of Transmit Interrupt Mask (available in all USART modes of operation)**
- **OVRE: Overrun Error Interrupt Mask**
- **FRAME: Framing Error Interrupt Mask**
- **PARE: Parity Error Interrupt Mask**
- **TIMEOUT: Time-out Interrupt Mask**
- **TXEMPTY: TXEMPTY Interrupt Mask**
- **ITER: Max Number of Repetitions Reached Interrupt Mask**
- **TXBUFE: Buffer Empty Interrupt Mask (available in all USART modes of operation)**
- **RXBUFF: Buffer Full Interrupt Mask (available in all USART modes of operation)**
- **NACK: Non Acknowledge Interrupt Mask**
- **CTSIC: Clear to Send Input Change Interrupt Mask**
- **MANE: Manchester Error Interrupt Mask**

### 36.8.10 USART Interrupt Mask Register (SPI\_MODE)

**Name:** US\_IMR (SPI\_MODE)

**Address:** 0x40024010 (0), 0x40028010 (1), 0x4002C010 (2), 0x40030010 (3), 0x40034010 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	UNRE	TXEMPTY	–
7	6	5	4	3	2	1	0
–	–	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

This configuration is relevant only if USART\_MODE = 0xE or 0xF in “[USART Mode Register](#)” on page 754.

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is not enabled.

1: The corresponding interrupt is enabled.

- **RXRDY: RXRDY Interrupt Mask**
- **TXRDY: TXRDY Interrupt Mask**
- **OVRE: Overrun Error Interrupt Mask**
- **TXEMPTY: TXEMPTY Interrupt Mask**
- **UNRE: SPI Underrun Error Interrupt Mask**

### 36.8.11 USART Channel Status Register

**Name:** US\_CSR

**Address:** 0x40024014 (0), 0x40028014 (1), 0x4002C014 (2), 0x40030014 (3), 0x40034014 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	MANERR
23	22	21	20	19	18	17	16
CTS	–	–	–	CTSIC	–	–	–
15	14	13	12	11	10	9	8
–	–	NACK	RXBUFF	TXBUFE	ITER	TXEMPTY	TIMEOUT
7	6	5	4	3	2	1	0
PARE	FRAME	OVRE	ENDTX	ENDRX	RXBRK	TXRDY	RXRDY

For SPI specific configuration, see [“USART Channel Status Register \(SPI\\_MODE\)” on page 767](#).

- **RXRDY: Receiver Ready**

0: No complete character has been received since the last read of US\_RHR or the receiver is disabled. If characters were being received when the receiver was disabled, RXRDY changes to 1 when the receiver is enabled.

1: At least one complete character has been received and US\_RHR has not yet been read.

- **TXRDY: Transmitter Ready**

0: A character is in the US\_THR waiting to be transferred to the Transmit Shift Register, or an STTBRK command has been requested, or the transmitter is disabled. As soon as the transmitter is enabled, TXRDY becomes 1.

1: There is no character in the US\_THR.

- **RXBRK: Break Received/End of Break**

0: No Break received or End of Break detected since the last RSTSTA.

1: Break Received or End of Break detected since the last RSTSTA.

- **ENDRX: End of Receiver Transfer**

0: The End of Transfer signal from the Receive PDC channel is inactive.

1: The End of Transfer signal from the Receive PDC channel is active.

- **ENDTX: End of Transmitter Transfer**

0: The End of Transfer signal from the Transmit PDC channel is inactive.

1: The End of Transfer signal from the Transmit PDC channel is active.

- **OVRE: Overrun Error**

0: No overrun error has occurred since the last RSTSTA.

1: At least one overrun error has occurred since the last RSTSTA.

- **FRAME: Framing Error**

0: No stop bit has been detected low since the last RSTSTA.

1: At least one stop bit has been detected low since the last RSTSTA.

- **PARE: Parity Error**

0: No parity error has been detected since the last RSTSTA.

1: At least one parity error has been detected since the last RSTSTA.

- **TIMEOUT: Receiver Time-out**

0: There has not been a time-out since the last Start Time-out command (STTTO in US\_CR) or the Time-out Register is 0.

1: There has been a time-out since the last Start Time-out command (STTTO in US\_CR).

- **TXEMPTY: Transmitter Empty**

0: There are characters in either US\_THR or the Transmit Shift Register, or the transmitter is disabled.

1: There are no characters in US\_THR, nor in the Transmit Shift Register.

- **ITER: Max Number of Repetitions Reached**

0: Maximum number of repetitions has not been reached since the last RSTSTA.

1: Maximum number of repetitions has been reached since the last RSTSTA.

- **TXBUFE: Transmission Buffer Empty**

0: The signal Buffer Empty from the Transmit PDC channel is inactive.

1: The signal Buffer Empty from the Transmit PDC channel is active.

- **RXBUFF: Reception Buffer Full**

0: The signal Buffer Full from the Receive PDC channel is inactive.

1: The signal Buffer Full from the Receive PDC channel is active.

- **NACK: Non Acknowledge Interrupt**

0: Non Acknowledge has not been detected since the last RSTNACK.

1: At least one Non Acknowledge has been detected since the last RSTNACK.

- **CTSIC: Clear to Send Input Change Flag**

0: No input change has been detected on the CTS pin since the last read of US\_CSR.

1: At least one input change has been detected on the CTS pin since the last read of US\_CSR.

- **CTS: Image of CTS Input**

0: CTS is set to 0.

1: CTS is set to 1.

- **MANERR: Manchester Error**

0: No Manchester error has been detected since the last RSTSTA.

1: At least one Manchester error has been detected since the last RSTSTA.

### 36.8.12 USART Channel Status Register (SPI\_MODE)

**Name:** US\_CSR (SPI\_MODE)

**Address:** 0x40024014 (0), 0x40028014 (1), 0x4002C014 (2), 0x40030014 (3), 0x40034014 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	RXBUFF	TXBUFE	UNRE	TXEMPTY	–
7	6	5	4	3	2	1	0
–	–	OVRE	ENDTX	ENDRX	–	TXRDY	RXRDY

This configuration is relevant only if USART\_MODE = 0xE or 0xF in “[USART Mode Register](#)” on page 754.

- **RXRDY: Receiver Ready**

0: No complete character has been received since the last read of US\_RHR or the receiver is disabled. If characters were being received when the receiver was disabled, RXRDY changes to 1 when the receiver is enabled.

1: At least one complete character has been received and US\_RHR has not yet been read.

- **TXRDY: Transmitter Ready**

0: A character is in the US\_THR waiting to be transferred to the Transmit Shift Register or the transmitter is disabled. As soon as the transmitter is enabled, TXRDY becomes 1.

1: There is no character in the US\_THR.

- **OVRE: Overrun Error**

0: No overrun error has occurred since the last RSTSTA.

1: At least one overrun error has occurred since the last RSTSTA.

- **TXEMPTY: Transmitter Empty**

0: There are characters in either US\_THR or the Transmit Shift Register, or the transmitter is disabled.

1: There are no characters in US\_THR, nor in the Transmit Shift Register.

- **UNRE: Underrun Error**

0: No SPI underrun error has occurred since the last RSTSTA.

1: At least one SPI underrun error has occurred since the last RSTSTA.

### 36.8.13 USART Receive Holding Register

**Name:** US\_RHR

**Address:** 0x40024018 (0), 0x40028018 (1), 0x4002C018 (2), 0x40030018 (3), 0x40034018 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
RXSYNH	–	–	–	–	–	–	RXCHR
7	6	5	4	3	2	1	0
RXCHR							

- **RXCHR: Received Character**

Last character received if RXRDY is set.

- **RXSYNH: Received Sync**

0: Last Character received is a Data.

1: Last Character received is a Command.



### 36.8.14 USART Transmit Holding Register

**Name:** US\_THR

**Address:** 0x4002401C (0), 0x4002801C (1), 0x4002C01C (2), 0x4003001C (3), 0x4003401C (4)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TXSYNH	–	–	–	–	–	–	TXCHR
7	6	5	4	3	2	1	0
TXCHR							

- **TXCHR: Character to be Transmitted**

Next character to be transmitted after the current character if TXRDY is not set.

- **TXSYNH: Sync Field to be Transmitted**

0: The next character sent is encoded as a data. Start Frame Delimiter is DATA SYNC.

1: The next character sent is encoded as a command. Start Frame Delimiter is COMMAND SYNC.

### 36.8.15 USART Baud Rate Generator Register

**Name:** US\_BRGR

**Address:** 0x40024020 (0), 0x40028020 (1), 0x4002C020 (2), 0x40030020 (3), 0x40034020 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	FP		
15	14	13	12	11	10	9	8
CD							
7	6	5	4	3	2	1	0
CD							

This register can only be written if the WPEN bit is cleared in “USART Write Protect Mode Register” on page 778.

#### • CD: Clock Divider

CD	USART_MODE ≠ ISO7816			USART_MODE = ISO7816
	SYNC = 0		SYNC = 1 or USART_MODE = SPI (Master or Slave)	
	OVER = 0	OVER = 1		
0	Baud Rate Clock Disabled			
1 to 65535	Baud Rate = Selected Clock/(16*CD)	Baud Rate = Selected Clock/(8*CD)	Baud Rate = Selected Clock/CD	Baud Rate = Selected Clock/(FI_DI_RATIO*CD)

#### • FP: Fractional Part

0: Fractional divider is disabled.

1–7: Baud rate resolution, defined by  $FP \times 1/8$ .

### 36.8.16 USART Receiver Time-out Register

**Name:** US\_RTOR

**Address:** 0x40024024 (0), 0x40028024 (1), 0x4002C024 (2), 0x40030024 (3), 0x40034024 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TO							
7	6	5	4	3	2	1	0
TO							

This register can only be written if the WPEN bit is cleared in [“USART Write Protect Mode Register” on page 778](#).

- **TO: Time-out Value**

0: The Receiver Time-out is disabled.

1–65535: The Receiver Time-out is enabled and the Time-out delay is TO x Bit Period.

### 36.8.17 USART Transmitter Timeguard Register

**Name:** US\_TTGR

**Address:** 0x40024028 (0), 0x40028028 (1), 0x4002C028 (2), 0x40030028 (3), 0x40034028 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
TG							

This register can only be written if the WPEN bit is cleared in [“USART Write Protect Mode Register” on page 778](#).

- **TG: Timeguard Value**

0: The Transmitter Timeguard is disabled.

1–255: The Transmitter timeguard is enabled and the timeguard delay is TG x Bit Period.

### 36.8.18 USART FI DI RATIO Register

**Name:** US\_FIDI

**Address:** 0x40024040 (0), 0x40028040 (1), 0x4002C040 (2), 0x40030040 (3), 0x40034040 (4)

**Access:** Read-write

**Reset:** 0x174

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
FI_DI_RATIO							
7	6	5	4	3	2	1	0
FI_DI_RATIO							

This register can only be written if the WPEN bit is cleared in “[USART Write Protect Mode Register](#)” on page 778.

- **FI\_DI\_RATIO: FI Over DI Ratio Value**

0: If ISO7816 mode is selected, the Baud Rate Generator generates no signal.

1–2047: If ISO7816 mode is selected, the baud rate is the clock provided on SCK divided by FI\_DI\_RATIO.

### 36.8.19 USART Number of Errors Register

**Name:** US\_NER

**Address:** 0x40024044 (0), 0x40028044 (1), 0x4002C044 (2), 0x40030044 (3), 0x40034044 (4)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
NB_ERRORS							

This register is relevant only if USART\_MODE = 0x4 or 0x6 in [“USART Mode Register” on page 754](#).

- **NB\_ERRORS: Number of Errors**

Total number of errors that occurred during an ISO7816 transfer. This register automatically clears when read.

### 36.8.20 USART IrDA FILTER Register

**Name:** US\_IF

**Address:** 0x4002404C (0), 0x4002804C (1), 0x4002C04C (2), 0x4003004C (3), 0x4003404C (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
IRDA_FILTER							

This register is relevant only if USART\_MODE = 0x8 in “[USART Mode Register](#)” on page 754.

This register can only be written if the WPEN bit is cleared in “[USART Write Protect Mode Register](#)” on page 778.

- **IRDA\_FILTER: IrDA Filter**

The IRDA\_FILTER value must be defined to meet the following criteria:

$$t_{\text{MCK}} * (\text{IRDA\_FILTER} + 3) < 1.41 \mu\text{s}$$

### 36.8.21 USART Manchester Configuration Register

**Name:** US\_MAN

**Address:** 0x40024050 (0), 0x40028050 (1), 0x4002C050 (2), 0x40030050 (3), 0x40034050 (4)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	DRIFT	ONE	RX_MPOL	–	–	RX_PP	
23	22	21	20	19	18	17	16
–	–	–	–	RX_PL			
15	14	13	12	11	10	9	8
–	–	–	TX_MPOL	–	–	TX_PP	
7	6	5	4	3	2	1	0
–	–	–	–	TX_PL			

This register can only be written if the WPEN bit is cleared in “USART Write Protect Mode Register” on page 778.

- **TX\_PL: Transmitter Preamble Length**

0: The Transmitter Preamble pattern generation is disabled

1–15: The Preamble Length is TX\_PL x Bit Period

- **TX\_PP: Transmitter Preamble Pattern**

The following values assume that TX\_MPOL field is not set:

Value	Name	Description
00	ALL_ONE	The preamble is composed of ‘1’s
01	ALL_ZERO	The preamble is composed of ‘0’s
10	ZERO_ONE	The preamble is composed of ‘01’s
11	ONE_ZERO	The preamble is composed of ‘10’s

- **TX\_MPOL: Transmitter Manchester Polarity**

0: Logic Zero is coded as a zero-to-one transition, Logic One is coded as a one-to-zero transition.

1: Logic Zero is coded as a one-to-zero transition, Logic One is coded as a zero-to-one transition.

- **RX\_PL: Receiver Preamble Length**

0: The receiver preamble pattern detection is disabled

1–15: The detected preamble length is RX\_PL x Bit Period

- **RX\_PP: Receiver Preamble Pattern detected**

The following values assume that RX\_MPOL field is not set:

Value	Name	Description
00	ALL_ONE	The preamble is composed of ‘1’s



01	ALL_ZERO	The preamble is composed of '0's
10	ZERO_ONE	The preamble is composed of '01's
11	ONE_ZERO	The preamble is composed of '10's

- **RX\_MPOL: Receiver Manchester Polarity**

0: Logic Zero is coded as a zero-to-one transition, Logic One is coded as a one-to-zero transition.

1: Logic Zero is coded as a one-to-zero transition, Logic One is coded as a zero-to-one transition.

- **ONE: Must Be Set to 1**

Bit 29 must always be set to 1 when programming the US\_MAN register.

- **DRIFT: Drift Compensation**

0: The USART can not recover from an important clock drift

1: The USART can recover from clock drift. The 16X clock mode must be enabled.

### 36.8.22 USART Write Protect Mode Register

**Name:** US\_WPMR

**Address:** 0x400240E4 (0), 0x400280E4 (1), 0x4002C0E4 (2), 0x400300E4 (3), 0x400340E4 (4)

**Access:** Read-write

**Reset:** See [Table 36-15](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	WPEN

- **WPEN: Write Protect Enable**

0: Disables the Write Protect if WPKEY corresponds to 0x555341 (“USA” in ASCII).

1: Enables the Write Protect if WPKEY corresponds to 0x555341 (“USA” in ASCII).

Protects the registers:

- “USART Mode Register” on page 754
- “USART Baud Rate Generator Register” on page 770
- “USART Receiver Time-out Register” on page 771
- “USART Transmitter Timeguard Register” on page 772
- “USART FI DI RATIO Register” on page 773
- “USART IrDA FILTER Register” on page 775
- “USART Manchester Configuration Register” on page 776

- **WPKEY: Write Protect KEY**

Value	Name	Description
0x555341	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

### 36.8.23 USART Write Protect Status Register

**Name:** US\_WPSR

**Address:** 0x400240E8 (0), 0x400280E8 (1), 0x4002C0E8 (2), 0x400300E8 (3), 0x400340E8 (4)

**Access:** Read-only

**Reset:** See [Table 36-15](#)

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	WPVS

- **WPVS: Write Protect Violation Status**

0: No Write Protect Violation has occurred since the last read of the US\_WPSR.

1: A Write Protect Violation has occurred since the last read of the US\_WPSR. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protect Violation Source**

When WPVS is active, this field indicates the write-protected register (through address offset or code) in which a write access has been attempted.

**Note:** Reading US\_WPSR automatically clears all fields.

## 37. Timer Counter (TC)

### 37.1 Description

The Timer Counter (TC) includes three identical 16-bit Timer Counter channels.

Each channel can be independently programmed to perform a wide range of functions including frequency measurement, event counting, interval measurement, pulse generation, delay timing and pulse width modulation.

Each channel has three external clock inputs, five internal clock inputs and two multi-purpose input/output signals which can be configured by the user. Each channel drives an internal interrupt signal which can be programmed to generate processor interrupts.

The Timer Counter (TC) embeds a quadrature decoder logic connected in front of the timers and driven by TIOA0, TIOB0 and TIOB1 inputs. When enabled, the quadrature decoder performs the input lines filtering, decoding of quadrature signals and connects to the timers/counters in order to read the position and speed of the motor through the user interface.

The Timer Counter block has two global registers which act upon all TC channels.

The Block Control Register allows the channels to be started simultaneously with the same instruction.

The Block Mode Register defines the external clock inputs for each channel, allowing them to be chained.

Table 37-1 gives the assignment of the device Timer Counter clock inputs common to Timer Counter 0 to 2.

**Table 37-1. Timer Counter Clock Assignment**

Name	Definition
TIMER_CLOCK1	MCK/2
TIMER_CLOCK2	MCK/8
TIMER_CLOCK3	MCK/32
TIMER_CLOCK4	MCK/128
TIMER_CLOCK5 <sup>(1)</sup>	SLCK

Note: 1. When Slow Clock is selected for Master Clock (CSS = 0 in PMC Master Clock Register), TIMER\_CLOCK5 input is equivalent to Master Clock.

### 37.2 Embedded Characteristics

- Provides three 16-bit Timer Counter channels
- Wide range of functions including:
  - Frequency measurement
  - Event counting
  - Interval measurement
  - Pulse generation
  - Delay timing
  - Pulse Width Modulation
  - Up/down capabilities
  - Quadrature decoder logic
  - 2-bit gray up/down count for stepper motor
- Each channel is user-configurable and contains:
  - Three external clock inputs
  - Five Internal clock inputs
  - Two multi-purpose input/output signals acting as trigger event

- Internal interrupt signal
- Two global registers that act on all TC channels
- Configuration registers can be write protected

### 37.3 Block Diagram

Figure 37-1. Timer Counter Block Diagram

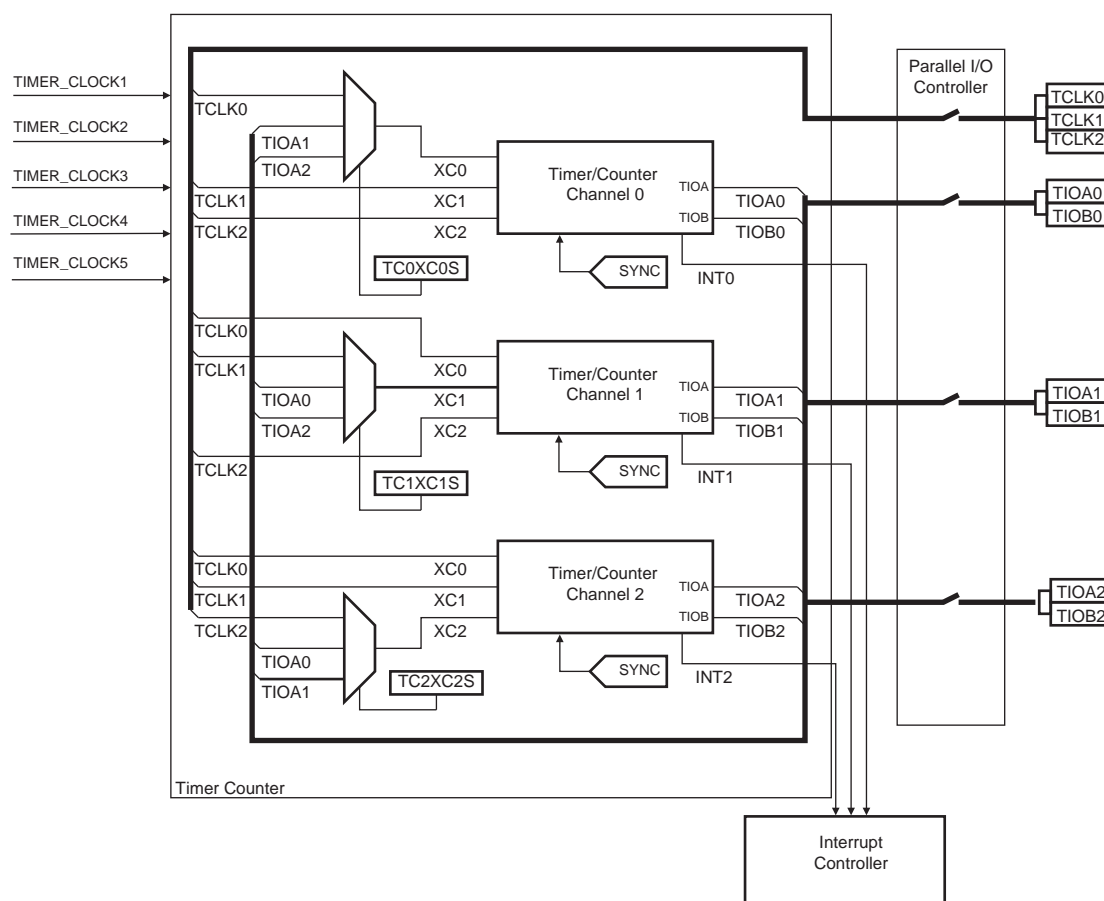


Table 37-2. Signal Name Description

Block/Channel	Signal Name	Description
Channel Signal	XC0, XC1, XC2	External Clock Inputs
	TIOA	Capture Mode: Timer Counter Input Waveform Mode: Timer Counter Output
	TIOB	Capture Mode: Timer Counter Input Waveform Mode: Timer Counter Input/Output
	INT	Interrupt Signal Output (internal signal)
	SYNC	Synchronization Input Signal (from configuration register)

## 37.4 Pin Name List

Table 37-3. TC pin list

Pin Name	Description	Type
TCLK0–TCLK2	External Clock Input	Input
TIOA0–TIOA2	I/O Line A	I/O
TIOB0–TIOB2	I/O Line B	I/O

## 37.5 Product Dependencies

### 37.5.1 I/O Lines

The pins used for interfacing the compliant external devices may be multiplexed with PIO lines. The programmer must first program the PIO controllers to assign the TC pins to their peripheral functions.

Table 37-4. I/O Lines

Instance	Signal	I/O Line	Peripheral
TC0	TCLK0	PB4	B
TC0	TCLK1	PB9	A
TC0	TCLK2	PB12	A
TC0	TIOA0	PA13	B
TC0	TIOA1	PB7	A
TC0	TIOA2	PB10	A
TC0	TIOB0	PA14	B
TC0	TIOB1	PB8	A
TC0	TIOB2	PB11	A
TC1	TCLK3	PB26	A
TC1	TCLK4	PA17	B
TC1	TCLK5	PA19	B
TC1	TIOA3	PB24	A
TC1	TIOA4	PA15	B
TC1	TIOA5	PA18	B
TC1	TIOB3	PB25	A
TC1	TIOB4	PA16	B
TC1	TIOB5	PA20	B

### 37.5.2 Power Management

The TC is clocked through the Power Management Controller (PMC), thus the programmer must first configure the PMC to enable the Timer Counter clock.

### 37.5.3 Interrupt

The TC has an interrupt line connected to the Interrupt Controller (IC). Handling the TC interrupt requires programming the IC before configuring the TC.

## 37.6 Functional Description

### 37.6.1 TC Description

The three channels of the Timer Counter are independent and identical in operation except when quadrature decoder is enabled. The registers for channel programming are listed in [Table 37-5 on page 801](#).

### 37.6.2 16-bit Counter

Each channel is organized around a 16-bit counter. The value of the counter is incremented at each positive edge of the selected clock. When the counter has reached the value 0xFFFF and passes to 0x0000, an overflow occurs and the COVFS bit in the TC Status Register (TC\_SR) is set.

The current value of the counter is accessible in real time by reading the TC Counter Value Register (TC\_CV). The counter can be reset by a trigger. In this case, the counter value passes to 0x0000 on the next valid edge of the selected clock.

### 37.6.3 Clock Selection

At block level, input clock signals of each channel can either be connected to the external inputs TCLK0, TCLK1 or TCLK2, or be connected to the internal I/O signals TIOA0, TIOA1 or TIOA2 for chaining by programming the TC Block Mode Register (TC\_BMR). See [Figure 37-2 “Clock Chaining Selection”](#).

Each channel can independently select an internal or external clock source for its counter:

- Internal clock signals: TIMER\_CLOCK1, TIMER\_CLOCK2, TIMER\_CLOCK3, TIMER\_CLOCK4, TIMER\_CLOCK5
- External clock signals: XC0, XC1 or XC2

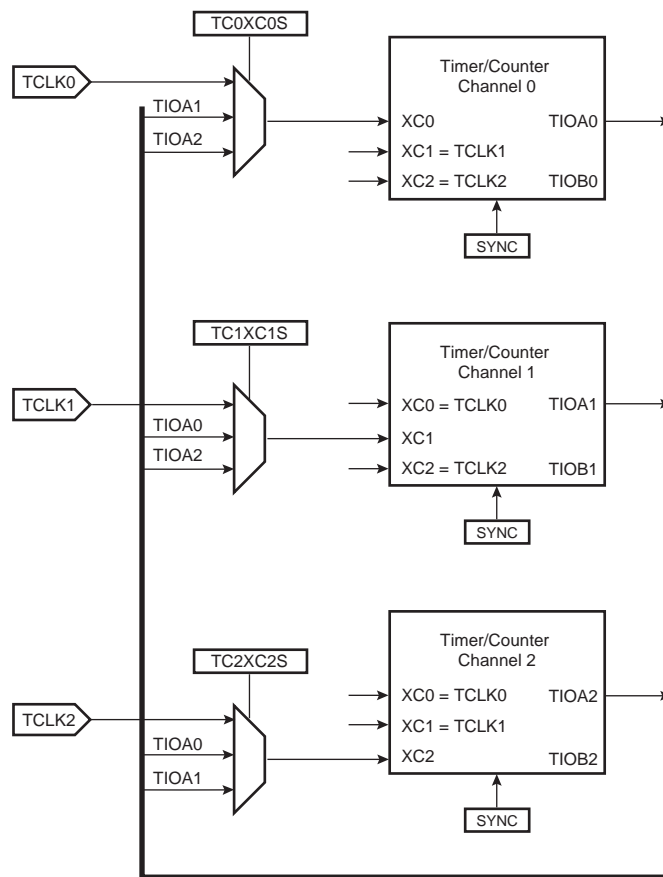
This selection is made by the TCCLKS bits in the TC Channel Mode Register (TC\_CMR).

The selected clock can be inverted with the CLKI bit in the TC\_CMR. This allows counting on the opposite edges of the clock.

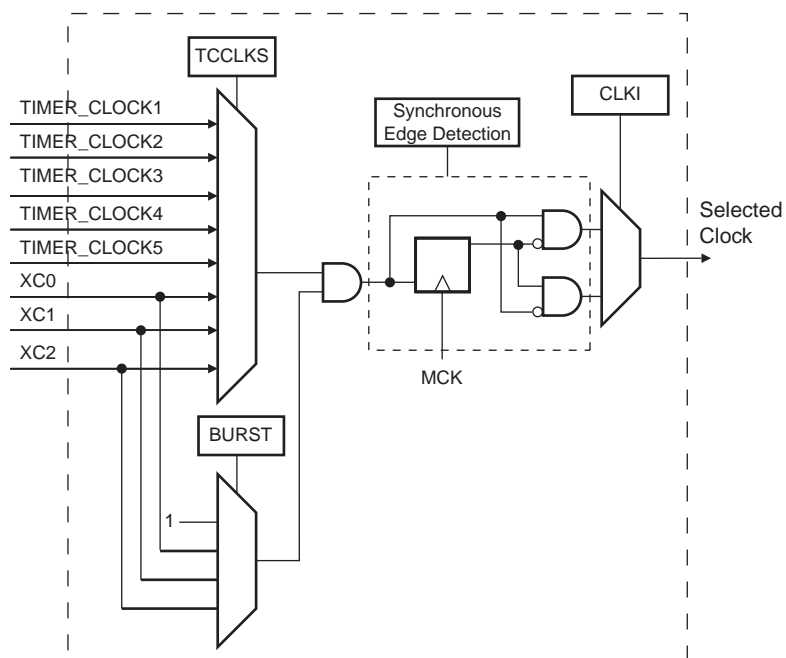
The burst function allows the clock to be validated when an external signal is high. The BURST parameter in the TC\_CMR defines this signal (none, XC0, XC1, XC2). See [Figure 37-3 “Clock Selection”](#).

**Note:** In all cases, if an external clock is used, the duration of each of its levels must be longer than the master clock period. The external clock frequency must be at least 2.5 times lower than the master clock

**Figure 37-2. Clock Chaining Selection**



**Figure 37-3. Clock Selection**



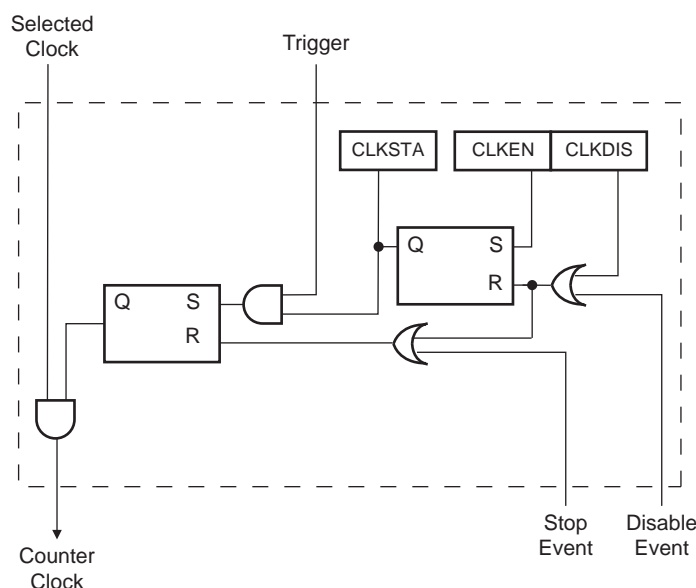


### 37.6.4 Clock Control

The clock of each counter can be controlled in two different ways: it can be enabled/disabled and started/stopped. See [Figure 37-4](#).

- The clock can be enabled or disabled by the user with the CLKEN and the CLKDIS commands in the TC Channel Control Register (TC\_CCR). In Capture Mode it can be disabled by an RB load event if LDBDIS is set to 1 in the TC\_CMR. In Waveform Mode, it can be disabled by an RC Compare event if CPCDIS is set to 1 in TC\_CMR. When disabled, the start or the stop actions have no effect: only a CLKEN command in the TC\_CCR can re-enable the clock. When the clock is enabled, the CLKSTA bit is set in the TC\_SR.
- The clock can also be started or stopped: a trigger (software, synchro, external or compare) always starts the clock. The clock can be stopped by an RB load event in Capture Mode (LDBSTOP = 1 in TC\_CMR) or a RC compare event in Waveform Mode (CPCSTOP = 1 in TC\_CMR). The start and the stop commands have effect only if the clock is enabled.

**Figure 37-4. Clock Control**



### 37.6.5 TC Operating Modes

Each channel can independently operate in two different modes:

- Capture Mode provides measurement on signals.
- Waveform Mode provides wave generation.

The TC Operating Mode is programmed with the WAVE bit in the TC Channel Mode Register.

In Capture Mode, TIOA and TIOB are configured as inputs.

In Waveform Mode, TIOA is always configured to be an output and TIOB is an output if it is not selected to be the external trigger.

### 37.6.6 Trigger

A trigger resets the counter and starts the counter clock. Three types of triggers are common to both modes, and a fourth external trigger is available to each mode.

Regardless of the trigger used, it will be taken into account at the following active edge of the selected clock. This means that the counter value can be read differently from zero just after a trigger, especially when a low frequency signal is selected as the clock.

The following triggers are common to both modes:

- Software Trigger: Each channel has a software trigger, available by setting SWTRG in TC\_CCR.
- SYNC: Each channel has a synchronization signal SYNC. When asserted, this signal has the same effect as a software trigger. The SYNC signals of all channels are asserted simultaneously by writing TC\_BCR (Block Control) with SYNC set.
- Compare RC Trigger: RC is implemented in each channel and can provide a trigger when the counter value matches the RC value if CPCTRG is set in the TC\_CMR.

The channel can also be configured to have an external trigger. In Capture Mode, the external trigger signal can be selected between TIOA and TIOB. In Waveform Mode, an external event can be programmed on one of the following signals: TIOB, XC0, XC1 or XC2. This external event can then be programmed to perform a trigger by setting bit ENETRIG in the TC\_CMR.

If an external trigger is used, the duration of the pulses must be longer than the master clock period in order to be detected.

### 37.6.7 Capture Operating Mode

This mode is entered by clearing the WAVE bit in the TC\_CMR.

Capture Mode allows the TC channel to perform measurements such as pulse timing, frequency, period, duty cycle and phase on TIOA and TIOB signals which are considered as inputs.

Figure 37-5 shows the configuration of the TC channel when programmed in Capture Mode.

### 37.6.8 Capture Registers A and B

Registers A and B (RA and RB) are used as capture registers. This means that they can be loaded with the counter value when a programmable event occurs on the signal TIOA.

The LDRA field in the TC\_CMR defines the TIOA selected edge for the loading of register A, and the LDRB field defines the TIOA selected edge for the loading of Register B.

RA is loaded only if it has not been loaded since the last trigger or if RB has been loaded since the last loading of RA.

RB is loaded only if RA has been loaded since the last trigger or the last loading of RB.

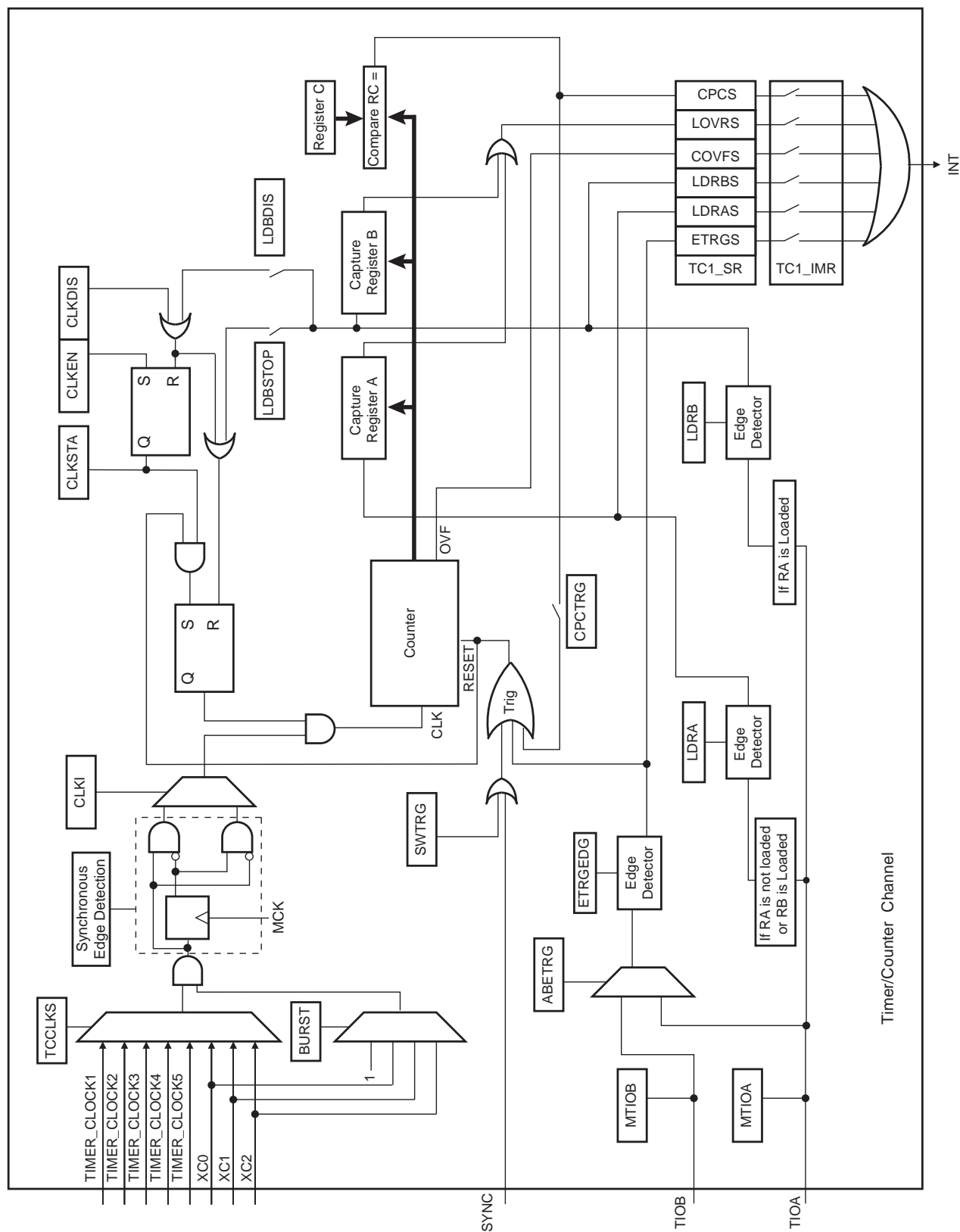
Loading RA or RB before the read of the last value loaded sets the Overrun Error Flag (LOVRS bit) in the TC\_SR. In this case, the old value is overwritten.

### 37.6.9 Trigger Conditions

In addition to the SYNC signal, the software trigger and the RC compare trigger, an external trigger can be defined.

The ABETRIG bit in the TC\_CMR selects TIOA or TIOB input signal as an external trigger. The External Trigger Edge Selection parameter (ETRGEDG field in TC\_CMR) defines the edge (rising, falling, or both) detected to generate an external trigger. If ETRGEDG = 0 (none), the external trigger is disabled.

Figure 37-5. Capture Mode



### 37.6.10 Waveform Operating Mode

Waveform operating mode is entered by setting the WAVE parameter in TC\_CMR (Channel Mode Register).

In Waveform Operating Mode the TC channel generates 1 or 2 PWM signals with the same frequency and independently programmable duty cycles, or generates different types of one-shot or repetitive pulses.

In this mode, TIOA is configured as an output and TIOB is defined as an output if it is not used as an external event (EEVT parameter in TC\_CMR).

Figure 37-6 shows the configuration of the TC channel when programmed in Waveform Operating Mode.

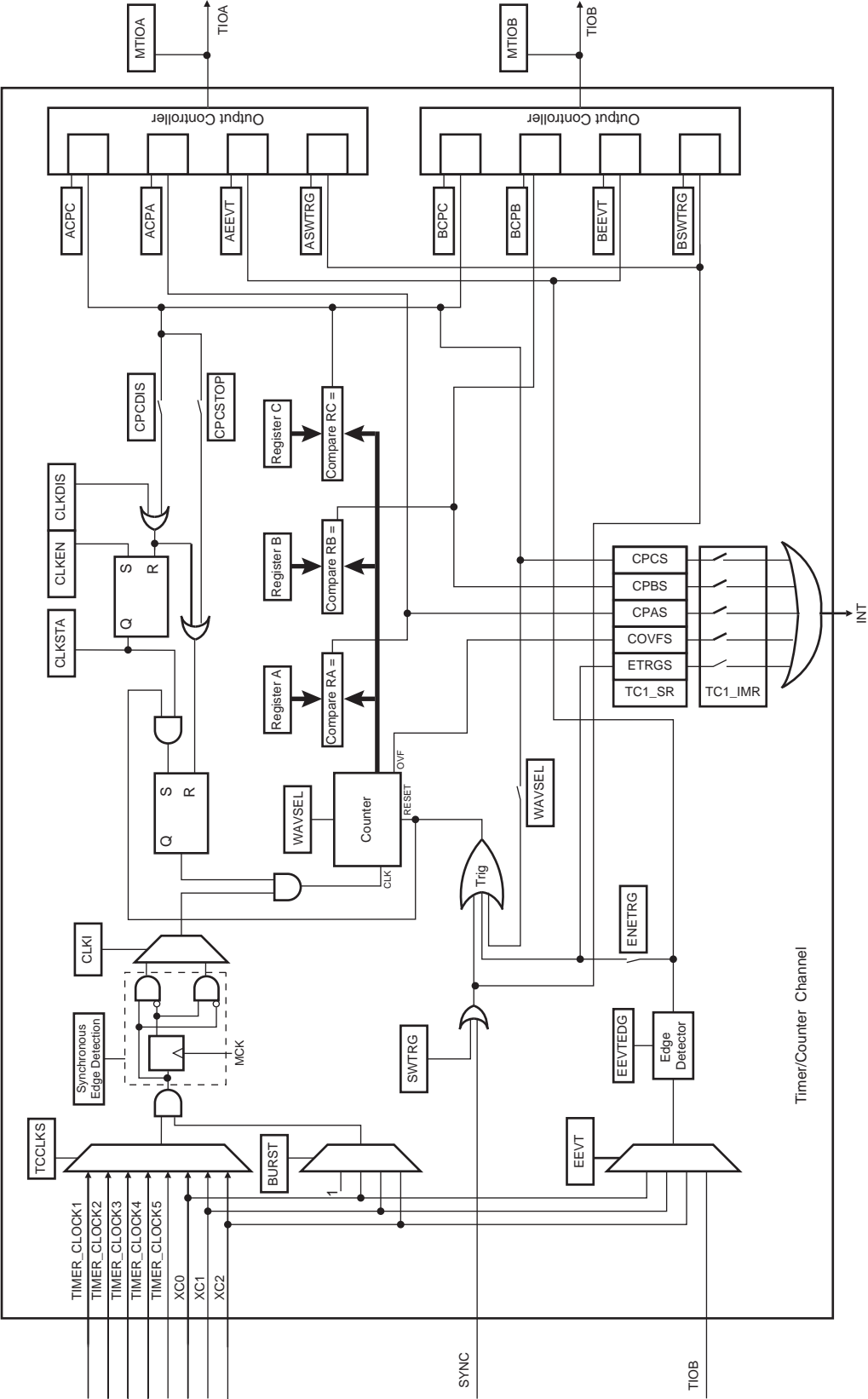
### 37.6.11 Waveform Selection

Depending on the WAVSEL parameter in TC\_CMR (Channel Mode Register), the behavior of TC\_CV varies.

With any selection, RA, RB and RC can all be used as compare registers.

RA Compare is used to control the TIOA output, RB Compare is used to control the TIOB output (if correctly configured) and RC Compare is used to control TIOA and/or TIOB outputs.

Figure 37-6. Waveform Mode



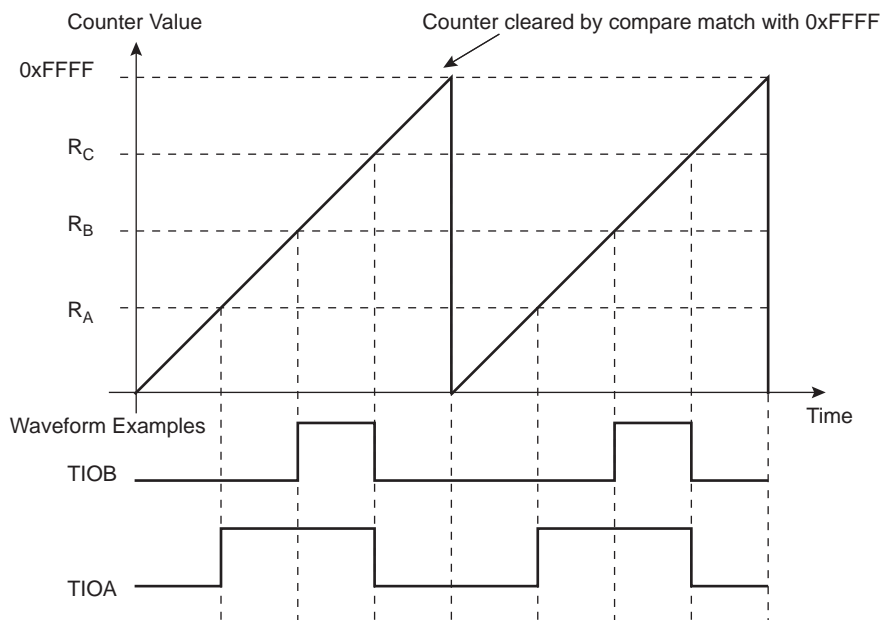
### 37.6.11.1 WAVSEL = 00

When WAVSEL = 00, the value of TC\_CV is incremented from 0 to 0xFFFF. Once 0xFFFF has been reached, the value of TC\_CV is reset. Incrementation of TC\_CV starts again and the cycle continues. See [Figure 37-7](#).

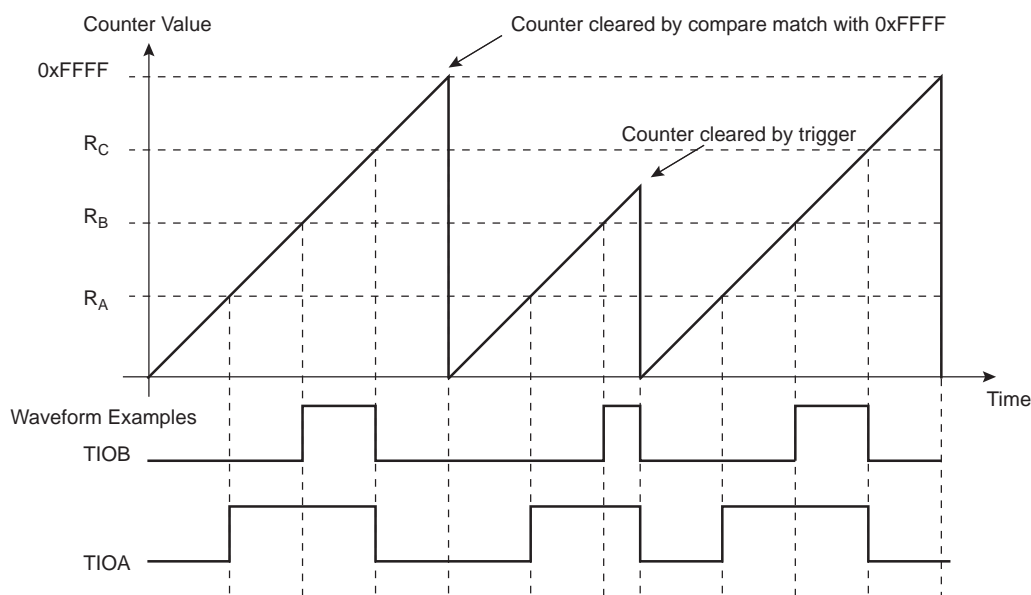
An external event trigger or a software trigger can reset the value of TC\_CV. It is important to note that the trigger may occur at any time. See [Figure 37-8](#).

RC Compare cannot be programmed to generate a trigger in this configuration. At the same time, RC Compare can stop the counter clock (CPCSTOP = 1 in TC\_CMR) and/or disable the counter clock (CPCDIS = 1 in TC\_CMR).

**Figure 37-7. WAVSEL = 00 without trigger**



**Figure 37-8. WAVSEL= 00 with Trigger**



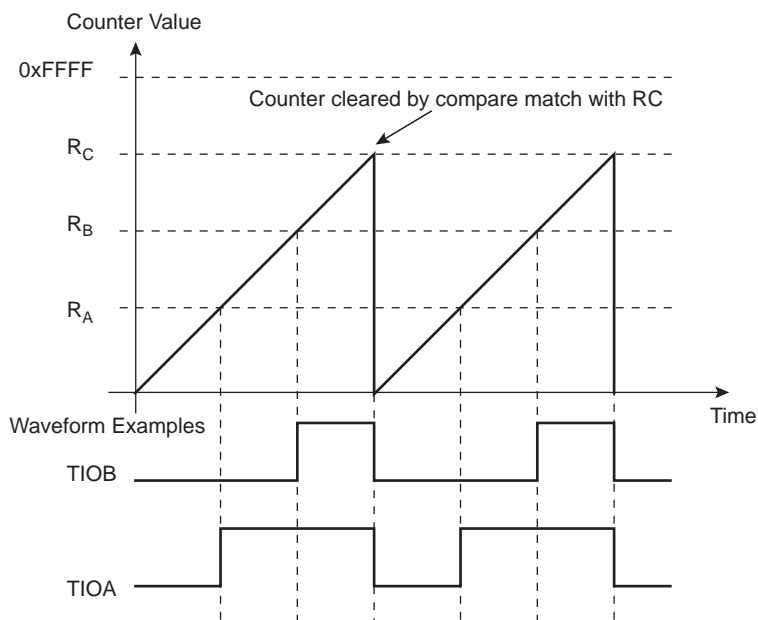
### 37.6.11.2 WAVSEL = 10

When WAVSEL = 10, the value of TC\_CV is incremented from 0 to the value of RC, then automatically reset on a RC Compare. Once the value of TC\_CV has been reset, it is then incremented and so on. See [Figure 37-9](#).

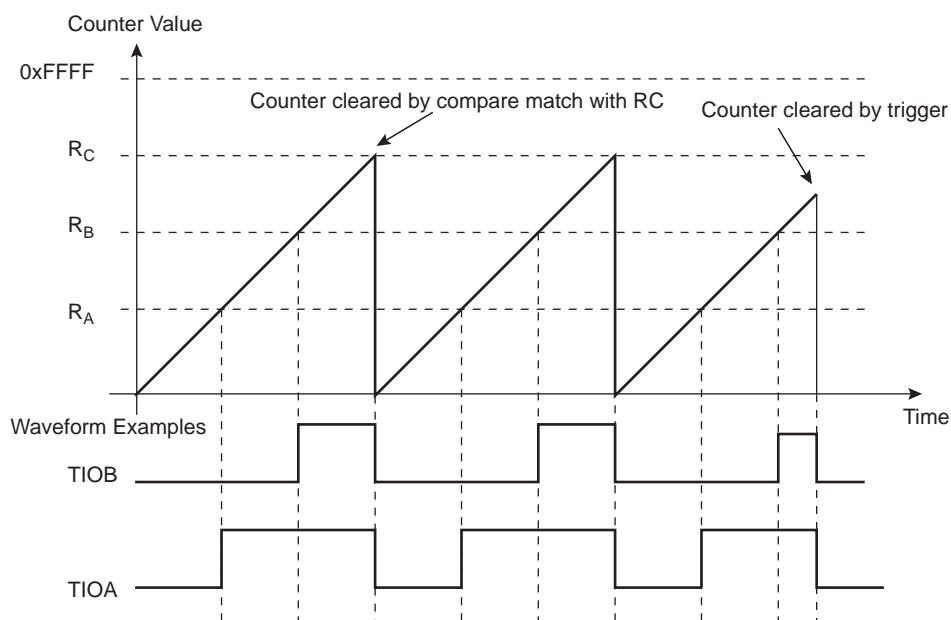
It is important to note that TC\_CV can be reset at any time by an external event or a software trigger if both are programmed correctly. See [Figure 37-10](#).

In addition, RC Compare can stop the counter clock (CPCSTOP = 1 in TC\_CMR) and/or disable the counter clock (CPCDIS = 1 in TC\_CMR).

**Figure 37-9. WAVSEL = 10 without Trigger**



**Figure 37-10. WAVSEL = 10 with Trigger**



### 37.6.11.3 WAVSEL = 01

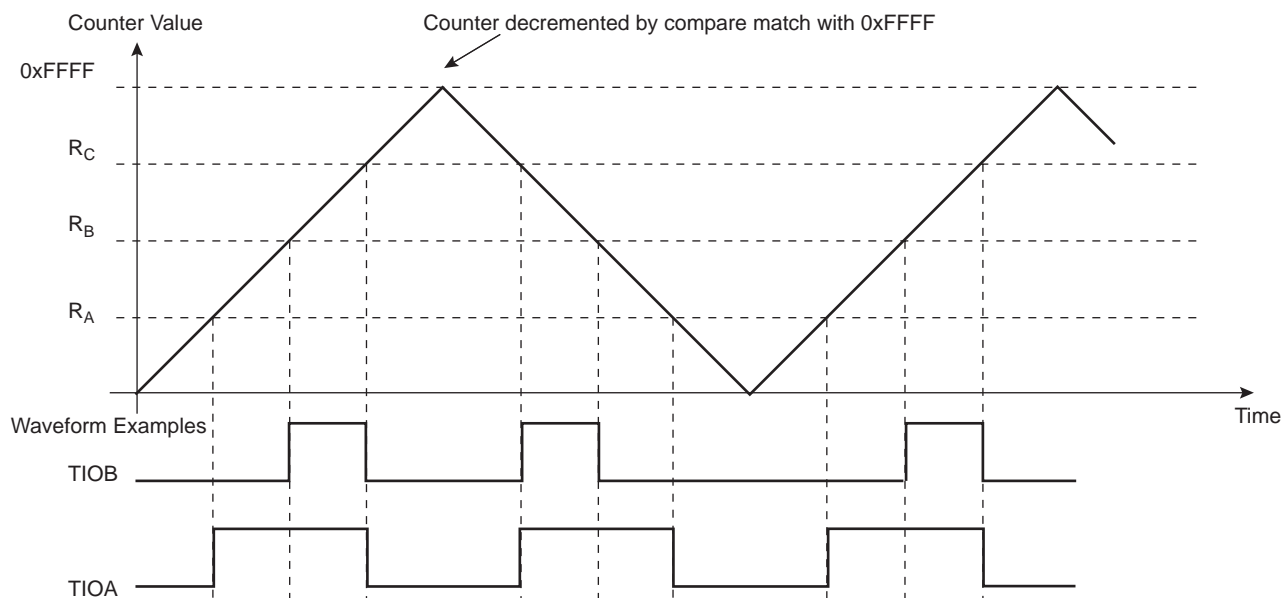
When WAVSEL = 01, the value of TC\_CV is incremented from 0 to 0xFFFF. Once 0xFFFF is reached, the value of TC\_CV is decremented to 0, then re-incremented to 0xFFFF and so on. See [Figure 37-11](#).

A trigger such as an external event or a software trigger can modify TC\_CV at any time. If a trigger occurs while TC\_CV is incrementing, TC\_CV then decrements. If a trigger is received while TC\_CV is decrementing, TC\_CV then increments. See [Figure 37-12](#).

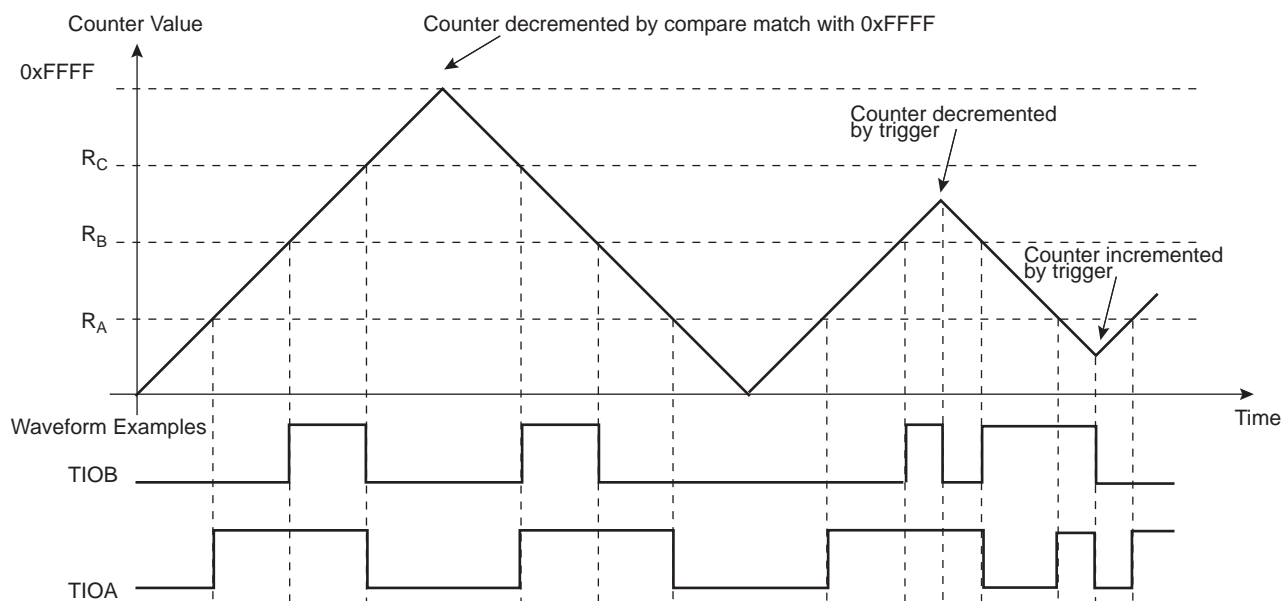
RC Compare cannot be programmed to generate a trigger in this configuration.

At the same time, RC Compare can stop the counter clock (CPCSTOP = 1) and/or disable the counter clock (CPCDIS = 1).

**Figure 37-11. WAVSEL = 01 without Trigger**



**Figure 37-12. WAVSEL = 01 with Trigger**





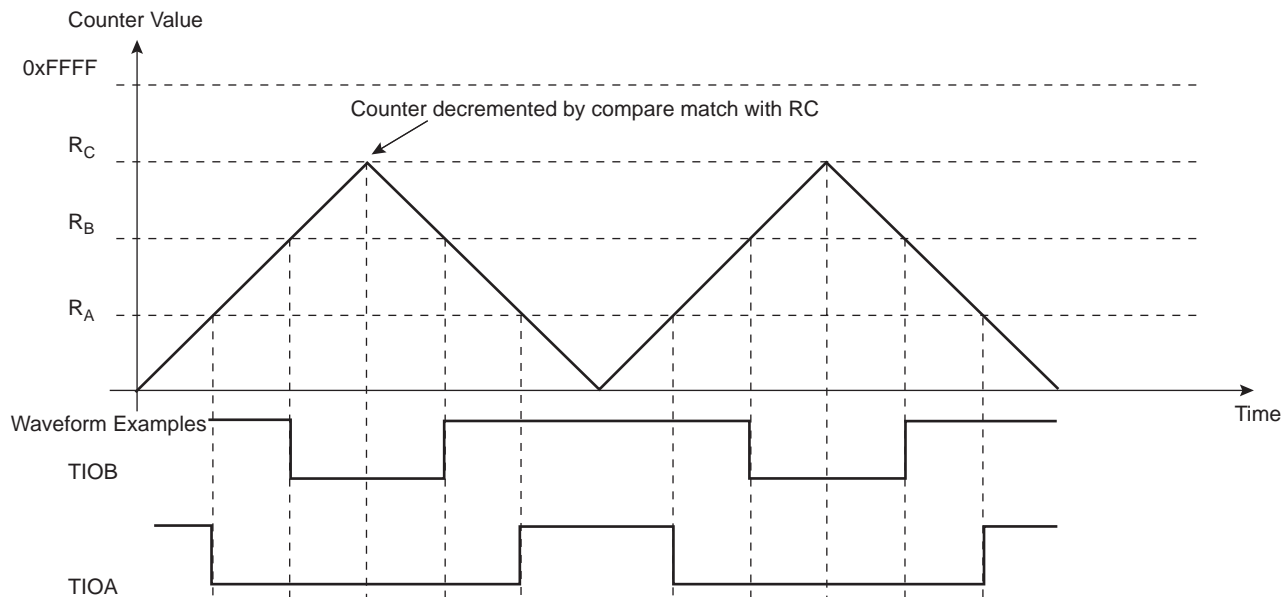
### 37.6.11.4 WAVSEL = 11

When WAVSEL = 11, the value of TC\_CV is incremented from 0 to RC. Once RC is reached, the value of TC\_CV is decremented to 0, then re-incremented to RC and so on. See Figure 37-13.

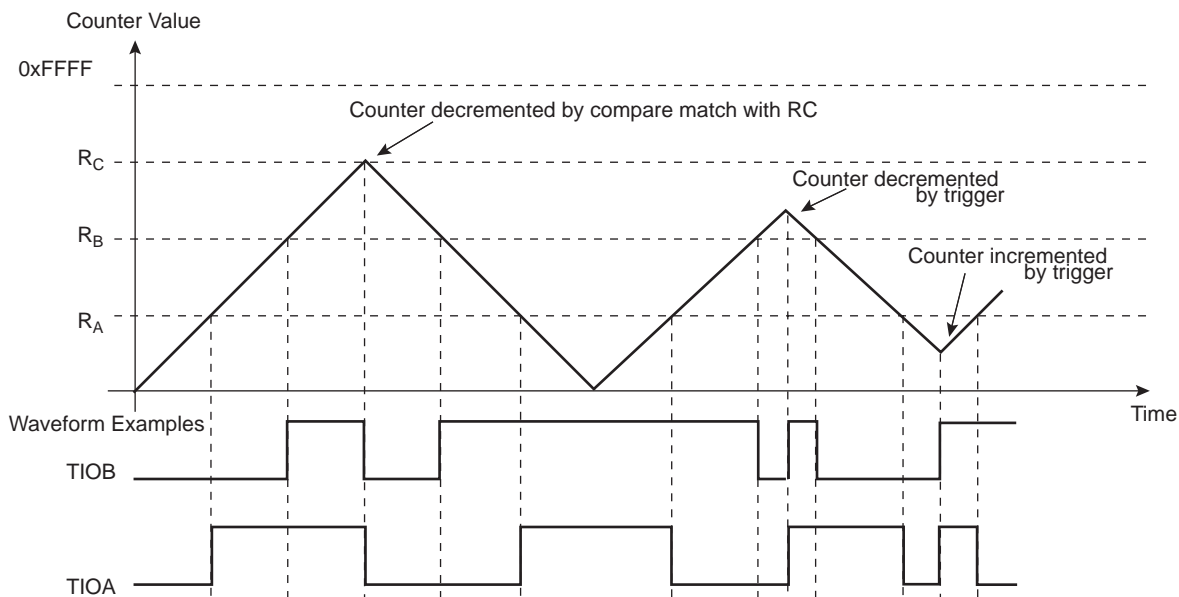
A trigger such as an external event or a software trigger can modify TC\_CV at any time. If a trigger occurs while TC\_CV is incrementing, TC\_CV then decrements. If a trigger is received while TC\_CV is decrementing, TC\_CV then increments. See Figure 37-14.

RC Compare can stop the counter clock (CPCSTOP = 1) and/or disable the counter clock (CPCDIS = 1).

**Figure 37-13. WAVSEL = 11 without Trigger**



**Figure 37-14. WAVSEL = 11 with Trigger**



### 37.6.12 External Event/Trigger Conditions

An external event can be programmed to be detected on one of the clock sources (XC0, XC1, XC2) or TIOB. The external event selected can then be used as a trigger.

The EEVT parameter in TC\_CMR selects the external trigger. The EEVTEDG parameter defines the trigger edge for each of the possible external triggers (rising, falling or both). If EEVTEDG is cleared (none), no external event is defined.

If TIOB is defined as an external event signal (EEVT = 0), TIOB is no longer used as an output and the compare register B is not used to generate waveforms and subsequently no IRQs. In this case the TC channel can only generate a waveform on TIOA.

When an external event is defined, it can be used as a trigger by setting bit ENETRIG in the TC\_CMR.

As in Capture Mode, the SYNC signal and the software trigger are also available as triggers. RC Compare can also be used as a trigger depending on the parameter WAVSEL.

### 37.6.13 Output Controller

The output controller defines the output level changes on TIOA and TIOB following an event. TIOB control is used only if TIOB is defined as output (not as an external event).

The following events control TIOA and TIOB: software trigger, external event and RC compare. RA compare controls TIOA and RB compare controls TIOB. Each of these events can be programmed to set, clear or toggle the output as defined in the corresponding parameter in TC\_CMR.

### 37.6.14 Quadrature Decoder Logic

#### 37.6.14.1 Description

The quadrature decoder logic is driven by TIOA0, TIOB0, TIOB1 input pins and drives the timer/counter of channel 0 and 1. Channel 2 can be used as a time base in case of speed measurement requirements (refer to [Figure 37-15 “Predefined Connection of the Quadrature Decoder with Timer Counters”](#)).

When writing a 0 to bit QDEN of the TC\_BMR, the quadrature decoder logic is totally transparent.

TIOA0 and TIOB0 are to be driven by the two dedicated quadrature signals from a rotary sensor mounted on the shaft of the off-chip motor.

A third signal from the rotary sensor can be processed through pin TIOB1 and is typically dedicated to be driven by an index signal if it is provided by the sensor. This signal is not required to decode the quadrature signals PHA, PHB.

Field TCCLKS of TC\_CMRx must be configured to select XC0 input (i.e., 0x101). Field TC0XC0S has no effect as soon as quadrature decoder is enabled.

Either speed or position/revolution can be measured. Position channel 0 accumulates the edges of PHA, PHB input signals giving a high accuracy on motor position whereas channel 1 accumulates the index pulses of the sensor, therefore the number of rotations. Concatenation of both values provides a high level of precision on motion system position.

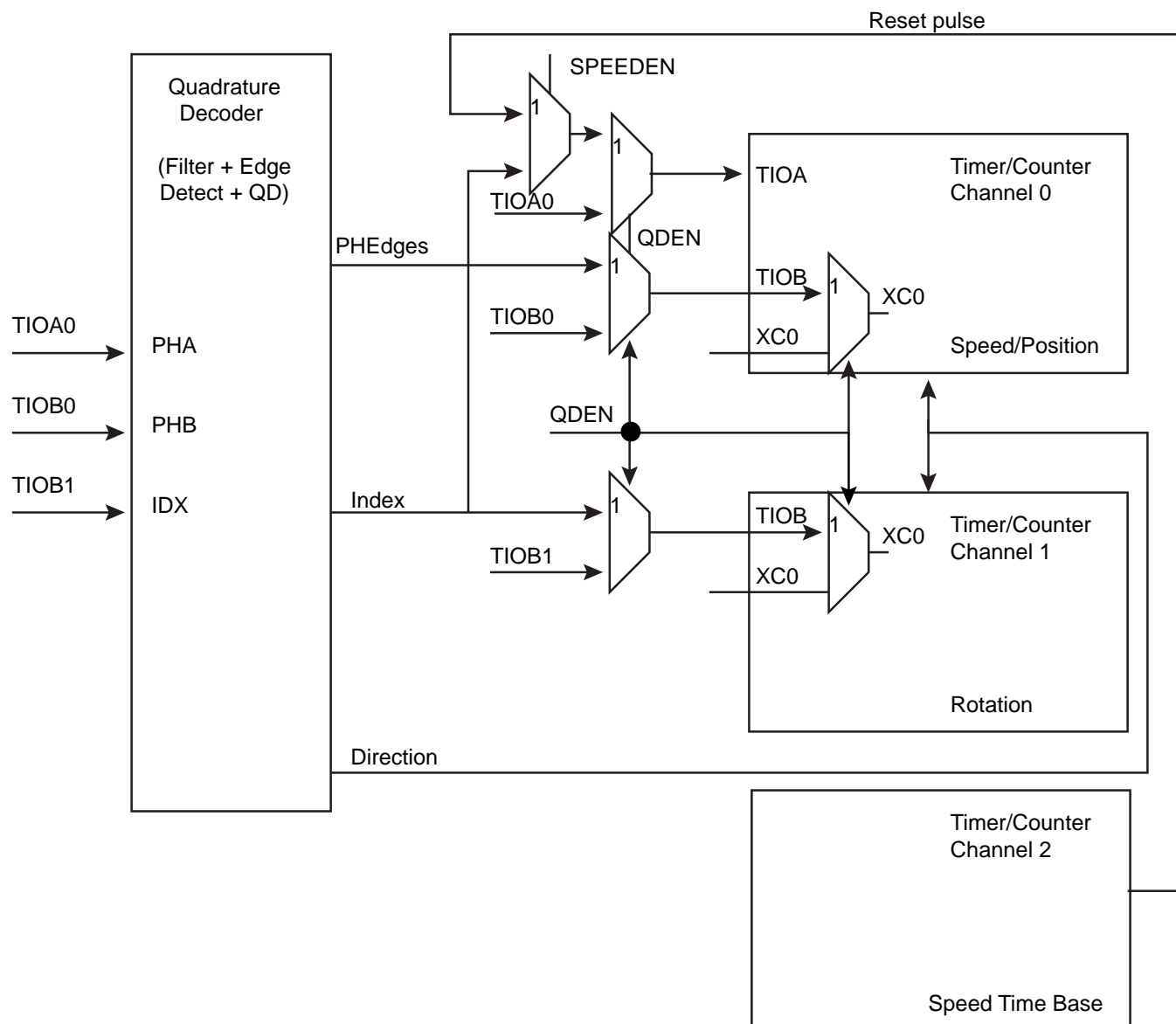
In speed mode, position cannot be measured but revolution can be measured.

Inputs from the rotary sensor can be filtered prior to down-stream processing. Accommodation of input polarity, phase definition and other factors are configurable.

Interruptions can be generated on different events.

A compare function (using TC\_RC register) is available on channel 0 (speed/position) or channel 1 (rotation) and can generate an interrupt by means of the CPCS flag in the TC\_SRx.

**Figure 37-15. Predefined Connection of the Quadrature Decoder with Timer Counters**



#### 37.6.14.2 Input Pre-processing

Input pre-processing consists of capabilities to take into account rotary sensor factors such as polarities and phase definition followed by configurable digital filtering.

Each input can be negated and swapping PHA, PHB is also configurable.

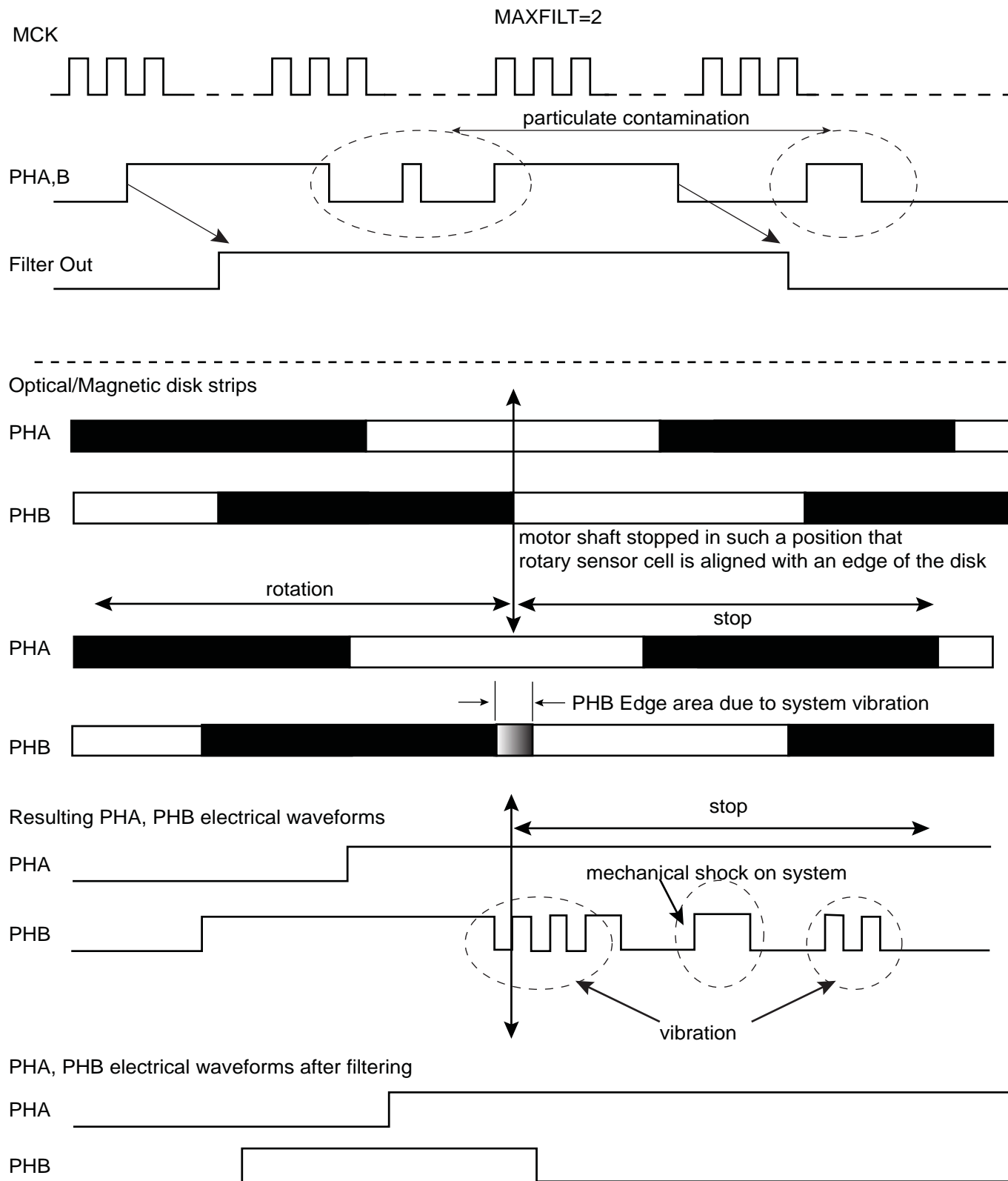
The MAXFILT field in the TC\_BMR is used to configure a minimum duration for which the pulse is stated as valid. When the filter is active, pulses with a duration lower than  $\text{MAXFILT} + 1 \times \text{tMCK ns}$  are not passed to down-stream logic.

Filters can be disabled using the FILTER bit in the TC\_BMR.

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Figure 37-17. Filtering Examples



### 37.6.14.3 Direction Status and Change Detection

After filtering, the quadrature signals are analyzed to extract the rotation direction and edges of the two quadrature signals detected in order to be counted by timer/counter logic downstream.

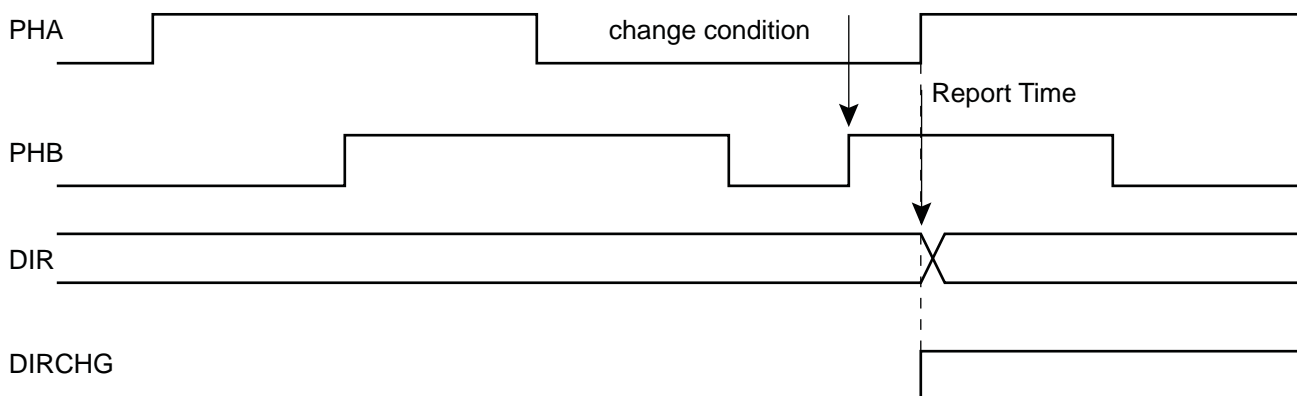
The direction status can be directly read at anytime in the TC\_QISR. The polarity of the direction flag status depends on the configuration written in TC\_BMR. INVA, INVB, INVIDX, SWAP modify the polarity of DIR flag.

Any change in rotation direction is reported in the TC\_QISR and can generate an interrupt.

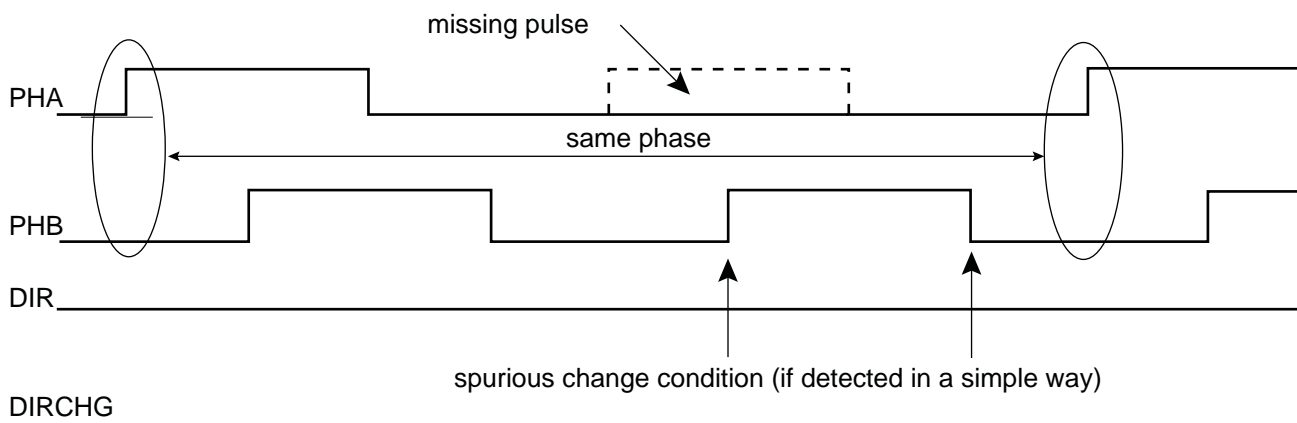
The direction change condition is reported as soon as two consecutive edges on a phase signal have sampled the same value on the other phase signal and there is an edge on the other signal. The two consecutive edges of one phase signal sampling the same value on other phase signal is not sufficient to declare a direction change, for the reason that particulate contamination may mask one or more reflective bars on the optical or magnetic disk of the sensor. (Refer to [Figure 37-18 "Rotation Change Detection"](#) for waveforms.)

**Figure 37-18. Rotation Change Detection**

Direction Change under normal conditions



No direction change due to particulate contamination masking a reflective bar

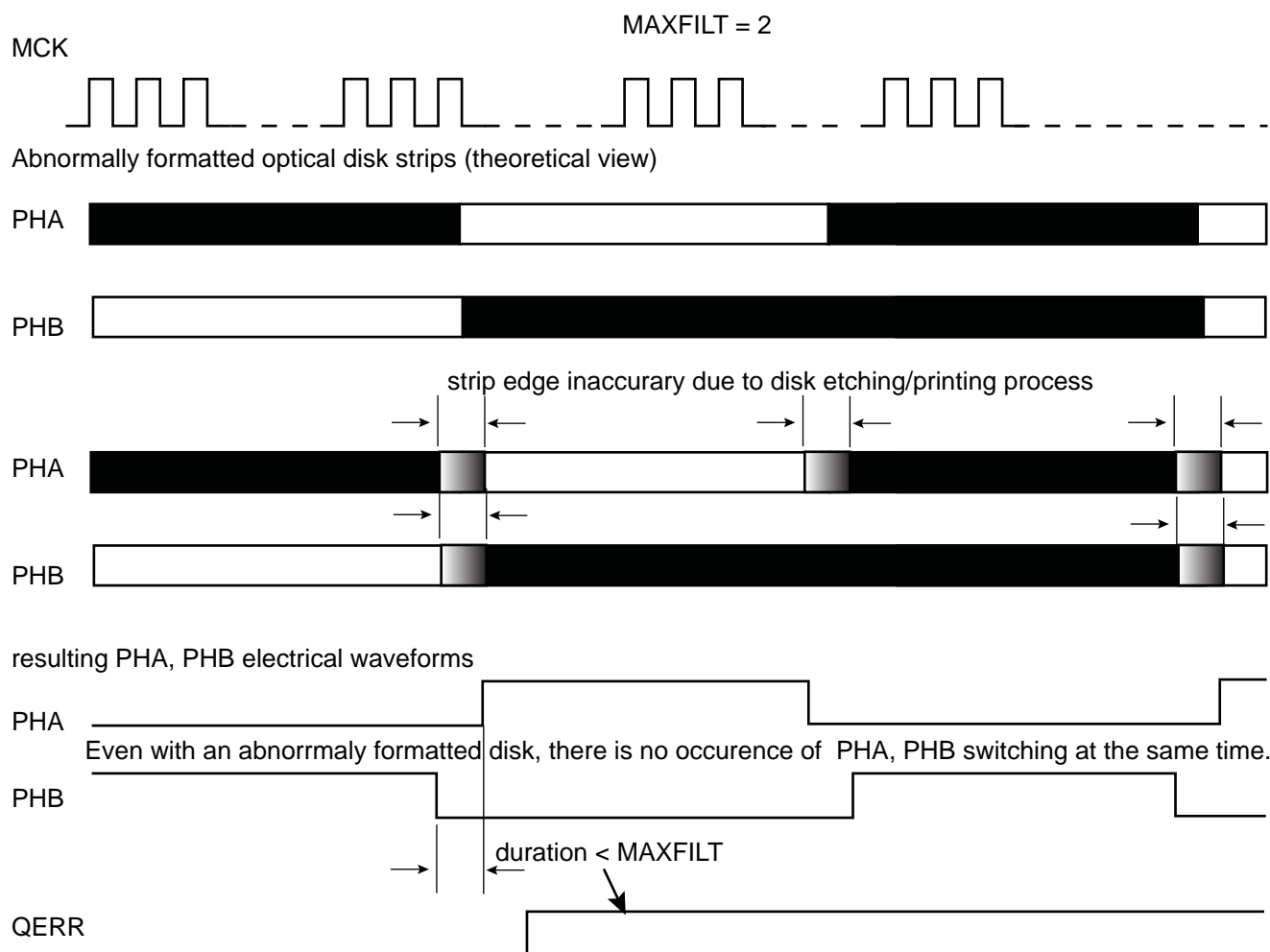


The direction change detection is disabled when QDTRANS is set in the TC\_BMR. In this case the DIR flag report must not be used.

A quadrature error is also reported by the quadrature decoder logic via the QERR flag in the TC\_QISR. This error is reported if the time difference between two edges on PHA, PHB is lower than a predefined value. This predefined value

is configurable and corresponds to  $(MAXFILT + 1) * tMCK$  ns. After being filtered there is no reason to have two edges closer than  $(MAXFILT + 1) * tMCK$  ns under normal mode of operation.

**Figure 37-19. Quadrature Error Detection**



MAXFILT must be tuned according to several factors such as the system clock frequency (MCK), type of rotary sensor and rotation speed to be achieved.

#### 37.6.14.4 Position and Rotation Measurement

When the POSEN bit is set in the TC\_BMR, the motor axis position is processed on channel 0 (by means of the PHA, PHB edge detections) and the number of motor revolutions are recorded on channel 1 if the IDX signal is provided on the TIOB1 input. The position measurement can be read in the TC\_CV0 register and the rotation measurement can be read in the TC\_CV1 register.

Channel 0 and 1 must be configured in capture mode (WAVE = 0 in TC\_CMR0).

In parallel, the number of edges are accumulated on timer/counter channel 0 and can be read on the TC\_CV0 register.

Therefore, the accurate position can be read on both TC\_CV registers and concatenated to form a 32-bit word.

The timer/counter channel 0 is cleared for each increment of IDX count value.

Depending on the quadrature signals, the direction is decoded and allows to count up or down in timer/counter channels 0 and 1. The direction status is reported on TC\_QISR.

#### 37.6.14.5 Speed Measurement

When SPEEDEN is set in the TC\_BMR, the speed measure is enabled on channel 0.

A time base must be defined on channel 2 by writing the TC\_RC2 period register. Channel 2 must be configured in waveform mode (WAVE bit set) in TC\_CMR2 register. The WAVSEL field must be defined with 0x10 to clear the counter by comparison and matching with TC\_RC value. Field ACPC must be defined at 0x11 to toggle TIOA output.

This time base is automatically fed back to TIOA of channel 0 when QDEN and SPEEDEN are set.

Channel 0 must be configured in capture mode (WAVE = 0 in TC\_CMR0). The ABETRGR bit of TC\_CMR0 must be configured at 1 to select TIOA as a trigger for this channel.

EDGTRG can be set to 0x01, to clear the counter on a rising edge of the TIOA signal and field LDRA must be set accordingly to 0x01, to load TC\_RA0 at the same time as the counter is cleared (LDRB must be set to 0x01). As a consequence, at the end of each time base period the differentiation required for the speed calculation is performed.

The process must be started by configuring bits CLKEN and SWTRG in the TC\_CCR.

The speed can be read on field RA in register TC\_RA0.

Channel 1 can still be used to count the number of revolutions of the motor.

### 37.6.15 2-bit Gray Up/Down Counter for Stepper Motor

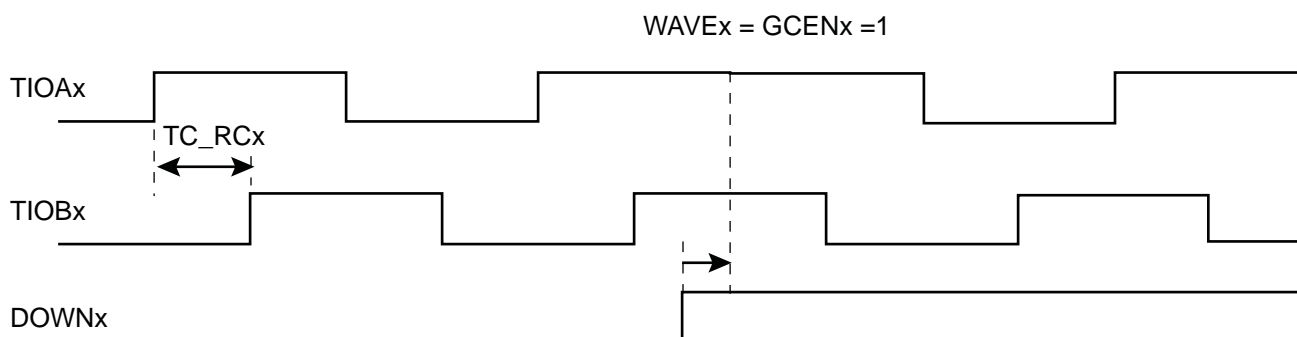
Each channel can be independently configured to generate a 2-bit gray count waveform on corresponding TIOA, TIOB outputs by means of the GCEN bit in TC\_SMMRx.

Up or Down count can be defined by writing bit DOWN in TC\_SMMRx.

It is mandatory to configure the channel in WAVE mode in the TC\_CMR.

The period of the counters can be programmed in TC\_RCx.

Figure 37-20. 2-bit Gray Up/Down Counter



## 37.7 Register Write Protection

To prevent any single software error from corrupting TC behavior, certain registers in the address space can be write-protected by setting the WPEN bit in the [TC Write Protection Mode Register](#) (TC\_WPMR).

The following registers can be protected:

- [TC Block Mode Register](#)
- [TC Channel Mode Register: Capture Mode](#)
- [TC Channel Mode Register: Waveform Mode](#)
- [TC Stepper Motor Mode Register](#)
- [TC Register A](#)
- [TC Register B](#)
- [TC Register C](#)



## 37.8 Timer Counter (TC) User Interface

**Table 37-5. Register Mapping**

Offset <sup>(1)</sup>	Register	Name	Access	Reset
0x00 + channel * 0x40 + 0x00	Channel Control Register	TC_CCR	Write-only	–
0x00 + channel * 0x40 + 0x04	Channel Mode Register	TC_CMR	Read/Write	0
0x00 + channel * 0x40 + 0x08	Stepper Motor Mode Register	TC_SMMR	Read/Write	0
0x00 + channel * 0x40 + 0x0C	Reserved			
0x00 + channel * 0x40 + 0x10	Counter Value	TC_CV	Read-only	0
0x00 + channel * 0x40 + 0x14	Register A	TC_RA	Read/Write <sup>(2)</sup>	0
0x00 + channel * 0x40 + 0x18	Register B	TC_RB	Read/Write <sup>(2)</sup>	0
0x00 + channel * 0x40 + 0x1C	Register C	TC_RC	Read/Write	0
0x00 + channel * 0x40 + 0x20	Status Register	TC_SR	Read-only	0
0x00 + channel * 0x40 + 0x24	Interrupt Enable Register	TC_IER	Write-only	–
0x00 + channel * 0x40 + 0x28	Interrupt Disable Register	TC_IDR	Write-only	–
0x00 + channel * 0x40 + 0x2C	Interrupt Mask Register	TC_IMR	Read-only	0
0xC0	Block Control Register	TC_BCR	Write-only	–
0xC4	Block Mode Register	TC_BMR	Read/Write	0
0xC8	QDEC Interrupt Enable Register	TC_QIER	Write-only	–
0xCC	QDEC Interrupt Disable Register	TC_QIDR	Write-only	–
0xD0	QDEC Interrupt Mask Register	TC_QIMR	Read-only	0
0xD4	QDEC Interrupt Status Register	TC_QISR	Read-only	0
0xD8	Reserved	–	–	–
0xE4	Write Protection Mode Register	TC_WPMR	Read/Write	0
0xE8–0xFC	Reserved	–	–	–

Notes: 1. Channel index ranges from 0 to 2.  
2. Read-only if WAVE = 0

### 37.8.1 TC Channel Control Register

**Name:** TC\_CCRx [x=0..2]

**Address:** 0x40010000 (0)[0], 0x40010040 (0)[1], 0x40010080 (0)[2], 0x40014000 (1)[0], 0x40014040 (1)[1], 0x40014080 (1)[2]

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	SWTRG	CLKDIS	CLKEN

- **CLKEN: Counter Clock Enable Command**

0: No effect.

1: Enables the clock if CLKDIS is not 1.

- **CLKDIS: Counter Clock Disable Command**

0: No effect.

1: Disables the clock.

- **SWTRG: Software Trigger Command**

0: No effect.

1: A software trigger is performed: the counter is reset and the clock is started.

### 37.8.2 TC Channel Mode Register: Capture Mode

**Name:** TC\_CMRx [x=0..2] (WAVE = 0)

**Address:** 0x40010004 (0)[0], 0x40010044 (0)[1], 0x40010084 (0)[2], 0x40014004 (1)[0], 0x40014044 (1)[1], 0x40014084 (1)[2]

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	LDRB		LDRA	
15	14	13	12	11	10	9	8
WAVE	CPCTRG	–	–	–	ABETRG	ETRGEDG	
7	6	5	4	3	2	1	0
LDBDIS	LDBSTOP	BURST		CLKI	TCCLKS		

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- TCCLKS: Clock Selection**

Value	Name	Description
0	TIMER_CLOCK1	Clock selected: internal TIMER_CLOCK1 clock signal (from PMC)
1	TIMER_CLOCK2	Clock selected: internal TIMER_CLOCK2 clock signal (from PMC)
2	TIMER_CLOCK3	Clock selected: internal TIMER_CLOCK3 clock signal (from PMC)
3	TIMER_CLOCK4	Clock selected: internal TIMER_CLOCK4 clock signal (from PMC)
4	TIMER_CLOCK5	Clock selected: internal TIMER_CLOCK5 clock signal (from PMC)
5	XC0	Clock selected: XC0
6	XC1	Clock selected: XC1
7	XC2	Clock selected: XC2

- CLKI: Clock Invert**

0: Counter is incremented on rising edge of the clock.

1: Counter is incremented on falling edge of the clock.

- BURST: Burst Signal Selection**

Value	Name	Description
0	NONE	The clock is not gated by an external signal.
1	XC0	XC0 is ANDed with the selected clock.
2	XC1	XC1 is ANDed with the selected clock.
3	XC2	XC2 is ANDed with the selected clock.

- LDBSTOP: Counter Clock Stopped with RB Loading**

0: Counter clock is not stopped when RB loading occurs.

1: Counter clock is stopped when RB loading occurs.

- **LDBDIS: Counter Clock Disable with RB Loading**

0: Counter clock is not disabled when RB loading occurs.

1: Counter clock is disabled when RB loading occurs.

- **ETRGEDG: External Trigger Edge Selection**

Value	Name	Description
0	NONE	The clock is not gated by an external signal.
1	RISING	Rising edge
2	FALLING	Falling edge
3	EDGE	Each edge

- **ABETRG: TIOA or TIOB External Trigger Selection**

0: TIOB is used as an external trigger.

1: TIOA is used as an external trigger.

- **CPCTRG: RC Compare Trigger Enable**

0: RC Compare has no effect on the counter and its clock.

1: RC Compare resets the counter and starts the counter clock.

- **WAVE: Waveform Mode**

0: Capture Mode is enabled.

1: Capture Mode is disabled (Waveform Mode is enabled).

- **LDRA: RA Loading Edge Selection**

Value	Name	Description
0	NONE	None
1	RISING	Rising edge of TIOA
2	FALLING	Falling edge of TIOA
3	EDGE	Each edge of TIOA

- **LDRB: RB Loading Edge Selection**

Value	Name	Description
0	NONE	None
1	RISING	Rising edge of TIOA
2	FALLING	Falling edge of TIOA
3	EDGE	Each edge of TIOA

### 37.8.3 TC Channel Mode Register: Waveform Mode

**Name:** TC\_CMRx [x=0..2] (WAVE = 1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
BSWTRG		BEEVT		BCPC		BCPB	
23	22	21	20	19	18	17	16
ASWTRG		AEEVT		ACPC		ACPA	
15	14	13	12	11	10	9	8
WAVE	WAVSEL		ENETRГ	EEVT		EEVTEDG	
7	6	5	4	3	2	1	0
CPCDIS	CPCSTOP	BURST		CLKI	TCCLKS		

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

#### • TCCLKS: Clock Selection

Value	Name	Description
0	TIMER_CLOCK1	Clock selected: internal TIMER_CLOCK1 clock signal (from PMC)
1	TIMER_CLOCK2	Clock selected: internal TIMER_CLOCK2 clock signal (from PMC)
2	TIMER_CLOCK3	Clock selected: internal TIMER_CLOCK3 clock signal (from PMC)
3	TIMER_CLOCK4	Clock selected: internal TIMER_CLOCK4 clock signal (from PMC)
4	TIMER_CLOCK5	Clock selected: internal TIMER_CLOCK5 clock signal (from PMC)
5	XC0	Clock selected: XC0
6	XC1	Clock selected: XC1
7	XC2	Clock selected: XC2

#### • CLKI: Clock Invert

0: Counter is incremented on rising edge of the clock.

1: Counter is incremented on falling edge of the clock.

#### • BURST: Burst Signal Selection

Value	Name	Description
0	NONE	The clock is not gated by an external signal.
1	XC0	XC0 is ANDed with the selected clock.
2	XC1	XC1 is ANDed with the selected clock.
3	XC2	XC2 is ANDed with the selected clock.

#### • CPCSTOP: Counter Clock Stopped with RC Compare

0: Counter clock is not stopped when counter reaches RC.

1: Counter clock is stopped when counter reaches RC.

#### • CPCDIS: Counter Clock Disable with RC Compare

0: Counter clock is not disabled when counter reaches RC.

1: Counter clock is disabled when counter reaches RC.

- **EEVTEDG: External Event Edge Selection**

Value	Name	Description
0	NONE	None
1	RISING	Rising edge
2	FALLING	Falling edge
3	EDGE	Each edge

- **EEVT: External Event Selection**

Signal selected as external event.

Value	Name	Description	TIOB Direction
0	TIOB	TIOB <sup>(1)</sup>	Input
1	XC0	XC0	Output
2	XC1	XC1	Output
3	XC2	XC2	Output

Note: 1. If TIOB is chosen as the external event signal, it is configured as an input and no longer generates waveforms and subsequently no IRQs.

- **ENETRГ: External Event Trigger Enable**

0: The external event has no effect on the counter and its clock.

1: The external event resets the counter and starts the counter clock.

Note: Whatever the value programmed in ENETRГ, the selected external event only controls the TIOA output and TIOB if not used as input (trigger event input or other input used).

- **WAVSEL: Waveform Selection**

Value	Name	Description
0	UP	UP mode without automatic trigger on RC Compare
1	UPDOWN	UPDOWN mode without automatic trigger on RC Compare
2	UP_RC	UP mode with automatic trigger on RC Compare
3	UPDOWN_RC	UPDOWN mode with automatic trigger on RC Compare

- **WAVE: Waveform Mode**

0: Waveform Mode is disabled (Capture Mode is enabled).

1: Waveform Mode is enabled.

- **ACPA: RA Compare Effect on TIOA**

Value	Name	Description
0	NONE	None

1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **ACPC: RC Compare Effect on TIOA**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **AEEVT: External Event Effect on TIOA**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **ASWTRG: Software Trigger Effect on TIOA**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **BCPB: RB Compare Effect on TIOB**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **BCPC: RC Compare Effect on TIOB**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **BEEVT: External Event Effect on TIOB**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle

- **BSWTRG: Software Trigger Effect on TIOB**

Value	Name	Description
0	NONE	None
1	SET	Set
2	CLEAR	Clear
3	TOGGLE	Toggle



### 37.8.4 TC Stepper Motor Mode Register

**Name:** TC\_SMMRx [x=0..2]

**Address:** 0x40010008 (0)[0], 0x40010048 (0)[1], 0x40010088 (0)[2], 0x40014008 (1)[0], 0x40014048 (1)[1], 0x40014088 (1)[2]

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	DOWN	GCEN

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- **GCEN: Gray Count Enable**

0: TIOAx [x=0..2] and TIOBx [x=0..2] are driven by internal counter of channel x.

1: TIOAx [x=0..2] and TIOBx [x=0..2] are driven by a 2-bit gray counter.

- **DOWN: DOWN Count**

0: Up counter.

1: Down counter.

37.8.5 TC Counter Value Register

**Name:** TC\_CVx [x=0..2]  
**Address:** 0x40010010 (0)[0], 0x40010050 (0)[1], 0x40010090 (0)[2], 0x40014010 (1)[0], 0x40014050 (1)[1], 0x40014090 (1)[2]  
**Access:** Read-only

31	30	29	28	27	26	25	24
CV							
23	22	21	20	19	18	17	16
CV							
15	14	13	12	11	10	9	8
CV							
7	6	5	4	3	2	1	0
CV							

- **CV: Counter Value**  
CV contains the counter value in real time.

### 37.8.6 TC Register A

**Name:** TC\_RAx [x=0..2]

**Address:** 0x40010014 (0)[0], 0x40010054 (0)[1], 0x40010094 (0)[2], 0x40014014 (1)[0], 0x40014054 (1)[1], 0x40014094 (1)[2]

**Access:** Read-only if WAVE = 0, Read/Write if WAVE = 1

31	30	29	28	27	26	25	24
RA							
23	22	21	20	19	18	17	16
RA							
15	14	13	12	11	10	9	8
RA							
7	6	5	4	3	2	1	0
RA							

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- **RA: Register A**

RA contains the Register A value in real time.

### 37.8.7 TC Register B

**Name:** TC\_RBx [x=0..2]

**Address:** 0x40010018 (0)[0], 0x40010058 (0)[1], 0x40010098 (0)[2], 0x40014018 (1)[0], 0x40014058 (1)[1], 0x40014098 (1)[2]

**Access:** Read-only if WAVE = 0, Read/Write if WAVE = 1

31	30	29	28	27	26	25	24
RB							
23	22	21	20	19	18	17	16
RB							
15	14	13	12	11	10	9	8
RB							
7	6	5	4	3	2	1	0
RB							

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- **RB: Register B**

RB contains the Register B value in real time.

### 37.8.8 TC Register C

**Name:** TC\_RCx [x=0..2]

**Address:** 0x4001001C (0)[0], 0x4001005C (0)[1], 0x4001009C (0)[2], 0x4001401C (1)[0], 0x4001405C (1)[1], 0x4001409C (1)[2]

**Access:** Read/Write

31	30	29	28	27	26	25	24
RC							
23	22	21	20	19	18	17	16
RC							
15	14	13	12	11	10	9	8
RC							
7	6	5	4	3	2	1	0
RC							

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- **RC: Register C**

RC contains the Register C value in real time.

### 37.8.9 TC Status Register

**Name:** TC\_SRx [x=0..2]

**Address:** 0x40010020 (0)[0], 0x40010060 (0)[1], 0x400100A0 (0)[2], 0x40014020 (1)[0], 0x40014060 (1)[1], 0x400140A0 (1)[2]

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	MTIOB	MTIOA	CLKSTA
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	CPBS	CPAS	LOVRS	COVFS

- **COVFS: Counter Overflow Status**

0: No counter overflow has occurred since the last read of the Status Register.

1: A counter overflow has occurred since the last read of the Status Register.

- **LOVRS: Load Overrun Status**

0: Load overrun has not occurred since the last read of the Status Register or WAVE = 1.

1: RA or RB have been loaded at least twice without any read of the corresponding register since the last read of the Status Register, if WAVE = 0.

- **CPAS: RA Compare Status**

0: RA Compare has not occurred since the last read of the Status Register or WAVE = 0.

1: RA Compare has occurred since the last read of the Status Register, if WAVE = 1.

- **CPBS: RB Compare Status**

0: RB Compare has not occurred since the last read of the Status Register or WAVE = 0.

1: RB Compare has occurred since the last read of the Status Register, if WAVE = 1.

- **CPCS: RC Compare Status**

0: RC Compare has not occurred since the last read of the Status Register.

1: RC Compare has occurred since the last read of the Status Register.

- **LDRAS: RA Loading Status**

0: RA Load has not occurred since the last read of the Status Register or WAVE = 1.

1: RA Load has occurred since the last read of the Status Register, if WAVE = 0.

- **LDRBS: RB Loading Status**

0: RB Load has not occurred since the last read of the Status Register or WAVE = 1.

1: RB Load has occurred since the last read of the Status Register, if WAVE = 0.

- **ETRGS: External Trigger Status**

0: External trigger has not occurred since the last read of the Status Register.

1: External trigger has occurred since the last read of the Status Register.

- **CLKSTA: Clock Enabling Status**

0: Clock is disabled.

1: Clock is enabled.

- **MTIOA: TIOA Mirror**

0: TIOA is low. If WAVE = 0, this means that TIOA pin is low. If WAVE = 1, this means that TIOA is driven low.

1: TIOA is high. If WAVE = 0, this means that TIOA pin is high. If WAVE = 1, this means that TIOA is driven high.

- **MTIOB: TIOB Mirror**

0: TIOB is low. If WAVE = 0, this means that TIOB pin is low. If WAVE = 1, this means that TIOB is driven low.

1: TIOB is high. If WAVE = 0, this means that TIOB pin is high. If WAVE = 1, this means that TIOB is driven high.

### 37.8.10 TC Interrupt Enable Register

**Name:** TC\_IERx [x=0..2]

**Address:** 0x40010024 (0)[0], 0x40010064 (0)[1], 0x400100A4 (0)[2], 0x40014024 (1)[0], 0x40014064 (1)[1], 0x400140A4 (1)[2]

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	CPBS	CPAS	LOVRS	COVFS

- **COVFS: Counter Overflow**

0: No effect.

1: Enables the Counter Overflow Interrupt.

- **LOVRS: Load Overrun**

0: No effect.

1: Enables the Load Overrun Interrupt.

- **CPAS: RA Compare**

0: No effect.

1: Enables the RA Compare Interrupt.

- **CPBS: RB Compare**

0: No effect.

1: Enables the RB Compare Interrupt.

- **CPCS: RC Compare**

0: No effect.

1: Enables the RC Compare Interrupt.

- **LDRAS: RA Loading**

0: No effect.

1: Enables the RA Load Interrupt.

- **LDRBS: RB Loading**

0: No effect.

1: Enables the RB Load Interrupt.

- **ETRGS: External Trigger**

0: No effect.

1: Enables the External Trigger Interrupt.



### 37.8.11 TC Interrupt Disable Register

**Name:** TC\_IDRx [x=0..2]

**Address:** 0x40010028 (0)[0], 0x40010068 (0)[1], 0x400100A8 (0)[2], 0x40014028 (1)[0], 0x40014068 (1)[1], 0x400140A8 (1)[2]

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	CPBS	CPAS	LOVRS	COVFS

- **COVFS: Counter Overflow**

0: No effect.

1: Disables the Counter Overflow Interrupt.

- **LOVRS: Load Overrun**

0: No effect.

1: Disables the Load Overrun Interrupt (if WAVE = 0).

- **CPAS: RA Compare**

0: No effect.

1: Disables the RA Compare Interrupt (if WAVE = 1).

- **CPBS: RB Compare**

0: No effect.

1: Disables the RB Compare Interrupt (if WAVE = 1).

- **CPCS: RC Compare**

0: No effect.

1: Disables the RC Compare Interrupt.

- **LDRAS: RA Loading**

0: No effect.

1: Disables the RA Load Interrupt (if WAVE = 0).

- **LDRBS: RB Loading**

0: No effect.

1: Disables the RB Load Interrupt (if WAVE = 0).

- **ETRGS: External Trigger**

0: No effect.

1: Disables the External Trigger Interrupt.

### 37.8.12 TC Interrupt Mask Register

**Name:** TC\_IMRx [x=0..2]

**Address:** 0x4001002C (0)[0], 0x4001006C (0)[1], 0x400100AC (0)[2], 0x4001402C (1)[0], 0x4001406C (1)[1], 0x400140AC (1)[2]

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
ETRGS	LDRBS	LDRAS	CPCS	CPBS	CPAS	LOVRS	COVFS

- **COVFS: Counter Overflow**

0: The Counter Overflow Interrupt is disabled.

1: The Counter Overflow Interrupt is enabled.

- **LOVRS: Load Overrun**

0: The Load Overrun Interrupt is disabled.

1: The Load Overrun Interrupt is enabled.

- **CPAS: RA Compare**

0: The RA Compare Interrupt is disabled.

1: The RA Compare Interrupt is enabled.

- **CPBS: RB Compare**

0: The RB Compare Interrupt is disabled.

1: The RB Compare Interrupt is enabled.

- **CPCS: RC Compare**

0: The RC Compare Interrupt is disabled.

1: The RC Compare Interrupt is enabled.

- **LDRAS: RA Loading**

0: The Load RA Interrupt is disabled.

1: The Load RA Interrupt is enabled.

- **LDRBS: RB Loading**

0: The Load RB Interrupt is disabled.

1: The Load RB Interrupt is enabled.

- **ETRGS: External Trigger**

0: The External Trigger Interrupt is disabled.

1: The External Trigger Interrupt is enabled.

### 37.8.13 TC Block Control Register

**Name:** TC\_BCR

**Address:** 0x400100C0 (0), 0x400140C0 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	SYNC

- **SYNC: Synchro Command**

0: No effect.

1: Asserts the SYNC signal which generates a software trigger simultaneously for each of the channels.

### 37.8.14 TC Block Mode Register

**Name:** TC\_BMR

**Address:** 0x400100C4 (0), 0x400140C4 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	MAXFILT	
23	22	21	20	19	18	17	16
MAXFILT				FILTER	–	IDXPHB	SWAP
15	14	13	12	11	10	9	8
INVIDX	INVB	INVA	EDGPHA	QDTRANS	SPEEDEN	POSEN	QDEN
7	6	5	4	3	2	1	0
–	–	TC2XC2S		TC1XC1S		TC0XC0S	

This register can only be written if the WPEN bit is cleared in the [TC Write Protection Mode Register](#).

- TC0XC0S: External Clock Signal 0 Selection**

Value	Name	Description
0	TCLK0	Signal connected to XC0: TCLK0
1	–	Reserved
2	TIOA1	Signal connected to XC0: TIOA1
3	TIOA2	Signal connected to XC0: TIOA2

- TC1XC1S: External Clock Signal 1 Selection**

Value	Name	Description
0	TCLK1	Signal connected to XC1: TCLK1
1	–	Reserved
2	TIOA0	Signal connected to XC1: TIOA0
3	TIOA2	Signal connected to XC1: TIOA2

- TC2XC2S: External Clock Signal 2 Selection**

Value	Name	Description
0	TCLK2	Signal connected to XC2: TCLK2
1	–	Reserved
2	TIOA0	Signal connected to XC2: TIOA0
3	TIOA1	Signal connected to XC2: TIOA1

- QDEN: Quadrature Decoder ENabled**

0: Disabled.

1: Enables the quadrature decoder logic (filter, edge detection and quadrature decoding).

Quadrature decoding (direction change) can be disabled using QDTRANS bit.

One of the POSEN or SPEEDEN bits must be also enabled.

- **POSEN: POSition ENabled**

0: Disable position.

1: Enables the position measure on channel 0 and 1.

- **SPEEDEN: SPEED ENabled**

0: Disabled.

1: Enables the speed measure on channel 0, the time base being provided by channel 2.

- **QDTRANS: Quadrature Decoding TRANSPARENT**

0: Full quadrature decoding logic is active (direction change detected).

1: Quadrature decoding logic is inactive (direction change inactive) but input filtering and edge detection are performed.

- **EDGPHA: EDGe on PHA count mode**

0: Edges are detected on PHA only.

1: Edges are detected on both PHA and PHB.

- **INVA: INVerted phA**

0: PHA (TIOA0) is directly driving quadrature decoder logic.

1: PHA is inverted before driving quadrature decoder logic.

- **INVB: INVerted phB**

0: PHB (TIOB0) is directly driving quadrature decoder logic.

1: PHB is inverted before driving quadrature decoder logic.

- **SWAP: SWAP PHA and PHB**

0: No swap between PHA and PHB.

1: Swap PHA and PHB internally, prior to driving quadrature decoder logic.

- **INVIDX: INVerted InDeX**

0: IDX (TIOA1) is directly driving quadrature logic.

1: IDX is inverted before driving quadrature logic.

- **IDXPBH: InDeX pin is PHB pin**

0: IDX pin of the rotary sensor must drive TIOA1.

1: IDX pin of the rotary sensor must drive TIOB0.

- **FILTER: Glitch Filter**

0: IDX,PHA, PHB input pins are not filtered.

1: IDX,PHA, PHB input pins are filtered using MAXFILT value.

- **MAXFILT: MAXimum FILTer**

1.. 63: Defines the filtering capabilities.

Pulses with a period shorter than MAXFILT+1 MCK clock cycles are discarded.

### 37.8.15 TC QDEC Interrupt Enable Register

**Name:** TC\_QIER

**Address:** 0x400100C8 (0), 0x400140C8 (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	QERR	DIRCHG	IDX

- **IDX: InDeX**

0: No effect.

1: Enables the interrupt when a rising edge occurs on IDX input.

- **DIRCHG: DIRection CHanGe**

0: No effect.

1: Enables the interrupt when a change on rotation direction is detected.

- **QERR: Quadrature ERROr**

0: No effect.

1: Enables the interrupt when a quadrature error occurs on PHA,PHB.

### 37.8.16 TC QDEC Interrupt Disable Register

**Name:** TC\_QIDR

**Address:** 0x400100CC (0), 0x400140CC (1)

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	QERR	DIRCHG	IDX

- **IDX: InDeX**

0: No effect.

1: Disables the interrupt when a rising edge occurs on IDX input.

- **DIRCHG: DIRection CHanGe**

0: No effect.

1: Disables the interrupt when a change on rotation direction is detected.

- **QERR: Quadrature ERROr**

0: No effect.

1: Disables the interrupt when a quadrature error occurs on PHA, PHB.

### 37.8.17 TC QDEC Interrupt Mask Register

**Name:** TC\_QIMR

**Address:** 0x400100D0 (0), 0x400140D0 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	QERR	DIRCHG	IDX

- **IDX: InDeX**

0: The interrupt on IDX input is disabled.

1: The interrupt on IDX input is enabled.

- **DIRCHG: DIRection CHAnGe**

0: The interrupt on rotation direction change is disabled.

1: The interrupt on rotation direction change is enabled.

- **QERR: Quadrature ERROr**

0: The interrupt on quadrature error is disabled.

1: The interrupt on quadrature error is enabled.



### 37.8.18 TC QDEC Interrupt Status Register

**Name:** TC\_QISR

**Address:** 0x400100D4 (0), 0x400140D4 (1)

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	DIR
7	6	5	4	3	2	1	0
–	–	–	–	–	QERR	DIRCHG	IDX

- **IDX: InDeX**

0: No Index input change since the last read of TC\_QISR.

1: The IDX input has changed since the last read of TC\_QISR.

- **DIRCHG: DIRection CHanGe**

0: No change on rotation direction since the last read of TC\_QISR.

1: The rotation direction changed since the last read of TC\_QISR.

- **QERR: Quadrature ERROr**

0: No quadrature error since the last read of TC\_QISR.

1: A quadrature error occurred since the last read of TC\_QISR.

- **DIR: DIRection**

Returns an image of the actual rotation direction.

### 37.8.19 TC Write Protection Mode Register

**Name:** TC\_WPMR

**Address:** 0x400100E4 (0), 0x400140E4 (1)

**Access:** Read/Write

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

- **WPEN: Write Protect Enable**

0: Disables the Write Protect if WPKEY corresponds to 0x54494D (“TIM” in ASCII).

1: Enables the Write Protect if WPKEY corresponds to 0x54494D (“TIM” in ASCII).

See [Section 37.7 “Register Write Protection”](#) for list of write-protected registers.

- **WPKEY: Write Protect KEY**

Value	Name	Description
0x54494D	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

## 38. Pulse Width Modulation Controller (PWM)

### 38.1 Description

The PWM macrocell controls several channels independently. Each channel controls one square output waveform. Characteristics of the output waveform such as period, duty-cycle and polarity are configurable through the user interface. Each channel selects and uses one of the clocks provided by the clock generator. The clock generator provides several clocks resulting from the division of the PWM macrocell master clock.

All PWM macrocell accesses are made through APB mapped registers.

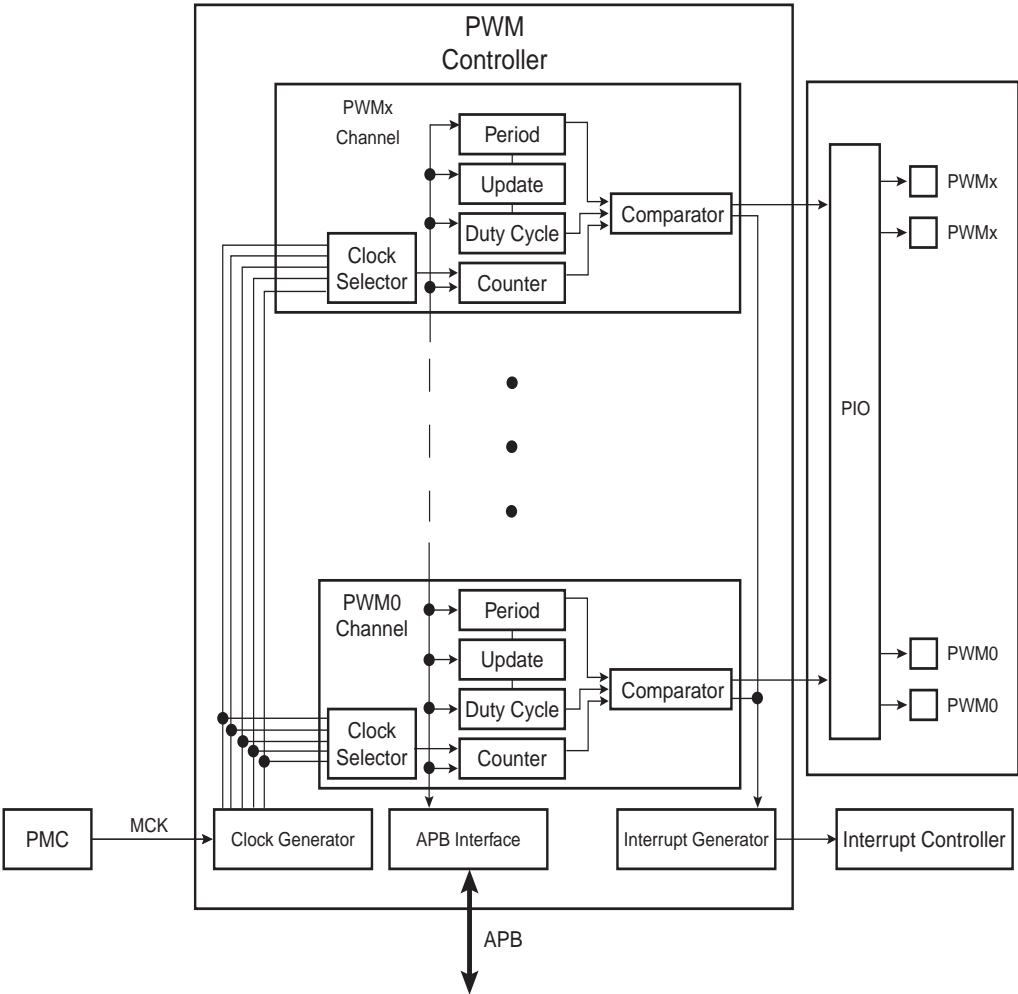
Channels can be synchronized, to generate non overlapped waveforms. All channels integrate a double buffering system in order to prevent an unexpected output waveform while modifying the period or the duty-cycle.

### 38.2 Embedded characteristics

- 4 Channels
- One 16-bit Counter Per Channel
- Common Clock Generator Providing Thirteen Different Clocks
  - A Modulo n Counter Providing Eleven Clocks
  - Two Independent Linear Dividers Working on Modulo n Counter Outputs
- Independent Channels
  - Independent Enable Disable Command for Each Channel
  - Independent Clock Selection for Each Channel
  - Independent Period and Duty Cycle for Each Channel
  - Double Buffering of Period or Duty Cycle for Each Channel
  - Programmable Selection of The Output Waveform Polarity for Each Channel
  - Programmable Center or Left Aligned Output Waveform for Each Channel Block Diagram

### 38.3 Block Diagram

Figure 38-1. Pulse Width Modulation Controller Block Diagram



### 38.4 I/O Lines Description

Each channel outputs one waveform on one external I/O line.

Table 38-1. I/O Line Description

Name	Description	Type
PWMx	PWM Waveform Output for channel x	Output

## 38.5 Product Dependencies

### 38.5.1 I/O Lines

The pins used for interfacing the PWM may be multiplexed with PIO lines. The programmer must first program the PIO controller to assign the desired PWM pins to their peripheral function. If I/O lines of the PWM are not used by the application, they can be used for other purposes by the PIO controller.

All of the PWM outputs may or may not be enabled. If an application requires only four channels, then only four PIO lines will be assigned to PWM outputs.

**Table 38-2. I/O Lines**

Instance	Signal	I/O Line	Peripheral
PWM	PWM0	PC0	B
PWM	PWM0	PC6	A
PWM	PWM1	PC1	B
PWM	PWM1	PC7	A
PWM	PWM2	PC2	B
PWM	PWM2	PC8	A
PWM	PWM3	PC3	B
PWM	PWM3	PC9	A

### 38.5.2 Power Management

The PWM is not continuously clocked. The programmer must first enable the PWM clock in the Power Management Controller (PMC) before using the PWM. However, if the application does not require PWM operations, the PWM clock can be stopped when not needed and be restarted later. In this case, the PWM will resume its operations where it left off.

All the PWM registers except PWM\_CDTY and PWM\_CPRD can be read without the PWM peripheral clock enabled. All the registers can be written without the peripheral clock enabled.

### 38.5.3 Interrupt Sources

The PWM interrupt line is connected on one of the internal sources of the Interrupt Controller. Using the PWM interrupt requires the Interrupt Controller to be programmed first. Note that it is not recommended to use the PWM interrupt line in edge sensitive mode.

**Table 38-3. Peripheral IDs**

Instance	ID
PWM	41

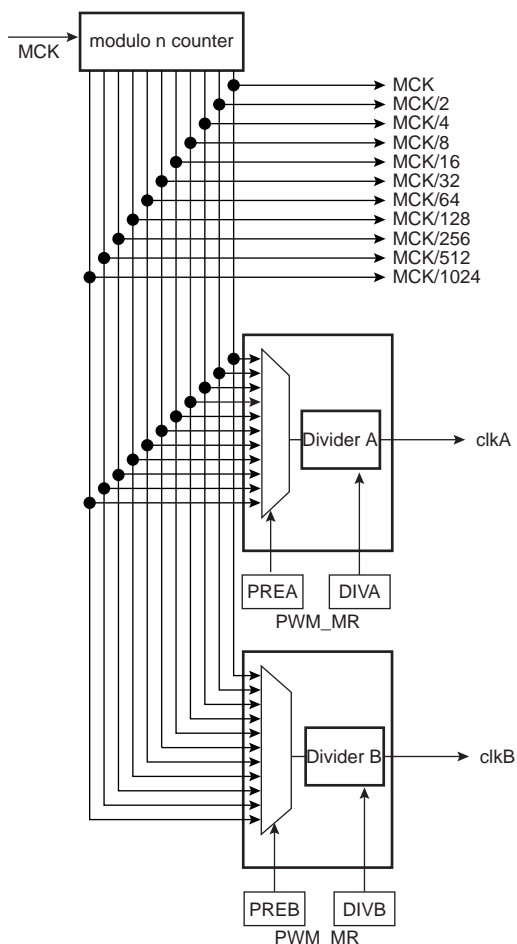
## 38.6 Functional Description

The PWM macrocell is primarily composed of a clock generator module and 4 channels.

- Clocked by the system clock, MCK, the clock generator module provides 13 clocks.
- Each channel can independently choose one of the clock generator outputs.
- Each channel generates an output waveform with attributes that can be defined independently for each channel through the user interface registers.

### 38.6.1 PWM Clock Generator

Figure 38-2. Functional View of the Clock Generator Block Diagram



**Caution:** Before using the PWM macrocell, the programmer must first enable the PWM clock in the Power Management Controller (PMC).

The PWM macrocell master clock, MCK, is divided in the clock generator module to provide different clocks available for all channels. Each channel can independently select one of the divided clocks.

The clock generator is divided in three blocks:

- A modulo n counter which provides 11 clocks:  $F_{MCK}$ ,  $F_{MCK}/2$ ,  $F_{MCK}/4$ ,  $F_{MCK}/8$ ,  $F_{MCK}/16$ ,  $F_{MCK}/32$ ,  $F_{MCK}/64$ ,  $F_{MCK}/128$ ,  $F_{MCK}/256$ ,  $F_{MCK}/512$ ,  $F_{MCK}/1024$
- Two linear dividers (1, 1/2, 1/3,... 1/255) that provide two separate clocks: clkA and clkB

Each linear divider can independently divide one of the clocks of the modulo n counter. The selection of the clock to be divided is made according to the PREA (PREB) field of the PWM Mode register (PWM\_MR). The resulting clock clkA (clkB) is the clock selected divided by DIVA (DIVB) field value in the PWM Mode register (PWM\_MR).

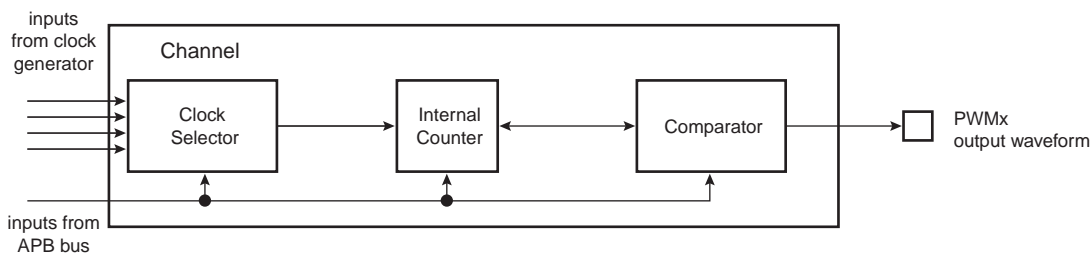
After a reset of the PWM controller, DIVA (DIVB) and PREA (PREB) in the PWM Mode register are set to 0. This implies that after reset clkA (clkB) are turned off.

At reset, all clocks provided by the modulo n counter are turned off except clock “clk”. This situation is also true when the PWM master clock is turned off through the Power Management Controller.

## 38.6.2 PWM Channel

### 38.6.2.1 Block Diagram

Figure 38-3. Functional View of the Channel Block Diagram



Each of the 4 channels is composed of three blocks:

- A clock selector which selects one of the clocks provided by the clock generator described in [Section 38.6.1 “PWM Clock Generator” on page 830](#).
- An internal counter clocked by the output of the clock selector. This internal counter is incremented or decremented according to the channel configuration and comparators events. The size of the internal counter is **16** bits.
- A comparator used to generate events according to the internal counter value. It also computes the PWMx output waveform according to the configuration.

### 38.6.2.2 Waveform Properties

The different properties of output waveforms are:

- the **internal clock selection**. The internal channel counter is clocked by one of the clocks provided by the clock generator described in the previous section. This channel parameter is defined in the CPRE field of the PWM\_CMRx register. This field is reset at 0.
- the **waveform period**. This channel parameter is defined in the CPRD field of the PWM\_CPRDx register.
  - If the waveform is left aligned, then the output waveform period depends on the counter source clock and can be calculated:

By using the Master Clock (MCK) divided by an X given prescaler value

(with X being 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, or 1024), the resulting period formula will be:

$$\frac{(X \times CPRD)}{MCK}$$

By using a Master Clock divided by one of both DIVA or DIVB divider, the formula becomes, respectively:

$$\frac{(X \times CPRD \times DIVA)}{MCK} \text{ or } \frac{(X \times CPRD \times DIVB)}{MCK}$$

If the waveform is center aligned then the output waveform period depends on the counter source clock and can be calculated:

By using the Master Clock (MCK) divided by an X given prescaler value

(with X being 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, or 1024). The resulting period formula will be:

$$\frac{(2 \times X \times CPRD)}{MCK}$$

By using a Master Clock divided by one of both DIVA or DIVB divider, the formula becomes, respectively:

$$\frac{(2 \times X \times CPRD \times DIVA)}{MCK} \text{ or } \frac{(2 \times X \times CPRD \times DIVB)}{MCK}$$

- the **waveform duty cycle**. This channel parameter is defined in the CDTY field of the PWM\_CDTYx register. If the waveform is left aligned then:

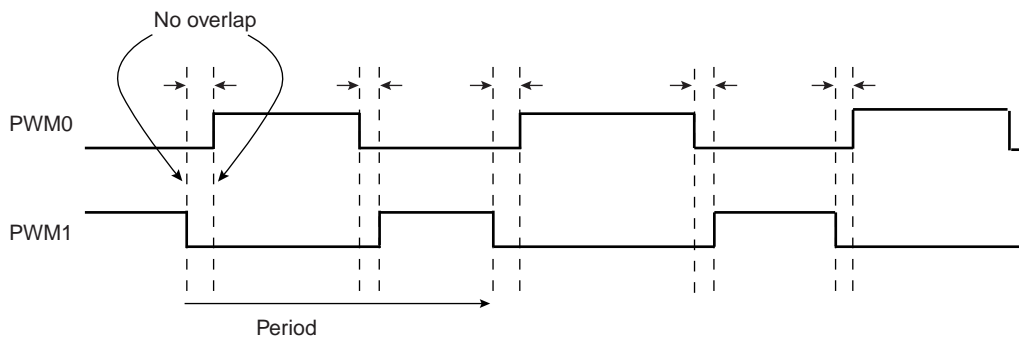
$$\text{duty cycle} = \frac{(\text{period} - 1/\text{fchannel\_x\_clock} \times CDTY)}{\text{period}}$$

If the waveform is center aligned, then:

$$\text{duty cycle} = \frac{((\text{period}/2) - 1/\text{fchannel\_x\_clock} \times CDTY))}{(\text{period}/2)}$$

- the **waveform polarity**. At the beginning of the period, the signal can be at high or low level. This property is defined in the CPOL field of the PWM\_CMRx register. By default the signal starts by a low level.
- the **waveform alignment**. The output waveform can be left or center aligned. Center aligned waveforms can be used to generate non overlapped waveforms. This property is defined in the CALG field of the PWM\_CMRx register. The default mode is left aligned.

**Figure 38-4. Non Overlapped Center Aligned Waveforms**



Note: 1. See [Figure 38-5 on page 833](#) for a detailed description of center aligned waveforms.

When center aligned, the internal channel counter increases up to CPRD and decreases down to 0. This ends the period.

When left aligned, the internal channel counter increases up to CPRD and is reset. This ends the period.

Thus, for the same CPRD value, the period for a center aligned channel is twice the period for a left aligned channel.

Waveforms are fixed at 0 when:

- CDTY = CPRD and CPOL = 0
- CDTY = 0 and CPOL = 1

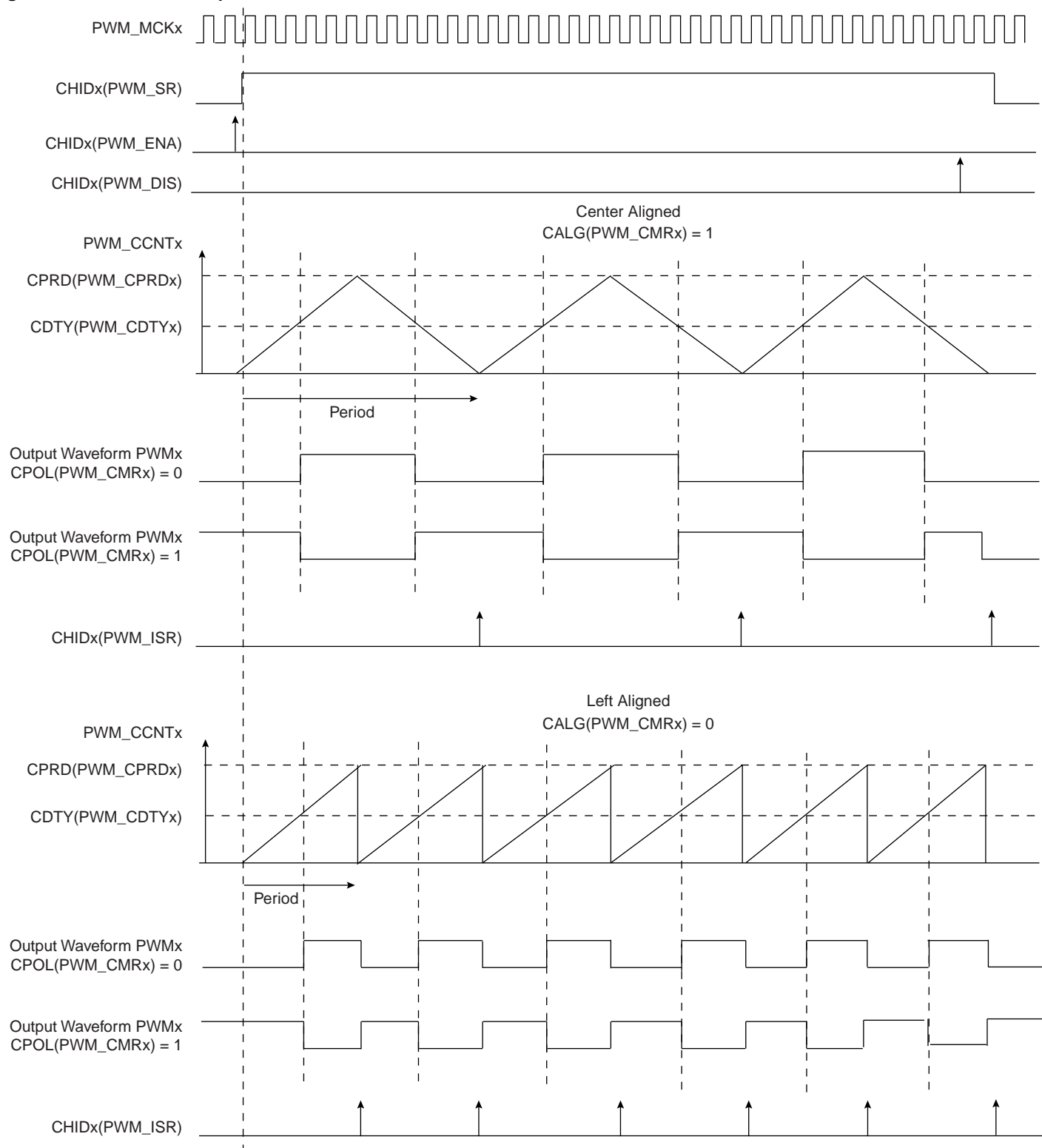
Waveforms are fixed at 1 (once the channel is enabled) when:

- CDTY = 0 and CPOL = 0
- CDTY = CPRD and CPOL = 1

The waveform polarity must be set before enabling the channel. This immediately affects the channel output level. Changes on channel polarity are not taken into account while the channel is enabled.



**Figure 38-5. Waveform Properties**



## 38.6.3 PWM Controller Operations

### 38.6.3.1 Initialization

Before enabling the output channel, this channel must have been configured by the software application:

- Configuration of the clock generator if DIVA and DIVB are required
- Selection of the clock for each channel (CPRE field in the PWM\_CMRx register)
- Configuration of the waveform alignment for each channel (CALG field in the PWM\_CMRx register)
- Configuration of the period for each channel (CPRD in the PWM\_CPRDx register). Writing in PWM\_CPRDx Register is possible while the channel is disabled. After validation of the channel, the user must use PWM\_CUPDx Register to update PWM\_CPRDx as explained below.
- Configuration of the duty cycle for each channel (CDTY in the PWM\_CDTYx register). Writing in PWM\_CDTYx Register is possible while the channel is disabled. After validation of the channel, the user must use PWM\_CUPDx Register to update PWM\_CDTYx as explained below.
- Configuration of the output waveform polarity for each channel (CPOL in the PWM\_CMRx register)
- Enable Interrupts (Writing CHIDx in the PWM\_IER register)
- Enable the PWM channel (Writing CHIDx in the PWM\_ENA register)

It is possible to synchronize different channels by enabling them at the same time by means of writing simultaneously several CHIDx bits in the PWM\_ENA register.

- In such a situation, all channels may have the same clock selector configuration and the same period specified.

### 38.6.3.2 Source Clock Selection Criteria

The large number of source clocks can make selection difficult. The relationship between the value in the Period Register (PWM\_CPRDx) and the Duty Cycle Register (PWM\_CDTYx) can help the user in choosing. The event number written in the Period Register gives the PWM accuracy. The Duty Cycle quantum cannot be lower than  $1/PWM\_CPRD_x$  value. The higher the value of PWM\_CPRDx, the greater the PWM accuracy.

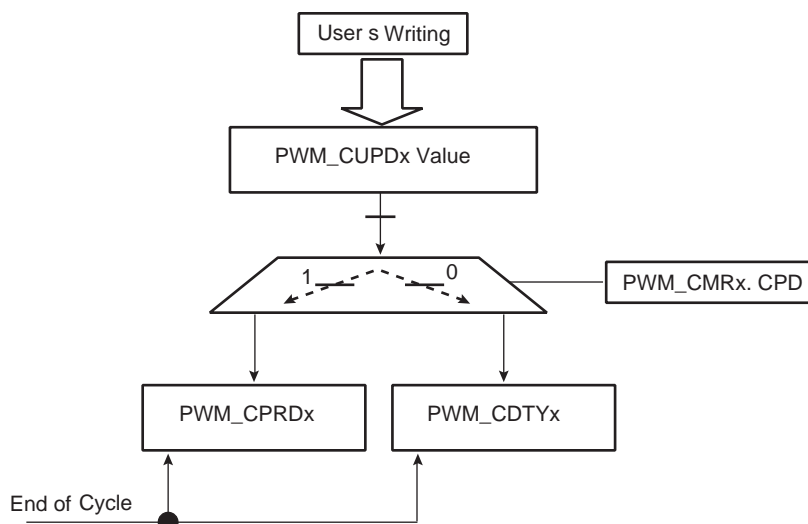
For example, if the user sets 15 (in decimal) in PWM\_CPRDx, the user is able to set a value between 1 up to 14 in PWM\_CDTYx Register. The resulting duty cycle quantum cannot be lower than 1/15 of the PWM period.

### 38.6.3.3 Changing the Duty Cycle or the Period

It is possible to modulate the output waveform duty cycle or period.

To prevent unexpected output waveform, the user must use the update register (PWM\_CUPDx) to change waveform parameters while the channel is still enabled. The user can write a new period value or duty cycle value in the update register (PWM\_CUPDx). This register holds the new value until the end of the current cycle and updates the value for the next cycle. Depending on the CPD field in the PWM\_CMRx register, PWM\_CUPDx either updates PWM\_CPRDx or PWM\_CDTYx. Note that even if the update register is used, the period must not be smaller than the duty cycle.

**Figure 38-6. Synchronized Period or Duty Cycle Update**



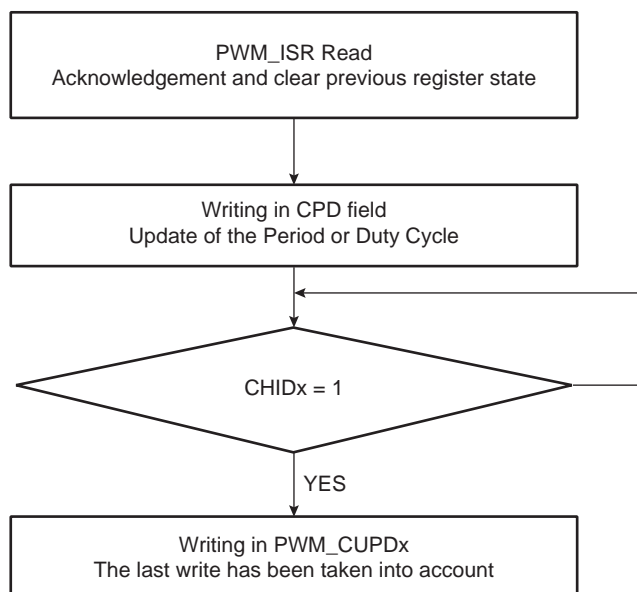
To prevent overwriting the PWM\_CUPDx by software, the user can use status events in order to synchronize his software. Two methods are possible. In both, the user must enable the dedicated interrupt in PWM\_IER at PWM Controller level.

The first method (polling method) consists of reading the relevant status bit in PWM\_ISR Register according to the enabled channel(s). See [Figure 38-7](#).

The second method uses an Interrupt Service Routine associated with the PWM channel.

Note: Reading the PWM\_ISR register automatically clears CHIDx flags.

**Figure 38-7. Polling Method**



Note: Polarity and alignment can be modified only when the channel is disabled.

#### 38.6.3.4 Interrupts

Depending on the interrupt mask in the PWM\_IMR register, an interrupt is generated at the end of the corresponding channel period. The interrupt remains active until a read operation in the PWM\_ISR register occurs.

A channel interrupt is enabled by setting the corresponding bit in the PWM\_IER register. A channel interrupt is disabled by setting the corresponding bit in the PWM\_IDR register.

## 38.7 Pulse Width Modulation Controller (PWM) User Interface

Table 38-4. Register Mapping<sup>(2)</sup>

Offset	Register	Name	Access	Reset
0x00	PWM Mode Register	PWM_MR	Read-write	0
0x04	PWM Enable Register	PWM_ENA	Write-only	-
0x08	PWM Disable Register	PWM_DIS	Write-only	-
0x0C	PWM Status Register	PWM_SR	Read-only	0
0x10	PWM Interrupt Enable Register	PWM_IER	Write-only	-
0x14	PWM Interrupt Disable Register	PWM_IDR	Write-only	-
0x18	PWM Interrupt Mask Register	PWM_IMR	Read-only	0
0x1C	PWM Interrupt Status Register	PWM_ISR	Read-only	0
0x20 - 0xFC	Reserved	—	—	—
0x100 - 0x1FC	Reserved			
0x200 + ch_num * 0x20 + 0x00	PWM Channel Mode Register	PWM_CMR	Read-write	0x0
0x200 + ch_num * 0x20 + 0x04	PWM Channel Duty Cycle Register	PWM_CDTY	Read-write	0x0
0x200 + ch_num * 0x20 + 0x08	PWM Channel Period Register	PWM_CPRD	Read-write	0x0
0x200 + ch_num * 0x20 + 0x0C	PWM Channel Counter Register	PWM_CCNT	Read-only	0x0
0x200 + ch_num * 0x20 + 0x10	PWM Channel Update Register	PWM_CUPD	Write-only	-

2. Some registers are indexed with “ch\_num” index ranging from 0 to 3.

### 38.7.1 PWM Mode Register

**Name:** PWM\_MR

**Address:** 0x48008000

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	PREB			
23	22	21	20	19	18	17	16
DIVB							
15	14	13	12	11	10	9	8
–	–	–	–	PREA			
7	6	5	4	3	2	1	0
DIVA							

- DIVA, DIVB: CLKA, CLKB Divide Factor**

Value	Name	Description
0	CLK_OFF	CLKA, CLKB clock is turned off
1	CLK_DIV1	CLKA, CLKB clock is clock selected by PREA, PREB
2-255	–	CLKA, CLKB clock is clock selected by PREA, PREB divided by DIVA, DIVB factor.

- PREA, PREB**

Value	Name	Description
0000	MCK	Master Clock
0001	MCKDIV2	Master Clock divided by 2
0010	MCKDIV4	Master Clock divided by 4
0011	MCKDIV8	Master Clock divided by 8
0100	MCKDIV16	Master Clock divided by 16
0101	MCKDIV32	Master Clock divided by 32
0110	MCKDIV64	Master Clock divided by 64
0111	MCKDIV128	Master Clock divided by 128
1000	MCKDIV256	Master Clock divided by 256
1001	MCKDIV512	Master Clock divided by 512
1010	MCKDIV1024	Master Clock divided by 1024

Values which are not listed in the table must be considered as “reserved”.

### 38.7.2 PWM Enable Register

**Name:** PWM\_ENA

**Address:** 0x48008004

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID**

0 = No effect.

1 = Enable PWM output for channel x.

### 38.7.3 PWM Disable Register

**Name:** PWM\_DIS

**Address:** 0x48008008

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID**

0 = No effect.

1 = Disable PWM output for channel x.



### 38.7.4 PWM Status Register

**Name:** PWM\_SR

**Address:** 0x4800800C

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID**

0 = PWM output for channel x is disabled.

1 = PWM output for channel x is enabled.

### 38.7.5 PWM Interrupt Enable Register

**Name:** PWM\_IER

**Address:** 0x48008010

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID.**

0 = No effect.

1 = Enable interrupt for PWM channel x.

### 38.7.6 PWM Interrupt Disable Register

**Name:** PWM\_IDR

**Address:** 0x48008014

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID.**

0 = No effect.

1 = Disable interrupt for PWM channel x.

### 38.7.7 PWM Interrupt Mask Register

**Name:** PWM\_IMR

**Address:** 0x48008018

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID.**

0 = Interrupt for PWM channel x is disabled.

1 = Interrupt for PWM channel x is enabled.

### 38.7.8 PWM Interrupt Status Register

**Name:** PWM\_ISR

**Address:** 0x4800801C

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	CHID3	CHID2	CHID1	CHID0

- **CHIDx: Channel ID**

0 = No new channel period has been achieved since the last read of the PWM\_ISR register.

1 = At least one new channel period has been achieved since the last read of the PWM\_ISR register.

Note: Reading PWM\_ISR automatically clears CHIDx flags.

### 38.7.9 PWM Channel Mode Register

**Name:** PWM\_CMR[0..3]

**Address:** 0x48008200 [0], 0x48008220 [1], 0x48008240 [2], 0x48008260 [3]

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	CPD	CPOL	CALG
7	6	5	4	3	2	1	0
–	–	–	–	CPRE			

#### • CPRE: Channel Pre-scaler

Value	Name	Description
0000	MCK	Master Clock
0001	MCKDIV2	Master Clock divided by 2
0010	MCKDIV4	Master Clock divided by 4
0011	MCKDIV8	Master Clock divided by 8
0100	MCKDIV16	Master Clock divided by 16
0101	MCKDIV32	Master Clock divided by 32
0110	MCKDIV64	Master Clock divided by 64
0111	MCKDIV128	Master Clock divided by 128
1000	MCKDIV256	Master Clock divided by 256
1001	MCKDIV512	Master Clock divided by 512
1010	MCKDIV1024	Master Clock divided by 1024
1011	CLKA	Clock A
1100	CLKB	Clock B

Values which are not listed in the table must be considered as “reserved”.

#### • CALG: Channel Alignment

0 = The period is left aligned.

1 = The period is center aligned.

#### • CPOL: Channel Polarity

0 = The output waveform starts at a low level.

1 = The output waveform starts at a high level.

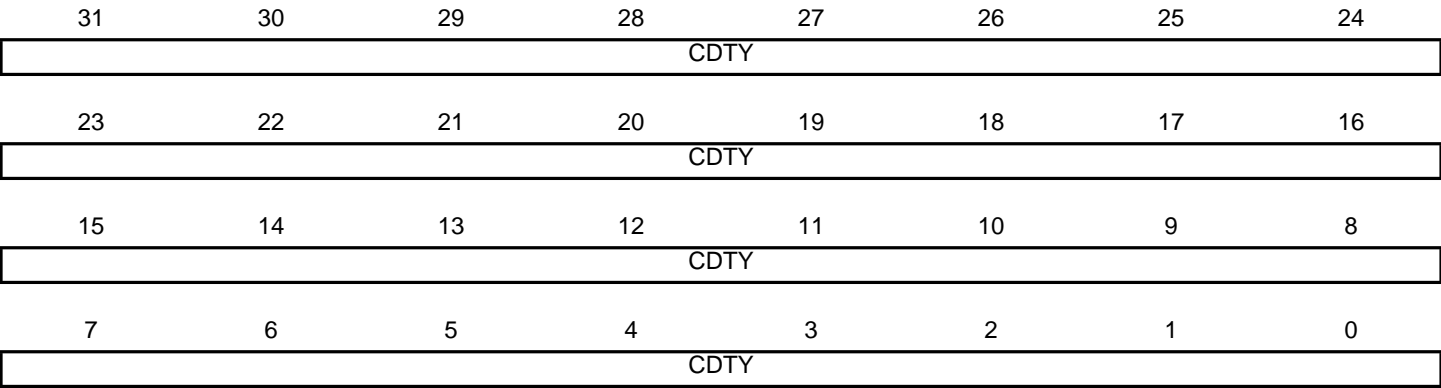
#### • CPD: Channel Update Period

0 = Writing to the PWM\_CUPDx will modify the duty cycle at the next period start event.

1 = Writing to the PWM\_CUPDx will modify the period at the next period start event.

38.7.10 PWM Channel Duty Cycle Register

**Name:** PWM\_CDTY[0..3]  
**Address:** 0x48008204 [0], 0x48008224 [1], 0x48008244 [2], 0x48008264 [3]  
**Access:** Read/Write



Only the first 16 bits (internal channel counter size) are significant.

- **CDTY: Channel Duty Cycle**  
Defines the waveform duty cycle. This value must be defined between 0 and CPRD (PWM\_CPRx).

### 38.7.11 PWM Channel Period Register

**Name:** PWM\_CPRD[0..3]

**Address:** 0x48008208 [0], 0x48008228 [1], 0x48008248 [2], 0x48008268 [3]

**Access:** Read/Write

31	30	29	28	27	26	25	24
CPRD							
23	22	21	20	19	18	17	16
CPRD							
15	14	13	12	11	10	9	8
CPRD							
7	6	5	4	3	2	1	0
CPRD							

Only the first 16 bits (internal channel counter size) are significant.

#### • CPRD: Channel Period

If the waveform is left-aligned, then the output waveform period depends on the counter source clock and can be calculated:

- By using the Master Clock (MCK) divided by an X given prescaler value (with X being 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, or 1024). The resulting period formula will be:

$$\frac{(X \times CPRD)}{MCK}$$

- By using a Master Clock divided by one of both DIVA or DIVB divider, the formula becomes, respectively:

$$\frac{(CPRD \times DIVA)}{MCK} \text{ or } \frac{(CPRD \times DIVB)}{MCK}$$

If the waveform is center-aligned, then the output waveform period depends on the counter source clock and can be calculated:

- By using the Master Clock (MCK) divided by an X given prescaler value (with X being 1, 2, 4, 8, 16, 32, 64, 128, 256, 512, or 1024). The resulting period formula will be:

$$\frac{(2 \times X \times CPRD)}{MCK}$$

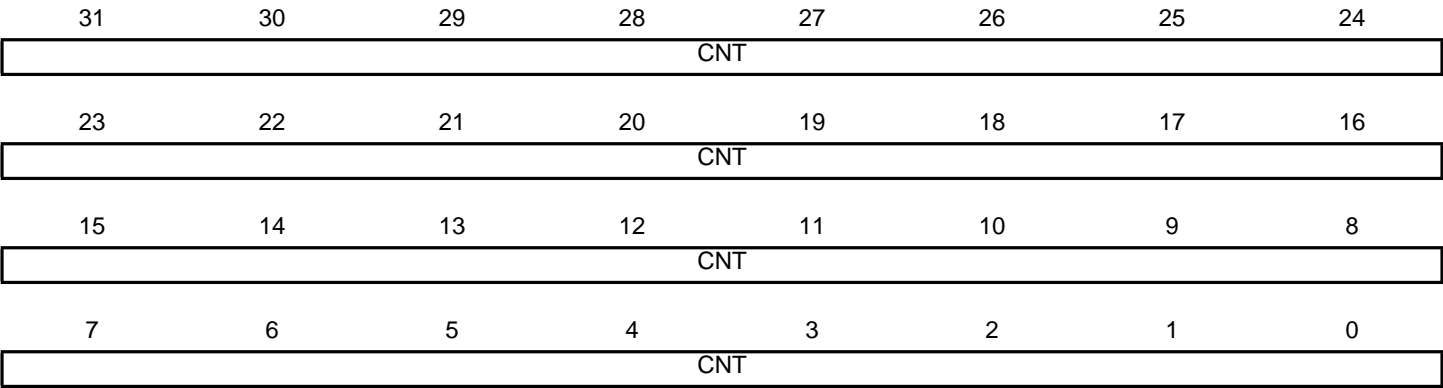
- By using a Master Clock divided by one of both DIVA or DIVB divider, the formula becomes, respectively:

$$\frac{(2 \times CPRD \times DIVA)}{MCK} \text{ or } \frac{(2 \times CPRD \times DIVB)}{MCK}$$



38.7.12 PWM Channel Counter Register

**Name:** PWM\_CCNT[0..3]  
**Address:** 0x4800820C [0], 0x4800822C [1], 0x4800824C [2], 0x4800826C [3]  
**Access:** Read-only



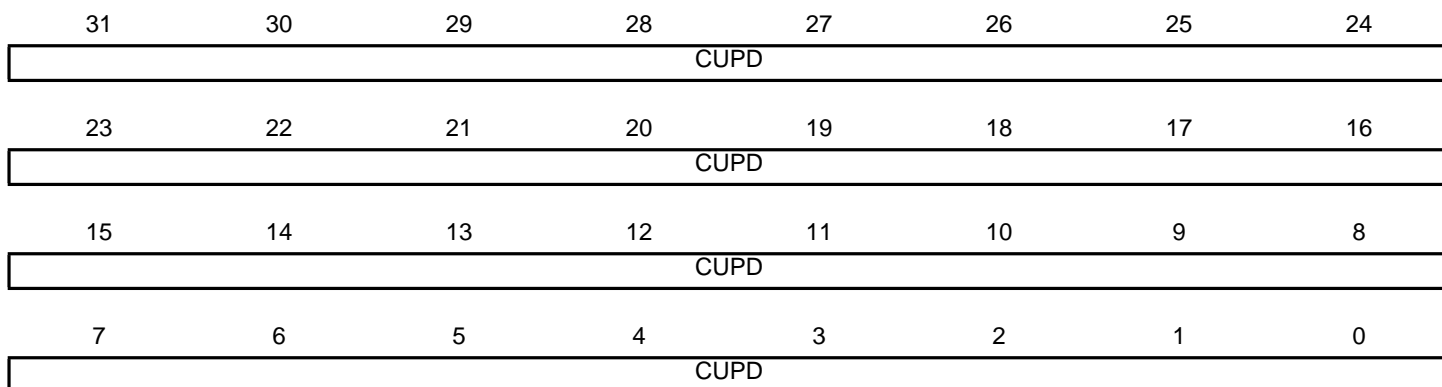
- **CNT: Channel Counter Register**  
Internal counter value. This register is reset when:
  - The channel is enabled (writing CHIDx in the PWM\_ENA register).
  - The counter reaches CPRD value defined in the PWM\_CPRDx register if the waveform is left aligned.

### 38.7.13 PWM Channel Update Register

**Name:** PWM\_CUPD[0..3]

**Address:** 0x48008210 [0], 0x48008230 [1], 0x48008250 [2], 0x48008270 [3]

**Access:** Write-only



CUPD: Channel Update Register

This register acts as a double buffer for the period or the duty cycle. This prevents an unexpected waveform when modifying the waveform period or duty-cycle.

Only the first 16 bits (internal channel counter size) are significant.

When CPD field of PWM\_CMRx register = 0, the duty-cycle (CDTY of PWM\_CDTYx register) is updated with the CUPD value at the beginning of the next period.

When CPD field of PWM\_CMRx register = 1, the period (CPRD of PWM\_CPRDx register) is updated with the CUPD value at the beginning of the next period.

## 39. Segment Liquid Crystal Display Controller (SLCDC)

### 39.1 Description

An LCD consists of several segments (pixels or complete symbols) which can be visible or invisible. A segment has two electrodes with liquid crystal between them. When a voltage above a threshold voltage is applied across the liquid crystal, the segment becomes visible.

The voltage must alternate to avoid an electrophoresis effect in the liquid crystal, which degrades the display. Hence the waveform across a segment must not have a DC component.

The SLCDC controller is intended for monochrome passive liquid crystal display (LCD) with up to 6 common terminals and up to 50 segment terminals.

The SLCDC is programmable to support many different requirements such as:

- Adjusting the driving time of the LCD pads in order to save power and increase the controllability of the DC offset
- Driving smaller LCD (down to 1 common by 1 segment)
- Adjusting the SLCDC frequency in order to obtain the best compromise between frequency and consumption and adapt it to the LCD driver
- The segments can be assigned in a user defined pattern to ease the usage of the digital functions multiplexed on these pins

#### 39.1.1 Definition of Terms.

Table 39-1. List of Terms

Term	Description
LCD	A passive display panel with terminals leading directly to a segment
Segment	The least viewing element (pixel) which can be on or off
Common(s)	Denotes how many segments are connected to a segment terminal
Duty	$1/(\text{Number of common terminals on an actual LCD display})$
Bias	$1/(\text{Number of voltage levels used driving a LCD display} - 1)$
Frame Rate	Number of times the LCD segments are energized per second.

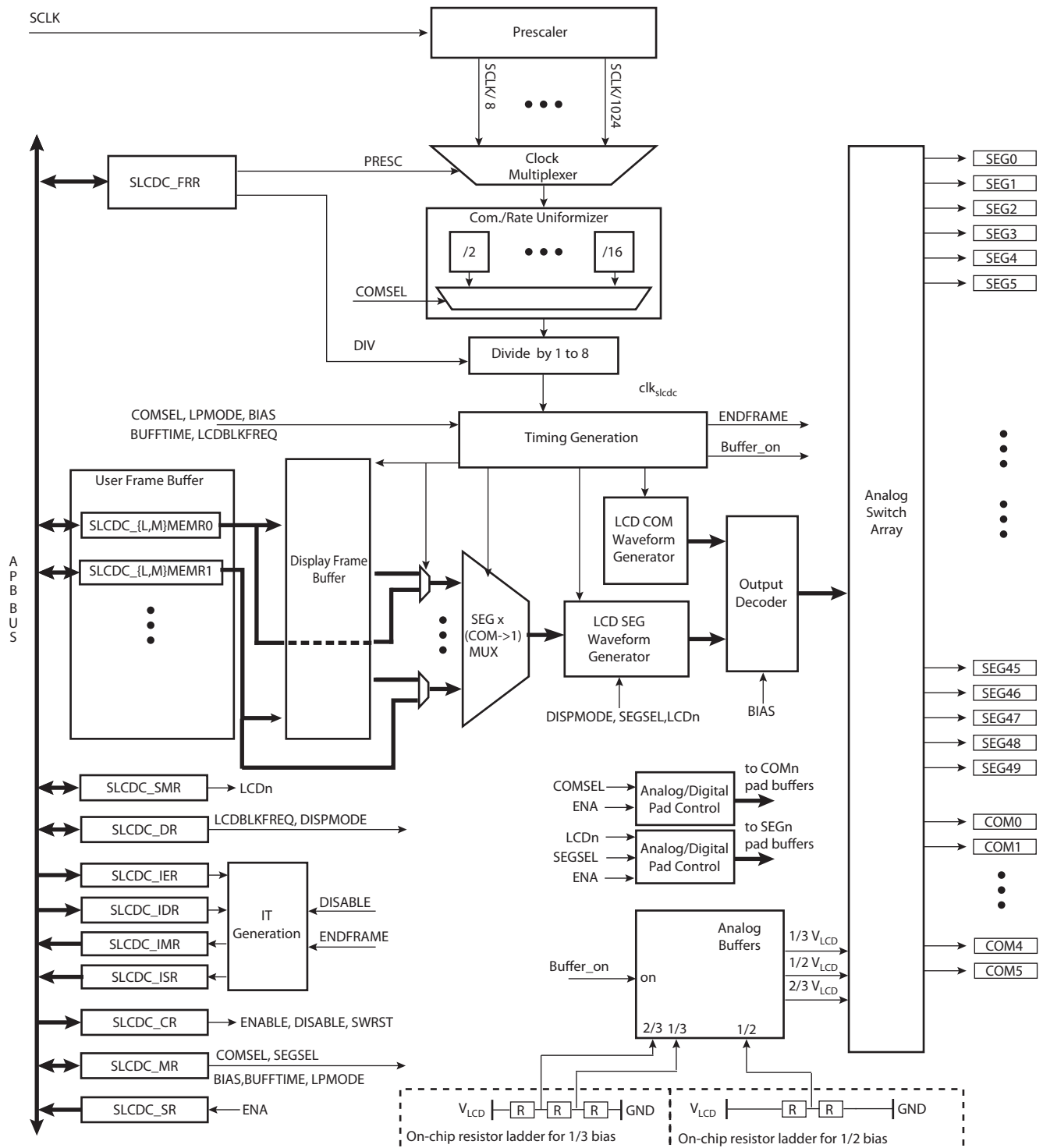
## 39.2 Embedded Characteristics

The SLCDC controller provides the following capabilities:

- Display Capacity: Up to 50 Segments and 6 Common Terminals
- Support from Static to 1/6 Duty
- Support Static, 1/2, 1/3 Bias
- Two LCD Supply Sources:
  - Internal: The On-chip LCD Power Supply
  - External
- LCD Output Voltage Software Selectable from 2.4V to VDDIN in 16 Steps (Control Embedded in the Supply Controller)
- Flexible Selection of Frame Frequency
- Two Interrupt Sources: End Of Frame and Disable
- Versatile Display Modes
- Equal Source and Sink Capability to Maximize LCD Life Time
- Segment and Common Pins not Needed for Driving the Display Can be Used as Ordinary I/O Pins
- Segments Lay Out can be Fully Defined by User to Optimize Usage of Multiplexed Digital Functions
- Latching of Display Data Gives Full Freedom in Register Updates
- Power Saving Modes for Extremely Low Power Consumption

## 39.3 Block Diagram

Figure 39-1. LCD Macrocell Block Diagram



## 39.4 I/O Lines Description

Table 39-2. I/O Lines Description

Name	Description	Type
SEG [49:0]	Segments control signals	Output
COM [5:0]	Commons control signals	Output

## 39.5 Product Dependencies

### 39.5.1 I/O Lines

The pins used for interfacing the SLCD Controller may be multiplexed with PIO lines. Refer to product block diagram.

If I/O lines of the SLCD Controller are not used by the application, they can be used for other purposes by the PIO Controller.

By default (SLCDC\_SMR0/1 registers cleared) the assignment of the segment controls and commons are automatically done depending on COMSEL and SEGSEL in SLCDC\_MR. As example, if 10 segments are programmed in the SEGSEL bitfield, they are automatically assigned to SEG[9:0] whereas remaining SEG pins are automatically selected to be driven by the multiplexed digital functions.

Anyway, the user can define a new lay out pattern for the segment assignment by programming the SLCDC\_SMR0/1 registers in order to optimize the usage of multiplexed digital function. If at least 1 bit is set in SLCDC\_SMR0/1 registers, the corresponding I/O line will be driven by an LCD segment whereas any cleared bit of this register will select the corresponding multiplexed digital function.

Table 39-3. I/O Lines

Instance	Signal	I/O Line	Peripheral
SLCDC	COM0	PA0	X1
SLCDC	COM1	PA1	X1
SLCDC	COM2	PA2	X1
SLCDC	COM3	PA3	X1
SLCDC	COM4/AD1	PA4	X1
SLCDC	COM5/AD2	PA5	X1
SLCDC	SEG0	PA6	X1
SLCDC	SEG1	PA7	X1
SLCDC	SEG2	PA8	X1
SLCDC	SEG3	PA9	X1
SLCDC	SEG4	PA10	X1
SLCDC	SEG5	PA11	X1
SLCDC	SEG6/AD0	PA12	X1
SLCDC	SEG7	PA13	X1
SLCDC	SEG8	PA14	X1
SLCDC	SEG9	PA15	X1
SLCDC	SEG10	PA16	X1
SLCDC	SEG11	PA17	X1
SLCDC	SEG12	PA18	X1

Table 39-3. I/O Lines

SLCDC	SEG13	PA19	X1
SLCDC	SEG14	PA20	X1
SLCDC	SEG15	PA21	X1
SLCDC	SEG16	PA22	X1
SLCDC	SEG17	PA23	X1
SLCDC	SEG18	PA24	X1
SLCDC	SEG19	PA25	X1
SLCDC	SEG20	PA26	X1
SLCDC	SEG21	PA27	X1
SLCDC	SEG22	PA28	X1
SLCDC	SEG23	PA29	X1
SLCDC	SEG24	PB6	X1
SLCDC	SEG25	PB7	X1
SLCDC	SEG26	PB8	X1
SLCDC	SEG27	PB9	X1
SLCDC	SEG28	PB10	X1
SLCDC	SEG29	PB11	X1
SLCDC	SEG30	PB12	X1
SLCDC	SEG31/AD3	PB13	X1
SLCDC	SEG32	PB14	X1
SLCDC	SEG33	PB15	X1
SLCDC	SEG34	PB16	X1
SLCDC	SEG35	PB17	X1
SLCDC	SEG36	PB18	X1
SLCDC	SEG37	PB19	X1
SLCDC	SEG38	PB20	X1
SLCDC	SEG39	PB21	X1
SLCDC	SEG40	PB22	X1
SLCDC	SEG41/AD4	PB23	X1
SLCDC	SEG42	PB24	X1
SLCDC	SEG43	PB25	X1
SLCDC	SEG44	PB26	X1
SLCDC	SEG45	PB27	X1
SLCDC	SEG46	PB28	X1
SLCDC	SEG47	PB29	X1
SLCDC	SEG48	PB30	X1
SLCDC	SEG49/AD5	PB31	X1

### 39.5.2 Power Management

The SLCD Controller is clocked by the slow clock (SCLK). All the timings are based upon a typical value of 32 kHz for SCLK. The power management of the SLCD controller is handled by the Shutdown Controller.

The SLCD Controller is supplied by 3V domain.

### 39.5.3 Interrupt Sources

The SLCD Controller interrupt line is connected to one of the internal sources of the Interrupt Controller. Using the SLCD Controller interrupt requires prior programming of the Interrupt Controller.

**Table 39-4. Peripheral IDs**

Instance	ID
SLCDC	32

### 39.5.4 Number of Segments and Commons

The product, embeds 50 segments and 6 Commons.



## 39.6 Functional Description

After the initialization sequence the SLCDC is ready to be enabled (SLCDC\_CR) in order to enter the display phase (where it is possible to do more than display data written in the SLCDC memory) up to the disable sequence.

- Initialization Sequence:
  1. Select the LCD supply source in the shutdown controller
    - Internal: the On-chip LCD Power Supply is selected,
    - External: the external supply source has to be between 2.5 to 3.6V
  2. Select the clock division (SLCDC\_FRR) to use a proper frame rate
  3. Enter the number of common and segments terminals (SLCDC\_MR)
  4. Select the bias in compliance with the LCD manufacturer data sheet (SLCDC\_MR)
  5. Enter buffer driving time (SLCDC\_MR)
  6. Define the segments remapping pattern if required (SLCDC\_SMR0/1)
- During the Display Phase:
  1. Data may be written at any time in the SLCDC memory, they are automatically latched and displayed at the next LCD frame
  2. It is possible to:
    - Adjust contrast
    - Adjust the frame frequency
    - Adjust buffer driving time
    - Reduce the SLCDC consumption by entering in low-power waveform at any time
    - Use the large set of display features such as blinking, inverted blink, etc.
- Disable Sequence:

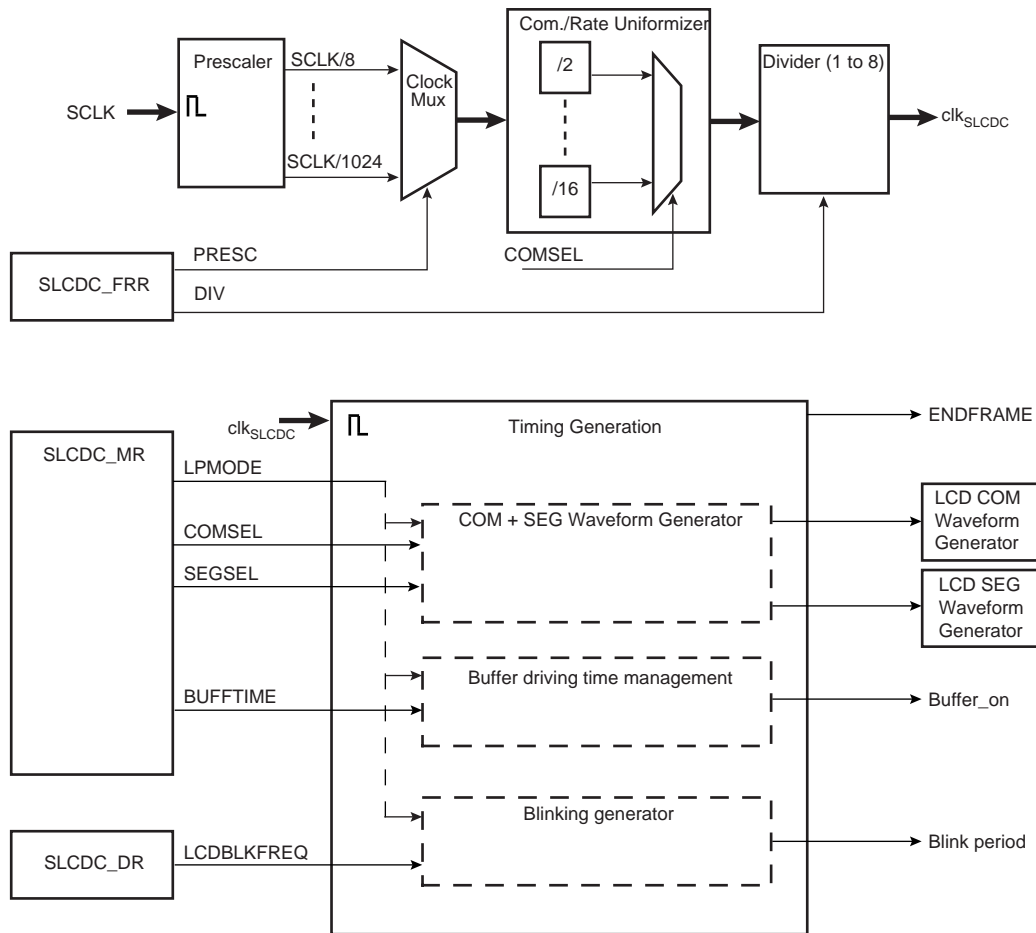
There are two ways to disable the SLCDC

1. By using the LCDDIS (LCD Disable) bit. (In this case, SLCDC configuration and memory content are kept.)
2. Or by using the SWRST (Software Reset) bit.

## 39.6.1 lock Generation

### 39.6.1.1 Block Diagram

Figure 39-2. Clock Generation Block Diagram

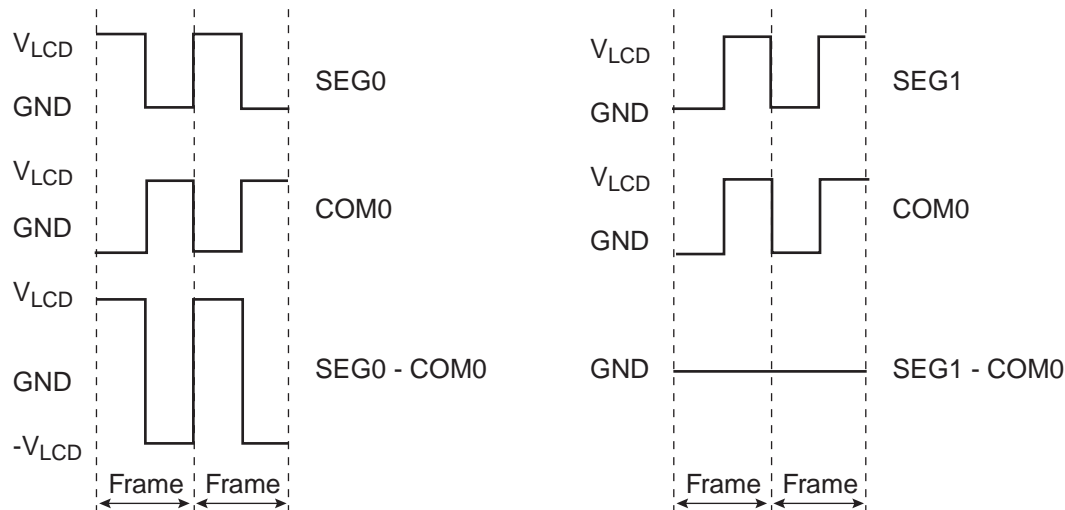


## 39.6.2 Waveform Generation

### 39.6.2.1 Static Duty and Bias

This kind of display is driven with the waveform shown in [Figure 39-3](#). SEG0 - COM0 is the voltage across a segment that is on, and SEG1 - COM0 is the voltage across a segment that is off.

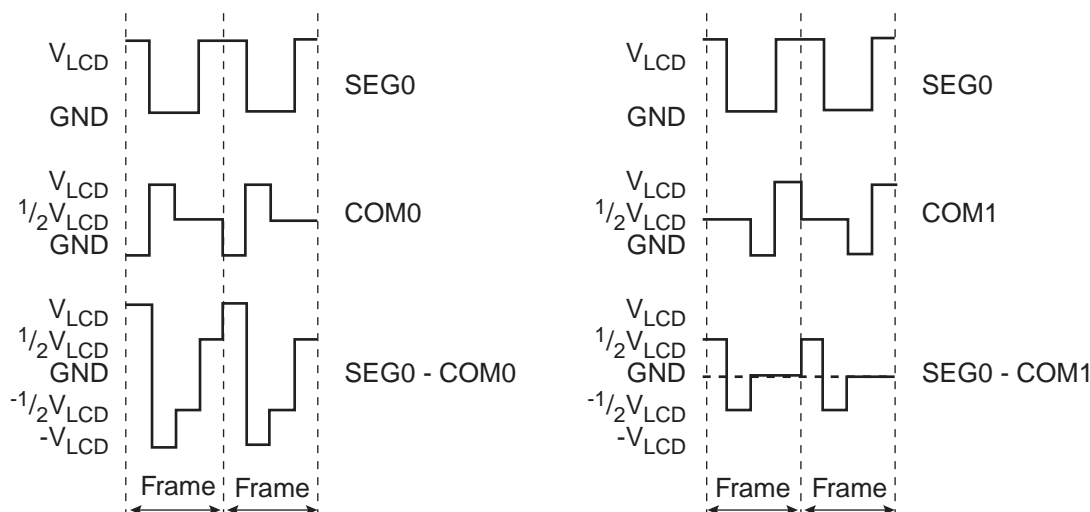
Figure 39-3. Driving an LCD with One Common Terminal



### 39.6.2.2 1/2 Duty and 1/2 Bias

For an LCD with two common terminals (1/2 duty) a more complex waveform must be used to control segments individually. Although 1/3 bias can be selected, 1/2 bias is most common for these displays. In the waveform shown in [Figure 39-4](#), SEG0 - COM0 is the voltage across a segment that is on, and SEG0 - COM1 is the voltage across a segment that is off.

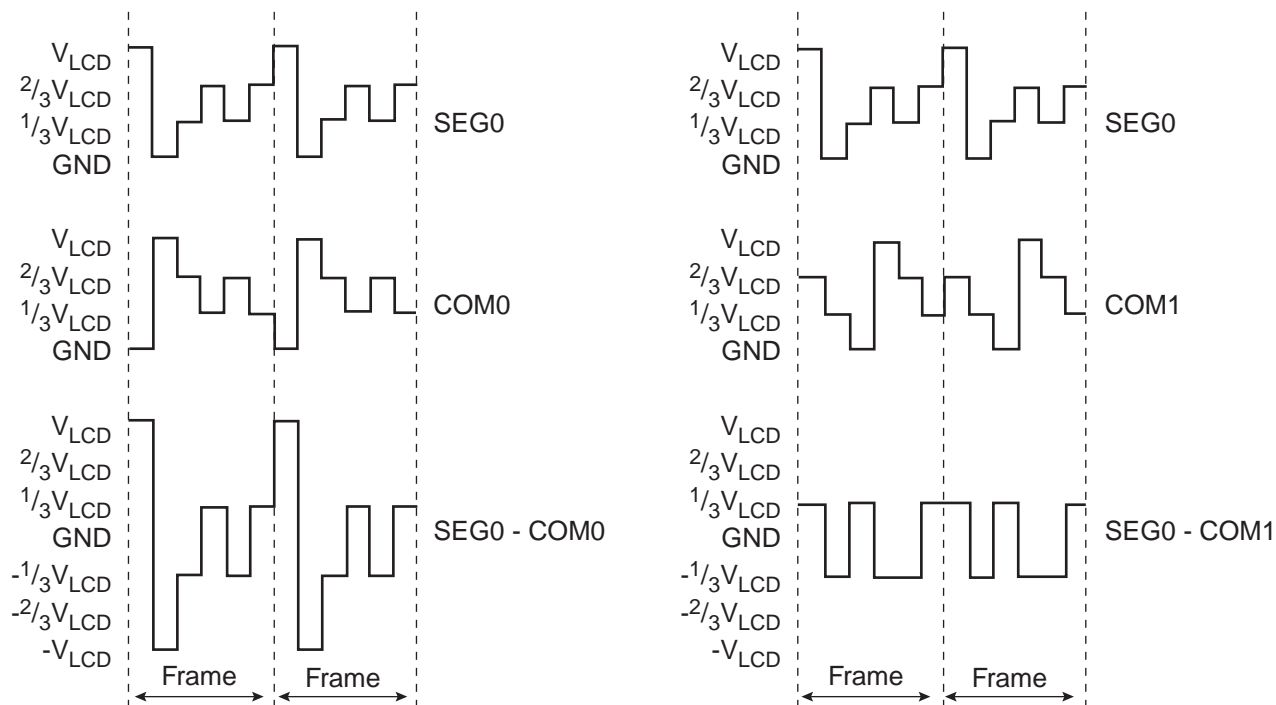
Figure 39-4. Driving an LCD with Two Common Terminals



### 39.6.2.3 1/3 Duty and 1/3 Bias

1/3 bias is usually recommended for an LCD with three common terminals (1/3 duty). In the waveform shown in Figure 39-5, SEG0 - COM0 is the voltage across a segment that is on and SEG0-COM1 is the voltage across a segment that is off.

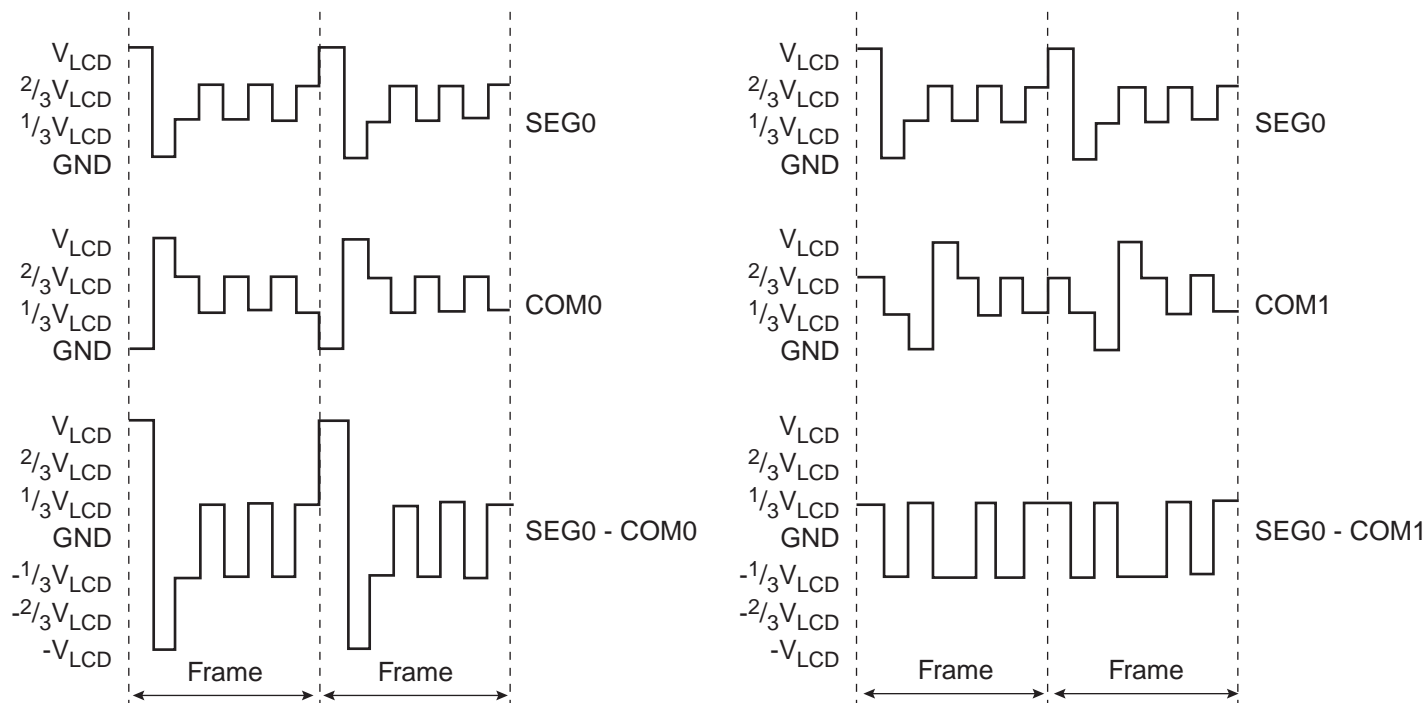
**Figure 39-5. Driving an LCD with Three Common Terminals**



### 39.6.2.4 1/4 Duty and 1/3 Bias

1/3 bias is optimal for LCD displays with four common terminals (1/4 duty). In the waveform shown in [Figure 39-6](#), SEG0 - COM0 is the voltage across a segment that is on and SEG0 - COM1 is the voltage across a segment that is off.

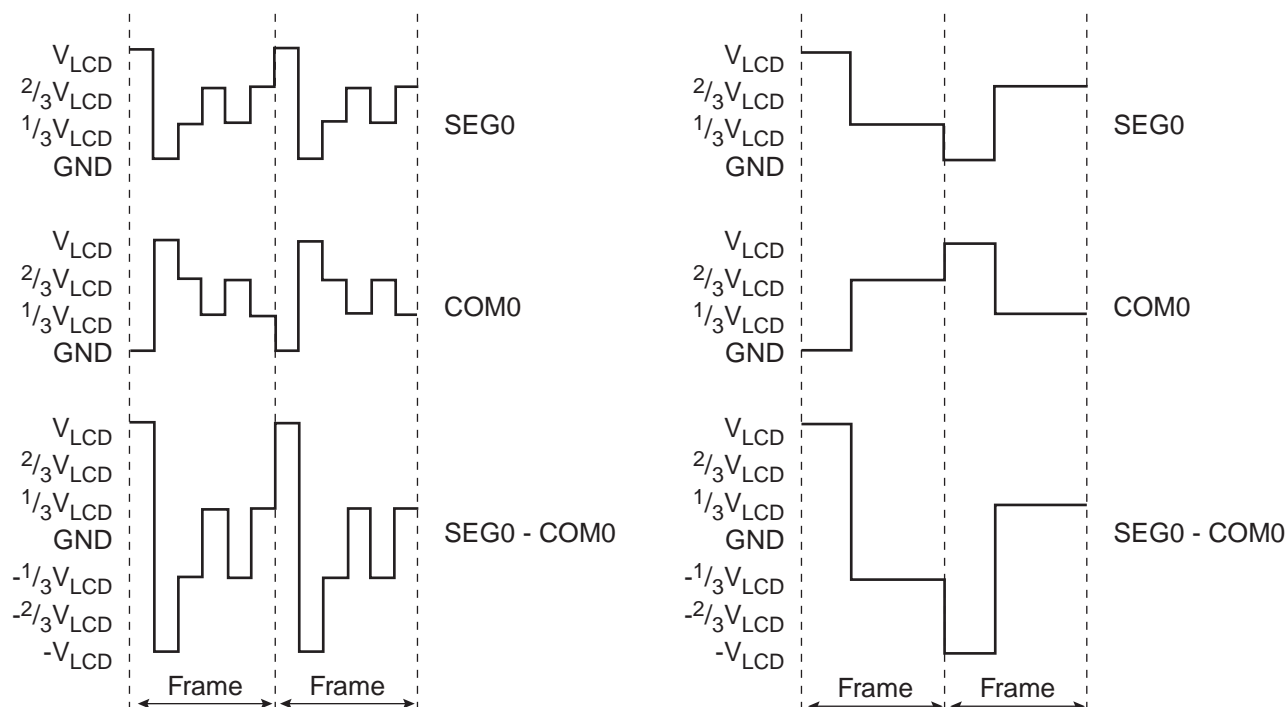
**Figure 39-6. Driving an LCD with Four Common Terminals**



### 39.6.2.5 Low Power Waveform

To reduce toggle activity and hence power consumption, a low power waveform can be selected by writing LPMODE to one. The default and low power waveform is shown in Figure 39-7 for 1/3 duty and 1/3 bias. For other selections of duty and bias, the effect is similar.

**Figure 39-7. Default and Low Power Waveform**



Note: Refer to the LCD specification to verify that low power waveforms are supported.

### 39.6.2.6 Frame Rate

The Frame Rate register (SLCDC\_FRR) enables the generation of the frequency used by the SLCD Controller. It is done by a prescaler (division by 8, 16, 32, 64, 128, 256, 512 and 1024) followed by a finer divider (division by 1, 2, 3, 4, 5, 6, 7 or 8).

To calculate the proper frame frequency needed, the equation below must be taken into account:

$$f_{frame} = \frac{f_{SCLK}}{(PRESC \cdot DIV \cdot NCOM)}$$

Where:

$f_{SCLK}$  = slow clock frequency

$f_{frame}$  = frame frequency

PRESC = prescaler value (8, 16, 32, 64, 128, 256, 512 or 1024)

DIV = divider value (1, 2, 3, 4, 5, 6, 7, or 8)

NCOM = depends of number of commons and is defined in Table 39-5 below:

**Table 39-5. NCOM**

Number of Commons	NCOM
1	16
2	16
3	15
4	16
5	15
6	18

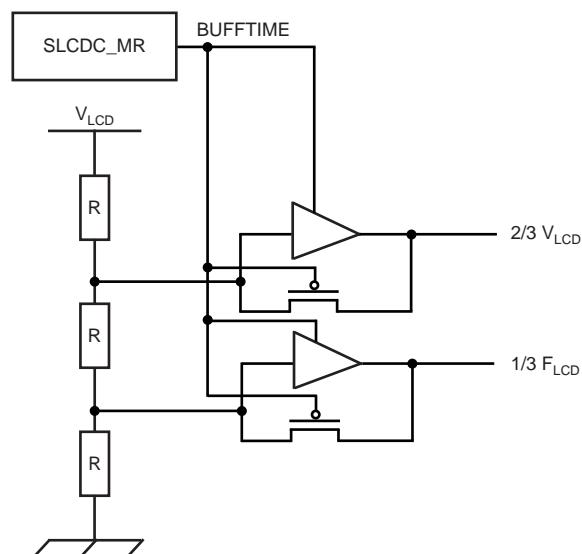
### 39.6.2.7 Buffer Driving Time

Intermediate voltage levels are generated from buffer drivers. The buffers are active the amount of time specified by BUFTIME[3:0] in SLCDC\_MR, then buffers are bypassed.

Shortening the drive time will reduce power consumption, but displays with high internal resistance or capacitance may need longer drive time to achieve sufficient contrast.

Example for bias = 1/3.

**Figure 39-8. Buffer Driving**

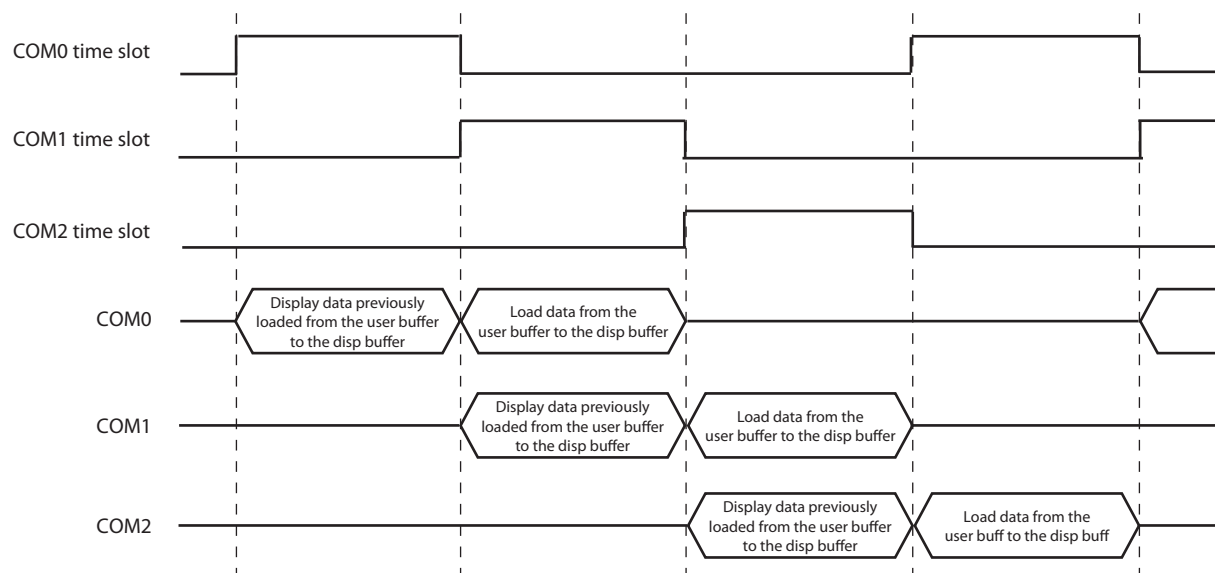


### 39.6.3 Number of Commons, Segments and Bias

It is important to note that the selection of the number of commons, segments and the bias is only taken into account when the SLCDC is disabled.

### 39.6.4 SLCDC memory

Figure 39-9. Memory Management



When a bit in the display memory (SLCDC\_LMEMRx SLCDC\_MMEMRx registers) is written to one, the corresponding segment is energized (on), and non-energized when a bit in the display memory is written to zero.

At the beginning of each common, the display buffer is updated. The value of the previous common is latched in the display memory (its value is transferred from the user buffer to the frame buffer).

The advantages of this solution are:

- Ability to access the user buffer at any time in the frame, in any display mode and even in low power waveform
- Ability to change only one pixel without reloading the picture

### 39.6.5 Display Features

In order to improve the flexibility of SLCDC the following set of display modes are embedded:

1. Force Mode Off: All pixels are turned off and the memory content is kept.
2. Force Mode On: All pixels are turned on and the memory content is kept.
3. Inverted Mode: All pixels are set in the inverted state as defined in SLCDC memory and the memory content is kept.
4. Two Blinking Modes:
  - Standard Blinking Mode: All pixels are alternately turned off to the predefined state in SLCDC memory at LCDBLKREQ frequency.
  - Inverted Blinking Mode: All pixels are alternately turned off to the predefined opposite state in SLCDC memory at LCDBLKREQ frequency.
5. Buffer Swap Mode: All pixels are alternatively assigned to the state defined in the user buffer then to the state defined in the display buffer.



### 39.6.6 Buffer Swap Mode

This mode allows to assign all pixels to two states alternatively without reloading the user buffer at each change.

The means to alternatively display two states is as follows:

1. Initially, the SLCDC must be in normal mode or in a standard blinking mode.
2. Data corresponding to the first pixel state is written in the user buffer (through the SLCDC\_MEM registers).
3. Wait two ENDFRAME events (to be sure that the user buffer is entirely transferred in the display buffer).
4. SLCDC\_DR must be programmed with DISPMODE = 6 (User Buffer Only Load Mode). This mode blocks the automatic transfer from the user buffer to the display buffer.
5. Wait ENDFRAME event. (The display mode is internally updated at the beginning of each frame.)
6. Data corresponding to the second pixel state is written in the user buffer (through the SLCDC\_MEM registers). So, now the first pixel state is in the display buffer and the second pixel state is in the user buffer.
7. SLCDC\_DR must be programmed with DISPMODE = 7 (buffer swap mode) and LCDBLKREQ must be programmed with the wanted blinking frequency (if not previously done).

Now, each state is alternatively displayed at LCDBLKREQ frequency.

Except for the phase dealing with the storage of the two display states, the management of the Buffer Swap Mode is the same as the standard blinking mode.

### 39.6.7 Disable Sequence

There are two ways to disable the SLCDC:

1. By using the disable bit. (In this case, register configuration and SLCDC memory are kept.)
2. Or by using the software reset bit that acts like a hardware reset.

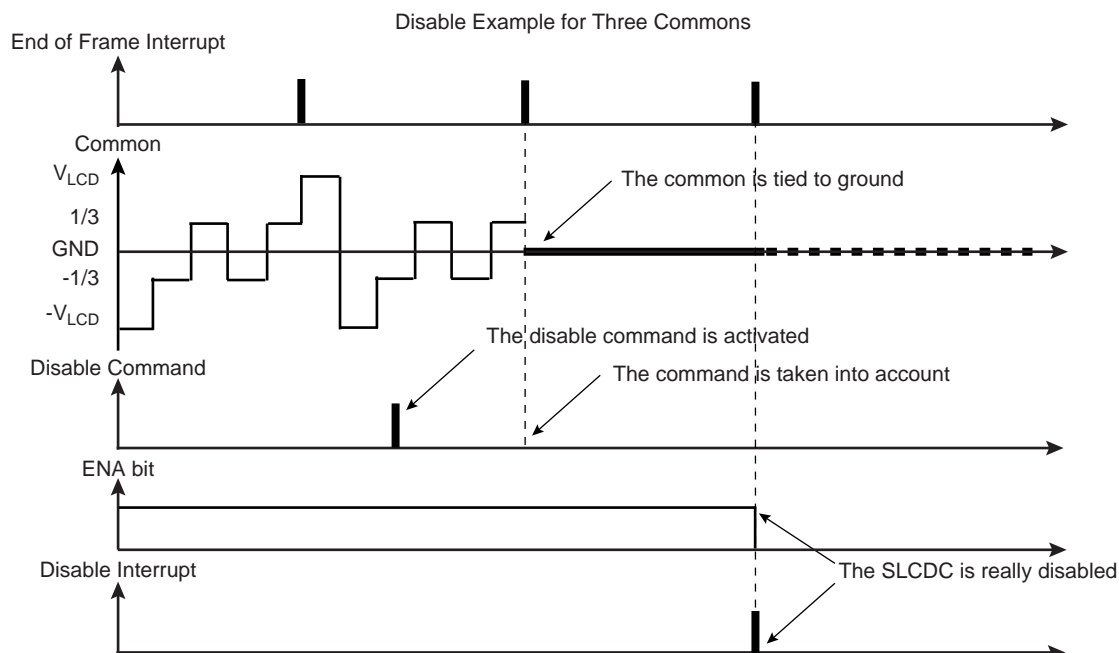
In both cases, no DC voltage should be left across any segment.

#### 39.6.7.1 Disable Bit

When the LCD Disable Command is activated during a frame, the next frame will be generated in "All Ground" Mode (whereby all commons and segments will be tied to ground).

At the end of this 'All Ground' frame, the disable bit is reset and the disable interrupt is asserted. This indicates that the SLCDC is really disabled and that the LCD can be switched off.

**Figure 39-10. Disabling Sequence**

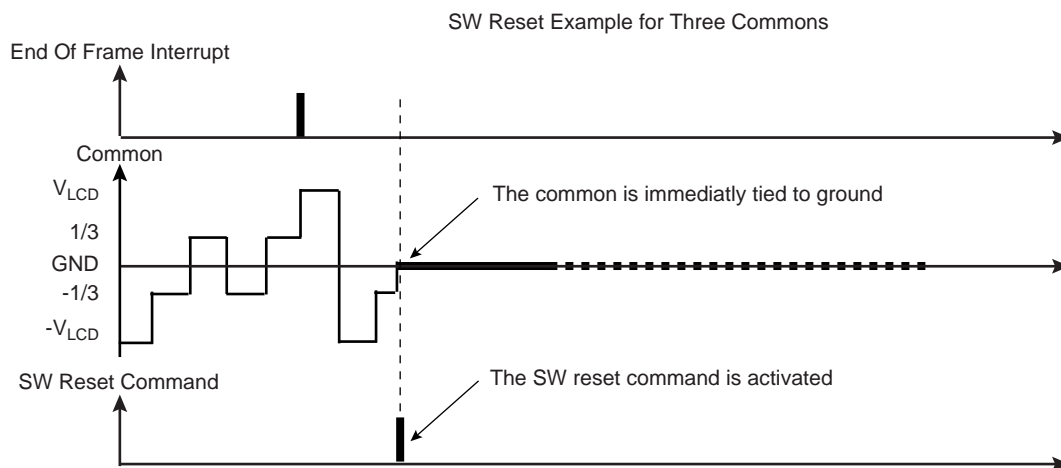


### 39.6.7.2 Software Reset

When the LCD software reset command is activated during a frame it is immediately taken into account and all commons and segments are tied to ground.

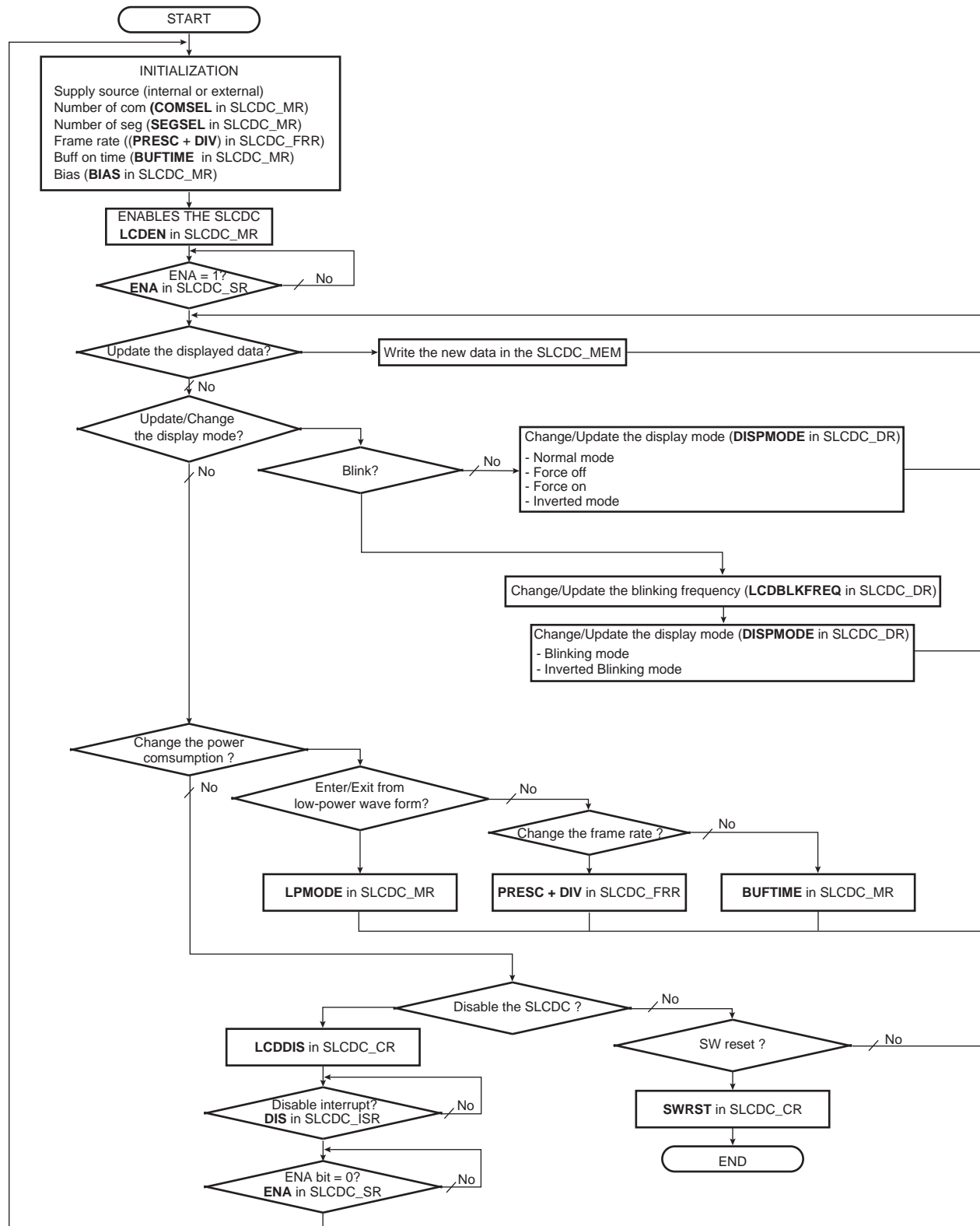
Note that in the case of a software reset, the disable interrupt is not asserted.

**Figure 39-11. Software Reset**



### 39.6.8 Flowchart

Figure 39-12.SLCDC Flow Chart



### 39.6.9 User Buffer Organization

The pixels to be displayed are written into SLCDC\_LMEMRx SLCDC\_MMEMRx registers. There are up to two 32-bit registers for each common terminal. The following table provides the address mapping of all commons/segments to be displayed.

If the segment map registers (SLCDC\_SMR0/1) are cleared and the number of segments to handle (SEGSEL bitfield in SLCDC\_MR) is lower or equal to 32, the registers SLCDC\_MMEMRx are not required to be programmed and can be left cleared (default value).

In case segments are remapped, the SLCDC\_MMEMRx registers are not required to be programmed if SLCDC\_SMR1 register is cleared (i.e. no segment remapped on SEG32 to SEG49 I/O pins). In this case SLCDC\_MMEMRx registers must be cleared.

In the same way if all segments are remapped on the upper part of the SEG terminals (SEG32 to SEG49) there is no need to program SLCDC\_LMEMRx registers (they must be cleared).

When segment remap is used (SLCDC\_SMR0/1 registers differ from 0), the unmapped segments must be kept cleared to limit internal signal switching.

		SEG0	--	SEG31	SEG32	--	SEG49	Memory address
SLCDC_MMEMR5	COM5				X	--	X	0x22C
SLCDC_LMEMR5	COM5	X	--	X				0x228
SLCDC_MMEMR4	COM4				X	--	X	0x224
SLCDC_LMEMR4	COM4	X	--	X				0x220
SLCDC_MMEMR3	COM3				X	--	X	0x21C
SLCDC_LMEMR3	COM3	X	--	X				0x218
SLCDC_MMEMR2	COM2				X	--	X	0x214
SLCDC_LMEMR2	COM2	X	--	X				0x210
SLCDC_MMEMR1	COM1				X	--	X	0x20C
SLCDC_LMEMR1	COM1	X	--	X				0x208
SLCDC_MMEMR0	COM0				X	--	X	0x204
SLCDC_LMEMR0	COM0	X	--	X				0x200

### 39.6.10 Segments Mapping Function

By default the segments pins (SEG0:49) are automatically assigned according to the SEGSEL configuration in the SLCDC\_MR register. The unused SEG I/O pins are forced to be driven by a digital peripheral or can be used as I/O through the PIO controller.

The automatic assignment is performed if the segment mapping function is not used (SLCDC\_SMR0/1 registers are cleared). The following table provides such assignments.

SEGSEL	I/O Port in Use as Segment Driver	I/O Port Pin if SLCDC_SMR0/1=0
0	SEG0	SEG1:49
1	SEG0:1	SEG2:49
...	...	...
48	SEG0:48	SEG49
49	SEG0:49	None

Programming is straightforward in this mode but it prevents flexibility of use of the digital peripheral multiplexed on SEG0:49 especially when the number of segments to drive is close to the maximum (50).

As example if the SEGSEL is set to 48, only digital peripheral associated to SEG49 can be used and none of the other digital peripherals multiplexed on SEG0:48 I/O can be used.

To offer a very flexible selection of digital peripherals multiplexed on SEG0:49 the user can manually configure the SEG I/O pins to be driven by the SLCDC.

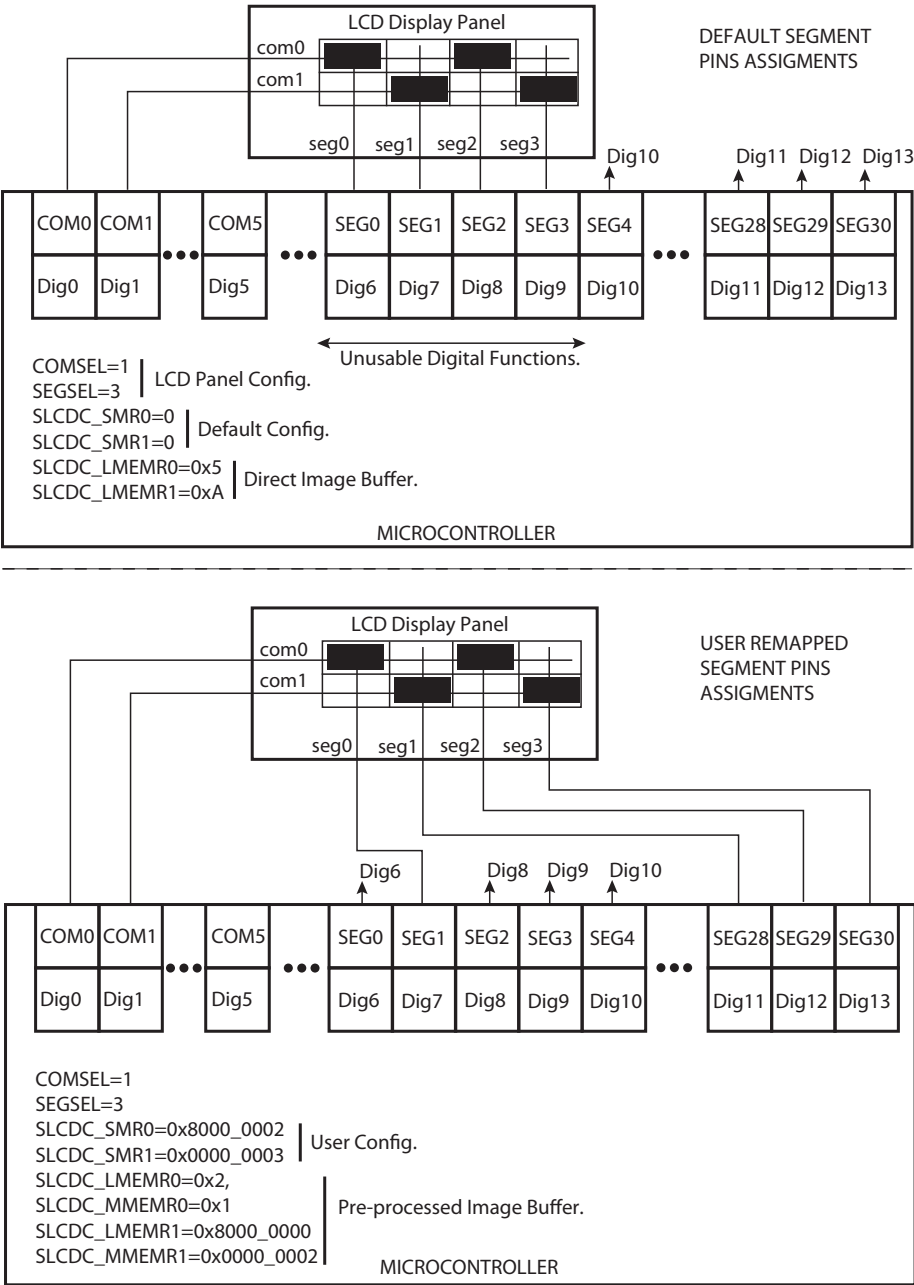
This is done by programming SLCDC\_SMR0/1 registers. As soon as they differ from 0 the segment remapping mode is used.

When configuring a logic 1 at index n (n = 0..49) into SLCDC\_SMR0/1 registers, the SLCDC forces the SEGn I/O pin to be driven by a segment waveform. In this mode the SEGSEL value configuration in SLCDC\_MR register is not taken into account.

In such case the SLCDC use the pixel value programmed at the same index in the SLCDC\_LMEMRx or SLCDC\_MMEMRx registers to generate the waveform carried on SEGn.

So, in remapping mode the software must carefully dispatch the pixels along SLCDC\_LMEMRx or SLCDC\_MMEMRx registers according to what is programmed into SLCDC\_SMR0/1 registers.

Figure 39-13.Segments Remapping Example



### 39.6.11 Write Protection Registers

To prevent any single software error that may corrupt SLCDC behavior, certain address spaces can be write-protected by setting the WPEN bit in the [“SLCDC Write Protect Mode Register”](#) (SLCDC\_WPMR) .

If a write access to the protected registers is detected, then the WPVS flag in the SLCDC Write Protect Status Register (SLCDC\_WPSR) is set and the field WPVSR indicates in which register the write access has been attempted.

The WPVS flag is automatically reset by reading the SLCDC Write Protect Status Register (SLCDC\_WPSR).

The protected registers are:

[“SLCDC Mode Register”](#) on page 874

[“SLCDC Frame Rate Register”](#) on page 876

[“SLCDC Display Register”](#) on page 877

[“SLCDC Segment Map Register 0”](#) on page 883

[“SLCDC Segment Map Register 1”](#) on page 884

## 39.7 Waveform specifications

### 39.7.1 DC characteristics

Refer to the DC Characteristics section of the product datasheet.

### 39.7.2 LCD Contrast

The peak value ( $V_{LCD}$ ) on the output waveform determines the LCD Contrast. VLCD is controlled by software in 16 steps from 2.4V to VDDIN.

This is a function of the supply controller.

## 39.8 Segment LCD Controller (SLCDC) User Interface

Table 39-6. Register Mapping

Offset	Register	Name	Access	Reset
0x0	SLCDC Control Register	SLCDC_CR	Write-only	-
0x4	SLCDC Mode Register	SLCDC_MR	Read-write	0x0
0x8	SLCDC Frame Rate Register	SLCDC_FRR	Read-write	0x0
0xC	SLCDC Display Register	SLCDC_DR	Read-write	0x0
0x10	SLCDC Status Register	SLCDC_SR	Read-only	0x0
0x20	SLCDC Interrupt Enable Register	SLCDC_IER	Write-only	-
0x24	SLCDC Interrupt Disable Register	SLCDC_IDR	Write-only	-
0x28	SLCDC Interrupt Mask Register	SLCDC_IMR	Write-only	-
0x2C	SLCDC Interrupt Status Register	SLCDC_ISR	Read-only	0x0
0x30	SLCDC Segment Map Register 0	SLCDC_SMR0	Read-write	0x0
0x34	SLCDC Segment Map Register 1	SLCDC_SMR1	Read-write	0x0
0x38-0xE4	Reserved	-	-	-
0xE4	SLCDC Write Protect Mode Register	SLCDC_WPMR	Read-write	0x0
0xE8	SLCDC Write Protect Status Register	SLCDC_WPSR	Read-only	0x0
0xEC-0xF8	Reserved	-	-	-
0xFC	Reserved	-	-	-
0x200 + com*0x8 + 0x0	SLCDC LSB Memory Register	SLCDC_LMEMR	Read-write	0x0
0x200 + com*0x8 + 0x4	SLCDC MSB Memory Register	SLCDC_MMEMR	Read-write	0x0



### 39.8.1 SLCDC Control Register

**Name:** SLCDC\_CR  
**Address:** 0x4003C000  
**Access:** Write-only  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	-
23	22	21	20	19	18	17	16
-	-	-	-	-	-	-	-
15	14	13	12	11	10	9	8
-	-	-	-	-	-	-	-
7	6	5	4	3	2	1	0
-	-	-	-	SWRST	-	LCDDIS	LCDEN

- **LCDEN: Enable the LCDC**

0 = No effect.

1 = The SLCDC is enabled

- **LCDDIS: Disable LCDC**

0 = No effect.

1 = The SLCDC is disabled.

Note: LCDDIS is taken into account at the beginning of the next frame.

- **SWRST: Software Reset**

0 = No effect.

1 = Equivalent to a power-up reset. When this command is performed, the SLCDC immediately ties all segments end commons lines to values corresponding to a “ground voltage”.

### 39.8.2 SLCDC Mode Register

**Name:** SLCDC\_MR  
**Address:** 0x4003C004  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
-	-	-	-	-	-	-	LPMODE
23	22	21	20	19	18	17	16
-	-	BIAS			BUFTIME		
15	14	13	12	11	10	9	8
-	-	SEGSEL					
7	6	5	4	3	2	1	0
-	-	-	-	-	COMSEL		

- **COMSEL: Selection of the Number of Commons**

(Taken into account only if the SLCDC is disabled)

Value	Name	Description
0x0	COM_0	COM0 is driven by SLCDC, COM1:5 are driven by digital function
0x1	COM_0TO1	COM0:1 are driven by SLCDC, COM2:5 are driven by digital function
0x2	COM_0TO2	COM0:2 are driven by SLCDC, COM3:5 are driven by digital function
0x3	COM_0TO3	COM0:3 are driven by SLCDC, COM4:5 are driven by digital function
0x4	COM_0TO4	COM0:4 are driven by SLCDC, COM5 is driven by digital function
0x5	COM_0TO5	COM0:5 are driven by SLCDC, No COM pin driven by digital function

- **SEGSEL: Selection of the Number of Segments**

(Taken into account only if the SLCDC is disabled)

SEGSEL must be programmed with the number of segments of the display panel minus 1.

If segment remapping function is not used (i.e. SLCDC\_SMRx equal 0) the SEGn [n = 0..49] I/O pins where n is greater than SEGSEL are forced to be driven by digital function. When segments remapping function is used, SEGn pins are driven by SLCDC only if corresponding PIXELn configuration bit is set in SLCDC\_SMR1/0 registers.

- **BUFTIME: Buffer On-Time**

(Taken into account from the next begin of frame)

Value	Name	Description
0x0	OFF	Nominal drive time is 0% of SCLK period
0x1	X2_SCLK_PERIOD	Nominal drive time is 2 periods of SCLK clock
0x2	X4_SCLK_PERIOD	Nominal drive time is 4 periods of SCLK clock
0x3	X8_SCLK_PERIOD	Nominal drive time is 8 periods of SCLK clock
0x4	X16_SCLK_PERIOD	Nominal drive time is 16 periods of SCLK clock
0x5	X32_SCLK_PERIOD	Nominal drive time is 32 periods of SCLK clock
0x6	X64_SCLK_PERIOD	Nominal drive time is 64 periods of SCLK clock

Value	Name	Description
0x7	X128_SCLK_PERIOD	Nominal drive time is 128 periods of SCLK clock
0x8	PERCENT_50	Nominal drive time is 50% of SCLK period
0x9	PERCENT_100	Nominal drive time is 100% of SCLK period

- **BIAS: LCD Display Configuration**

(Taken into account when the SLCDC is disabled.)

Value	Name	Description
0x0	STATIC	static
0x1	BIAS_1_2	bias 1/2
0x2	BIAS_1_3	bias 1/3

Note: BIAS is only taken into account when the SLCDC is disabled.

- **LPMODE: Low Power Mode** (Taken into account from the next begin of frame.)

0 = Normal Mode.

1 = Low Power Waveform is enabled.

### 39.8.3 SLCDC Frame Rate Register

**Name:** SLCDC\_FRR

**Address:** 0x4003C008

**Access:** Read-write

**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	DIV		
7	6	5	4	3	2	1	0
–	–	–	–	–	PRESC		

- PRESC: Clock Prescaler**

(Taken into account from the next begin of frame.)

Value	Name	Description
0x0	SCLK_DIV8	slow clock is divided by 8
0x1	SCLK_DIV16	slow clock is divided by 16
0x2	SCLK_DIV32	slow clock is divided by 32
0x3	SCLK_DIV64	slow clock is divided by 64
0x4	SCLK_DIV128	slow clock is divided by 128
0x5	SCLK_DIV256	slow clock is divided by 256
0x6	SCLK_DIV512	slow clock is divided by 512
0x7	SCLK_DIV1024	slow clock is divided by 1024

- DIV: Clock Division**

(Taken into account from the next begin of frame.)

Value	Name	Description
0x0	PRESC_CLK_DIV1	clock output from prescaler is divided by 1
0x1	PRESC_CLK_DIV2	clock output from prescaler is divided by 2
0x2	PRESC_CLK_DIV3	clock output from prescaler is divided by 3
0x3	PRESC_CLK_DIV4	clock output from prescaler is divided by 4
0x4	PRESC_CLK_DIV5	clock output from prescaler is divided by 5
0x5	PRESC_CLK_DIV6	clock output from prescaler is divided by 6
0x6	PRESC_CLK_DIV7	clock output from prescaler is divided by 7
0x7	PRESC_CLK_DIV8	clock output from prescaler is divided by 8

### 39.8.4 SLCDC Display Register

**Name:** SLCDC\_DR  
**Address:** 0x4003C00C  
**Access:** Read-write  
**Reset:** 0x00000000

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
LCDBLKREQ							
7	6	5	4	3	2	1	0
–	–	–	–	–	DISPMODE		

- **DISPMODE: Display Mode Register**

(Taken into account from the next begin of frame.)

Value	Name	Description
0x0	NORMAL	Normal Mode: Latched data are displayed.
0x1	FORCE_OFF	Force Off Mode: All pixels are invisible. (The SLCDC memory is unchanged.)
0x2	FORCE_ON	Force On Mode All pixels are visible. (The SLCDC memory is unchanged.)
0x3	BLINKING	Blinking Mode: All pixels are alternately turned off to the predefined state in SLCDC memory at LCDBLKREQ frequency. (The SLCDC memory is unchanged.)
0x4	INVERTED	Inverted Mode: All pixels are set in the inverted state as defined in SLCDC memory. (The SLCDC memory is unchanged.)
0x5	INVERTED_BLINK	Inverted Blinking Mode: All pixels are alternately turned off to the predefined opposite state in SLCDC memory at LCDBLKREQ frequency. (The SLCDC memory is unchanged.)
0x6	USER_BUFFER_LOAD	User Buffer Only Load Mode: Blocks the automatic transfer from User Buffer to Display Buffer.
0x7	BUFFERS_SWAP	Buffer Swap Mode: All pixels are alternatively assigned to the state defined in the User Buffer, then to the state defined in the Display Buffer at LCDBLKREQ frequency.

- **LCDBLKREQ: LCD Blinking Frequency Selection**

(Taken into account from the next begin of frame.)

Blinking frequency = Frame Frequency/LCDBLKREQ[7:0].

Note: 0 written in LCDBLKREQ stops blinking.

### 39.8.5 SLCDC Status Register

**Name:** SLCDC\_SR

**Address:** 0x4003C010

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
—	—	—	—	—	—	—	—
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	ENA

- **ENA: Enable Status** (Automatically Set/Reset)

0 = The SLCDC is disabled.

1 = The SLCDC is enabled.

### 39.8.6 SLCDC Interrupt Enable Register

**Name:** SLCDC\_IER

**Address:** 0x4003C020

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	DIS	–	ENDFRAME

- **ENDFRAME:** End of Frame Interrupt Enable

- **DIS:** Disable Interrupt Enable

0 = No effect.

1 = Enables the corresponding interrupt.

### 39.8.7 SLCDC Interrupt Disable Register

**Name:** SLCDC\_IDR

**Address:** 0x4003C024

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	DIS	–	ENDFRAME

- **ENDFRAME:** End of Frame Interrupt Disable

- **DIS:** Disable Interrupt Disable

0 = No effect.

1 = Disables the corresponding interrupt.



### 39.8.8 SLCDC Interrupt Mask Register

**Name:** SLCDC\_IMR

**Address:** 0x4003C028

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	DIS	–	ENDFRAME

- **ENDFRAME:** End of Frame Interrupt Mask

- **DIS:** Disable Interrupt Mask

0 = The corresponding interrupt is not enabled.

1 = The corresponding interrupt is enabled.

### 39.8.9 SLCDC Interrupt Status Register

**Name:** LCDC\_ISR  
**Address:** 0x4003C02C  
**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	DIS	–	ENDFRAME

- **ENDFRAME: End of Frame Interrupt Mask**

0 = No End of Frame occurred since the last read.

1 = End of Frame occurred since the last read.

- **DIS: Disable Interrupt Mask**

0 = The SLCDC is enabled.

1 = The SLCDC is disabled.

### 39.8.10 SLCDC Segment Map Register 0

**Name:** SLCDC\_SMR0

**Address:** 0x4003C030

**Access:** Read-write

31	30	29	28	27	26	25	24
LCD31	LCD30	LCD29	LCD28	LCD27	LCD26	LCD25	LCD24
23	22	21	20	19	18	17	16
LCD23	LCD22	LCD21	LCD20	LCD19	LCD18	LCD17	LCD16
15	14	13	12	11	10	9	8
LCD15	LCD14	LCD13	LCD12	LCD11	LCD10	LCD9	LCD8
7	6	5	4	3	2	1	0
LCD7	LCD6	LCD5	LCD4	LCD3	LCD2	LCD1	LCD0

- **LCDx: LCD Segment Mapped on SEGx I/O pin**

(Can be programmed only if SLCDC is disabled)

0 = The corresponding I/O pin will be driven either by SLCDC or digital function according to SEGSEL bitfield configuration in the SLCDC\_MR register.

1 = A LCD segment will be driven on corresponding I/O pin.

### 39.8.11 SLCDC Segment Map Register 1

**Name:** SLCDC\_SMR1

**Address:** 0x4003C034

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	LCD49	LCD48
15	14	13	12	11	10	9	8
LCD47	LCD46	LCD45	LCD44	LCD43	LCD42	LCD41	LCD40
7	6	5	4	3	2	1	0
LCD39	LCD38	LCD37	LCD36	LCD35	LCD34	LCD33	LCD32

- **LCDx: LCD Segment Mapped on SEGx I/O pin**

(Can be programmed only if SLCDC is disabled)

0 = The corresponding I/O pin will be driven either by SLCDC or digital function according to SEGSEL bitfield configuration in the SLCDC\_MR register.

1 = A LCD segment will be driven on corresponding I/O pin.

### 39.8.12 SLCDC Write Protect Mode Register

**Name:** SLCDC\_WPMR

**Address:** 0x4003C0E4

**Access:** Read-write

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

- **WPEN: Write Protect Enable**

0 = Disables the Write Protect if WPKEY corresponds to 0x4C4344 ("LCD" in ASCII).

1 = Enables the Write Protect if WPKEY corresponds to 0x4C4344 ("LCD" in ASCII).

Protects the registers:

- **WPKEY: Write Protect KEY**

Should be written at value 0x4C4344 ("LCD" in ASCII). Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0.

Protects the registers:

["SLCDC Mode Register" on page 874](#)

["SLCDC Frame Rate Register" on page 876](#)

["SLCDC Display Register" on page 877](#)

["SLCDC Segment Map Register 0" on page 883](#)

["SLCDC Segment Map Register 1" on page 884](#)

### 39.8.13 SLCDC Write Protect Status Register

**Name:** SLCDC\_WPSR

**Address:** 0x4003C0E8

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

- **WPVS: Write Protect Violation Status**

0 = No Write Protect Violation has occurred since the last read of the SLCDC\_WPSR register.

1 = A Write Protect Violation has occurred since the last read of the SLCDC\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

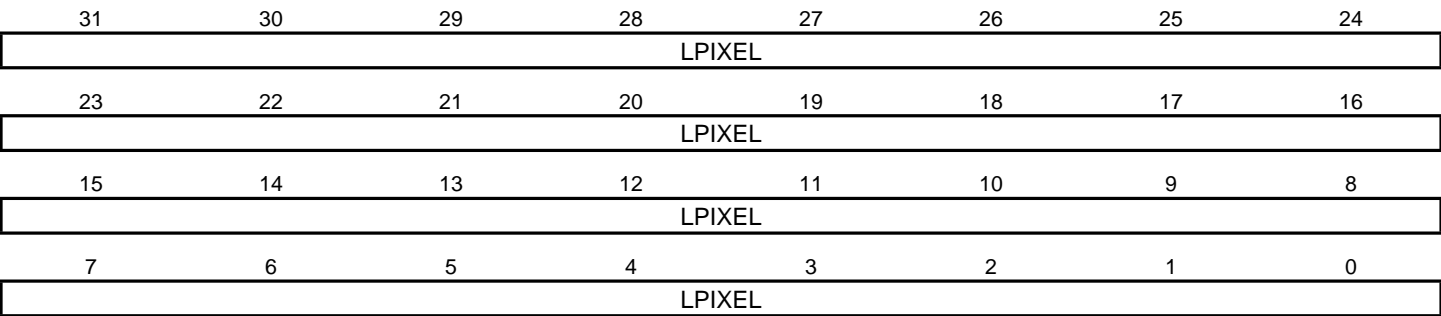
- **WPVSR: Write Protect Violation Source**

When WPVS is active, this field indicates the write-protected register (through address offset or code) in which a write access has been attempted.

Reading SLCDC\_WPSR automatically clears all fields.

39.8.14 SLCDC LSB Memory Register

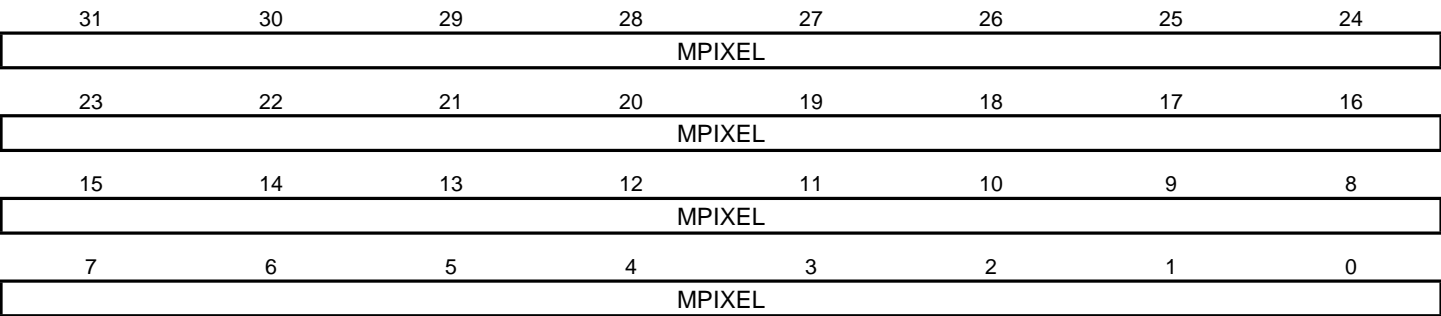
**Name:** SLCDC\_LMEMRx [x = 0..5]  
**Address:** 0x4003C200 [0], 0x4003C208 [1], 0x4003C210 [2], 0x4003C218 [3], 0x4003C220 [4], 0x4003C228 [5]  
**Access:** Read-write



- **LPIXEL : LSB Pixels pattern associated to COMx terminal**  
0 = The pixel associated to COMx terminal is not visible (if non inverted display mode is used).  
1 = The pixel associated to COMx terminal is visible (if non inverted display mode is used).
- Note: LPIXEL[n] (n = 0..31) drives SEGn terminal.

39.8.15 SLCDC MSB Memory Register

**Name:** SLCDC\_MMEMRx [x = 0..5]  
**Address:** 0x4003C204 [0], 0x4003C20C [1], 0x4003C214 [2], 0x4003C21C [3], 0x4003C224 [4], 0x4003C22C [5]  
**Access Type:** Read-write



- MPIXEL : MSB Pixels pattern associated to COMx terminal**  
0 = The pixel associated to COMx terminal is not visible (if non inverted display mode is used).  
1 = The pixel associated to COMx terminal is visible (if non inverted display mode is used).
- Note: MPIXEL[n] (n = 32..49) drives SEGn terminal.



## 40. Analog-to-Digital Converter (ADC)

### 40.1 Description

The ADC is based on a 10-bit Analog-to-Digital Converter (ADC) managed by an ADC Controller. Refer to [Figure 40-1, "Analog-to-Digital Converter Block Diagram"](#). It also integrates an 8-to-1 analog multiplexer, making possible the analog-to-digital conversions of 8 analog lines. The conversions extend from 0V to the voltage carried on pin ADVREF or the voltage provided by the internal reference voltage which can be programmed in the Analog Control register (ADC\_ACR). The selection of reference voltage source is defined by ONREF and FORCEREF bits in the Mode register (ADC\_MR).

The ADC supports the 8-bit or 10-bit resolution mode. The 8-bit resolution mode prevents using the 16-bit peripheral DMA transfer into memory when only 8-bit resolution is required by the application. Note that using this low resolution mode does not increase the conversion rate.

Conversion results are reported in a common register for all channels, as well as in a channel-dedicated register.

The 11-bit and 12-bit resolution modes are obtained by averaging multiple samples to decrease quantization noise. For 11-bit mode, four samples are used, which gives an effective sample rate of 1/4 of the actual sample frequency. For 12-bit mode, 16 samples are used, giving an effective sample rate of 1/16th of the actual sample frequency. This allows conversion speed to be traded for better accuracy.

The last channel is internally connected to a temperature sensor. The processing of this channel can be fully configured for efficient downstream processing due to the slow frequency variation of the value carried on such a sensor. The seventh channel is reserved for measurement of VDDBU voltage.

The software trigger, the external trigger on rising edge of the ADTRG pin or internal triggers from Timer Counter output(s) are configurable.

The main comparison circuitry allows automatic detection of values below a threshold, higher than a threshold, in a given range or outside the range. Thresholds and ranges are fully configurable.

The ADC also integrates a sleep mode and a conversion sequencer, and connects with a PDC channel. These features reduce both power consumption and processor intervention.

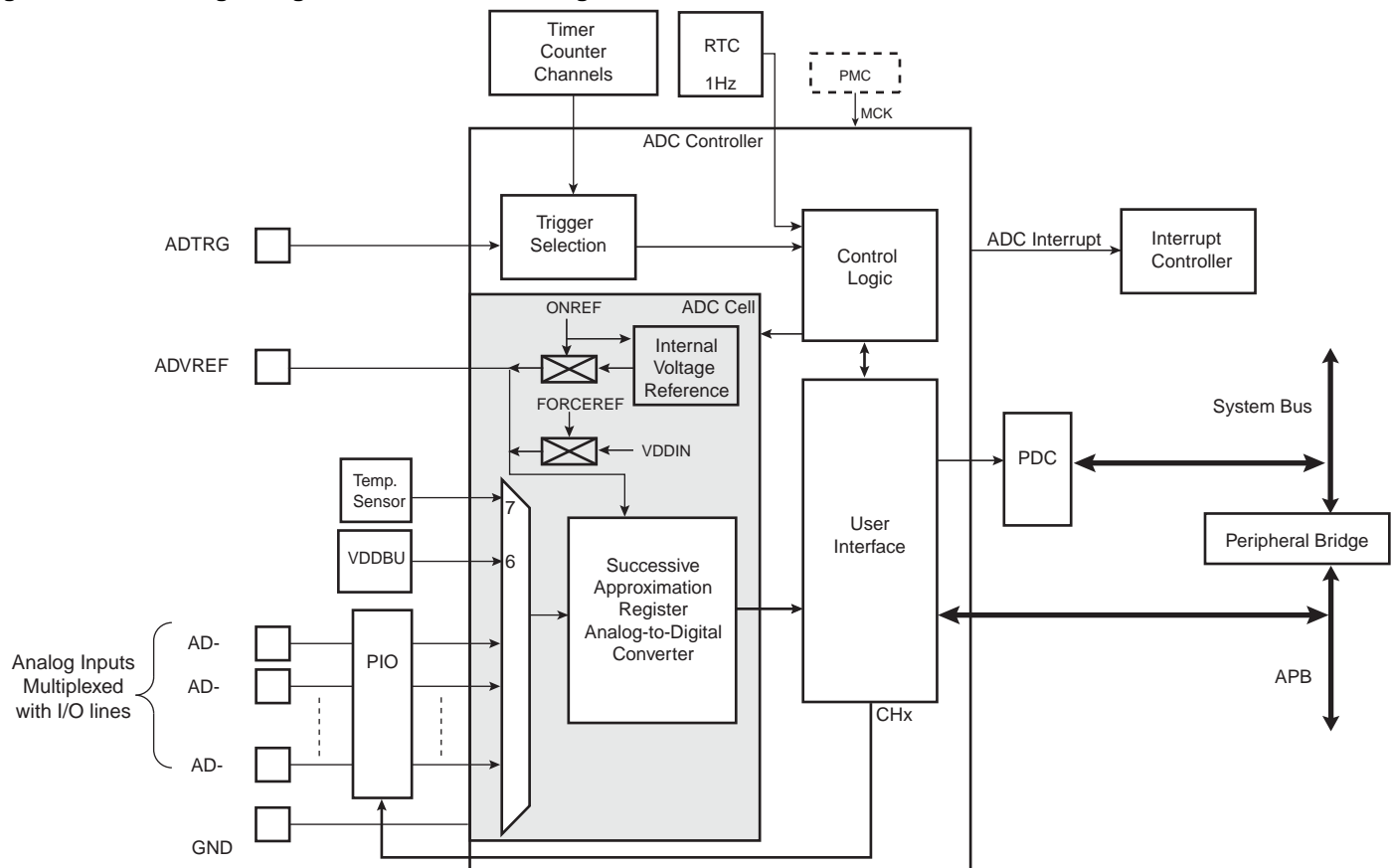
Finally, the user can configure ADC timings, such as startup time and tracking time.

## 40.2 Embedded Characteristics

- 10-bit Resolution with Enhanced Mode up to 12 Bits
- 500 KHz Conversion Rate
- Digital Averaging Function Provides Enhanced Resolution Mode up to 12 Bits
- On-chip Temperature Sensor Management
- Wide Range of Power Supply Operation
- Selectable External Voltage Reference or Programmable Internal Reference
- Integrated Multiplexer Offering Up to 8 Independent Analog Inputs
- Individual Enable and Disable of Each Channel
- Hardware or Software Trigger
  - External Trigger Pin
  - Timer Counter Outputs (Corresponding TIOA Trigger)
- PDC Support
- Possibility of ADC Timings Configuration
- Two Sleep Modes and Conversion Sequencer
  - Automatic Wakeup on Trigger and Back to Sleep Mode after Conversions of all Enabled Channels
  - Possibility of Customized Channel Sequence
- Standby Mode for Fast Wakeup Time Response
  - Power Down Capability
- Automatic Window Comparison of Converted Values
- Register Write Protection

### 40.3 Block Diagram

**Figure 40-1. Analog-to-Digital Converter Block Diagram**



## 40.4 Signal Description

### Table 40-1. ADC Pin Description

Pin Name	Description
ADVREF	External reference voltage
AD0 - AD7 <sup>(1)(2)</sup>	Analog input channels
ADTRG	External trigger

Note:

1. AD7 is not an actual pin; it is internally connected to a temperature sensor.
2. AD6 is not an actual pin; it is internally connected to VDDBU.

## 40.5 Product Dependencies

### 40.5.1 Power Management

The ADC Controller is not continuously clocked. The programmer must first enable the ADC Controller MCK in the Power Management Controller (PMC) before using the ADC Controller. However, if the application does not require ADC operations, the ADC Controller clock can be stopped when not needed and restarted when necessary. Configuring the ADC Controller does not require the ADC Controller clock to be enabled.

### 40.5.2 Interrupt Sources

The ADC interrupt line is connected on one of the internal sources of the Interrupt Controller. Using the ADC interrupt requires the interrupt controller to be programmed first.

### 40.5.3 Analog Inputs

The analog input pins can be multiplexed with PIO lines. In this case, the assignment of the ADC input is automatically done as soon as the corresponding channel is enabled by writing the Channel Enable register (ADC\_CHER). By default, after reset, the PIO line is configured as a digital input with its pull-up enabled, and the ADC input is connected to the GND.

### 40.5.4 Temperature Sensor

The temperature sensor is internally connected to channel index 7 of the ADC.

The temperature sensor provides an output voltage  $V_T$  that is proportional to the absolute temperature (PTAT). To activate the temperature sensor, the TEMPON bit in the Temperature Sensor Mode register (ADC\_TEMPMR) must be set. After setting the bit, the startup time of the temperature sensor must be achieved prior to initiating any measurement.

### 40.5.5 I/O Lines

The pin ADTRG may be shared with other peripheral functions through the PIO Controller. In this case, the PIO Controller should be set accordingly to assign the pin ADTRG to the ADC function.

### 40.5.6 Timer Triggers

Timer Counters may or may not be used as hardware triggers depending on user requirements. Thus, some or all of the timer counters may be unconnected.

### 40.5.7 Conversion Performances

For performance and electrical characteristics of the ADC, see the Electrical Characteristics.

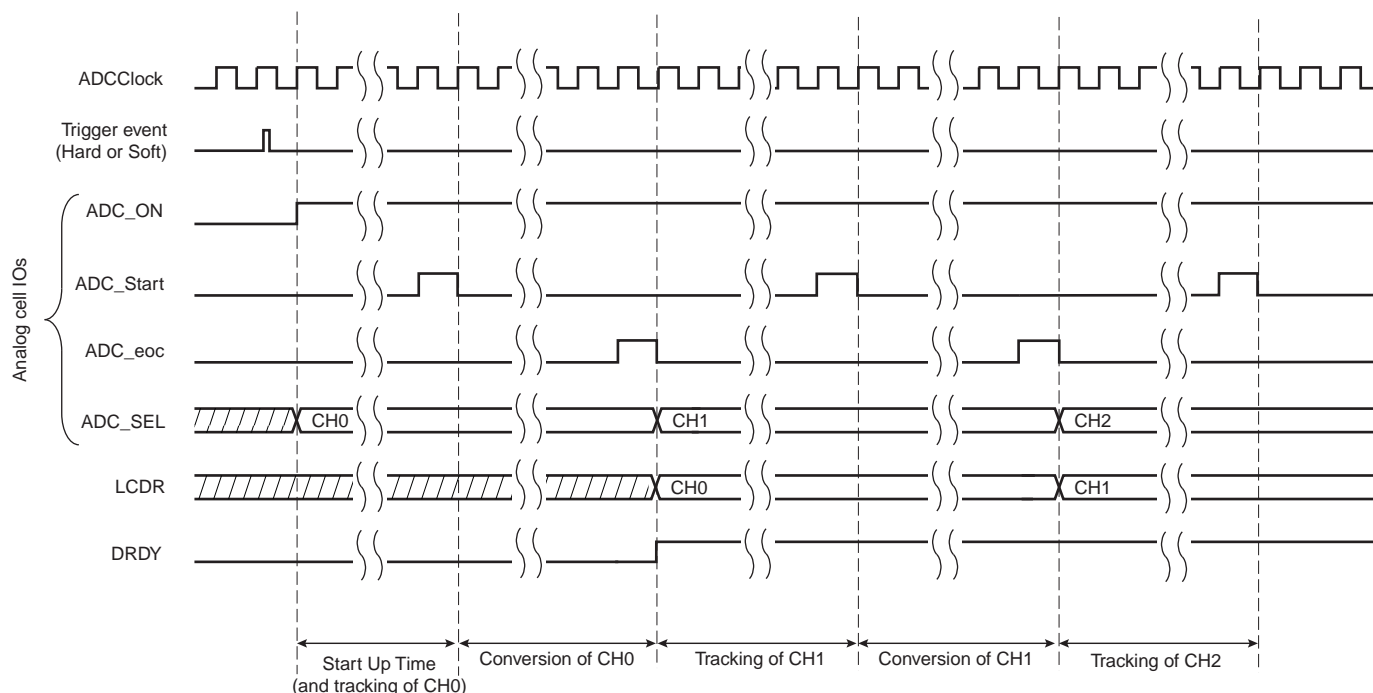
## 40.6 Functional Description

### 40.6.1 Analog-to-digital Conversion

The ADC uses the ADC clock to perform conversions. Converting a single analog value to a 10-bit digital data requires tracking clock cycles as defined in the field TRACKTIM of the [“ADC Mode Register” on page 907](#). The ADC clock frequency is selected in the PRESCAL field of ADC\_MR.

The ADC clock range is between  $MCK/2$ , if PRESCAL is 0, and  $MCK/512$ , if PRESCAL is set to 255 (0xFF). PRESCAL must be programmed in order to provide an ADC clock frequency according to the parameters given in the Electrical Characteristics section.

**Figure 40-2. Sequence of ADC Conversions**



#### 40.6.2 Conversion Reference

The conversion is performed on a full range between 0V and the reference voltage. The reference voltage is defined by the external pin ADVREF, or programmed using the internal reference voltage that is configured in ADC\_ACR. Analog inputs between these voltages convert to values based on a linear conversion.

#### 40.6.3 Conversion Resolution

The ADC supports 8-bit or 10-bit resolutions. The 8-bit selection is performed by setting the LOWRES bit in ADC\_MR. By default, after a reset, the resolution is the highest and the DATA field in the data registers is fully used. By setting the LOWRES bit, the ADC switches to the lowest resolution and the conversion results can be read in the lowest significant bits of the data registers. The two highest bits of the DATA field in the corresponding Channel Data register (ADC\_CDR) and of the LDATA field in the Last Converted Data register (ADC\_LCDR) read 0.

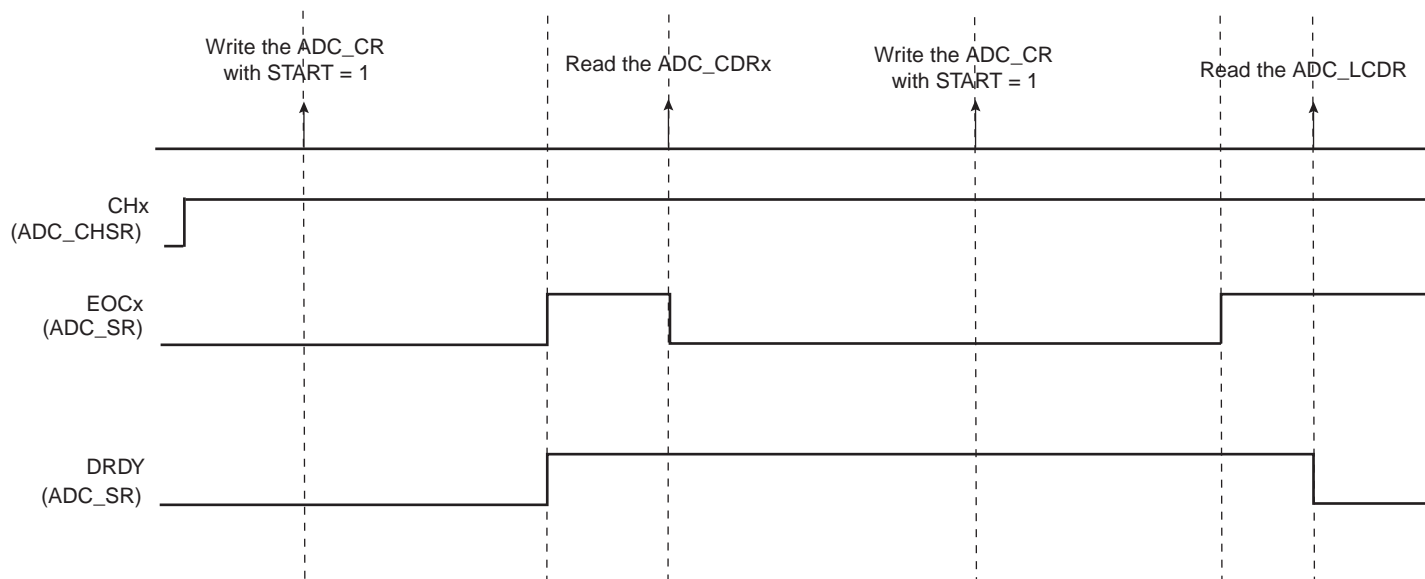
#### 40.6.4 Conversion Results

When a conversion is completed, the resulting 10-bit digital value is stored in ADC\_CDRx of the current channel and in ADC\_LCDR. By setting the TAG option in the Extended Mode register (ADC\_EMR), ADC\_LCDR presents the channel number associated with the last converted data in the CHNB field.

The EOCx bit and DRDY in the Interrupt Status register (ADC\_ISR) are set. In the case of a connected PDC channel, DRDY rising triggers a data request. In any case, both EOC and DRDY can trigger an interrupt.

Reading one ADC\_CDR clears the corresponding EOC bit. Reading ADC\_LCDR clears the DRDY bit.

**Figure 40-3. EOCx and DRDY Flag Behavior**

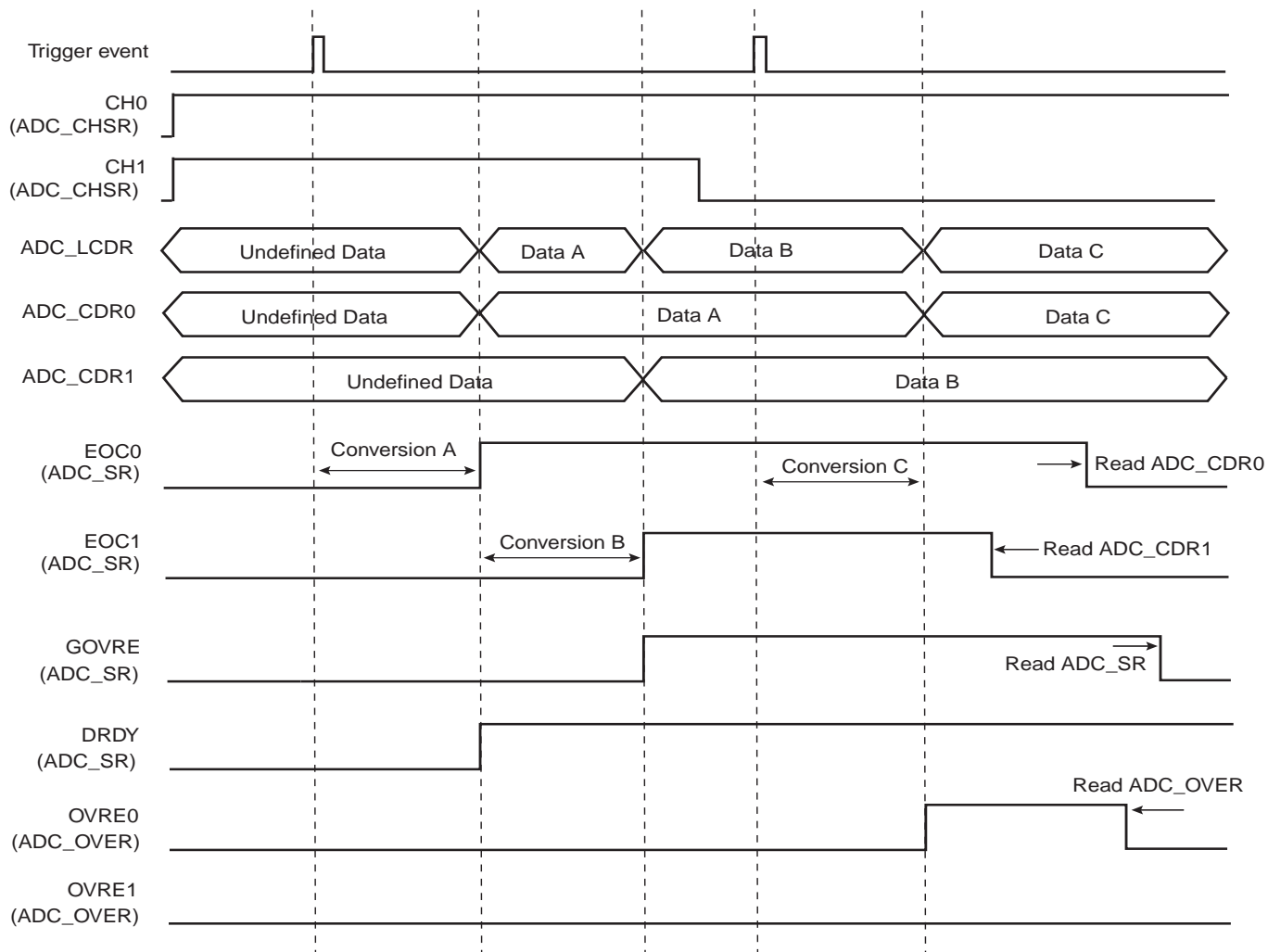


If ADC\_CDR is not read before further incoming data is converted, the corresponding Overrun Error (OVREx) flag is set in the Overrun Status register (ADC\_OVER).

Likewise, new data converted when DRDY is high sets the GOVRE bit (General Overrun Error) in ADC\_ISR.

The OVREx flag is automatically cleared when ADC\_OVER is read, and GOVRE flag is automatically cleared when ADC\_ISR is read.

**Figure 40-4. GOVRE and OVREx Flag Behavior**



**Warning:** If the corresponding channel is disabled during a conversion or if it is disabled and then reenabled during a conversion, its associated data and its corresponding EOC and OVRE flags in ADC\_ISR are unpredictable.

### 40.6.5 Conversion Triggers

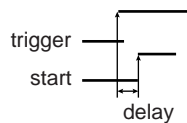
Conversions of the active analog channels are started with a software or hardware trigger. The software trigger is provided by writing the Control register (ADC\_CR) with the START bit at 1.

The hardware trigger can be one of the TIOA outputs of the Timer Counter channels or the external trigger input of the ADC (ADTRG). The hardware trigger is selected with the TRGSEL field in ADC\_MR. The selected hardware trigger is enabled with the TRGEN bit in ADC\_MR.

The minimum time between two consecutive trigger events must be strictly greater than the duration time of the longest conversion sequence as configured in the Mode register (ADC\_MR), the Channel Status register (ADC\_CHSR) and the Channel Sequence 1 register (ADC\_SEQR1).

If a hardware trigger is selected, the start of a conversion is triggered after a delay which starts at each rising edge of the selected signal. Due to asynchronous handling, the delay may vary in a range of 2 MCK clock periods to 1 ADC clock period.

**Figure 40-5. Hardware Trigger Delay**



If one of the TIOA outputs is selected, the corresponding Timer Counter channel must be programmed in waveform mode.

Only one start command is necessary to initiate a conversion sequence on all the channels. The ADC hardware logic automatically performs the conversions on the active channels, then waits for a new request. The Channel Enable (ADC\_CHER) and Channel Disable (ADC\_CHDR) registers permit the analog channels to be enabled or disabled independently.

If the ADC is used with a PDC, only the transfers of converted data from enabled channels are performed and the resulting data buffers should be interpreted accordingly.

### 40.6.6 Sleep Mode and Conversion Sequencer

The ADC sleep mode maximizes power saving by automatically deactivating the ADC when it is not being used for conversions. Sleep mode is selected by setting the SLEEP bit in ADC\_MR.

The Sleep mode is managed by a conversion sequencer, which automatically processes the conversions of all channels at lowest power consumption.

This mode can be used when the minimum period of time between two successive trigger events is greater than the startup period of the ADC (see the ADC Characteristics section).

When a start conversion request occurs, the ADC is automatically activated. As the analog cell requires a start-up time, the logic waits during this time and starts the conversion on the enabled channels. When all conversions are complete, the ADC is deactivated until the next trigger. Triggers occurring during the sequence are not taken into account.

The conversion sequencer allows automatic processing with minimum processor intervention and optimized power consumption. Conversion sequences can be performed periodically using a Timer/Counter output. By using the PDC, the periodic acquisition of several samples can be processed automatically without any intervention of the processor.

The sequence can be customized by programming ADC\_SEQR1 and setting to 1 the USEQ bit of ADC\_MR. The user can choose a specific order of channels and can program up to 8 conversions by sequence. The user is totally free to create his own sequence, by writing channel numbers in ADC\_SEQR1. Not only can channel numbers be written in any sequence, they can be repeated several times. Only enabled sequence bits are converted.

If all ADC channels (i.e. 8) are used on an application board, there is no restrictions in the use of the user sequence. However, if some ADC channels are not enabled for conversion but rather used as pure digital inputs, the respective indexes of these channels cannot be used in the user sequence fields in ADC\_SEQR1. For example, if channel 3 is



disabled (`ADC_CSR[3] = 0`), `ADC_SEQR1` fields `USCH1` up to `USCH8` must not contain the value 3. Thus the length of the user sequence may be limited by this behavior.

As an example, if only four channels over 8 (`CH0` up to `CH3`) are selected for ADC conversions, the user sequence length cannot exceed four channels. Each trigger event may launch up to four successive conversions of any combination of channels 0 up to 3 but no more (i.e., in this case the sequence `CH0, CH0, CH1, CH1, CH1` is impossible).

A sequence that repeats the same channel several times requires more enabled channels than channels actually used for conversion. For example, the sequence `CH0, CH0, CH1, CH1` requires four enabled channels (four free channels on application boards) whereas only `CH0, CH1` are really converted.

#### 40.6.7 Comparison Window

The ADC Controller features automatic comparison functions. It compares converted values to a low threshold, a high threshold or both, depending on the `CMPMODE` function chosen in the Extended Mode register (`ADC_EMR`). The comparison can be done on all channels or only on the channel specified in `CMPSEL` field of `ADC_EMR`. To compare all channels, the `CMPALL` bit of `ADC_EMR` should be set.

Moreover, a filtering option can be set by writing the number of consecutive comparison errors needed to raise the flag. This number can be written and read in the `CMPFILTER` field of `ADC_EMR`.

The flag can be read on the `COMPE` bit of `ADC_ISR` and can trigger an interrupt.

The high threshold and the low threshold can be read/write in the Compare Window register (`ADC_CWR`).

If the comparison window is to be used with the `LOWRES` bit in `ADC_MR` set to 1, the thresholds do not need to be adjusted as adjustment will be done internally. Whether or not the `LOWRES` bit is set, thresholds must always be configured in consideration of the maximum ADC resolution.

#### 40.6.8 ADC Timings

Each ADC has its own minimal startup time that is programmed through the field `STARTUP` in `ADC_MR`.

A minimal tracking time is necessary for the ADC to guarantee the best converted final value between two channel selections. This time has to be programmed through the `TRACKTIM` field in `ADC_MR`.

**Warning:** No input buffer amplifier to isolate the source is included in the ADC. This must be taken into consideration to program a precise value in the `TRACKTIM` field. See the ADC Characteristics section.

#### 40.6.9 Temperature Sensor

The temperature sensor is internally connected to channel index 7. For temperature measurement, the `TEMPON` bit must be written to 1 in `ADC_TEMPMR`.

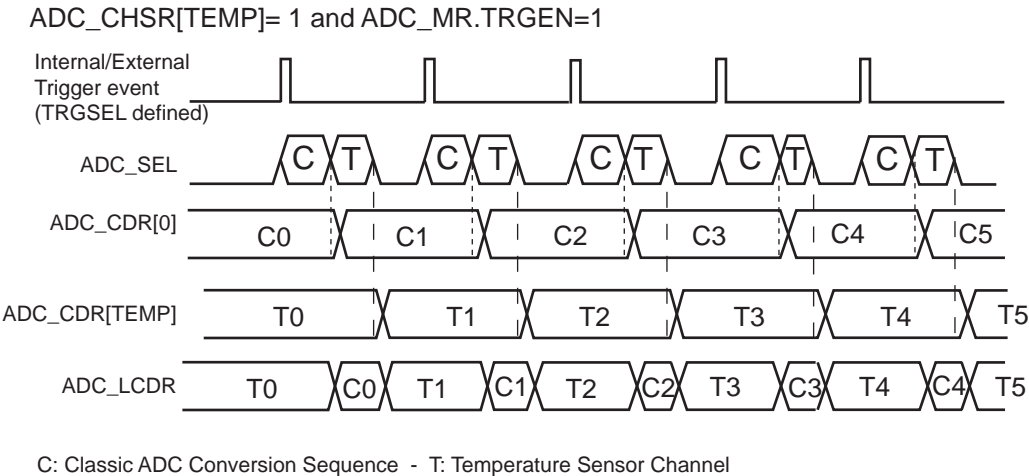
The temperature measurement can be made in different ways through the ADC Controller. The different methods of measurement depend on the configuration bits `TRGEN` in `ADC_MR` and `CH7` in `ADC_CHSR`.

The temperature measurement can be triggered like the other channels by enabling its associated conversion channel index 7, writing to 1 the bit `CH7` of `ADC_CHER`.

The manual start can only be performed if `TRGEN` bit in `ADC_MR` is disabled. When the `START` bit in `ADC_CR` is written to 1, the temperature sensor channel conversion will be scheduled together with the other enabled channels (if any). The result of the conversion is placed in the `ADC_CDR7` register and the associated flag `EOC7` is set in `ADC_ISR`.

If the `TRGEN` bit is set in `ADC_MR`, the channel of the temperature sensor is periodically converted together with other enabled channels and the result is placed on `ADC_LCDR` and `ADC_CDR7`. Thus the temperature conversion result is part of the Peripheral DMA Controller buffer. The temperature channel can be enabled/disabled at anytime but this may not be optimal for downstream processing.

Figure 40-6. Non-optimized Temperature Conversion



Assuming ADC\_CHSR[0] = 1 and ADC\_CHSR[TEMP] = 1  
where TEMP is the index of the temperature sensor channel

trig.event1→						
	DMA Buffer	0	ADC_CDR[0]		DMA Transfer	
	Structure	0	ADC_CDR[TEMP]		Base Address (BA)	
trig.event2→		0	ADC_CDR[0]		BA + 0x02	
		0	ADC_CDR[TEMP]		BA + 0x04	
trig.event3→		0	ADC_CDR[0]		BA + 0x06	
		0	ADC_CDR[TEMP]		BA + 0x08	
		0	ADC_CDR[0]		BA + 0x0A	

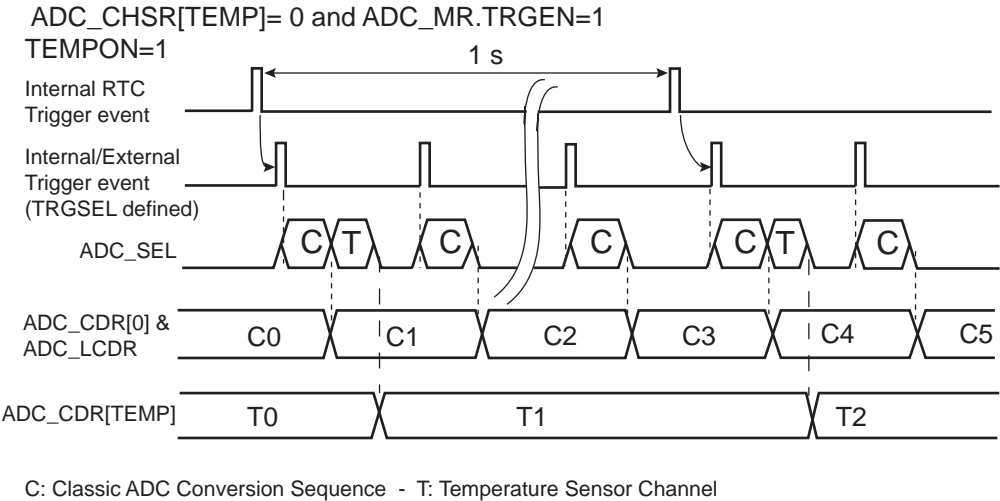
The temperature factor has a slow variation rate and is potentially different from other conversion channels. As a result, the ADC Controller triggers the measurement differently when TEMPON is set in ADC\_TEMPMR but CH7 is not set in the ADC\_CHSR.

Under these conditions, the measurement is triggered every second by means of an internal trigger generated by RTC, always enabled and totally independent of the internal/external triggers. The RTC event will be taken into account on the next internal/external trigger event as described in [Figure 40-7, "Optimized Temperature Conversion Combined With Classical Conversions"](#). The internal/external trigger is selected through the TRGSEL field of ADC\_MR.

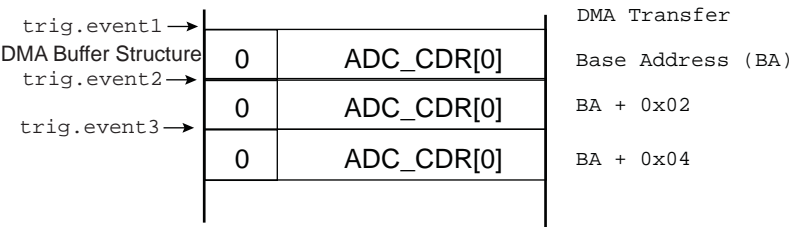
In this mode of operation, the temperature sensor is only powered for a period of time covering the startup time and conversion time (refer to [Figure 40-8, "Temperature Conversion Only"](#) for more details).

Every second, a conversion is scheduled for channel 7 but the result of the conversion is only uploaded in ADC\_CDR7 and not in ADC\_LCDR . Therefore there is no change in the structure of the Peripheral DMA Controller buffer due to the conversion of the temperature channel; only the enabled channels are kept in the buffer. The end of conversion of the temperature channel is reported by means of EOC7 flag in ADC\_ISR.

Figure 40-7. Optimized Temperature Conversion Combined With Classical Conversions

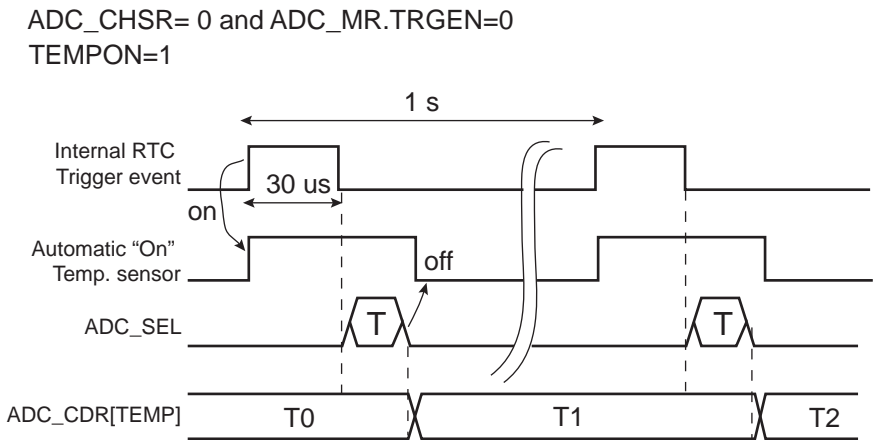


Assuming ADC\_CHSR[0] = 1



If TEMPON=1, TRGEN is disabled and none of the channels are enabled in ADC\_CHSR (ADC\_CHSR=0), then only channel 7 is converted at a rate of one conversion per second (see [Figure 40-8, "Temperature Conversion Only"](#)). This mode of operation, when combined with the sleep mode operation of the ADC Controller, provides a low-power mode for temperature measurement. This assumes there is no other ADC conversion to schedule at a high sampling rate or simply no other channel to convert.

Figure 40-8. Temperature Conversion Only



Moreover, it is possible to raise a flag only if there is a predefined change in the temperature. The user can define a range of temperature or a threshold in the Temperature Compare Window register (ADC\_TEMPCWR), and the mode of comparison that can be programmed into ADC\_TEMPMPR by means of the TEMPCMPMOD field. These values define the way the TEMPCHG flag is raised in ADC\_ISR.

The TEMPCHG flag can be used to generate a temperature-dependent interrupt instead of the end-of-conversion interrupt. In particular, the interrupt is generated only if the temperature sensor, as measured by the ADC, reports a temperature value below, above, inside or outside programmable thresholds (see ADC\_TEMPMPR).

In any case, if TEMPON is set, the temperature can be read at anytime in ADC\_CDR7 without any specific software intervention.

#### 40.6.10 VDDBU Measurement

The seventh ADC channel (CH6) of the ADC Controller is reserved for measurement of the VDDBU power supply pin. For this channel, setting up, starting conversion, and other tasks must be performed the same way as for all other channels. VDDBU is measured without any attenuation. This means that for VDDBU greater than the voltage reference applied to the ADC, the digital output clamps to the maximum value.

#### 40.6.11 Enhanced Resolution Mode and Digital Averaging Function

The enhanced resolution mode is enabled if LOWRES is cleared in ADC\_MR, and the OSR field is set to 1, 2 in ADC\_EMR. The enhancement is based on a digital averaging function.

FREERUN must be set to 0 when digital averaging is used (OSR differs from 0 in ADC\_EMR).

There is no averaging on the last index channel if the measure is triggered by an RTC event (see [Section 40.6.9 "Temperature Sensor"](#)).

In enhanced resolution mode, the ADC Controller trades conversion speed for quantization noise by averaging multiple samples, thus providing a digital low-pass filter function.

If 1-bit enhancement resolution is selected (OSR=1 in ADC\_EMR), the ADC effective sample rate is the maximum ADC sample rate divided by 4; therefore, the oversampling ratio is 4.

When the 2-bit enhancement resolution is selected (OSR=2 in ADC\_EMR), the ADC effective sample rate is the maximum ADC sample rate divided by 16 (oversampling ratio is 16).

The selected oversampling ratio applies to all enabled channels except for the temperature sensor channel when triggered by an RTC event.

The average result is valid into the ADC\_CDRx register (x corresponding to the index of the channel) only if EOCn flag is set in ADC\_ISR and OVREN flag is cleared in ADC\_OVER. The average result for all channels is valid in ADC\_LCDR only if DRDY is set and GOVRE is cleared in ADC\_ISR.

Note that ADC\_CDRx are not buffered. Therefore, when an averaging sequence is ongoing, the value in these registers changes after each averaging sample. On the other hand, overrun flags in ADC\_OVER rise as soon as the first sample of an averaging sequence is received and thus the previous averaged value is not read (even if the new averaged value is not ready).

In consequence, when an overrun flag rises in ADC\_OVER, the previous unread data is lost but the data has not been overwritten by the new averaged value, as the averaging sequence concerning this channel may still be on-going.

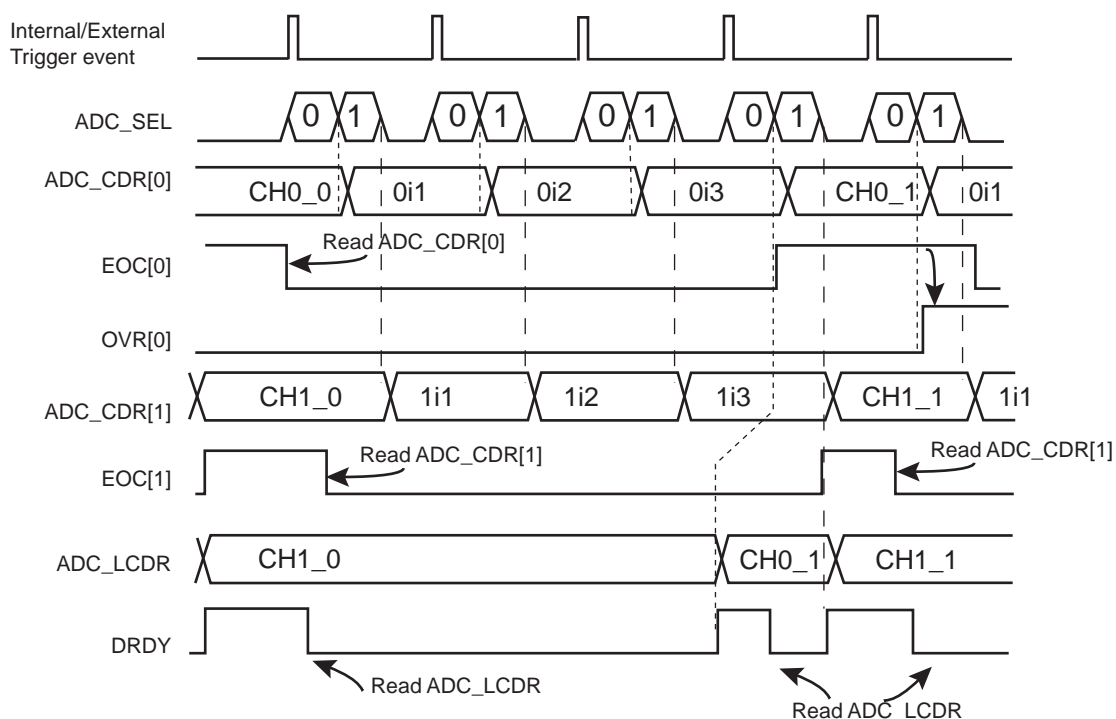
##### 40.6.11.1 Averaging Function versus Trigger Events

The samples can be defined in different ways for the averaging function depending on the configuration of the ASTE bit in ADC\_EMR and the USEQ bit in ADC\_MR.

When USEQ is cleared, there are two ways to generate the averaging through the trigger event. If ASTE bit is cleared in ADC\_EMR, every trigger event generates one sample for each enabled channel as described in [Figure 40-9, "Digital Averaging Function Waveforms over Multiple Trigger Events"](#). Therefore, four trigger events are requested to get the result of averaging if OSR=1.

**Figure 40-9. Digital Averaging Function Waveforms over Multiple Trigger Events**

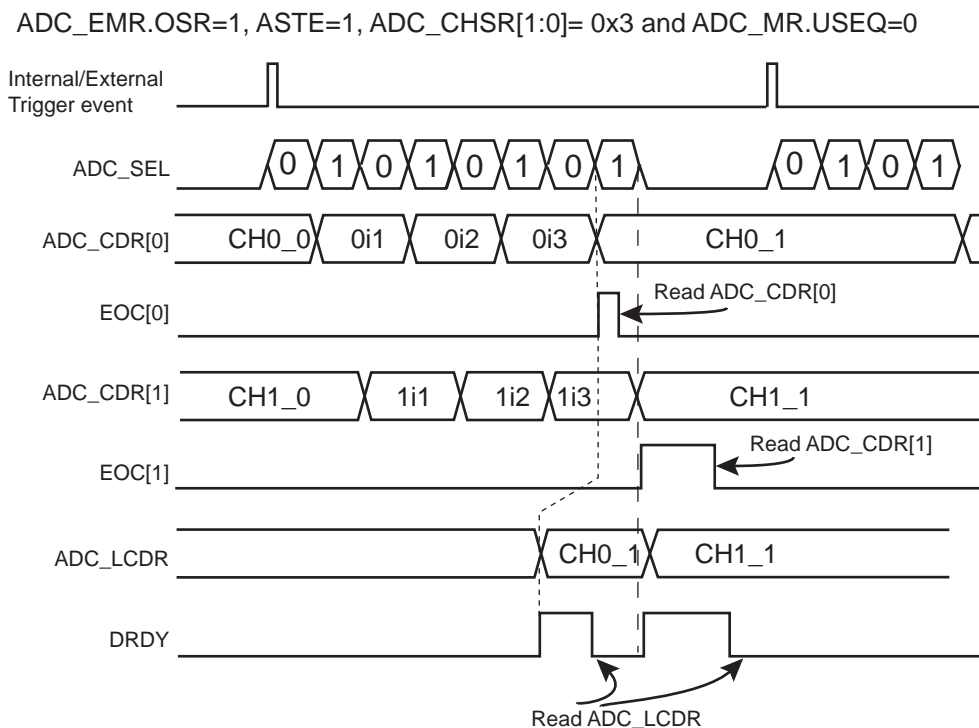
ADC\_EMR.OSR=1 ASTE=0, ADC\_CHSR[1:0]= 0x3 and ADC\_MR.USEQ=0



Note: 0i1, 0i2, 0i3, 1i1, 1i2, 1i3 are intermediate results and CH0/1\_0/1 are final result of average function.

If  $ASTE = 1$  in  $ADC\_EMR$  and  $USEQ=0$  in  $ADC\_MR$ , then the sequence to be converted, defined in  $ADC\_CHSR$ , is automatically repeated  $n$  times where  $n$  corresponds to the oversampling ratio defined in the  $OSR$  field in  $ADC\_EMR$ . As a result, only one trigger is required to obtain the result of the averaging function as described in [Figure 40-9, "Digital Averaging Function Waveforms over Multiple Trigger Events"](#).

**Figure 40-10. Digital Averaging Function Waveforms on a Single Trigger Event**



When USEQ is set, the user can define the channel sequence to be converted by configuring ADC\_SEQRx and ADC\_CHER so that channels are not interleaved during the averaging period. Under these conditions, a sample is defined for each end of conversion as shown in [Figure 40-11, "Digital Averaging Function Waveforms on Single Trigger Event, Non-interleaved"](#).

Therefore if the same channel is configured to be converted four times consecutively, and OSR=1 in ADC\_EMR, the averaging result will be placed in the corresponding channel data register ADC\_CDRx and ADC\_LCDR for each trigger event.

In such a case, the ADC effective sample rate remains the maximum ADC sample rate divided by 4 or 16, depending on OSR.

When USEQ=1, ASTE=1 and OSR is different from 0, it is important to note that the user sequence must follow a specific pattern. The user sequence must be programmed so that it generates a stream of conversion, where a given channel is successively converted with an integer multiple depending on the value of OSR. Up to four channels can be converted in this specific mode.

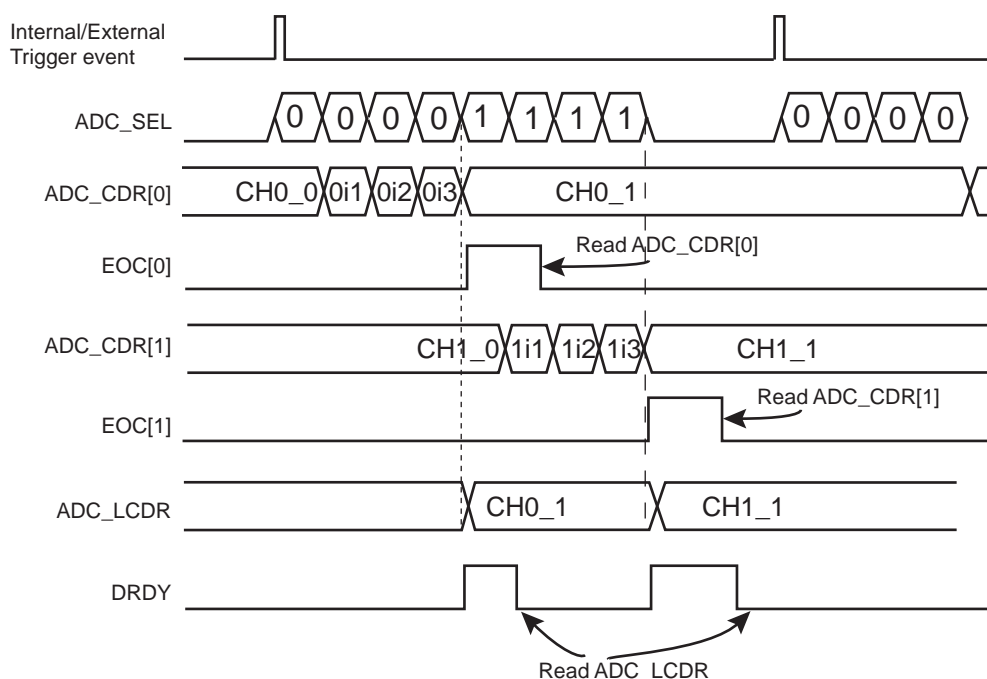
When OSR=1, each channel to convert must be repeated four times consecutively in the sequence, so the first four single bits enabled in ADC\_CHSR must have the associated channel index programmed to the same value in ADC\_SEQ1/2. Therefore, for OSR=1, a maximum of four channels can be converted (the user sequence allows a maximum of 16 conversions for each trigger event).

When OSR=2, a channel to convert must be repeated 16 times consecutively in the sequence, so all fields must be enabled in the ADC\_CHSR register, and their associated channel index programmed to the same value in ADC\_SEQ1/2. Therefore, for OSR=2, only one channel can be converted (the user sequence allows a maximum of 16 conversions for each trigger event).

OSR=3 and OSR=4 are prohibited when USEQ=1 and ASTE=1.

**Figure 40-11. Digital Averaging Function Waveforms on Single Trigger Event, Non-interleaved**

ADC\_EMR.OSR=1, ASTE=1, ADC\_CHSR[7:0]=0xFF and ADC\_MR.USEQ=1  
ADC\_SEQ1R = 0x1111\_0000



Note: 0i1, 0i2, 0i3, 1i1, 1i2, 1i3 are intermediate results and CH0/1\_0/1 are final results of average function.

#### 40.6.11.2 Oversampling Digital Output Range

When an oversampling is performed, the maximum value that can be read on ADC\_CDRx or ADC\_LCDR is not the full scale value, even if the maximum voltage is supplied on the analog input. This is due to the digital averaging algorithm. For example, when OSR=1, four samples are accumulated and the result is then right-shifted by 1 (divided by 2).

The maximum output value carried on ADC\_CDRx or ADC\_LCDR depends on the OSR value configured in ADC\_EMR.

**Table 40-4. Oversampling Digital Output Range Values**

Resolution	Samples	Shift	Full Scale Value	Maximum Value
8-bit	1	0	255	255
10-bit	1	0	1023	1023
11-bit	4	1	2047	2046
12-bit	16	2	4095	4092

#### 40.6.12 Buffer Structure

The PDC read channel is triggered each time a new data is stored in ADC\_LCDR. The same structure of data is repeatedly stored in ADC\_LCDR each time a trigger event occurs. Depending on the user mode of operation (ADC\_MR, ADC\_CHSR, ADC\_SEQR1), the structure differs. Each data read to the PDC buffer, carried on a half-word (16-bit), consists of the last converted data right-aligned. When TAG is set in ADC\_EMR, the four most significant bits carry the channel number, thus simplifying post-processing in the PDC buffer or improved checking of the PDC buffer integrity.

## 40.7 Register Write Protection

To prevent any single software error from corrupting ADC behavior, certain registers in the address space can be write-protected by setting the WPEN bit in the “[ADC Write Protection Mode Register](#)” (ADC\_WPMR).

If a write access to a write-protected register is detected, the WPVS flag in the “[ADC Write Protection Status Register](#)” (ADC\_WPSR) is set and the field WPVSR indicates the register in which the write access has been attempted.

The WPVS bit is automatically cleared after reading the ADC\_WPSR.

The following registers can be write-protected:

- “[ADC Mode Register](#)” on page 907
- “[ADC Channel Sequence 1 Register](#)” on page 909
- “[ADC Channel Enable Register](#)” on page 910
- “[ADC Channel Disable Register](#)” on page 911
- “[ADC Temperature Sensor Mode Register](#)” on page 918
- “[ADC Temperature Compare Window Register](#)” on page 919
- “[ADC Extended Mode Register](#)” on page 921
- “[ADC Compare Window Register](#)” on page 923
- “[ADC Analog Control Register](#)” on page 925



## 40.8 Analog-to-Digital Converter (ADC) User Interface

**Table 40-5. Register Mapping**

Offset	Register	Name	Access	Reset
0x00	Control Register	ADC_CR	Write-only	–
0x04	Mode Register	ADC_MR	Read/Write	0x00000000
0x08	Channel Sequence 1 Register	ADC_SEQR1	Read/Write	0x00000000
0x0C	Reserved	–	–	–
0x10	Channel Enable Register	ADC_CHER	Write-only	–
0x14	Channel Disable Register	ADC_CHDR	Write-only	–
0x18	Channel Status Register	ADC_CHSR	Read-only	0x00000000
0x1C	Reserved	–	–	–
0x20	Last Converted Data Register	ADC_LCDR	Read-only	0x00000000
0x24	Interrupt Enable Register	ADC_IER	Write-only	–
0x28	Interrupt Disable Register	ADC_IDR	Write-only	–
0x2C	Interrupt Mask Register	ADC_IMR	Read-only	0x00000000
0x30	Interrupt Status Register	ADC_ISR	Read-only	0x00000000
0x34	Temperature Sensor Mode Register	ADC_TEMPMR	Read/Write	0x00000000
0x38	Temperature Compare Window Register	ADC_TEMPCWR	Read/Write	0x00000000
0x3C	Overrun Status Register	ADC_OVER	Read-only	0x00000000
0x40	Extended Mode Register	ADC_EMR	Read/Write	0x00000000
0x44	Compare Window Register	ADC_CWR	Read/Write	0x00000000
0x50	Channel Data Register 0	ADC_CDR0	Read-only	0x00000000
0x54	Channel Data Register 1	ADC_CDR1	Read-only	0x00000000
...	...	...	...	...
0x6C	Channel Data Register 7	ADC_CDR7	Read-only	0x00000000
0x70 - 0x90	Reserved	–	–	–
0x94	Analog Control Register	ADC_ACR	Read/Write	0x00000000
0x98 - 0xAC	Reserved	–	–	–
0xC4 - 0xE0	Reserved	–	–	–
0xE4	Write Protection Mode Register	ADC_WPMR	Read/Write	0x00000000
0xE8	Write Protection Status Register	ADC_WPSR	Read-only	0x00000000
0xEC - 0xF8	Reserved	–	–	–
0xFC	Reserved	–	–	–
0x100 - 0x124	Reserved for PDC registers	–	–	–

Note: If an offset is not listed in the table it must be considered as “reserved”.

### 40.8.1 ADC Control Register

**Name:** ADC\_CR

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	START	SWRST

- **SWRST: Software Reset**

0: No effect.

1: Resets the ADC simulating a hardware reset.

- **START: Start Conversion**

0: No effect.

1: Begins analog-to-digital conversion.

## 40.8.2 ADC Mode Register

**Name:** ADC\_MR

**Access:** Read/Write

31	30	29	28	27	26	25	24
USEQ	–	–	–	TRACKTIM			
23	22	21	20	19	18	17	16
–	–	–	–	STARTUP			
15	14	13	12	11	10	9	8
PRESCAL							
7	6	5	4	3	2	1	0
FREERUN	–	SLEEP	LOWRES	TRGSEL		TRGEN	

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register” on page 926](#).

### • TRGEN: Trigger Enable

Value	Name	Description
0	DIS	Hardware triggers are disabled. Starting a conversion is only possible by software.
1	EN	Hardware trigger selected by TRGSEL field is enabled.

### • TRGSEL: Trigger Selection

Value	Name	Description
0	ADC_TRIG0	External trigger ADTRG
1	ADC_TRIG1	Timer Counter Channel 0 Output
2	ADC_TRIG2	Timer Counter Channel 1 Output
3	ADC_TRIG3	Timer Counter Channel 2 Output
4	ADC_TRIG4	Timer Counter Channel 3 Output
5	ADC_TRIG5	Timer Counter Channel 4 Output
6	ADC_TRIG6	Timer Counter Channel 5 Output
7	–	Reserved

### • LOWRES: Resolution

Value	Name	Description
0	BITS_10	10-bit resolution. For higher resolution by averaging, refer to <a href="#">Section 40.8.15 “ADC Extended Mode Register”</a>
1	BITS_8	8-bit resolution

### • SLEEP: Sleep Mode

Value	Name	Description
0	NORMAL	Normal Mode: The ADC core and reference voltage circuitry are kept ON between conversions
1	SLEEP	Sleep Mode: The ADC core and reference voltage circuitry are OFF between conversions

- **FREERUN: Free Run Mode**

Value	Name	Description
0	OFF	Normal Mode
1	ON	Free Run Mode: Never wait for any trigger.

Note: FREERUN must be set to 0 when digital averaging is used (OSR differs from 0 in ADC\_EMR register).

- **PRESCAL: Prescaler Rate Selection**

$$\text{ADCClock} = \text{MCK} / ((\text{PRESCAL} + 1) * 2)$$

- **STARTUP: Start Up Time**

Value	Name	Description
0	SUT0	0 periods of ADCClock
1	SUT8	8 periods of ADCClock
2	SUT16	16 periods of ADCClock
3	SUT24	24 periods of ADCClock
4	SUT64	64 periods of ADCClock
5	SUT80	80 periods of ADCClock
6	SUT96	96 periods of ADCClock
7	SUT112	112 periods of ADCClock
8	SUT512	512 periods of ADCClock
9	SUT576	576 periods of ADCClock
10	SUT640	640 periods of ADCClock
11	SUT704	704 periods of ADCClock
12	SUT768	768 periods of ADCClock
13	SUT832	832 periods of ADCClock
14	SUT896	896 periods of ADCClock
15	SUT960	960 periods of ADCClock

- **TRACKTIM: Tracking Time**

$$\text{Tracking Time} = (\text{TRACKTIM} + 1) * \text{ADCClock periods.}$$

- **USEQ: User Sequence Enable**

Value	Name	Description
0	NUM_ORDER	Normal Mode: The controller converts channels in a simple numeric order depending only on the channel index.
1	REG_ORDER	User Sequence Mode: The sequence respects what is defined in ADC_SEQR1 and can be used to convert the same channel several times.

### 40.8.3 ADC Channel Sequence 1 Register

**Name:** ADC\_SEQR1

**Access:** Read/Write

31	30	29	28	27	26	25	24
USCH8				USCH7			
23	22	21	20	19	18	17	16
USCH6				USCH5			
15	14	13	12	11	10	9	8
USCH4				USCH3			
7	6	5	4	3	2	1	0
USCH2				USCH1			

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **USCHx: User Sequence Number x**

The sequence number x (USCHx) can be programmed by the channel number CHy where y is the value written in this field. The allowed range is 0 up to 7. So it is only possible to use the sequencer from CH0 to CH7.

This register activates only if ADC\_MR(USEQ) field is set to 1.

Any USCHx field is taken into account only if ADC\_CHSR(CHx) register field reads logical 1; else any value written in USCHx does not add the corresponding channel in the conversion sequence.

Configuring the same value in different fields leads to multiple samples of the same channel during the conversion sequence. This can be done consecutively, or not, depending on user needs.

#### 40.8.4 ADC Channel Enable Register

**Name:** ADC\_CHER

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **CHx: Channel x Enable**

0: No effect.

1: Enables the corresponding channel.

**Note:** If USEQ = 1 in ADC\_MR, CHx corresponds to the xth channel of the sequence described in ADC\_SEQR1.

### 40.8.5 ADC Channel Disable Register

**Name:** ADC\_CHDR

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **CHx: Channel x Disable**

0: No effect.

1: Disables the corresponding channel.

**Warning:** If the corresponding channel is disabled during a conversion or if it is disabled then reenabled during a conversion, its associated data and its corresponding EOC and OVRE flags in ADC\_ISR are unpredictable.

## 40.8.6 ADC Channel Status Register

**Name:** ADC\_CHSR

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
CH7	CH6	CH5	CH4	CH3	CH2	CH1	CH0

- **CHx: Channel x Status**

0: The corresponding channel is disabled.

1: The corresponding channel is enabled.



### 40.8.7 ADC Last Converted Data Register

**Name:** ADC\_LCDR

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
CHNB				LDATA			
7	6	5	4	3	2	1	0
LDATA							

- **LDATA: Last Data Converted**

The analog-to-digital conversion data is placed into this register at the end of a conversion and remains until a new conversion is completed.

- **CHNB: Channel Number**

Indicates the last converted channel when the TAG option is set to 1 in ADC\_EMR. If the TAG option is not set, CHNB = 0.

#### 40.8.8 ADC Interrupt Enable Register

**Name:** ADC\_IER

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	RXBUFF	ENDRX	COMPE	GOVRE	DRDY
23	22	21	20	19	18	17	16
–	–	–	–	TEMPCHG	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
EOC7	EOC6	EOC5	EOC4	EOC3	EOC2	EOC1	EOC0

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Enables the corresponding interrupt.

- **EOCx: End of Conversion Interrupt Enable x**
- **TEMPCHG: Temperature Change Interrupt Enable**
- **DRDY: Data Ready Interrupt Enable**
- **GOVRE: General Overrun Error Interrupt Enable**
- **COMPE: Comparison Event Interrupt Enable**
- **ENDRX: End of Receive Buffer Interrupt Enable**
- **RXBUFF: Receive Buffer Full Interrupt Enable**

### 40.8.9 ADC Interrupt Disable Register

**Name:** ADC\_IDR

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	RXBUFF	ENDRX	COMPE	GOVRE	DRDY
23	22	21	20	19	18	17	16
–	–	–	–	TEMPCHG	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
EOC7	EOC6	EOC5	EOC4	EOC3	EOC2	EOC1	EOC0

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Disables the corresponding interrupt.

- **EOCx: End of Conversion Interrupt Disable x**
- **TEMPCHG: Temperature Change Interrupt Disable**
- **DRDY: Data Ready Interrupt Disable**
- **GOVRE: General Overrun Error Interrupt Disable**
- **COMPE: Comparison Event Interrupt Disable**
- **ENDRX: End of Receive Buffer Interrupt Disable**
- **RXBUFF: Receive Buffer Full Interrupt Disable**

#### 40.8.10 ADC Interrupt Mask Register

**Name:** ADC\_IMR

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	RXBUFF	ENDRX	COMPE	GOVRE	DRDY
23	22	21	20	19	18	17	16
–	–	–	–	TEMPCHG	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
EOC7	EOC6	EOC5	EOC4	EOC3	EOC2	EOC1	EOC0

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is disabled.

1: The corresponding interrupt is enabled.

- **EOCx: End of Conversion Interrupt Mask x**
- **TEMPCHG: Temperature Change Interrupt Mask**
- **DRDY: Data Ready Interrupt Mask**
- **GOVRE: General Overrun Error Interrupt Mask**
- **COMPE: Comparison Event Interrupt Mask**
- **ENDRX: End of Receive Buffer Interrupt Mask**
- **RXBUFF: Receive Buffer Full Interrupt Mask**

### 40.8.11 ADC Interrupt Status Register

**Name:** ADC\_ISR

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	RXBUFF	ENDRX	COMPE	GOVRE	DRDY
23	22	21	20	19	18	17	16
–	–	–	–	TEMPCHG	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
EOC7	EOC6	EOC5	EOC4	EOC3	EOC2	EOC1	EOC0

- **EOCx: End of Conversion x**

0: The corresponding analog channel is disabled, or the conversion is not finished. This flag is cleared when reading the corresponding ADC\_CDRx registers.

1: The corresponding analog channel is enabled and conversion is complete.

- **TEMPCHG: Temperature Change**

0: There is no comparison match (defined in ADC\_TEMPCWR) since the last read of ADC\_ISR.

1: The temperature value reported on ADC\_CDR7 has changed since the last read of ADC\_ISR, according to what is defined in ADC\_TEMPMR and ADC\_TEMPCWR.

- **DRDY: Data Ready**

0: No data has been converted since the last read of ADC\_LCDR.

1: At least one data has been converted and is available in ADC\_LCDR.

- **GOVRE: General Overrun Error**

0: No general overrun error occurred since the last read of ADC\_ISR.

1: At least one general overrun error has occurred since the last read of ADC\_ISR.

- **COMPE: Comparison Error**

0: No comparison error since the last read of ADC\_ISR.

1: At least one comparison error (defined in ADC\_EMR and ADC\_CWR) has occurred since the last read of ADC\_ISR.

- **ENDRX: End of RX Buffer**

0: The receive counter register has not reached 0 since the last write in ADC\_RCR or ADC\_RNCR.

1: The receive counter register has reached 0 since the last write in ADC\_RCR or ADC\_RNCR.

- **RXBUFF: RX Buffer Full**

0: ADC\_RCR or ADC\_RNCR have a value other than 0.

1: Both ADC\_RCR and ADC\_RNCR have a value of 0.

#### 40.8.12 ADC Temperature Sensor Mode Register

**Name:** ADC\_TEMP\_MR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	TEMPCMPMOD		–	–	–	TEMPON

This register can only be written if the WPEN bit is cleared in “[ADC Write Protection Mode Register](#)”.

- **TEMPON: Temperature Sensor ON**

0: The temperature sensor is not enabled.

1: The temperature sensor is enabled and the measurements are triggered.

- **TEMPCMPMOD: Temperature Comparison Mode**

Value	Name	Description
0	LOW	Generates an event when the converted data is lower than the low threshold of the window.
1	HIGH	Generates an event when the converted data is higher than the high threshold of the window.
2	IN	Generates an event when the converted data is in the comparison window.
3	OUT	Generates an event when the converted data is out of the comparison window.

### 40.8.13 ADC Temperature Compare Window Register

**Name:** ADC\_TEMPCWR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	THIGHTHRES			
23	22	21	20	19	18	17	16
THIGHTHRES							
15	14	13	12	11	10	9	8
–	–	–	–	TLOWTHRES			
7	6	5	4	3	2	1	0
TLOWTHRES							

This register can only be written if the WPEN bit is cleared in the [“ADC Write Protection Mode Register”](#) .

- **TLOWTHRES: Temperature Low Threshold**

Low threshold associated to compare settings of ADC\_TEMPMPR.

- **THIGHTHRES: Temperature High Threshold**

High threshold associated to compare settings of ADC\_TEMPMPR.

#### 40.8.14 ADC Overrun Status Register

**Name:** ADC\_OVER

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
OVRE7	OVRE6	OVRE5	OVRE4	OVRE3	OVRE2	OVRE1	OVRE0

- **OVREx: Overrun Error x**

0: No overrun error on the corresponding channel since the last read of ADC\_OVER.

1: There has been an overrun error on the corresponding channel since the last read of ADC\_OVER.

**Note:** An overrun error does not always mean that the unread data has been replaced by a new valid data. Refer to [Section 40.6.11 "Enhanced Resolution Mode and Digital Averaging Function"](#) for details.



### 40.8.15 ADC Extended Mode Register

**Name:** ADC\_EMR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	TAG
23	22	21	20	19	18	17	16
–	–	–	ASTE	–	–	OSR	
15	14	13	12	11	10	9	8
–	–	CMPFILTER		–	–	CMPALL	–
7	6	5	4	3	2	1	0
CMPSEL				–	–	CMPMODE	

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **CMPMODE: Comparison Mode**

Value	Name	Description
0	LOW	Generates an event when the converted data is lower than the low threshold of the window.
1	HIGH	Generates an event when the converted data is higher than the high threshold of the window.
2	IN	Generates an event when the converted data is in the comparison window.
3	OUT	Generates an event when the converted data is out of the comparison window.

- **CMPSEL: Comparison Selected Channel**

If CMPALL = 0: CMPSEL indicates which channel has to be compared.

If CMPALL = 1: No effect.

- **CMPALL: Compare All Channels**

0: Only channel indicated in CMPSEL field is compared.

1: All channels are compared.

- **CMPFILTER: Compare Event Filtering**

Number of consecutive compare events necessary to raise the flag = CMPFILTER+1

When programmed to 0, the flag rises as soon as an event occurs.

- **OSR: Over Sampling Rate**

Value	Name	Description
0	NO_AVERAGE	No averaging. ADC sample rate is maximum.
1	OSR4	1-bit enhanced resolution by averaging. ADC sample rate divided by 4.
2	OSR16	2-bit enhanced resolution by averaging. ADC sample rate divided by 16.

This field is active if LOWRES is cleared in [“ADC Mode Register” on page 907](#).

Note: FREERUN (see ADC\_MR register) must be set to 0 when digital averaging is used.

- **ASTE: Averaging on Single Trigger Event**

Value	Name	Description
0	MULTI_TRIG_AVERAGE	The average requests several trigger events.
1	SINGLE_TRIG_AVERAGE	The average requests only one trigger event.

- **TAG: TAG of the ADC\_LDCR register**

0: Sets CHNB to zero in ADC\_LDCR.

1: Appends the channel number to the conversion result in ADC\_LDCR.

#### 40.8.16 ADC Compare Window Register

**Name:** ADC\_CWR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	HIGHTHRES			
23	22	21	20	19	18	17	16
HIGHTHRES							
15	14	13	12	11	10	9	8
–	–	–	–	LOWTHRES			
7	6	5	4	3	2	1	0
LOWTHRES							

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **LOWTHRES: Low Threshold**

Low threshold associated to compare settings of ADC\_EMR.

If LOWRES is set in ADC\_MR, only the 10 LSB of LOWTHRES must be programmed. The 2 LSB will be automatically discarded to match the value carried on ADC\_CDR (8-bit).

- **HIGHTHRES: High Threshold**

High threshold associated to compare settings of ADC\_EMR.

If LOWRES is set in ADC\_MR, only the 10 LSB of HIGHTHRES must be programmed. The 2 LSB will be automatically discarded to match the value carried on ADC\_CDR (8-bit).

#### 40.8.17 ADC Channel Data Register

**Name:** ADC\_CDRx [x=0..7]

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	DATA			
7	6	5	4	3	2	1	0
DATA							

- **DATA: Converted Data**

The analog-to-digital conversion data is placed into this register at the end of a conversion and remains until a new conversion is completed. The Channel Data Register (CDR) is only loaded if the corresponding analog channel is enabled.

#### 40.8.18 ADC Analog Control Register

**Name:** ADC\_ACR

**Access:** Read/Write

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	ONREF	FORCEREF	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	IRVS				IRVCE	–	–

This register can only be written if the WPEN bit is cleared in [“ADC Write Protection Mode Register”](#).

- **IRVCE: Internal Reference Voltage Change Enable**

0 (STUCK\_AT\_DEFAULT): The internal reference voltage is stuck at the default value (see the Electrical Characteristics for further details).

1 (SELECTION): The internal reference voltage is defined by field IRVS.

- **IRVS: Internal Reference Voltage Selection**

See the “Programmable Voltage Reference Selection Values” table in the Electrical Characteristics for further details.

- **FORCEREF: Force Internal Reference Voltage**

0: The internal voltage reference is defined.

1: The internal voltage reference is forced to VDDIO (ONREF must be set to 0).

- **ONREF: Internal Voltage Reference ON**

0: The external pin ADVREF defines the voltage reference.

1: The internal voltage reference is selected (FORCEREF must be set to 0).

#### 40.8.19 ADC Write Protection Mode Register

**Name:** ADC\_WPMR

**Access:** Read/Write

**Reset:** See [Table 40-5](#)

31	30	29	28	27	26	25	24
WPKEY							
23	22	21	20	19	18	17	16
WPKEY							
15	14	13	12	11	10	9	8
WPKEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPEN

For more information on Write Protection registers, refer to [Section 40.7 "Register Write Protection"](#).

- **WPEN: Write Protect Enable**

0 = Disables the write protection if WPKEY corresponds to 0x414443 ("ADC" in ASCII).

1 = Enables the write protection if WPKEY corresponds to 0x414443 ("ADC" in ASCII).

See [Section 40.7 "Register Write Protection"](#) for the list of registers that can be protected.

- **WPKEY: Write Protect Key**

Value	Name	Description
0x414443	PASSWD	Writing any other value in this field aborts the write operation of the WPEN bit. Always reads as 0

#### 40.8.20 ADC Write Protection Status Register

**Name:** ADC\_WPSR

**Access:** Read-only

**Reset:** See [Table 40-5](#)

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
WPVSR							
15	14	13	12	11	10	9	8
WPVSR							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	WPVS

For more information on Write Protection registers, refer to [Section 40.7 "Register Write Protection"](#).

- **WPVS: Write Protection Violation Status**

0: No write protection violation has occurred since the last read of the ADC\_WPSR register.

1: A write protection violation has occurred since the last read of the ADC\_WPSR register. If this violation is an unauthorized attempt to write a protected register, the associated violation is reported into field WPVSR.

- **WPVSR: Write Protection Violation Source**

When WPVS = 1, WPVSR indicates the register address offset at which a write access has been attempted.

## 41. Advanced Encryption Standard (AES)

### 41.1 Description

The Advanced Encryption Standard (AES) is compliant with the American *FIPS (Federal Information Processing Standard) Publication 197* specification.

The AES supports all five confidentiality modes of operation for symmetrical key block cipher algorithms (ECB, CBC, OFB, CFB and CTR), as specified in the *NIST Special Publication 800-38A Recommendation*. It is compatible with all these modes via Peripheral DMA Controller channels, minimizing processor intervention for large buffer transfers.

The 128-bit/192-bit/256-bit key is stored in four/six/eight 32-bit registers (AES\_KEYWRx) which are all write-only.

The 128-bit input data and initialization vector (for some modes) are each stored in four 32-bit registers (AES\_IDATARx and AES\_IVRx) which are all write-only.

As soon as the initialization vector, the input data and the key are configured, the encryption/decryption process may be started. Then the encrypted/decrypted data are ready to be read out on the four 32-bit output data registers (AES\_ODATARx) or through the PDC channels.

### 41.2 Embedded Characteristics

- Compliant with *FIPS Publication 197, Advanced Encryption Standard (AES)*
- 128-bit/192-bit/256-bit Cryptographic Key
- 12/14/16 Clock Cycles Encryption/Decryption Processing Time with a 128-bit/192-bit/256-bit Cryptographic Key
- Double Input Buffer Optimizes Runtime
- Support of the Five Standard Modes of Operation Specified in the *NIST Special Publication 800-38A, Recommendation for Block Cipher Modes of Operation - Methods and Techniques*:
  - Electronic Code Book (ECB)
  - Cipher Block Chaining (CBC) including CBC-MAC
  - Cipher Feedback (CFB)
  - Output Feedback (OFB)
  - Counter (CTR)
  - Galois Count Mode (GCM)
- 8-, 16-, 32-, 64- and 128-bit Data Sizes Possible in CFB Mode
- Last Output Data Mode Allows Optimized Message Authentication Code (MAC) Generation
- Connection to PDC Channel Capabilities Optimizes Data Transfers for all Operating Modes
  - One Channel for the Receiver, One Channel for the Transmitter
  - Next Buffer Support



## 41.3 Product Dependencies

### 41.3.1 Power Management

The AES may be clocked through the Power Management Controller (PMC), so the programmer must first to configure the PMC to enable the AES clock.

### 41.3.2 Interrupt

The AES interface has an interrupt line connected to the Interrupt Controller.

Handling the AES interrupt requires programming the Interrupt Controller before configuring the AES.

**Table 41-1. Peripheral IDs**

Instance	ID
AES	36

## 41.4 Functional Description

The Advanced Encryption Standard (AES) specifies a FIPS-approved cryptographic algorithm that can be used to protect electronic data. The AES algorithm is a symmetric block cipher that can encrypt (encipher) and decrypt (decipher) information.

Encryption converts data to an unintelligible form called ciphertext. Decrypting the ciphertext converts the data back into its original form, called plaintext. The CIPHER bit in the AES Mode Register (AES\_MR) allows selection between the encryption and the decryption processes.

The AES is capable of using cryptographic keys of 128/192/256 bits to encrypt and decrypt data in blocks of 128 bits. This 128-bit/192-bit/256-bit key is defined in the Key Registers (AES\_KEYWRx).

The input to the encryption processes of the CBC, CFB, and OFB modes includes, in addition to the plaintext, a 128-bit data block called the initialization vector (IV), which must be set in the Initialization Vector Registers (AES\_IVRx). The initialization vector is used in an initial step in the encryption of a message and in the corresponding decryption of the message. The Initialization Vector Registers are also used by the CTR mode to set the counter value.

### 41.4.1 Operation Modes

The AES supports the following modes of operation:

- ECB: Electronic Code Book
- CBC: Cipher Block Chaining
- OFB: Output Feedback
- CFB: Cipher Feedback
  - CFB8 (CFB where the length of the data segment is 8 bits)
  - CFB16 (CFB where the length of the data segment is 16 bits)
  - CFB32 (CFB where the length of the data segment is 32 bits)
  - CFB64 (CFB where the length of the data segment is 64 bits)
  - CFB128 (CFB where the length of the data segment is 128 bits)
- CTR: Counter
- GCM: Galois Counter Mode

The data pre-processing, post-processing and data chaining for the concerned modes are automatically performed. Refer to the *NIST Special Publication 800-38A Recommendation* for more complete information.

These modes are selected by setting the OPMOD field in the AES\_MR.

In CFB mode, five data sizes are possible (8, 16, 32, 64 or 128 bits), configurable by means of the CFBS field in the AES\_MR ([Section 41.7.2 “AES Mode Register” on page 944](#)).

In CTR mode, the size of the block counter embedded in the module is 16 bits. Therefore, there is a rollover after processing 1 megabyte of data. If the file to be processed is greater than 1 megabyte, this file must be split into fragments of 1 megabyte. Prior to loading the first fragment into AES\_IDATARx, AES\_IVRx must be cleared. For any fragment, after the transfer is completed and prior to transferring the next fragment, AES\_IVR0 must be programmed so that the fragment number (0 for the first fragment, 1 for the second one, and so on) is written in the 16 MSB of AES\_IVR0.

#### 41.4.2 Double Input Buffer

The input data register can be double-buffered to reduce the runtime of large files.

This mode allows writing a new message block when the previous message block is being processed. This is only possible when DMA accesses are performed (SMOD = 0x2).

The DUALBUFF bit in AES\_MR must be set to 1 to access the double buffer.

#### 41.4.3 Start Modes

The SMOD field in the AES\_MR allows selection of the encryption (or decryption) start mode.

##### 41.4.3.1 Manual Mode

The sequence order is as follows:

- Write the AES\_MR with all required fields, including but not limited to SMOD and OPMOD.
- Write the 128-bit/192-bit/256-bit key in the Key Registers (AES\_KEYWRx).
- Write the initialization vector (or counter) in the Initialization Vector Registers (AES\_IVRx).

Note: The Initialization Vector Registers concern all modes except ECB.

- Set the bit DATRDY (Data Ready) in the AES Interrupt Enable register (AES\_IER), depending on whether an interrupt is required or not at the end of processing.
- Write the data to be encrypted/decrypted in the authorized Input Data Registers (See [Table 41-2](#)).

**Table 41-2. Authorized Input Data Registers**

Operation Mode	Input Data Registers to Write
ECB	All
CBC	All
OFB	All
128-bit CFB	All
64-bit CFB	AES_IDATAR0 and AES_IDATAR1
32-bit CFB	AES_IDATAR0
16-bit CFB	AES_IDATAR0
8-bit CFB	AES_IDATAR0
CTR	All
GCM	All

Note: In 64-bit CFB mode, writing to AES\_IDATAR2 and AES\_IDATAR3 registers is not allowed and may lead to errors in processing.

Note: In 32-, 16- and 8-bit CFB modes, writing to AES\_IDATAR1, AES\_IDATAR2 and AES\_IDATAR3 registers is not allowed and may lead to errors in processing.

- Set the START bit in the AES Control register AES\_CR to begin the encryption or the decryption process.
- When processing completes, the DATRDY bit in the AES Interrupt Status Register (AES\_ISR) raises. If an interrupt has been enabled by setting the DATRDY bit in AES\_IER, the interrupt line of the AES is activated.

- When the software reads one of the Output Data Registers (AES\_ODATARx), the DATRDY bit is automatically cleared.

#### 41.4.3.2 Auto Mode

The Auto Mode is similar to the manual one, except that in this mode, as soon as the correct number of Input Data registers is written, processing is automatically started without any action in the Control Register.

#### 41.4.4 PDC Mode

The Peripheral DMA Controller (PDC) can be used in association with the AES to perform an encryption/decryption of a buffer without any action by the software during processing.

The field SMOD must be configured to 0x2 in AES\_MR.

The sequence order is as follows:

- Write the AES\_MR with all required fields, including but not limited to SMOD and OPMOD.
- Write the key in the Key Registers (AES\_KEYWRx).
- Write the initialization vector (or counter) in the Initialization Vector Registers (AES\_IVRx).

Note: The Initialization Vector Registers concern all modes except ECB.

- Set the Transmit Pointer Register (AES\_TPR) to the address where the data buffer to encrypt/decrypt is stored and the Receive Pointer Register (AES\_RPR) where it must be encrypted/decrypted.

Note: Transmit and receive buffers can be identical.

- Set the Transmit and the Receive Counter Registers (AES\_TCR and AES\_RCR) to the same value. This value must be a multiple of the data transfer type size (see [Table 41-3 "Data Transfer Type for the Different Operation Modes"](#)).

Note: The same requirements are necessary for the Next Pointer(s) and Counter(s) of the PDC (AES\_TNPR, AES\_RNPR, AES\_TNCR, AES\_RNCR).

- If not already done, set the bit ENDRX (or RXBUFF if the next pointers and counters are used) in the AES Interrupt Enable register (AES\_IER), depending on whether an interrupt is required or not at the end of processing.
- Enable the PDC in transmission and reception to start the processing (AES\_PTCR).

When the processing completes, the bit ENDRX (or RXBUFF) in the AES Interrupt Status Register (AES\_ISR) rises. If an interrupt has been enabled by setting the corresponding bit in AES\_IER, the interrupt line of the AES is activated.

When PDC is used, the data size to transfer (byte, half-word or word) depends on the AES mode of operations. This size is automatically configured by the AES.

**Table 41-3. Data Transfer Type for the Different Operation Modes**

Operation Mode	Data Transfer Type
ECB	Word
CBC	Word
OFB	Word
CFB 128-bit	Word
CFB 64-bit	Word
CFB 32-bit	Word
CFB 16-bit	Half-word
CFB 8-bit	Byte
CTR	Word
GCM	Word

### 41.4.5 Last Output Data Mode

This mode is used to generate cryptographic checksums on data (MAC) by means of cipher block chaining encryption algorithm (CBC-MAC algorithm for example).

After each end of encryption/decryption, the output data are available either on the output data registers for Manual and Auto mode or at the address specified in the receive buffer pointer for PDC mode (See [Table 41-4 "Last Output Data Mode Behavior versus Start Modes"](#) ).

The Last Output Data bit (LOD) in the AES Mode Register (AES\_MR) allows retrieval of only the last data of several encryption/decryption processes.

Therefore, there is no need to define a read buffer in PDC mode.

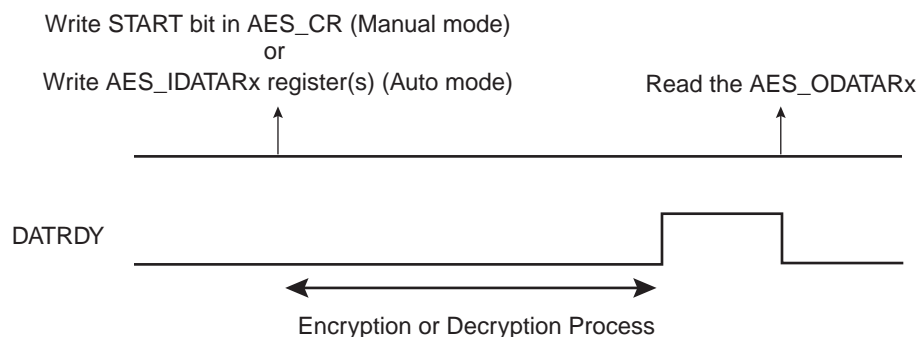
This data are only available on the Output Data Registers (AES\_ODATARx).

### 41.4.6 Manual and Auto Modes

#### 41.4.6.1 If LOD = 0

The DATRDY flag is cleared when at least one of the Output Data Registers is read (See [Figure 41-1](#)).

**Figure 41-1. Manual and Auto Modes with LOD = 0**

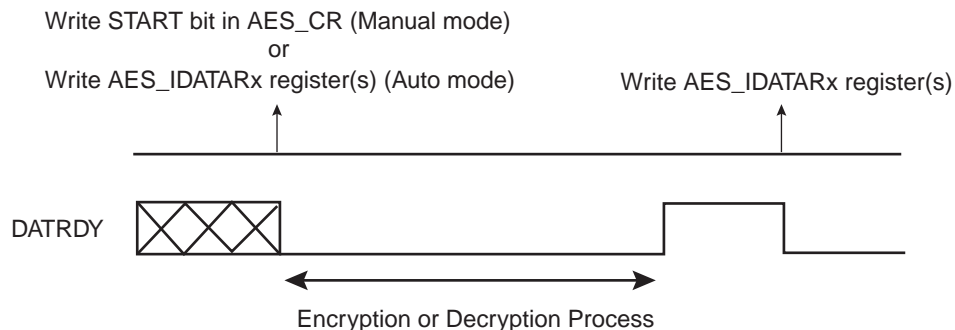


If the user does not want to read the output data registers between each encryption/decryption, the DATRDY flag will not be cleared. If the DATRDY flag is not cleared, the user cannot know the end of the following encryptions/decryptions.

#### 41.4.6.2 If LOD = 1

The DATRDY flag is cleared when at least one Input Data Register is written, so before the start of a new transfer (See [Figure 41-2](#)). No more Output Data Register reads are necessary between consecutive encryptions/decryptions.

**Figure 41-2. Manual and Auto Modes with LOD = 1**

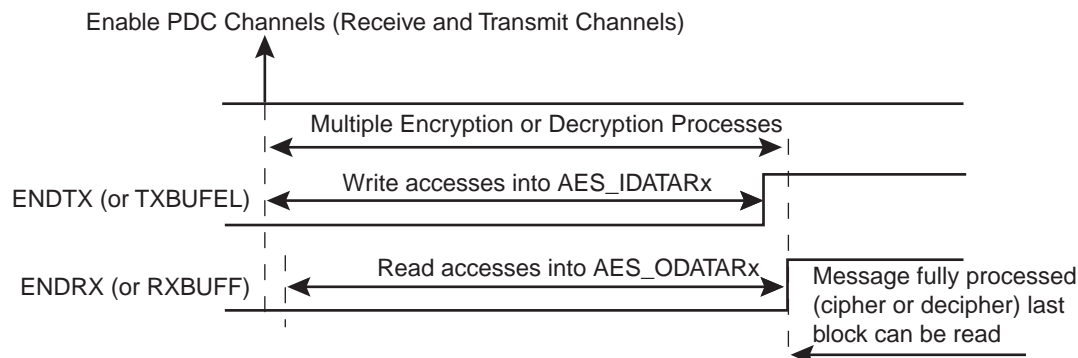


## 41.4.7 PDC Mode

### 41.4.7.1 If LOD = 0

This mode may be used for all AES operating modes except CBC-MAC where LOD = 1 mode is recommended. The end of the encryption/decryption is notified by the ENDRX (or RXBUFF) flag rise (see [Figure 41-3](#)).

**Figure 41-3. PDC transfer with LOD = 0**



### 41.4.7.2 If LOD = 1

This mode is recommended to process AES CBC-MAC operating mode.

The user must first wait for the ENDTX (or TXBUFE) flag to rise, then for DATRDY to ensure that the encryption/decryption is completed (see [Figure 41-4](#)).

In this case, no receive buffers are required.

The output data are only available on the Output Data Registers (AES\_ODATARx).

**Figure 41-4. PDC transfer with LOD = 1**

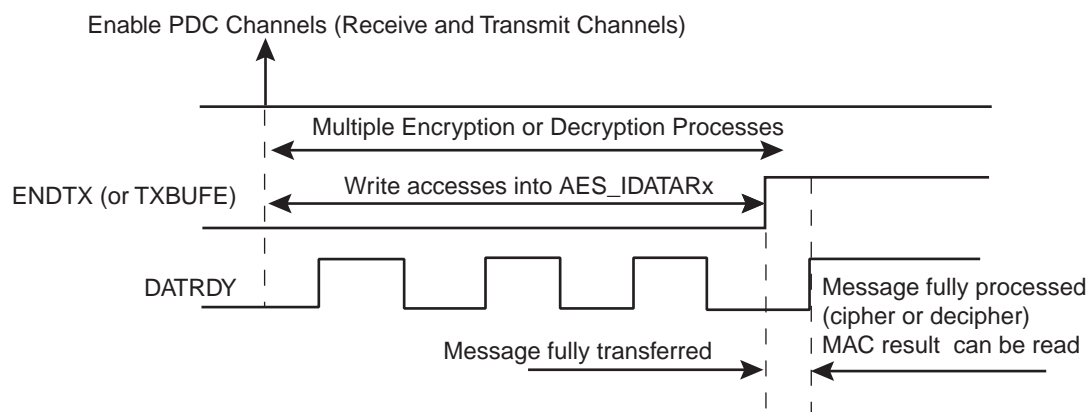


Table 41-4 summarizes the different cases.

**Table 41-4. Last Output Data Mode Behavior versus Start Modes**

Sequence	Manual and Auto Modes		PDC Mode	
	LOD = 0	LOD = 1	LOD = 0	LOD = 1
DATRDY Flag Clearing Condition <sup>(1)</sup>	At least one Output Data Register must be read	At least one Input Data Register must be written	Not used	Managed by the PDC
End of Encryption/Decryption Notification	DATRDY	DATRDY	ENDRX (or RXBUFF)	ENDTX (or TXBUFE) then DATRDY
Encrypted/Decrypted Data Result Location	In the Output Data Registers	In the Output Data Registers	At the address specified in the Receive Pointer Register (AES_RPR)	In the Output Data Registers

Note: 1. Depending on the mode, there are other ways of clearing the DATRDY flag. See [“AES Interrupt Status Register” on page 950](#).

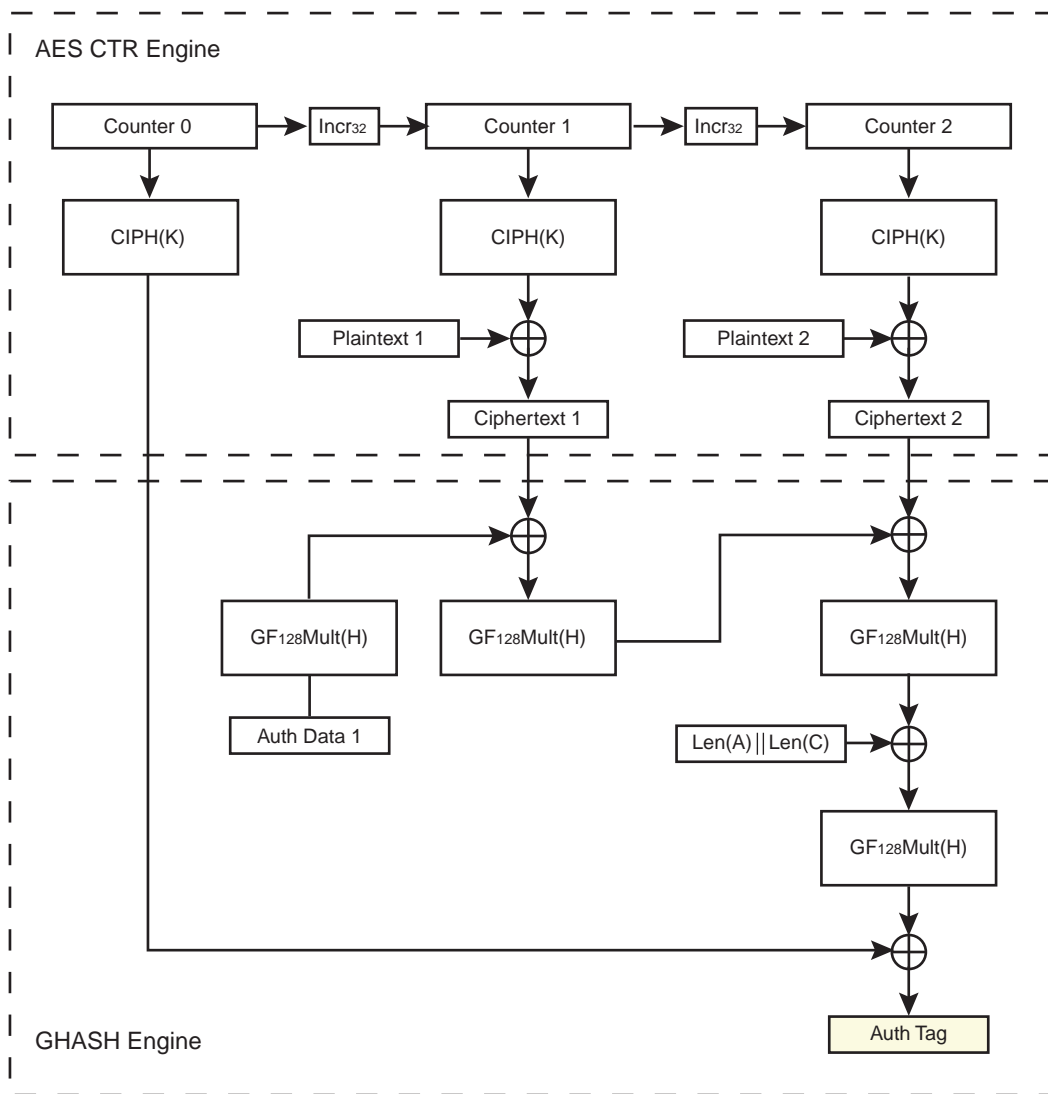
Warning: In PDC mode, reading to the Output Data registers before the last data transfer may lead to unpredictable results.

## 41.5 Galois Counter Mode (GCM)

### 41.5.1 Description

GCM comprises the AES engine in CTR mode along with a universal hash function (GHASH engine) that is defined over a binary Galois field to produce a message authentication tag. The GHASH engine processes data packets after the AES operation. GCM provides assurance of the confidentiality of data through the AES Counter mode of operation for encryption. Authenticity of the confidential data is assured through the GHASH engine. GCM can also provide assurance of data that is not encrypted. Refer to the [NIST Special Publication 800-38D Recommendation](#) for more complete information.

Figure 41-5. GCM Block Diagram



GCM mode can be used with or without the PDC master. Messages may be processed as a single complete packet of data or they may be broken into multiple packets of data over time.

GCM processing is computed on 128-bit input data fields. There is no support for unaligned data. The AES key length can be whatever length is supported by the AES module.

Whenever a new key (AES\_KEYWRx) is written to the hardware, the GCM Hash subkey (H) is automatically generated. The GCM Hash subkey (H) generation must be complete before doing any other action. The DATRDY bit of the

AES\_ISR indicates when the subkey generation is complete (with interrupt if configured). The generated H value is then available in the AES\_GCMH registers. The H field can be written to any value afterwards.

The recommended programming procedure when using PDC is described in [Section 41.5.2](#).

## 41.5.2 GCM Processing

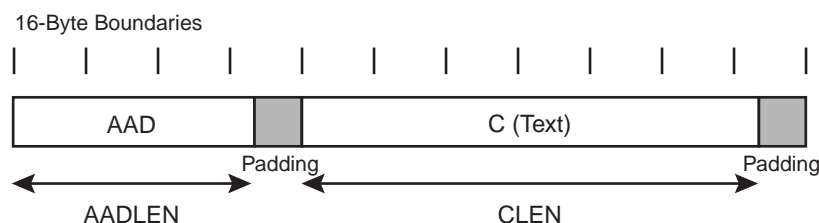
GCM processing comprises three phases:

1. Processing the Additional Authenticated Data (AAD, hash computation only).
2. Processing the ciphertext (C, hash computation + ciphering/deciphering).
3. Generating the TAG (using length of AAD, length of C and  $J_0$ , see NIST documentation for details).

The Tag generation can be done either automatically, after the end of AAD/C processing if TAG\_EN bit is set in the AES\_MR or done manually, using the GHASH field in AES\_GHASHRx (see [Section 41.5.2.1](#) and [Section 41.5.2.4](#) for details).

### 41.5.2.1 Processing a Complete Message with Tag Generation

**Figure 41-6. Full Message Alignment**



To process a complete message with Tag generation, perform the following steps:

1. In AES\_MR set OPMOD to GCM and GTAGEN to '1' (configuration as usual for the rest).
2. Set KEYW in AES\_KEYWRx and wait until DATRDY bit of AES\_ISR is at level '1' (GCM H subkey generation complete), use interrupt if needed.
3. Set IV in AES\_IVRx with ' $J_0 + 1$ ' value. As described in NIST documentation  $J_0 = IV \parallel 0_{31} \parallel 1$  when  $\text{len}(IV) = 96$  and  $J_0 = \text{GHASH}_H(IV \parallel 0_{s+64} \parallel [\text{len}(IV)]_{64})$  if  $\text{len}(IV) \neq 96$ . See [Section 41.5.2.5 "Processing a Message with only AAD \(GHASHH\)"](#) for  $J_0$  generation.
4. Set AADLEN field in AES\_AADLENR and CLEN field in AES\_CLENR.
5. Fill the IDATA field of AES\_IDATARx with the message to process according to the SMOD configuration used. If Manual Mode or Auto Mode is used, the DATRDY bit indicates when the data have been processed (however, no output data are generated when processing AAD).
6. Wait for TAGRDY to be at level '1' (use interrupt if needed), then read the GCM\_TAG field of AES\_GCMTAGRx to obtain the authentication tag of the message.

### 41.5.2.2 Processing a Complete Message without Tag Generation

Processing a message without generating the Tag can be used to customize the Tag generation, or to process a fragmented message. To manually generate the GCM Tag see [Section 41.5.2.4](#).

To process a complete message without Tag generation, perform the following steps:

1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
2. Set KEYW in AES\_KEYWRx and wait until DATRDY bit of AES\_ISR is at level '1' (GCM H subkey generation complete), use interrupt if needed. The H value can be read (or written to any value) in the AES\_GCMH registers.
3. Set IV in AES\_IVRx with ' $J_0 + 1$ ' value. As described in NIST documentation  $J_0 = IV \parallel 0_{31} \parallel 1$  when  $\text{len}(IV) = 96$  and  $J_0 = \text{GHASH}_H(IV \parallel 0_{s+64} \parallel [\text{len}(IV)]_{64})$  if  $\text{len}(IV) \neq 96$ . See [Section 41.5.2.5](#) for  $J_0$  generation example when  $\text{len}(IV) \neq 96$ .
4. Set AADLEN field in AES\_AADLENR and CLEN field in AES\_CLENR.



5. Fill the IDATA field of AES\_IDATARx with the message to process according to the SMOD configuration used. If Manual Mode or Auto Mode is used, the DATRDY bit indicates when the data have been processed (however, no output data are generated when processing AAD).
6. Make sure the last output data have been read if  $CLEN \neq 0$  (or wait for DATRDY), then read the GHASH field of AES\_GHASHRx to obtain the hash value after the last processed data.

#### 41.5.2.3 Processing a Fragmented Message

If needed, a message can be processed by fragments, in such case automatic GCM TAG generation is not supported.

To process a message by fragments, perform the following steps:

- First fragment:
  1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
  2. Set KEYW in AES\_KEYWRx and wait for DATRDY bit of AES\_ISR to be at level '1' (GCM H subkey generation complete), use interrupt if needed. The H value can be read (or written to any value) in the AES\_GCMH registers.
  3. Set IV in AES\_IVRx with 'J<sub>0</sub> + 1' value. As described in NIST documentation  $J_0 = IV \parallel 0_{31} \parallel 1$  when  $\text{len}(IV) = 96$  and  $J_0 = \text{GHASH}_H(IV \parallel 0_{3+64} \parallel [\text{len}(IV)]_{64})$  if  $\text{len}(IV) \neq 96$ . See [Section 41.5.2.5](#) for J<sub>0</sub> generation example when  $\text{len}(IV) \neq 96$ .
  4. Set AADLEN field in AES\_AADLENR and CLEN field in AES\_CLENR according to the length of the first fragment, or set the fields with the full message length, both configurations work.
  5. Fill the IDATA field of AES\_IDATARx with the first fragment of the message to process (aligned on 16-byte boundary) according to the SMOD configuration used. If Manual Mode or Auto Mode is used the DATRDY bit indicates when the data have been processed (however, no output data are generated when processing AAD).
  6. Make sure the last output data have been read if the fragment ends in C phase (or wait for DATRDY if the fragment ends in AAD phase), then read the GHASH field of AES\_GHASHRx to obtain the value of the hash after the last processed data and finally read the CTR field of the AES\_CTR to obtain the value of the CTR encryption counter (not needed when the fragment ends in AAD phase).
- Next fragment (or last fragment):
  1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
  2. Set KEYW in AES\_KEYWRx and wait until DATRDY bit of AES\_ISR is at level '1' (GCM H subkey complete), use interrupt if needed. The H value can be read (or written to any value) in the AES\_GCMH registers.
  3. Set IV in AES\_IVRx with "J<sub>0</sub> + 1" value, if the fragment begins in AAD phase or with 'LSB<sub>96</sub>(J<sub>0</sub>) || CTR' value, if the fragment begins in C phase (96 bit LSB of J<sub>0</sub> concatenated with saved CTR value from previous fragment).
  4. Set AADLEN field in AES\_AADLENR and CLEN field in AES\_CLENR according to the length of the current fragment, or set the fields with the remaining message length, both configurations work.
  5. Fill the GHASH field of AES\_GHASHRx with the value stored after the previous fragment.
  6. Fill the IDATA field of AES\_IDATARx with the current fragment of the message to process (aligned on 16 byte boundary) according to the SMOD configuration used. If Manual Mode or Auto Mode is used, the DATRDY bit indicates when the data have been processed (however, no output data are generated when processing AAD).
  7. Make sure the last output data have been read if the fragment ends in C phase (or wait for DATRDY if the fragment ends in AAD phase), then read the GHASH field of AES\_GHASHRx to obtain the value of the hash after the last processed data and finally read the CTR field of the AES\_CTR to obtain the value of the CTR encryption counter (not needed when the fragment ends in AAD phase).

Note: Step 1 and 2 are required only if the value of the concerned registers has been modified.

Once the last fragment has been processed, the GHASH value will allow manual generation of the GCM tag, see [Section 41.5.2.4](#) for details.

#### 41.5.2.4 Manual GCM Tag Generation

To generate a GCM Tag manually, perform the following steps:

Processing  $S = \text{GHASH}_H(A \parallel 0_v \parallel C \parallel 0_u \parallel [\text{len}(A)]_{64} \parallel [\text{len}(C)]_{64})$ :

1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
2. Set KEYW in AES\_KEYWRx and wait for DATRDY bit of AES\_ISR to be at level '1' (GCM H subkey complete), use interrupt if needed. The H value can be read (or written to any value) in the AES\_GCMH registers.
3. Set AADLEN field to 0x10 (16 bytes) in AES\_AADLENR and CLEN field to '0' in AES\_CLENR. This will allow running a single GHASH<sub>H</sub> on a 16-byte input data (see Figure 41-7).
4. Fill the GHASH field of AES\_GHASHRx with the state of the GHASH field stored at the end of the message processing.
5. Fill the IDATA field of AES\_IDATARx according to the SMOD configuration used with 'LEN(A)<sub>64</sub> || LEN(C)<sub>64</sub>' value as described in the NIST documentation and wait for DATRDY to be at level '1', use interrupt if needed.
6. Read the GHASH field of AES\_GHASHRx to obtain the current value of the hash.

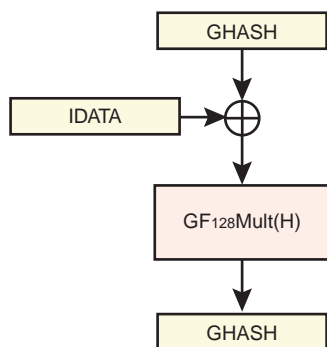
Processing T = GCTR<sub>K</sub>(J<sub>0</sub>, S):

7. In AES\_MR set OPMOD to CTR (configuration as usual for the rest).
8. Set the IV field in AES\_IVRx with 'J<sub>0</sub>' value.
9. Fill the IDATA field of AES\_IDATARx with the GHASH value read at step 6 and wait for DATRDY to be at level '1', use interrupt if needed.
10. Read the ODATA field of AES\_ODATARx to obtain the GCM TAG value.

Note: Step 4 is optional if the GHASH field is to be filled with value '0' (0 length packet for instance).

#### 41.5.2.5 Processing a Message with only AAD (GHASH<sub>H</sub>)

**Figure 41-7. Single GHASH<sub>H</sub> Block Diagram (AADLEN ≤ 0x10 and CLEN = 0)**



It is possible to process a message with only AAD setting the CLEN field to '0' in the AES\_CLENR, this can be used for J<sub>0</sub> generation when LEN(IV) ≠ 96 for instance.

Example: Processing J<sub>0</sub> when LEN(IV) ≠ 96

To Process J<sub>0</sub> = GHASH<sub>H</sub>(IV || 0<sub>s+64</sub> || [len(IV)]<sub>64</sub>) perform the following steps:

1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
2. Set KEYW in AES\_KEYWRx and wait until DATRDY bit of AES\_ISR is at level '1' (GCM H subkey complete), use interrupt if needed. The H value can be read (or written to any value) in the AES\_GCMH registers.
3. Set AADLEN field with 'LEN(IV || 0<sub>s+64</sub> || [len(IV)]<sub>64</sub>)' in AES\_AADLENR and CLEN field to '0' in AES\_CLENR. This will allow running a GHASH<sub>H</sub> only.
4. Fill the IDATA field of AES\_IDATARx with the message to process (IV || 0<sub>s+64</sub> || [len(IV)]<sub>64</sub>) according to the SMOD configuration used. If Manual Mode or Auto Mode is used, the DATRDY bit indicates when a GHASH<sub>H</sub> step is over, use interrupt if needed.
5. Read the GHASH field of AES\_GHASHRx to obtain the J<sub>0</sub> value.

Note: The GHASH value can be overwritten at any time by writing the GHASH field value of AES\_GHASHRx, used to perform a GHASH<sub>H</sub> with an initial value for GHASH (write GHASH field between step 3 and step 4 in this case).

#### 41.5.2.6 Processing a Single GF<sub>128</sub> Multiplication

The AES can also be used to process a single multiplication in the Galois Field on 128 bits (GF<sub>128</sub>) using a single GHASH<sub>H</sub> with custom H value (See [Figure 41-7](#)).

To run a GF<sub>128</sub> multiplication (A x B) perform the following steps:

1. In AES\_MR set OPMOD to GCM and GTAGEN to '0' (configuration as usual for the rest).
2. Set AADLEN field with 0x10 (16 bytes) in AES\_AADLENR and CLEN field to '0' in AES\_CLENR. This will allow running a single GHASH<sub>H</sub>.
3. Fill the H field of the AES\_GCMH registers with B value.
4. Fill the IDATA field of AES\_IDATARx with the A value according to the SMOD configuration used. If Manual Mode or Auto Mode is used, the DATRDY bit indicates when a GHASH<sub>H</sub> computation is over, use interrupt if needed.
5. Read the GHASH field of AES\_GHASHRx to obtain the result.

Note: GHASH field of AES\_GHASHRx can be initialized with a value C between step 3 and step 4 to run a ((A XOR C) x B) GF<sub>128</sub> multiplication

## 41.6 Security Features

### 41.6.1 Unspecified Register Access Detection

When an unspecified register access occurs, the URAD bit in the Interrupt Status Register (AES\_ISR) raises. Its source is then reported in the Unspecified Register Access Type field (URAT). Only the last unspecified register access is available through the URAT field.

Several kinds of unspecified register accesses can occur:

- Input Data Register written during the data processing when SMOD = IDATAR0\_START
- Output Data Register read during data processing
- Mode Register written during data processing
- Output Data Register read during sub-keys generation
- Mode Register written during sub-keys generation
- Write-only register read access

The URAD bit and the URAT field can only be reset by the SWRST bit in the AES\_CR.

## 41.7 Advanced Encryption Standard (AES) User Interface

Table 41-5. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	AES_CR	Write-only	–
0x04	Mode Register	AES_MR	Read-write	0x0
0x08–0x0C	Reserved	–	–	–
0x10	Interrupt Enable Register	AES_IER	Write-only	–
0x14	Interrupt Disable Register	AES_IDR	Write-only	–
0x18	Interrupt Mask Register	AES_IMR	Read-only	0x0
0x1C	Interrupt Status Register	AES_ISR	Read-only	0x0000001E
0x20	Key Word Register 0	AES_KEYWR0	Write-only	–
0x24	Key Word Register 1	AES_KEYWR1	Write-only	–
0x28	Key Word Register 2	AES_KEYWR2	Write-only	–
0x2C	Key Word Register 3	AES_KEYWR3	Write-only	–
0x30	Key Word Register 4	AES_KEYWR4	Write-only	–
0x34	Key Word Register 5	AES_KEYWR5	Write-only	–
0x38	Key Word Register 6	AES_KEYWR6	Write-only	–
0x3C	Key Word Register 7	AES_KEYWR7	Write-only	–
0x40	Input Data Register 0	AES_IDATAR0	Write-only	–
0x44	Input Data Register 1	AES_IDATAR1	Write-only	–
0x48	Input Data Register 2	AES_IDATAR2	Write-only	–
0x4C	Input Data Register 3	AES_IDATAR3	Write-only	–
0x50	Output Data Register 0	AES_ODATAR0	Read-only	0x0
0x54	Output Data Register 1	AES_ODATAR1	Read-only	0x0
0x58	Output Data Register 2	AES_ODATAR2	Read-only	0x0
0x5C	Output Data Register 3	AES_ODATAR3	Read-only	0x0
0x60	Initialization Vector Register 0	AES_IVR0	Write-only	–
0x64	Initialization Vector Register 1	AES_IVR1	Write-only	–
0x68	Initialization Vector Register 2	AES_IVR2	Write-only	–
0x6C	Initialization Vector Register 3	AES_IVR3	Write-only	–
0x70	Additional Authenticated Data Length Register	AES_AADLENR	Read-write	–
0x74	Plaintext/Ciphertext Length Register	AES_CLENR	Read-write	–
0x78	GCM Intermediate Hash Word Register 0	AES_GHASHR0	Read-write	–
0x7C	GCM Intermediate Hash Word Register 1	AES_GHASHR1	Read-write	–
0x80	GCM Intermediate Hash Word Register 2	AES_GHASHR2	Read-write	–
0x84	GCM Intermediate Hash Word Register 3	AES_GHASHR3	Read-write	–
0x88	GCM Authentication Tag Word Register 0	AES_TAGR0	Read-only	–
0x8C	GCM Authentication Tag Word Register 1	AES_TAGR1	Read-only	–

**Table 41-5. Register Mapping**

Offset	Register	Name	Access	Reset
0x90	GCM Authentication Tag Word Register 2	AES_TAGR2	Read-only	–
0x94	GCM Authentication Tag Word Register 3	AES_TAGR3	Read-only	–
0x98	GCM Encryption Counter Value Register	AES_CTRR	Read-only	–
0x9C	GCM H World Register 0	AES_GCMHR0	Read-write	–
0xA0	GCM H World Register 1	AES_GCMHR1	Read-write	–
0xA4	GCM H World Register 2	AES_GCMHR2	Read-write	–
0xA8	GCM H World Register 3	AES_GCMHR3	Read-write	–
0x70–0xFC	Reserved	–	–	–
0x100–0x124	Reserved for the PDC	–	–	–

### 41.7.1 AES Control Register

**Name:** AES\_CR

**Address:** 0x40000000

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	SWRST
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	START

- **START: Start Processing**

0: No effect

1: Starts manual encryption/decryption process.

- **SWRST: Software Reset**

0: No effect.

1: Resets the AES. A software triggered hardware reset of the AES interface is performed.

## 41.7.2 AES Mode Register

**Name:** AES\_MR

**Address:** 0x40000004

**Access:** Read-write

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
CKEY				—	CFBS		
15	14	13	12	11	10	9	8
LOD	OPMOD			KEYSIZE		SMOD	
7	6	5	4	3	2	1	0
PROCDLY				DUALBUFF	—	GTAGEN	CIPHER

- **CIPHER: Processing Mode**

0: Decrypts data.

1: Encrypts data.

- **GTAGEN: GCM Automatic Tag Generation Enable**

0: Automatic GCM Tag generation disabled.

1: Automatic GCM Tag generation enabled.

- **DUALBUFF: Dual Input Buffer**

Value	Name	Description
0x0	INACTIVE	AES_IDATARx cannot be written during processing of previous block.
0x1	ACTIVE	AES_IDATARx can be written during processing of previous block when SMOD = 0x2. It speeds up the overall runtime of large files.

- **PROCDLY: Processing Delay**

$$\text{Processing Time} = 12 \times (\text{PROCDLY} + 1)$$

The Processing Time represents the number of clock cycles that the AES needs in order to perform one encryption/decryption.

Note: The best performance is achieved with PROCDLY equal to 0.

- **SMOD: Start Mode**

Value	Name	Description
0x0	MANUAL_START	Manual Mode
0x1	AUTO_START	Auto Mode
0x2	IDATAR0_START	AES_IDATAR0 access only Auto Mode

Values which are not listed in the table must be considered as “reserved”.

If a PDC transfer is used, it is mandatory to set SMOD to 0x2.



- **KEYSIZE: Key Size**

Value	Name	Description
0x0	AES128	AES Key Size is 128 bits
0x1	AES192	AES Key Size is 192 bits
0x2	AES256	AES Key Size is 256 bits

Values which are not listed in the table must be considered as “reserved”.

- **OPMOD: Operation Mode**

Value	Name	Description
0x0	ECB	ECB: Electronic Code Book mode
0x1	CBC	CBC: Cipher Block Chaining mode
0x2	OFB	OFB: Output Feedback mode
0x3	CFB	CFB: Cipher Feedback mode
0x4	CTR	CTR: Counter mode (16-bit internal counter)
0x5	GCM	GCM: Galois Counter mode

Values which are not listed in the table must be considered as “reserved”.

For CBC-MAC operating mode, please set OPMOD to CBC and LOD to 1.

- **LOD: Last Output Data Mode**

0: No effect.

After each end of encryption/decryption, the output data are available either on the output data registers (Manual and Auto modes) or at the address specified in the Receive Pointer Register (AES\_RPR) for PDC mode.

In Manual and Auto modes, the DATRDY flag is cleared when at least one of the Output Data registers is read.

1: The DATRDY flag is cleared when at least one of the Input Data Registers is written.

No more Output Data Register reads is necessary between consecutive encryptions/decryptions (see [“Last Output Data Mode” on page 932](#)).

**Warning:** In PDC mode, reading to the Output Data registers before the last data encryption/decryption process may lead to unpredictable results.

- **CFBS: Cipher Feedback Data Size**

Value	Name	Description
0x0	SIZE_128BIT	128-bit
0x1	SIZE_64BIT	64-bit
0x2	SIZE_32BIT	32-bit
0x3	SIZE_16BIT	16-bit
0x4	SIZE_8BIT	8-bit

Values which are not listed in table must be considered as “reserved”.

- **CKEY: Key**

Value	Name	Description
0xE	PASSWD	This field must be written with 0xE the first time that AES_MR is programmed. For subsequent programming of the AES_MR, any value can be written, including that of 0xE. Always reads as 0.

### 41.7.3 AES Interrupt Enable Register

**Name:** AES\_IER

**Address:** 0x40000010

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	URAD
7	6	5	4	3	2	1	0
–	–	–	TXBUFE	RXBUFF	ENDTX	ENDRX	DATRDY

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Enables the corresponding interrupt.

- **DATRDY: Data Ready Interrupt Enable**
- **ENDRX: End of Receive Buffer Interrupt Enable**
- **ENDTX: End of Transmit Buffer Interrupt Enable**
- **RXBUFF: Receive Buffer Full Interrupt Enable**
- **TXBUFE: Transmit Buffer Empty Interrupt Enable**
- **URAD: Unspecified Register Access Detection Interrupt Enable**

#### 41.7.4 AES Interrupt Disable Register

**Name:** AES\_IDR  
**Address:** 0x40000014  
**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	URAD
7	6	5	4	3	2	1	0
–	–	–	TXBUFE	RXBUFF	ENDTX	ENDRX	DATRDY

The following configuration values are valid for all listed bit names of this register:

0: No effect.

1: Disables the corresponding interrupt.

- **DATRDY: Data Ready Interrupt Disable**
- **ENDRX: End of Receive Buffer Interrupt Disable**
- **ENDTX: End of Transmit Buffer Interrupt Disable**
- **RXBUFF: Receive Buffer Full Interrupt Disable**
- **TXBUFE: Transmit Buffer Empty Interrupt Disable**
- **URAD: Unspecified Register Access Detection Interrupt Disable**

### 41.7.5 AES Interrupt Mask Register

**Name:** AES\_IMR

**Address:** 0x40000018

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	URAD
7	6	5	4	3	2	1	0
–	–	–	TXBUFE	RXBUFF	ENDTX	ENDRX	DATRDY

The following configuration values are valid for all listed bit names of this register:

0: The corresponding interrupt is not enabled.

1: The corresponding interrupt is enabled.

- **DATRDY: Data Ready Interrupt Mask**
- **ENDRX: End of Receive Buffer Interrupt Mask**
- **ENDTX: End of Transmit Buffer Interrupt Mask**
- **RXBUFF: Receive Buffer Full Interrupt Mask**
- **TXBUFE: Transmit Buffer Empty Interrupt Mask**
- **URAD: Unspecified Register Access Detection Interrupt Mask**

## 41.7.6 AES Interrupt Status Register

**Name:** AES\_ISR  
**Address:** 0x4000001C  
**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	TAGRDY
15	14	13	12	11	10	9	8
URAT				–	–	–	URAD
7	6	5	4	3	2	1	0
–	–	–	TXBUFE	RXBUFF	ENDTX	ENDRX	DATRDY

### • DATRDY: Data Ready

0: Output data not valid.

1: Encryption or decryption process is completed.

DATRDY is cleared when a Manual encryption/decryption occurs (START bit in AES\_CR) or when a software triggered hardware reset of the AES interface is performed (SWRST bit in AES\_CR).

**LOD = 0 (AES\_MR):**

In Manual and Auto mode, the DATRDY flag can also be cleared when at least one of the Output Data Registers is read.

In PDC mode, DATRDY is set and cleared automatically.

**LOD = 1 (AES\_MR):**

In Manual and Auto mode, the DATRDY flag can also be cleared when at least one of the Input Data Registers is written.

In PDC mode, DATRDY is set and cleared automatically.

### • ENDRX: End of RX Buffer

0: The Receive Counter Register has not reached 0 since the last write in AES\_RCR or AES\_RNCR.

1: The Receive Counter Register has reached 0 since the last write in AES\_RCR or AES\_RNCR.

Note: This flag must be used only in PDC mode with LOD bit cleared.

### • ENDTX: End of TX Buffer

0: The Transmit Counter Register has not reached 0 since the last write in AES\_TCR or AES\_TNCR.

1: The Transmit Counter Register has reached 0 since the last write in AES\_TCR or AES\_TNCR.

Note: This flag must be used only in PDC mode with LOD bit set.

### • RXBUFF: RX Buffer Full

0: AES\_RCR or AES\_RNCR has a value other than 0.

1: Both AES\_RCR and AES\_RNCR have a value of 0.

Note: This flag must be used only in PDC mode with LOD bit cleared.

- **TXBUFE: TX Buffer Empty**

0: AES\_TCR or AES\_TNCR has a value other than 0.

1: Both AES\_TCR and AES\_TNCR have a value of 0.

Note: This flag must be used only in PDC mode with LOD bit set.

- **URAD: Unspecified Register Access Detection Status**

0: No unspecified register access has been detected since the last SWRST.

1: At least one unspecified register access has been detected since the last SWRST.

URAD bit is reset only by the SWRST bit in the AES\_CR.

- **URAT: Unspecified Register Access:**

Value	Name	Description
0x0	IDR_WR_PROCESSING	Input Data Register written during the data processing when SMOD = 0x2 mode.
0x1	ODR_RD_PROCESSING	Output Data Register read during the data processing.
0x2	MR_WR_PROCESSING	Mode Register written during the data processing.
0x3	ODR_RD_SUBKGEN	Output Data Register read during the sub-keys generation.
0x4	MR_WR_SUBKGEN	Mode Register written during the sub-keys generation.
0x5	WOR_RD_ACCESS	Write-only register read access.

Only the last Unspecified Register Access Type is available through the URAT field.

URAT field is reset only by the SWRST bit in the AES\_CR.

- **TAGRDY: GCM Tag Ready**

0: GCM Tag is not valid.

1: GCM Tag generation is complete.

### 41.7.7 AES Key Word Register x

**Name:** AES\_KEYWRx

**Address:** 0x40000020

**Access:** Write-only

31	30	29	28	27	26	25	24
KEYW							
23	22	21	20	19	18	17	16
KEYW							
15	14	13	12	11	10	9	8
KEYW							
7	6	5	4	3	2	1	0
KEYW							

- **KEYW: Key Word**

The four/six/eight 32-bit Key Word registers set the 128-bit/192-bit/256-bit cryptographic key used for encryption/decryption.

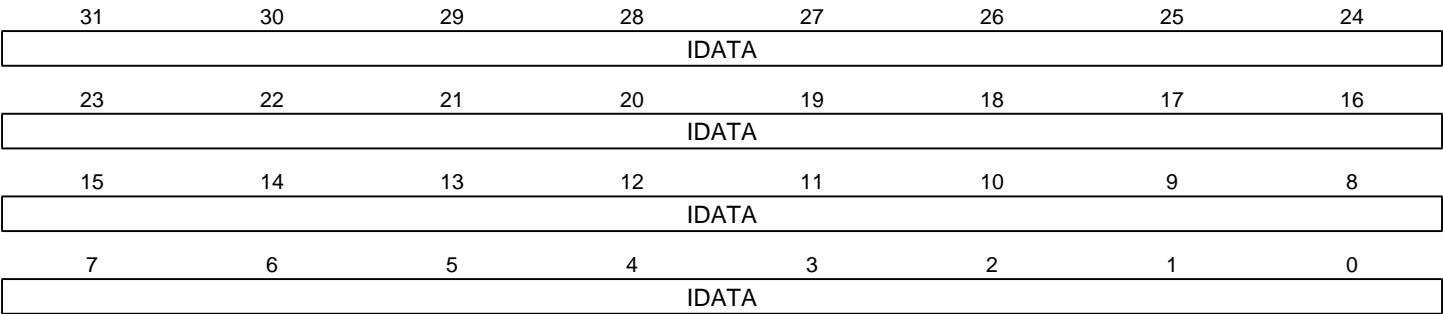
AES\_KEYWR0 corresponds to the first word of the key and respectively AES\_KEYWR3/AES\_KEYWR5/AES\_KEYWR7 to the last one.

These registers are write-only to prevent the key from being read by another application.



41.7.8 AES Input Data Register x

**Name:** AES\_IDATARx  
**Address:** 0x40000040  
**Access:** Write-only



• IDATA: Input Data Word

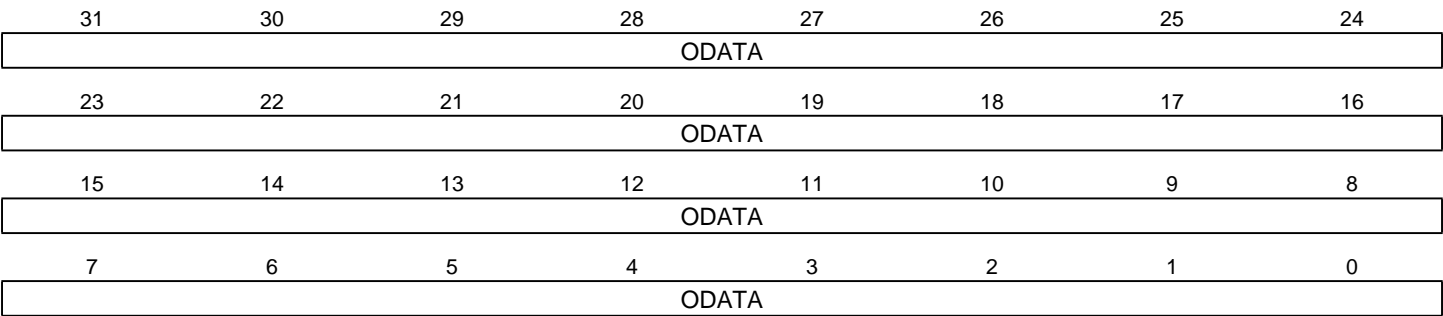
The four 32-bit Input Data registers set the 128-bit data block used for encryption/decryption. AES\_IDATAR0 corresponds to the first word of the data to be encrypted/decrypted, and AES\_IDATAR3 to the last one. These registers are write-only to prevent the input data from being read by another application.

41.7.9 AES Output Data Register x

Name: AES\_ODATARx

Address: 0x40000050

Access: Read-only



• ODATA: Output Data

The four 32-bit Output Data registers contain the 128-bit data block that has been encrypted/decrypted. AES\_ODATAR0 corresponds to the first word, AES\_ODATAR3 to the last one.

#### 41.7.10 AES Initialization Vector Register x

**Name:** AES\_IVRx  
**Address:** 0x40000060  
**Access:** Write-only

31	30	29	28	27	26	25	24
IV							
23	22	21	20	19	18	17	16
IV							
15	14	13	12	11	10	9	8
IV							
7	6	5	4	3	2	1	0
IV							

- **IV: Initialization Vector**

The four 32-bit Initialization Vector registers set the 128-bit Initialization Vector data block that is used by some modes of operation as an additional initial input.

AES\_IVR0 corresponds to the first word of the Initialization Vector, AES\_IVR3 to the last one.

These registers are write-only to prevent the Initialization Vector from being read by another application.

For CBC, OFB and CFB modes, the IV input value corresponds to the initialization vector.

For CTR mode, the IV input value corresponds to the initial counter value.

**Note:** These registers are not used in ECB mode and must not be written.

41.7.11 AES Additional Authenticated DataLength Register

Name: AES\_AADLENR

Address: 0x40000070

Access: Read-write



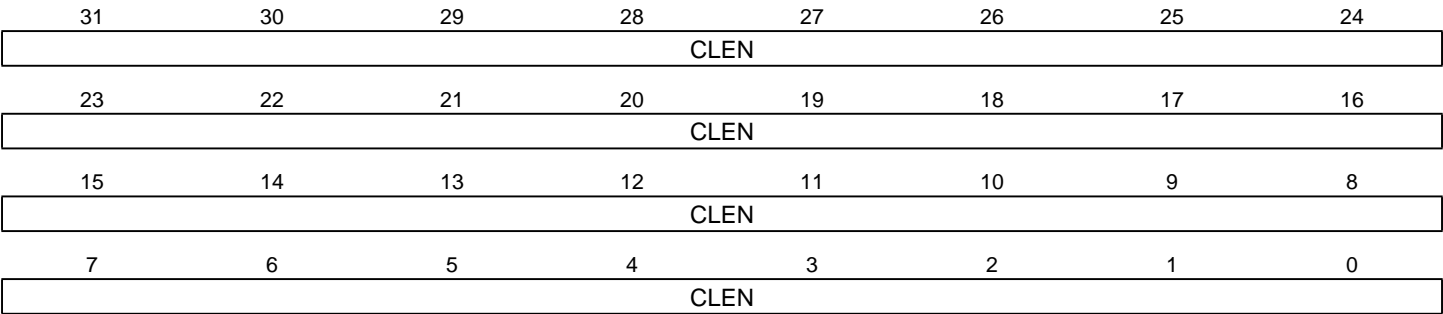
• AADLEN: AAD Length

Length in bytes of the AAD data that is to be processed.

Note: Note: The maximum byte length of the AAD portion of a message is limited to the 32-bit counter length.

41.7.12 AES Plaintext/Ciphertext Length Register

Name: AES\_CLENR  
Address: 0x40000074  
Access: Read-write



• CLEN: Plaintext/Ciphertext Length

Length in bytes of the plaintext/ciphertext (C) data that is to be processed.

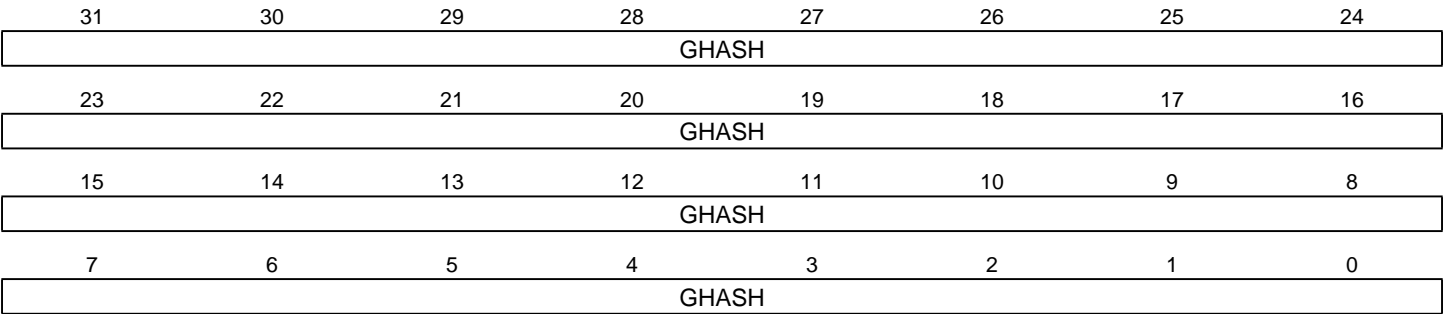
Note: Note: The maximum byte length of the C portion of a message is limited to the 32-bit counter length.

41.7.13 AES GCM Intermediate Hash Word Register x

Name: AES\_GHASHRx [x=0..3]

Address: 0x40000078

Access: Read-write



• GHASH: Intermediate GCM Hash Word x

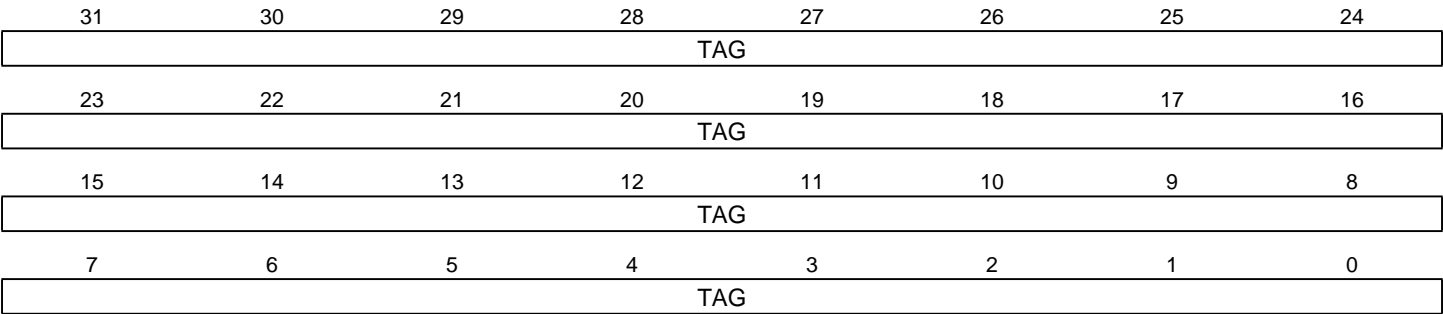
The four 32-bit Hash Word registers expose the intermediate GHASH value. May be read to save the current GHASH value so processing can later be resumed, presumably on a later message fragment. Must be written after writing the key and before issuing a “start” command on the next message fragment. Writing a new key to AES\_KEYWRx causes AES\_GHASHRx to be initialized with zeroes.

41.7.14 AES GCM Authentication Tag Word Register x

Name: AES\_TAGRx [x=0..3]

Address: 0x40000088

Access: Read-only



• TAG: GCM Authentication Tag x

The four 32-bit Tag registers contain the final 128-bit GCM Authentication tag “T” when GCM processing is complete. TAG0 corresponds to the first word, TAG3 to the last word.

41.7.15 AES GCM Encryption Counter Value Register

**Name:** AES\_CTRR  
**Address:** 0x40000098  
**Access:** Read-only

31	30	29	28	27	26	25	24
CTR							
23	22	21	20	19	18	17	16
CTR							
15	14	13	12	11	10	9	8
CTR							
7	6	5	4	3	2	1	0
CTR							

- **CTR: GCM Encryption Counter**  
Reports the current value of the 32-bit GCM counter.



#### 41.7.16 AES GCM H Word Register x

**Name:** AES\_GCMHRx [x=0..3]

**Address:** 0x4000009C

**Access:** Read-write

31	30	29	28	27	26	25	24
H							
23	22	21	20	19	18	17	16
H							
15	14	13	12	11	10	9	8
H							
7	6	5	4	3	2	1	0
H							

- **H: GCM H word x**

The four 32-bit H registers contain the 128-bit H value computed from the KEYW value. The H value can be overwritten by any value afterwards offering single GF128 multiplication possibilities.

## 42. Integrity Check Monitor (ICM)

### 42.1 Description

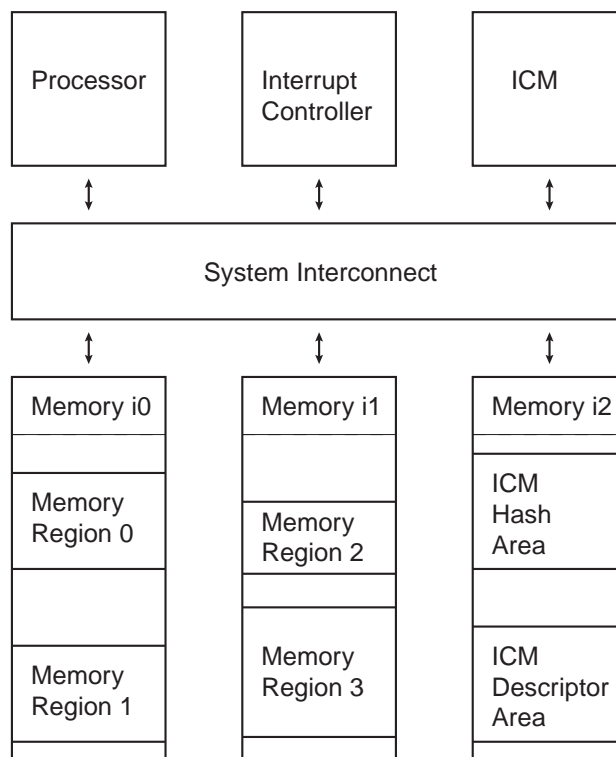
The Integrity Check Monitor (ICM) is an AHB DMA controller that performs hash calculation over multiple memory regions through the use of transfer descriptors located in memory (ICM Descriptor Area). The Hash function is based on the Secure Hash Algorithm (SHA). The ICM controller integrates two modes of operation. The first one is used to hash a list of memory regions and save the digests to memory (ICM Hash Area). The second operation mode is an active monitoring of the memory. In that mode, the hash function is evaluated and compared to the digest located at a predefined memory address (ICM Hash Area). If a mismatch occurs, an interrupt is raised. See [Figure 42-1](#) for an example of 4 regions active monitoring. Hash and Descriptor areas are located in Memory instance i2, and the four regions are split in memory instances i0 and i1.

The ICM SHA engine is compliant with the American *FIPS (Federal Information Processing Standard) Publication 180-2* specification.

The following terms are concise definitions of the ICM concepts used throughout this document:

- Region: A partition of instruction or data memory space.
- Region Descriptor: A data structure stored in memory, defining region attributes.
- Region Attributes: Region data start address, Region SHA engine processing mode, Write Back or Compare function mode.
- Context Registers: A set of registers containing the current configuration of the SHA engine and data address.
- Main List: A list of region descriptors. Each element associates the start address of a region with a set of attributes.
- Secondary List: A linked list defined on a per region basis that describes the memory layout of the region (when the region is non contiguous).

**Figure 42-1. Integrity Check Monitor Integrated in the System**



## 42.1.1 Product Dependencies

### 42.1.1.1 Power Management

The ICM may be clocked through the Power Management Controller (PMC), so the programmer must first configure the PMC to enable the ICM clock.

### 42.1.1.2 Interrupt

The ICM interface has an interrupt line connected to the Interrupt Controller.

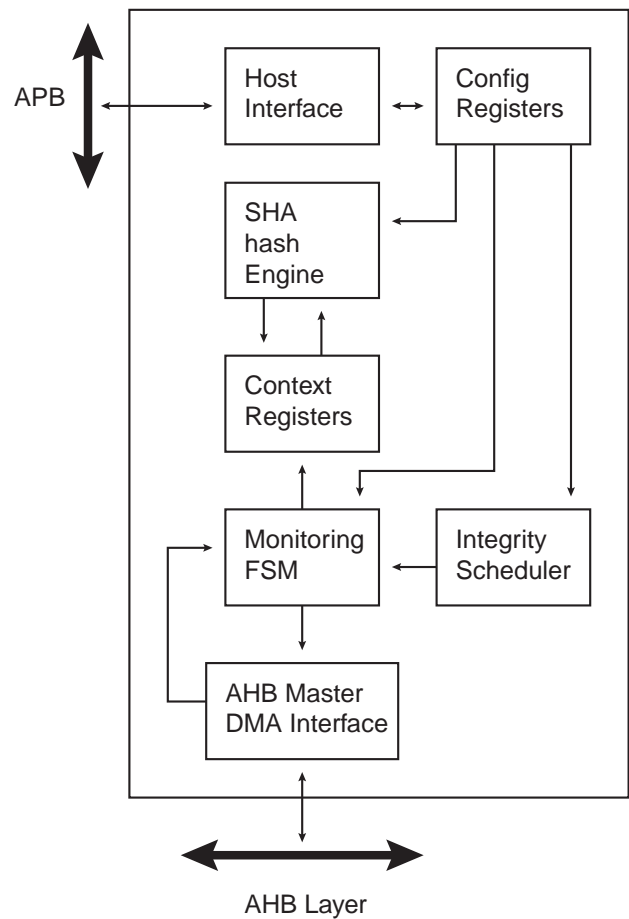
Handling the ICM interrupt requires programming the interrupt controller before configuring the ICM.

## 42.2 Embedded Characteristics

- DMA AHB master interface
- Supports up to 4 Non-Contiguous Memory Region Monitoring
- Supports block gathering through the use of linked list
- Supports Secure Hash Algorithm (SHA1, SHA224, SHA256, )
- Compliant with *FIPS Publication 180-2*
- Configurable Processing Period:
  - When SHA1 algorithm is processed, the runtime period is either 85 or 209 clock cycles.
  - When SHA256 or SHA224 algorithm is processed, the runtime period is either 72 or 194 clock cycles.
- Programmable Bus burden.

42.3 Block Diagram

Figure 42-2. Integrity Check Monitor Diagram

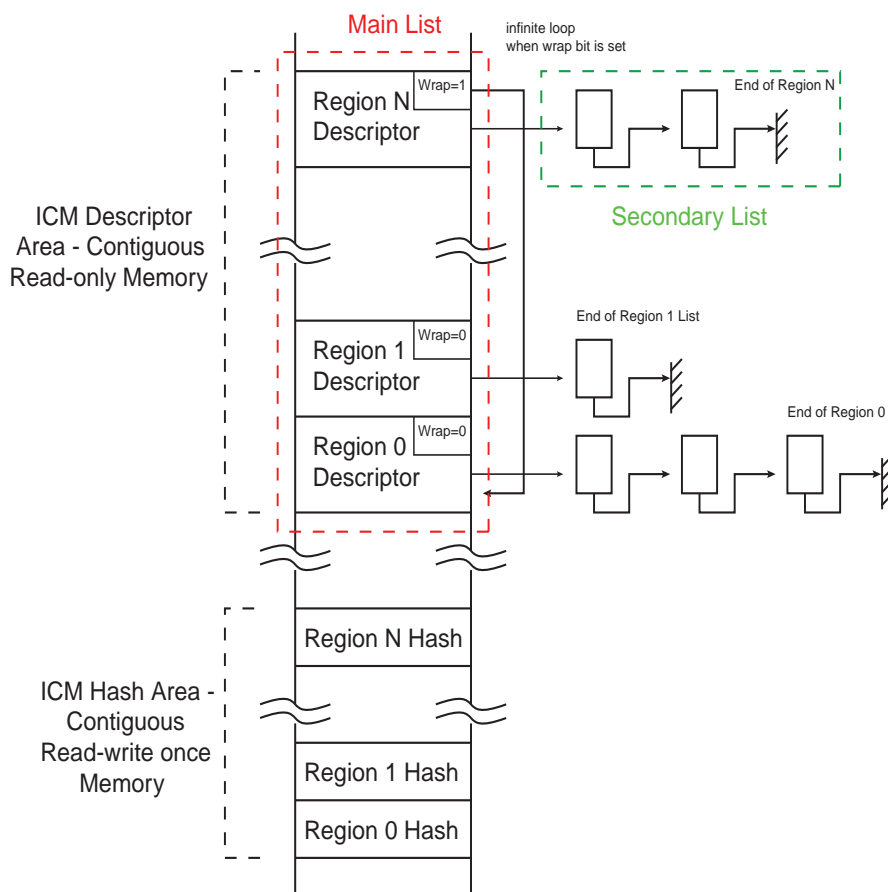


## 42.4 Functional Description

The Integrity Check Monitor (ICM) is a DMA controller that performs SHA-based memory hashing over memory regions. As shown in Figure 42-2, it integrates a DMA interface, a Monitoring Finite State Machine (FSM), an integrity scheduler, a set of context registers, a SHA engine, an APB interface and configuration registers.

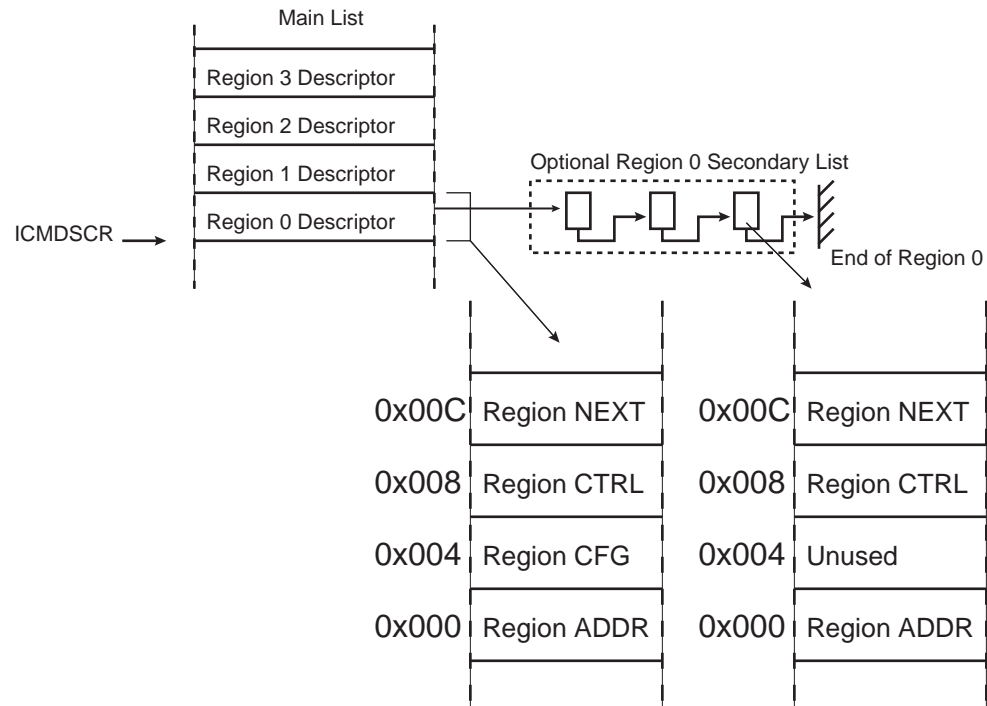
When the ICM module is enabled, it sequentially retrieves a circular list of region descriptors from the memory (Main List described in Figure 42-3). Up to 4 regions may be monitored. Each region descriptor is composed of four words indicating the layout of the memory region, see Figure 42-4. It also contains hashing engine configuration on a per region basis. As soon as the descriptor is loaded from the memory and context registers are updated with the data structure, the hashing operation starts. A programmable number of Blocks (see TRSIZE field of the ICM\_RCTRL structure member) is transferred from the memory to the SHA engine. When the desired number of blocks have been transferred, the digest is whether moved to memory (write-back function) or compared with a digest reference located in the system memory (compare function). If a digest mismatch occurs, an interrupt is triggered if unmasked. The ICM module walks through the Region descriptor list until the end of the list marked by an End of List bit set to one. To continuously monitor the list of regions, the Wrap bit must be set to one in the last data structure.

**Figure 42-3. ICM Region Descriptor and Hash Areas**



Each Region descriptor supports gathering of data through the use of the Secondary List. Unlike the Main List, the Secondary List cannot modify the configuration attributes of the Region. When the end of the Secondary List has been encountered, the ICM returns to the Main List. Memory integrity Monitoring can be considered as a background service and the mandatory bandwidth shall be very limited. In order to limit the ICM memory bandwidth, use the BBC field of the ICM\_CFG register to control ICM memory load.

Figure 42-4. Region Descriptor



ICM integrates a Secure Hash Algorithm Engine (SHA). This module requires a message padded according to FIPS180-2 specification. The SHA module produces an N-bit message digest each time a block is read and a processing period ends. N is 160 for SHA1, 224 for SHA224, 256 for SHA256.

#### 42.4.1 ICM Region Descriptor Structure

The ICM Region Descriptor Area is a contiguous Area of system memory that the controller and the processor can access. When the ICM controller is activated, the controller performs a descriptor fetch operation at  $*(ICM\_DSCR)$  address. If the Main List contains more than one descriptor (i.e. more than one region is to be monitored), the fetch address is  $*(ICM\_DSCR) + (RID \ll 4)$  where RID is the region identifier.

**Table 42-1. Region Descriptor Structure (Main List)**

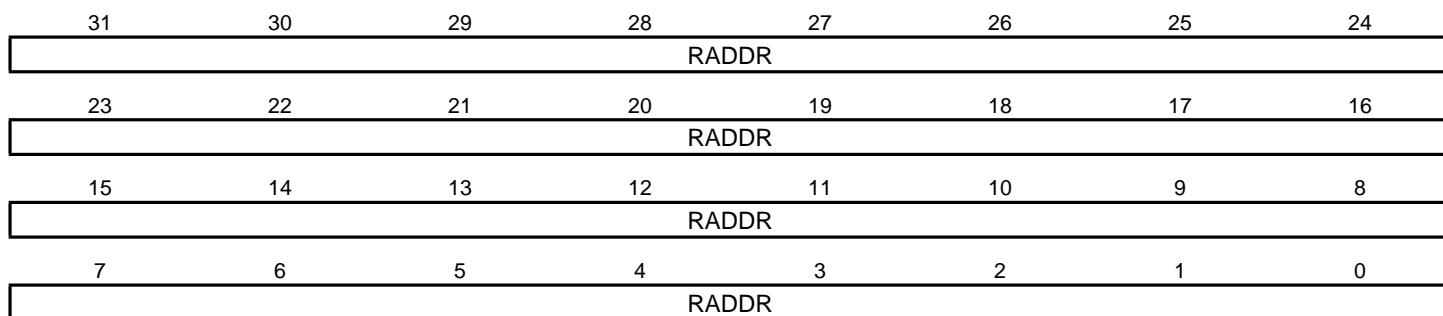
Offset	Structure Member	Name
$ICM\_DSCR + 0x000 + RID * (0x10)$	ICM Region Start Address	ICM_RADDR
$ICM\_DSCR + 0x004 + RID * (0x10)$	ICM Region Configuration	ICM_RCFG
$ICM\_DSCR + 0x008 + RID * (0x10)$	ICM Region Control	ICM_RCTRL
$ICM\_DSCR + 0x00C + RID * (0x10)$	ICM Region Next Address	ICM_RNEXT

#### 42.4.1.1 ICM Region Start Address Structure Member

**Name:** ICM\_RADDR

**Address:** ICM\_DSCR+0x000+RID\*(0x10)

**Access:** Read-write



- **RADDR: Region Start Address**

This field indicates the first byte address of the Region.



#### 42.4.1.2 ICM Region Configuration Structure Member

**Name:** ICM\_RCFG

**Address:** ICM\_DSCR+0x004+RID\*(0x10)

**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	MRPROT					
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	ALGO			–	PROCDLY	SUIEN	ECIEN
7	6	5	4	3	2	1	0
WCIEN	BEIEN	DMIEN	RHIEN	–	EOM	WRAP	CDWBN

- **CDWBN: Compare Digest or Write Back Digest**

0: The digest is written to the Hash area

1: The digest value is compared to the digest stored in the Hash area

- **WRAP: Wrap Command**

0: the next region descriptor address loaded is the current region identifier descriptor address incremented by 0x10

1: the next region descriptor address loaded is ICM\_DSCR.

- **EOM: End Of Monitoring**

0: The current descriptor does not terminate the monitoring.

1: The current descriptor terminates the Main List. WRAP bit value is discarded.

- **RHIEN: Region Hash Completed Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **DMIEN: Digest Mismatch Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **BEIEN: Bus Error Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **WCIEN: Wrap Condition Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **ECIEN: End Bit Condition Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **SUIEN: Monitoring Status Updated Condition Interrupt Enable**

0: Interrupt is disabled, the flag is not set when the condition is met.

1: Interrupt is enabled, the flag is set when the condition is met.

- **PROCDLY: Processing Delay**

Value	Name	Description
0	SHORTEST	SHA processing runtime is the shortest one
1	LONGEST	SHA processing runtime is the longest one

When SHA1 algorithm is processed, the runtime period is either 85 or 209 clock cycles.

When SHA256 or SHA224 algorithm is processed, the runtime period is either 72 or 194 clock cycles.

- **ALGO: SHA Algorithm**

Value	Name	Description
0	SHA1	SHA1 algorithm processed
1	SHA256	SHA256 algorithm processed
4	SHA224	SHA224 algorithm processed

Values which are not listed in the table must be considered as “reserved”.

- **MRPROT: Memory Region AHB Protection**

This field indicates the value of HPROT AHB signal when the ICM retrieves the memory region.

#### 42.4.1.3 ICM Region Control Structure Member

**Name:** ICM\_RCTRL

**Address:** ICM\_DSCR+0x008+RID\*(0x10)

**Access:** Read-write

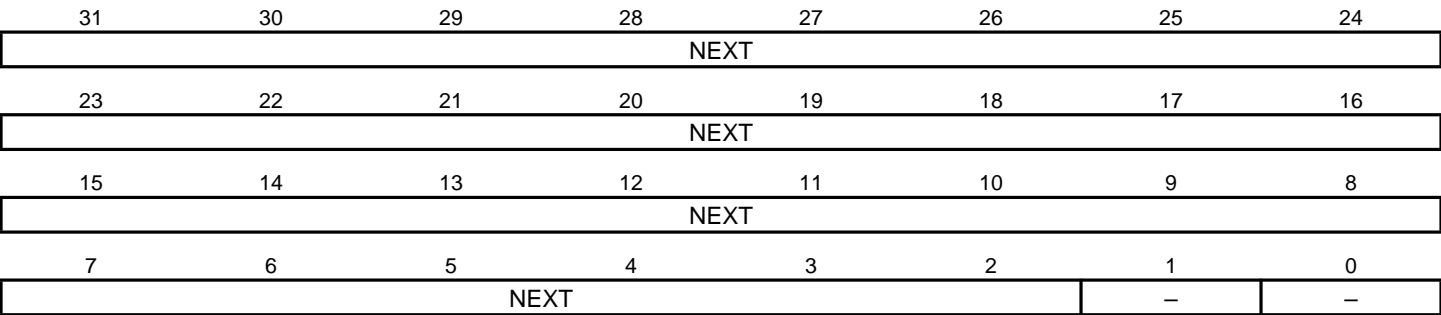
31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
TRSIZE							
7	6	5	4	3	2	1	0
TRSIZE							

- **TRSIZE: Transfer Size for the current chunk of data**

ICM performs a transfer of (TRSIZE+1) blocks of 512 bits.

42.4.1.4 ICM Region Next Address Structure Member

**Name:** ICM\_RNEXT  
**Address:** ICM\_DSCR+0x00C+RID\*(0x10)  
**Access:** Read-write



- **NEXT: Region Transfer Descriptor Next address**  
When set to 0, this field indicates that the current descriptor is the last descriptor of the Secondary List, otherwise it points at a new descriptor of the Secondary List.





### 42.4.3 ICM SHA Engine

The module can process SHA1, SHA224, SHA256, by means of a configuration field in the SHA\_MR register.

#### 42.4.3.1 Processing Period

The SHA engine processing period can be configured.

The short processing period allows to allocate bandwidth to the SHA module whereas the long processing period allocates more bandwidth on the system bus to other applications.

In SHA mode, the shortest processing period is 85 clock cycles + 2 clock cycles for start command synchronization. The longest period is 209 clock cycles + 2 clock cycles.

In SHA256 and SHA224 modes, the shortest processing period is 72 clock cycles + 2 clock cycles for start command synchronization. The longest period is 194 clock cycles + 2 clock cycles.

### 42.4.4 Security Features

When an undefined register access occurs, the URAD bit in the Interrupt Status Register (ICM\_ISR) raises if unmasked. Its source is then reported in the Undefined Access Status Register (ICM\_UASR). Only the first undefined register access is available through the URAT field.

Several kinds of unspecified register accesses can occur:

- Unspecified structure member set to one detected when the descriptor is loaded.
- Configuration register (ICM\_CFG) modified during active monitoring.
- Descriptor register (ICM\_DSCR) modified during active monitoring.
- Hash register (ICM\_HASH) modified during active monitoring.
- Write-only register read access.

The URAD bit and the URAT field can only be reset by the SWRST field in the ICM\_CTRL register.

### 42.4.5 ICM Automatic Monitoring Mode

The ASCD field of the ICM\_CFG register is used to activate the ICM Automatic Mode. When set to one, the ICM performs the following list of actions:

- The ICM controller walks through the Main List once, CDWBN context register is set to 0 internally (i.e. write back is activated), EOM context register is set to 0.
- When the Wrap condition is found, the ICM controller enters active monitoring. CDWBN field is now internally set to 1, EOM context register is set to 0. CDWBN and EOM fields of the descriptor structure are discarded.

## 42.5 Programming the ICM for Multiple Regions

Table 42-6. Region Attributes

Transfer Type	Main List	CDWBN Field	WRAP Field	EOM Field	NEXT Field	Comments
1) Single Region with contiguous list of blocks, digest written to memory, monitoring disabled	1 item	0	0	1	0	The Main List contains only one descriptor. The Secondary List is empty for that descriptor. The digest is computed and saved to memory.
2) Single Region with non contiguous list of blocks, digest written to memory, monitoring disabled	1 item	0	0	1	Secondary List address of the current region identifier	The Main List contains only one descriptor. The Secondary List describes the layout of the non contiguous region.
3) Single Region with contiguous list of blocks, digest comparison enabled, monitoring enabled	1 item	1	1	0	0	When the hash computation is terminated, the digest is compared with the one saved in memory.
4) Multiple Regions with contiguous list of blocks, digest written to memory, monitoring disabled.	More than one item	0	0	1 for the last, 0 otherwise	0	ICM walks through the list once.
5) Multiple Regions with contiguous list of blocks, digest comparison is enabled, monitoring is enabled.	More than one item	1	1 for the last, 0 otherwise	0	0	ICM performs active monitoring of the regions. If a mismatch occurs, an interrupt is raised.
6) Multiple Regions with non contiguous list of blocks, digest is written to memory, monitoring is disabled	More than one item	0	0	1	Secondary List address	ICM performs hashing and saves digests to the Hash area.
7) Multiple Regions with non contiguous list of blocks, digest comparison is enabled, monitoring is enabled	More than one item	1	1	0	Secondary List address	ICM performs data gathering on a per region basis



## 42.6 Integrity Check Monitor (ICM) User Interface

Table 42-7. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Configuration Register	ICM_CFG	Read-write	0x0
0x04	Control Register	ICM_CTRL	Write-only	–
0x08	Status Register	ICM_SR	Write-only	0x0
0x0C	Reserved	–	–	–
0x10	Interrupt Enable Register	ICM_IER	Write-only	–
0x14	Interrupt Disable Register	ICM_IDR	Write-only	–
0x18	Interrupt Mask Register	ICM_IMR	Read-only	0x0
0x1C	Interrupt Status Register	ICM_ISR	Read-only	0x0
0x20	Undefined Access Status Register	ICM_UASR	Read-only	0x0
0x24-0x2C	Reserved	–	–	–
0x30	Region Descriptor Area Start Address Register	ICM_DSCR	Read-write	0x0
0x34	Region Hash Area Start Address Register	ICM_HASH	Read-write	0x0
0x38	User Initial Hash Value 0 Register	ICM_UIHVAL0	Write-only	0x0
...	...	...	...	...
0x54	User Initial Hash Value 7	ICM_UIHVAL7	Write-only	0x0
0x94-0xFC	Reserved	–	–	–

## 42.6.1 ICM Configuration Register

**Name:** ICM\_CFG  
**Address:** 0x40044000  
**Access:** Read-write

31	30	29	28	27	26	25	24
–	–	DAPROT					
23	22	21	20	19	18	17	16
–	–	HAPROT					
15	14	13	12	11	10	9	8
UALGO			UIHASH	–	–	DUALBUFF	ASCD
7	6	5	4	3	2	1	0
BBC				–	SLBDIS	EOMDIS	WBDIS

- **WBDIS: Write Back Disable**

0: Write Back Operations are permitted.

1: Write Back Operations are forbidden. Context Register CDWBN is internally set to one and cannot be modified by a linked list element. The CDWBN field of the ICM\_RCFG structure member has no effect.

When ASCD field of the ICM\_CFG register is set, WBDIS field is discarded and has no effect.

- **EOMDIS: End of Monitoring Disable**

0: End of Monitoring is permitted

1: End of Monitoring is forbidden. The EOM field of the ICM\_RCFG structure member has no effect.

- **SLBDIS: Secondary List Branching Disable**

0: Branching to the Secondary List is permitted.

1: Branching to the Secondary List is forbidden. The NEXT field of the ICM\_RCFG structure member has no effect and is always considered as zero.

- **BBC: Bus Burden Control**

This field is used to control the burden of the ICM system bus. The number of system clock cycles between the end of the current processing and the next block transfer is set to  $2^{BBC}$ . Up to 32,768 cycles can be inserted.

- **ASCD: Automatic Switch To Compare Digest**

0: Automatic mode is disabled.

1: When this mode is enabled, the ICM controller automatically switches to active monitoring after the first Main List walk. Both CDWBN and WBDIS fields are discarded. The EOM field of ICM\_RCFG is taken into account.

- **DUALBUFF: Dual Input Buffer**

0: Dual Input buffer mode is disabled.

1: Dual Input buffer mode is enabled.

- **UIHASH: User Initial Hash Value**

0: When set to 0, the secure hash standard provides hash initial value.

1: When set to 1, the initial hash value is programmable. The ALGO field of the ICM\_RCFG structure member has no effect. UALGO is taken into account.

- **UALGO: User SHA Algorithm**

Value	Name	Description
0	SHA1	SHA1 algorithm processed
1	SHA256	SHA256 algorithm processed
4	SHA224	SHA224 algorithm processed

- **DAPROT: Region Descriptor Area Protection**

This field indicates the value of the HPROT AHB signal when the ICM module performs a read operation in the Region Descriptor Area.

- **HAPROT: Region Hash Area Protection**

This field indicates the value of the HPROT AHB signal when the ICM module performs a read operation in the Region Hash Area.

## 42.6.2 ICM Control Register

**Name:** ICM\_CTRL

**Address:** 0x40044004

**Access:** Write-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
RMEN				RMDIS			
7	6	5	4	3	2	1	0
REHASH				—	SWRST	DISABLE	ENABLE

- **ENABLE: ICM Enable**

0: No effect

1: When set to one, the ICM controller is activated.

- **DISABLE: ICM Disable Register**

0: No effect

1: The ICM controller is disabled. If a region is active, this region is terminated.

- **SWRST: Software Reset**

0: No effect.

1: Resets the ICM controller.

- **REHASH: Recompute Internal Hash**

0: No effect.

1: When REHASH[*i*] is set to one, Region *i* digest is re-computed. This bit is only available when Region monitoring is disabled.

- **RMDIS: Region Monitoring Disable**

0: No effect

1: When bit RMDIS[*i*] is set to one, the monitoring of Region with identifier *i* is disabled.

- **RMEN: Region Monitoring Enable**

0: No effect

1: When bit RMEN[*i*] is set to one, the monitoring of Region with identifier *i* is activated.

Monitoring is activated by default.

### 42.6.3 ICM Status Register

**Name:** ICM\_SR

**Address:** 0x40044008

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	—
23	22	21	20	19	18	17	16
—	—	—	—	—	—	—	—
15	14	13	12	11	10	9	8
RMDIS				RAWRMDIS			
7	6	5	4	3	2	1	0
—	—	—	—	—	—	—	ENABLE

- **ENABLE: ICM Controller Enable Register**

0: ICM controller is disabled.

1: ICM Controller is activated.

- **RAWRMDIS: RAW Region Monitoring Disabled Status**

0: RAW Region Monitoring is activated

1: RAW Region Monitoring is deactivated

- **RMDIS: Region Monitoring Disabled Status**

0: Region Monitoring is activated

1: Region Monitoring is deactivated

#### 42.6.4 ICM Interrupt Enable Register

**Name:** ICM\_IER

**Address:** 0x40044010

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	URAD
23	22	21	20	19	18	17	16
RSU				REC			
15	14	13	12	11	10	9	8
RWC				RBE			
7	6	5	4	3	2	1	0
RDM				RHC			

- **RHC: Region Hash Completed Interrupt Enable**

0: No effect

1: When RHC[*i*] is set to one, the Region *i* hash completed interrupt is enabled.

- **RDM: Region Digest Mismatch Interrupt Enable**

0: No effect

1: When RDM[*i*] is set to one, the Region *i* digest mismatch interrupt is enabled.

- **RBE: Region Bus Error Interrupt Enable**

0: No effect.

1: When RBE[*i*] is set to one, the Region *i* Bus Error interrupt is enabled.

- **RWC: Region Wrap Condition detected Interrupt Enable**

0: No effect.

1: When RWC[*i*] is set to one, the Region *i* Wrap Condition interrupt is enabled.

- **REC: Region End bit Condition Detected Interrupt Enable**

0: No effect.

1: When REC[*i*] is set to one, the region *i* End bit Condition interrupt is enabled.

- **RSU: Region Status Updated Interrupt Disable**

0: No effect.

1: When RSU[*i*] is set to one, the region *i* Status Updated interrupt is enabled.

- **URAD: Undefined Register Access Detection Interrupt Enable**

0: No effect.

1: The Undefined Register Access interrupt is enabled.

## 42.6.5 ICM Interrupt Disable Register

**Name:** ICM\_IDR

**Address:** 0x40044014

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	URAD
23	22	21	20	19	18	17	16
RSU				REC			
15	14	13	12	11	10	9	8
RWC				RBE			
7	6	5	4	3	2	1	0
RDM				RHC			

- **RHC: Region Hash Completed Interrupt Disable**

0: No effect

1: When RHC[*i*] is set to one, the Region *i* hash completed interrupt is disabled.

- **RDM: Region Digest Mismatch Interrupt Disable**

0: No effect

1: When RDM[*i*] is set to one, the Region *i* digest mismatch interrupt is disabled.

- **RBE: Region Bus Error Interrupt Disable**

0: No effect.

1: When RBE[*i*] is set to one, the Region *i* Bus Error interrupt is disabled.

- **RWC: Region Wrap Condition Detected Interrupt Disable**

0: No effect.

1: When RWC[*i*] is set to one, the Region *i* Wrap Condition interrupt is disabled.

- **REC: Region End bit Condition detected Interrupt Disable**

0: No effect.

1: When REC[*i*] is set to one, the region *i* End bit Condition interrupt is disabled.

- **RSU: Region Status Updated Interrupt Disable**

0: No effect.

1: When RSU[*i*] is set to one, the region *i* Status Updated interrupt is disabled.

- **URAD: Undefined Register Access Detection Interrupt Disable**

0: No effect.

1: Undefined Register Access detection interrupt is disabled.

## 42.6.6 ICM Interrupt Mask Register

**Name:** ICM\_IMR

**Address:** 0x40044018

**Access:** Read-only

31	30	29	28	27	26	25	24
—	—	—	—	—	—	—	URAD
23	22	21	20	19	18	17	16
RSU				REC			
15	14	13	12	11	10	9	8
RWC				RBE			
7	6	5	4	3	2	1	0
RDM				RHC			

- **RHC: Region Hash Completed Interrupt Mask**

0: When RHC[i] is set to zero, the interrupt is disabled for region i.

1: When RHC[i] is set to one, the interrupt is enabled for region i.

- **RDM: Region Digest Mismatch Interrupt Mask**

0: When RDM[i] is set to zero, the interrupt is disabled for region i.

1: When RDM[i] is set to one, the interrupt is enabled for region i.

- **RBE: Region Bus Error Interrupt Mask**

0: When RBE[i] is set to zero, the interrupt is disabled for region i.

1: When RBE[i] is set to one, the interrupt is enabled for region i.

- **RWC: Region Wrap Condition Detected Interrupt Mask**

0: When RWC[i] is set to zero, the interrupt is disabled for region i.

1: When RWC[i] is set to one, the interrupt is enabled for region i.

- **REC: Region End bit Condition Detected Interrupt Mask**

0: When REC[i] is set to zero, the interrupt is disabled for region i.

1: When REC[i] is set to one, the interrupt is enabled for region i.

- **RSU: Region Status Updated Interrupt Mask**

0: When RSU[i] is set to zero, the interrupt is disabled for region i.

1: When RSU[i] is set to one, the interrupt is enabled for region i.

- **URAD: Undefined Register Access Detection Interrupt Mask**

0: Interrupt is disabled

1: Interrupt is enabled.



## 42.6.7 ICM Interrupt Status Register

**Name:** ICM\_ISR  
**Address:** 0x4004401C  
**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	URAD
23	22	21	20	19	18	17	16
RSU				REC			
15	14	13	12	11	10	9	8
RWC				RBE			
7	6	5	4	3	2	1	0
RDM				RHC			

- **RHC: Region Hash Completed**

When RHC[*i*] is set, it indicates that the ICM has completed the region with identifier *i*.

- **RDM: Region Digest Mismatch**

When RDM[*i*] is set, it indicates that there is a digest comparison mismatch between the hash value of the region with identifier *i* and the reference value located in the Hash Area.

- **RBE: Region Bus Error**

When RBE[*i*] is set, it indicates that a bus error has been detected while hashing memory region *i*.

- **RWC: Region Wrap Condition Detected**

When RWC[*i*] is set, it indicates that a wrap condition has been detected.

- **REC: Region End bit Condition Detected**

When REC[*i*] is set, it indicates that an end bit condition has been detected.

- **RSU: Region Status Updated Detected**

When RSU[*i*] is set, it indicates that a region status updated condition has been detected.

- **URAD: Undefined Register Access Detection Status**

0: No undefined register access has been detected since the last SWRST.

1: At least one undefined register access has been detected since the last SWRST.

The URAD bit is only reset by the SWRST bit in the ICM\_CTRL control register.

The URAT field indicates the unspecified access type.

#### 42.6.8 ICM Undefined Access Status Register

**Name:** ICM\_UASR

**Address:** 0x40044020

**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	URAT		

- **URAT: Undefined Register Access Trace**

Value	Name	Description
0	UNSPEC_STRUCT_MEMBER	Unspecified structure member set to one detected when the descriptor is loaded.
1	ICM_CFG_MODIFIED	ICM_CFG modified during active monitoring.
2	ICM_DSCR_MODIFIED	ICM_DSCR modified during active monitoring.
3	ICM_HASH_MODIFIED	ICM_HASH modified during active monitoring
4	READ_ACCESS	Write-only register read access

Only the first Undefined Register Access Trace is available through the URAT field.

The URAT field is only reset by the SWRST bit in the ICM\_CTRL register.

### 42.6.9 ICM Descriptor Area Start Address Register

**Name:** ICM\_DSCR  
**Address:** 0x40044030  
**Access:** Read-write

31	30	29	28	27	26	25	24
DASA							
23	22	21	20	19	18	17	16
DASA							
15	14	13	12	11	10	9	8
DASA							
7	6	5	4	3	2	1	0
DASA	–	–	–	–	–	–	–

- **DASA: Descriptor Area Start Address**

The start address is a multiple of the total size of the data structure (64 bytes).

### 42.6.10 ICM Hash Area Start Address Register

**Name:** ICM\_HASH  
**Address:** 0x40044034  
**Access:** Read-write

31	30	29	28	27	26	25	24
HASA							
23	22	21	20	19	18	17	16
HASA							
15	14	13	12	11	10	9	8
HASA							
7	6	5	4	3	2	1	0
HASA	–	–	–	–	–	–	–

- **HASA: Hash Area Start Address**

This field points at the Hash memory location. The address must be a multiple of 128 bytes.

#### 42.6.11 ICM User Initial Hash Value Register

**Name:** ICM\_UIHVALx [x=0..7]

**Address:** 0x40044038

**Access:** Write-only

31	30	29	28	27	26	25	24
VAL							
23	22	21	20	19	18	17	16
VAL							
15	14	13	12	11	10	9	8
VAL							
7	6	5	4	3	2	1	0
VAL							

- **VAL: Initial Hash Value**

When UIHASH field of IMC\_CFG register is set, the Initial Hash Value is user-programmable.

To meet the desired standard, use the following example values.

For ICM\_UIHVAL0 bitfield:

Example	Comment
0x67452301	SHA1 algorithm
0xC1059ED8	SHA224 algorithm
0x6A09E667	SHA256 algorithm

For ICM\_UIHVAL1 bitfield:

Example	Comment
0xEFCDAB89	SHA1 algorithm
0x367CD507	SHA224 algorithm
0xBB67AE85	SHA256 algorithm

For ICM\_UIHVAL2 bitfield:

Example	Comment
0x98BADCFE	SHA1 algorithm
0x3070DD17	SHA224 algorithm
0x3C6EF372	SHA256 algorithm

For ICM\_UIHVAL3 bitfield:

Example	Comment
0x10325476	SHA1 algorithm
0xF70E5939	SHA224 algorithm
0xA54FF53A	SHA256 algorithm

For ICM\_UIHVAL4 bitfield:

Example	Comment
0xC3D2E1F0	SHA1 algorithm
0xFFC00B31	SHA224 algorithm
0x510E527F	SHA256 algorithm

For ICM\_UIHVAL5 bitfield:

Example	Comment
0x68581511	SHA224 algorithm
0x9B05688C	SHA256 algorithm

For ICM\_UIHVAL6 bitfield:

Example	Comment
0x64F98FA7	SHA224 algorithm
0x1F83D9AB	SHA256 algorithm

For ICM\_UIHVAL7 bitfield:

Example	Comment
0xBEFA4FA4	SHA224 algorithm
0x5BE0CD19	SHA256 algorithm

Example of Initial Value for SHA-1 Algorithm

Register Address	Addr 0x03/ Byte Lane [31:24]	Addr 0x02 Byte Lane [23:16]	Addr 0x01 Byte Lane [15:8]	Addr 0x00 Byte Lane [7:0]
0x000 ICM_UIHVAL0	01	23	45	67
0x004 ICM_UIHVAL1	89	ab	cd	ef
0x008 ICM_UIHVAL2	fe	dc	ba	98
0x00C ICM_UIHVAL3	76	54	32	10
0x010 ICM_UIHVAL4	f0	e1	d2	c3

## 43. Classical Public Key Cryptography Controller (CPKCC)

### 43.1 Description

The Classical Public Key Cryptography Controller (CPKCC) is an Atmel macrocell that processes public key cryptography algorithm calculus in both  $GF(p)$  and  $GF(2^n)$  fields. The ROMed CPKCL, the Classical Public Key Cryptography Library, is the library built on the top of the CPKCC.

The Classical Public Key Cryptography Library includes complete implementation of the following public key cryptography algorithms:

- RSA, DSA:
  - Modular Exponentiation with CRT up to 6144 bits
  - Modular Exponentiation without CRT up to 5408 bits
  - Prime generation
  - Utilities: GCD/modular Inverse, Divide, Modular reduction, Multiply, ...
- Elliptic Curves:
  - ECDSA up to 1504 bits
  - Point Multiply,
  - Point Add/Doubling
  - Elliptic Curves in  $GF(p)$  or  $GF(2^n)$
  - Choice of the curves parameters so compatibility with NIST Curves or others.
- Deterministic Random Number Generation (DRNG ANSI X9.31) for DSA

## 43.2 Product Dependencies

### 43.2.1 Power Management

The CPKCC is not continuously clocked. The CPCKCC interface is clocked through the Power Management Controller (PMC).

### 43.2.2 Interrupt Sources

The CPKCC has an interrupt line connected to the Nested Vector Interrupt Controller (NVIC). Handling interrupts requires programming the NVIC before configuring the CPKCC.

## 43.3 Functional Description

The CPKCC macrocell is managed by the CPKCL Library available in the ROM memory of the SAM4C. The user interface of the CPKCC is not described in this chapter.

The usage description of the CPKCC and its associated Library is provided in a separate document. Contact an Atmel Sales Representative for further details.

## 44. True Random Number Generator (TRNG)

### 44.1 Description

The True Random Number Generator (TRNG) passes the American *NIST Special Publication 800-22 and Diehard Random Tests Suites*.

The TRNG may be used as an entropy source for seeding an NIST approved DRNG (Deterministic RNG) as required by FIPS PUB 140-2 and 140-3.

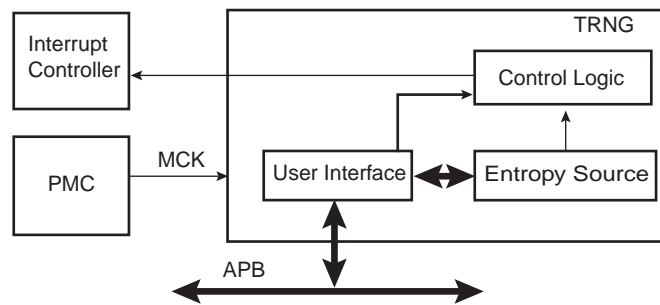
### 44.2 Embedded Characteristics

- Passed NIST Special Publication 800-22 Tests Suite
- Passed Diehard Random Tests Suite
- May be used as Entropy Source for seeding an NIST approved DRNG (Deterministic RNG) as required by FIPS PUB 140-2 and 140-3
- Provides a 32-bit Random Number Every 84 Clock Cycles



## 44.3 Block Diagram

Figure 44-1. TRNG Block Diagram



## 44.4 Product Dependencies

### 44.4.1 Power Management

The TRNG interface may be clocked through the Power Management Controller (PMC), thus the programmer must first configure the PMC to enable the TRNG user interface clock. The user interface clock is independent from any clock that may be used in the logic circuitry used for the source of entropy. The source of entropy can be enabled before enabling the user interface clock.

### 44.4.2 Interrupt

The TRNG interface has an interrupt line connected to the Interrupt Controller. In order to handle interrupts, the Interrupt Controller must be programmed before configuring the TRNG.

Table 44-1. Peripheral IDs

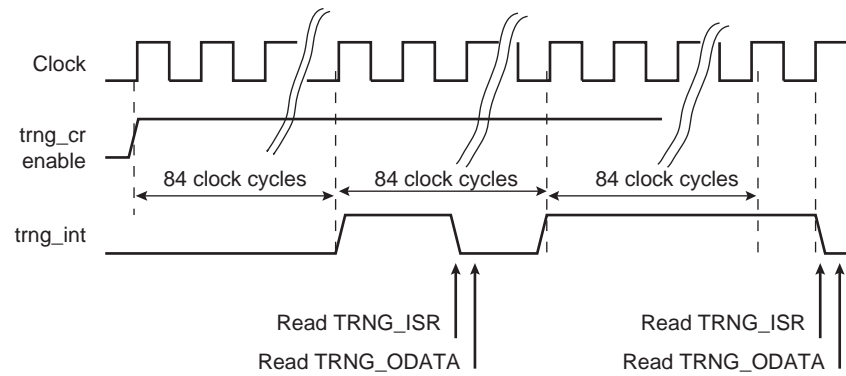
Instance	ID
TRNG	33

## 44.5 Functional Description

As soon as the TRNG is enabled (TRNG\_CTRL register), the generator provides one 32-bit value every 84 clock cycles. Interrupt `trng_int` can be enabled through the TRNG\_IER register (respectively disabled in TRNG\_IDR). This interrupt is set when a new random value is available and is cleared when the status register is read (TRNG\_SR register). The flag DATRDY of the status register (TRNG\_ISR) is set when the random data is ready to be read out on the 32-bit output data register (TRNG\_ODATA).

The normal mode of operation checks that the status register flag equals 1 before reading the output data register when a 32-bit random value is required by the software application.

**Figure 44-2. TRNG Data Generation Sequence**



## 44.6 True Random Number Generator (TRNG) User Interface

Table 44-2. Register Mapping

Offset	Register	Name	Access	Reset
0x00	Control Register	TRNG_CR	Write-only	–
0x10	Interrupt Enable Register	TRNG_IER	Write-only	–
0x14	Interrupt Disable Register	TRNG_IDR	Write-only	–
0x18	Interrupt Mask Register	TRNG_IMR	Read-only	0x0000_0000
0x1C	Interrupt Status Register	TRNG_ISR	Read-only	0x0000_0000
0x50	Output Data Register	TRNG_ODATA	Read-only	0x0000_0000

#### 44.6.1 TRNG Control Register

**Name:** TRNG\_CR

**Address:** 0x40048000

**Access:** Write-only

31	30	29	28	27	26	25	24
KEY							
23	22	21	20	19	18	17	16
KEY							
15	14	13	12	11	10	9	8
KEY							
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	ENABLE

- **ENABLE:** Enables the TRNG to provide random values

0: Disables the TRNG.

1: Enables the TRNG.

- **KEY: Security Key**

KEY = 0x524e47 (RNG in ASCII).

This key is to be written when the ENABLE bit is set or cleared.

## 44.6.2 TRNG Interrupt Enable Register

**Name:** TRNG\_IER

**Address:** 0x40048010

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	DATRDY

- **DATRDY: Data Ready Interrupt Enable**

0: No effect.

1: Enables the corresponding interrupt.

### 44.6.3 TRNG Interrupt Disable Register

**Name:** TRNG\_IDR

**Address:** 0x40048014

**Access:** Write-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	DATRDY

- **DATRDY: Data Ready Interrupt Disable**

0: No effect.

1: Disables the corresponding interrupt.

#### 44.6.4 TRNG Interrupt Mask Register

**Name:** TRNG\_IMR  
**Address:** 0x40048018  
**Reset:** 0x0000\_0000  
**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
–	–	–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	DATRDY

- **DATRDY: Data Ready Interrupt Mask**

0: The corresponding interrupt is not enabled.

1: The corresponding interrupt is enabled.

#### 44.6.5 TRNG Interrupt Status Register

**Name:** TRNG\_ISR  
**Address:** 0x4004801C  
**Reset:** 0x0000\_0000  
**Access:** Read-only

31	30	29	28	27	26	25	24
–	–	–	–	–	–	–	–
23	22	21	20	19	18	17	16
–	–	–	–	–	–	–	–
15	14	13	12	11	10	9	8
		–	–	–	–	–	–
7	6	5	4	3	2	1	0
–	–	–	–	–	–	–	DATRDY

- **DATRDY: Data Ready**

0: Output data is not valid or TRNG is disabled.

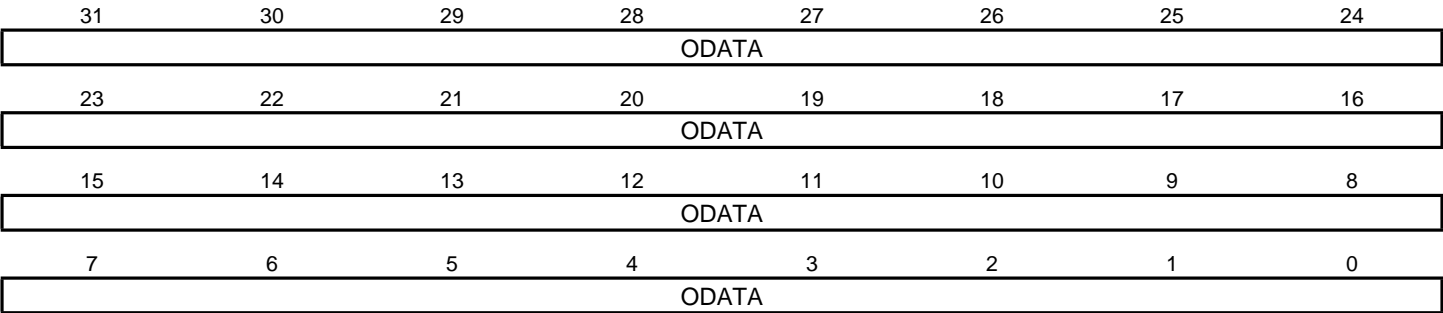
1: New Random value is completed.

DATRDY is cleared when this register is read.



44.6.6 TRNG Output Data Register

**Name:** TRNG\_ODATA  
**Address:** 0x40048050  
**Reset:** 0x0000\_0000  
**Access:** Read-only



- **ODATA: Output Data**  
The 32-bit Output Data register contains the 32-bit random data.

# 45. SAM4C Series Electrical Characteristics

## 45.1 Absolute Maximum Ratings

Table 45-1. Absolute Maximum Ratings\*

Storage Temperature.....	-60°C to + 150°C
Junction Temperature.....	-40°C to + 150°C
Voltage difference between two ground pins (among GND, GNDA and GNDREF).....	±50mV
Power Supply inputs with respect to ground pins:	
VDDCORE, VDDPLL.....	1.4V
VDDBU, VDDIO, VDDIN, VDDLCD.....	4.0V
Voltage on Digital Input Pins with Respect to Ground .....	-0.3V to VDDIO +0.3V
Total DC Output Current on all I/O lines 100-lead LQFP.....	100 mA

\*NOTICE: Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. **Exposure to absolute maximum rating conditions for extended periods may affect device reliability.**

## 45.2 Recommended Operating Conditions

**Table 45-2. Recommended Operating Conditions on Power supply Inputs**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
VDDCORE	Core logic power supply		1.08	1.20	1.32	V
VDDBU	Backup region power supply		1.62	3.3	3.6	V
VDDIO	I/Os power supply		1.62	3.3	3.6	V
VDDIN	Analog cells (Voltage Regulators, 10-bit ADC, temperature sensor) power supply		1.62	3.3	3.6	V
VDDLCD	LCD output buffers power supply		2.5	–	3.6	V
VDDPLL	PLLs and Main crystal oscillator power supply		1.08	–	1.32	V
$f_{MCK}$	Master clock frequency	VDDCORE @ 1.20V, $T_A = 85^\circ\text{C}$ VDDCORE @ 1.08V, $T_A = 85^\circ\text{C}$	–	–	120 100	MHz

**Table 45-3. Recommended Operating Conditions on Input Pins**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$AD[x]_{IN}$	Input voltage range on 10-bit ADC analog inputs	On AD[0..x]	0		min (VDDIN, VDDIO)	V
$V_{GPIO\_IN}$	Input voltage range on GPIOs	On any pin configured as a digital input	0		VDDIO	V

**Table 45-4. Recommended Thermal Operating Conditions**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$T_A$	Ambient temperature range		-40		+85	$^\circ\text{C}$
$T_J$	Junction temperature range		-40		100	
$R_{JA}$	Junction-to-ambient thermal resistance	LQFP 100 package		40		$^\circ\text{C}/\text{W}$
$P_D$	Power dissipation	$T_A = 70^\circ\text{C}$ $T_A = 85^\circ\text{C}$			750 375	mW

## 45.3 Electrical Parameters Usage

The tables that follow further on in [Section 45.4 “I/O Characteristics”](#), [Section 45.5 “Embedded Analog Peripherals Characteristics”](#), [Section 45.6 “Embedded Flash Characteristics”](#), and [Section 45.7 “Power Supply Current Consumption”](#) define the limiting values for several electrical parameters. Unless otherwise noted, these values are valid over the ambient temperature range  $T_A = [-40^{\circ}\text{C} + 85^{\circ}\text{C}]$ . Note that these limits may be affected by the board on which the MCU is mounted. Particularly, noisy supply and ground conditions must be avoided and care must be taken to provide:

- a PCB with a low impedance ground plane (unbroken ground planes are strongly recommended)
- low impedance decoupling of the MCU power supply inputs. A 100 nF Ceramic X7R (or X5R) capacitor placed very close to each power supply input is a minimum requirement. See special recommendations regarding integrated analog functions like voltage reference or voltage regulators in corresponding sections.
- low impedance power supply decoupling of external components. This recommendation aims at avoiding current spikes travelling into the PCB ground plane.

## 45.4 I/O Characteristics

### 45.4.1 I/O DC Characteristics

Table 45-5. I/O DC Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL}$	Low-level input voltage	$1.62V < V_{DDIO} < 3.6V$	–	–	$0.3 \times V_{DDIO}$	V
$V_{IH}$	High-level input voltage	$1.62V < V_{DDIO} < 3.6V$	$0.7 \times V_{DDIO}$	–	–	V
$V_{OH}$	High-level output voltage	$1.62V < V_{DDIO} < 3.6V$ $I_{OH} = 0$ $I_{OH} > 0$ (see $I_{OH}$ details below)	$V_{DDIO}$ $V_{DDIO} - 0.4$	–	–	V
$V_{OL}$	Low-level output voltage	$1.62V < V_{DDIO} < 3.6V$ $I_{OL} = 0$ $I_{OL} > 0$ (see $I_{OL}$ details below)	–	–	0 0.4	V
$I_{OH}$	High-level output current	PA0, PA8, PA29, PB13, PC0, PC5 pins	–	–	–	mA
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.4$	–	–	-7	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = V_{DDIO} - 0.4$	–	–	-7	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = V_{DDIO} - 0.4$	–	–	-11	
		Other GPIO pins, Low Drive				
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.4$			-3	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = V_{DDIO} - 0.4$			-5	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = V_{DDIO} - 0.4$			-6	
		Other GPIO pins, Medium Drive	–	–	–	
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.4$			-6	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = V_{DDIO} - 0.4$	–	–	-8	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = V_{DDIO} - 0.4$			-8	
		Relaxed Mode	–	–	–	
		PA0, PA8, PA29, PB13, PC0, PC5 pins				
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.6$			-12	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = 2.2V$	–	–	-22	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = 2.4V$			-26	
		Relaxed Mode	–	–	–	
		Other GPIO pins, Low Drive				
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.6$			-5	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = 2.2V$			-12	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = 2.4V$			-13	
		Relaxed Mode	–	–	–	
		Other GPIO pins, Medium Drive				
		$V_{DDIO} = 1.62V$ ; $V_{OH} = V_{DDIO} - 0.6$			-10	
		$V_{DDIO} = 3.3V$ ; $V_{OH} = 2.2V$	–	–	-20	
		$V_{DDIO} = 3.6V$ ; $V_{OH} = 2.4V$			-24	

Table 45-5. I/O DC Characteristics (Continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{OL}$	Low-level output current	1.62V < VDDIO < 3.6V; $V_{OL} = 0.4V$	–	–	–	mA
		PA0, PA8, PA29, PB13, PC0, PC5 pins			9	
		Other GPIO pins, Low Drive	–	–	8	
		Other GPIO pins, Medium Drive			10	
		Relaxed Mode: 1.62V < VDDIO < 3.6V; $V_{OL} = 0.6V$	–	–	–	
		PA0,PA8,PA29,PB13,PC0,PC5 pins			13	
		Other GPIO pins, Low Drive	–	–	10	
		Other GPIO pins, Medium Drive			13	
$V_{Hys}$	Hysteresis voltage	(Hysteresis mode enabled)	150	–	–	mV
$I_{IL}$	Input low leakage current	No pull-up or pull-down; $V_{IN} = GND$ ; $V_{DDIO}$ Max. (Typ: $T_A = 25^{\circ}C$ , Max: $T_A = 85^{\circ}C$ )	–			nA
		- PA0,PA8,PA29,PB13,PC0,PC5 pins		12	57	
		- PB16,PB17,PB18,PB19,PB20,PB21 pins		5	41	
		- Other pins		4	7	
$I_{IH}$	Input high leakage current	No pull-up or pull-down; $V_{IN} = VDD$ ; $V_{DDIO}$ Max. (Typ: $T_A = 25^{\circ}C$ , Max: $T_A = 85^{\circ}C$ )	–			nA
		- PA0, PA8, PA29, PB13, PC0, PC5 pins		15	150	
		- PB16, PB17, PB18, PB19, PB20, PB21 pins		1.7	9	
		- Other pins		5	14	
$R_{PULLUP}$	Pull-up resistor	Digital input mode	70	100	130	k $\Omega$
$R_{PULLDOWN}$	Pull-down resistor		70	100	130	k $\Omega$
$R_{ODT}$	On-die series termination resistor		–	36	–	$\Omega$
$C_{PAD}$	Input capacitance	I/O configured as digital input	–	–	5	pF

## 45.4.2 I/O AC Characteristics

Criteria used to define the maximum frequency of the I/Os:

- Output duty cycle (40%-60%)
- Minimum output swing: 100 mV to VDDIO - 100 mV
- Minimum output swing: 100 mV to VDDIO - 100 mV
- Addition of rising and falling time inferior to 75% of the period

**Table 45-6. I/O Characteristics**

Symbol	Parameter	Conditions		Min	Max	Unit
FreqMax1	Pin Group 1 <sup>(1)</sup> Maximum output frequency	10 pF	$V_{DDIO} = 3.3V$	—	70	MHz
		30 pF		—	45	
FreqMax2	Pin Group 2 <sup>(2)</sup> Maximum output frequency	10 pF	$V_{DDIO} = 3.3V$	—	35	MHz
		25 pF		—	15	
FreqMax3	Pin Group 3 <sup>(3)</sup> Maximum output frequency	10 pF	$V_{DDIO} = 3.3V$	—	70	MHz
		25 pF		—	35	

Notes: 1. Pin Group 1 = PA0, PA8, PA29, PB13, PC0, PC5 pins

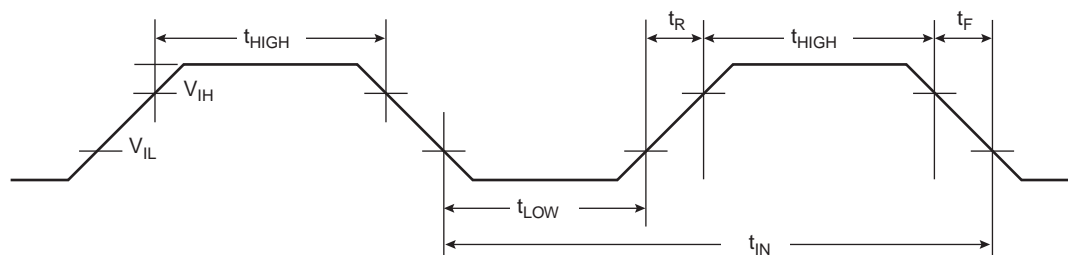
2. Other pins, low drive settings

3. Other pins, medium drive settings

Table 45-7 provides the input characteristics of the I/O lines. In particular, these values apply when the XIN input is used as a clock input of the device (oscillator set in bypass mode). They do not apply for the XIN32 input which is made for slow signals with frequencies up to 50 kHz.

**Table 45-7. Input Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Units
$f_{IN}$	Input frequency		—	—	50	MHz
$t_{IN}$	Input period		20	—	—	ns
$t_{HIGH}$	Time at high level		8	—	—	ns
$t_{LOW}$	Time at low level		8	—	—	ns
$t_R$	Rise time		—	—	2.2	ns
$t_F$	Fall time		—	—	2.2	ns



### 45.4.3 SPI Characteristics

Figure 45-1. SPI Master Mode with (CPOL= NCPHA = 0) or (CPOL= NCPHA= 1)

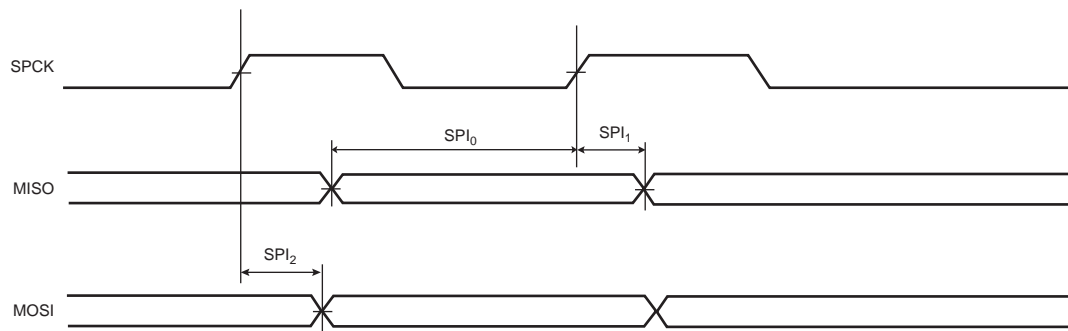


Figure 45-2. SPI Master Mode with (CPOL = 0 and NCPHA=1) or (CPOL=1 and NCPHA= 0)

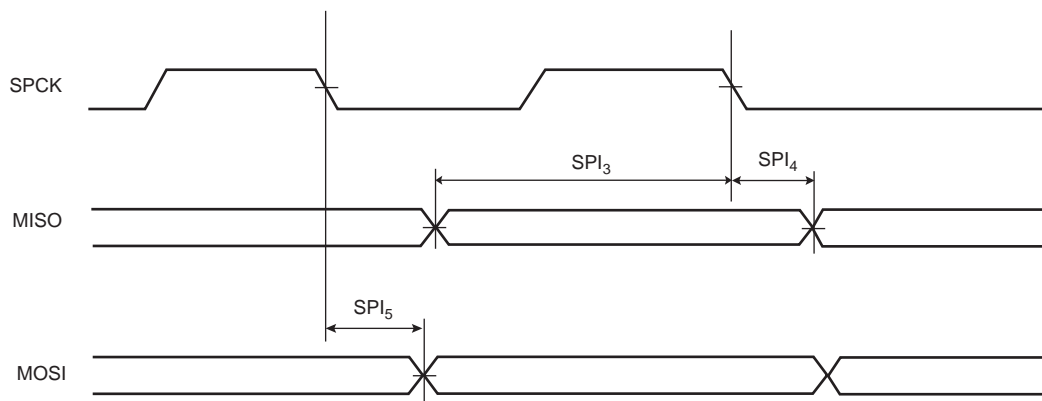
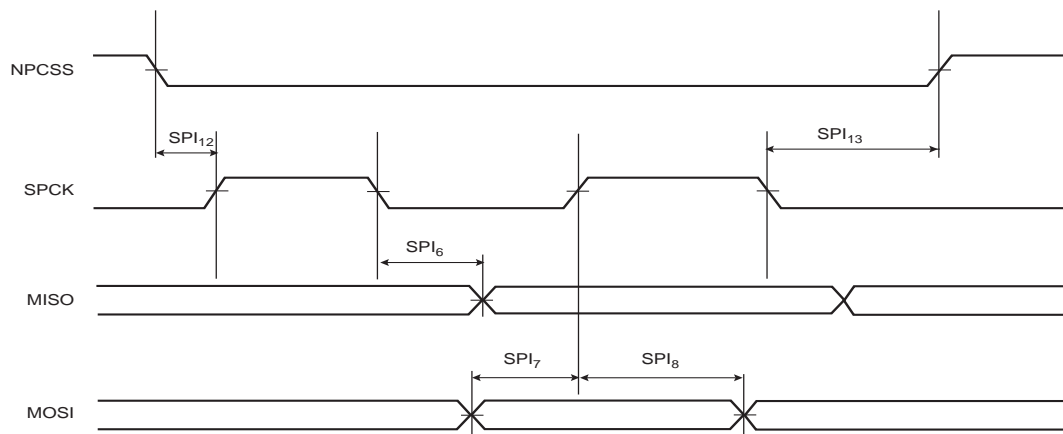
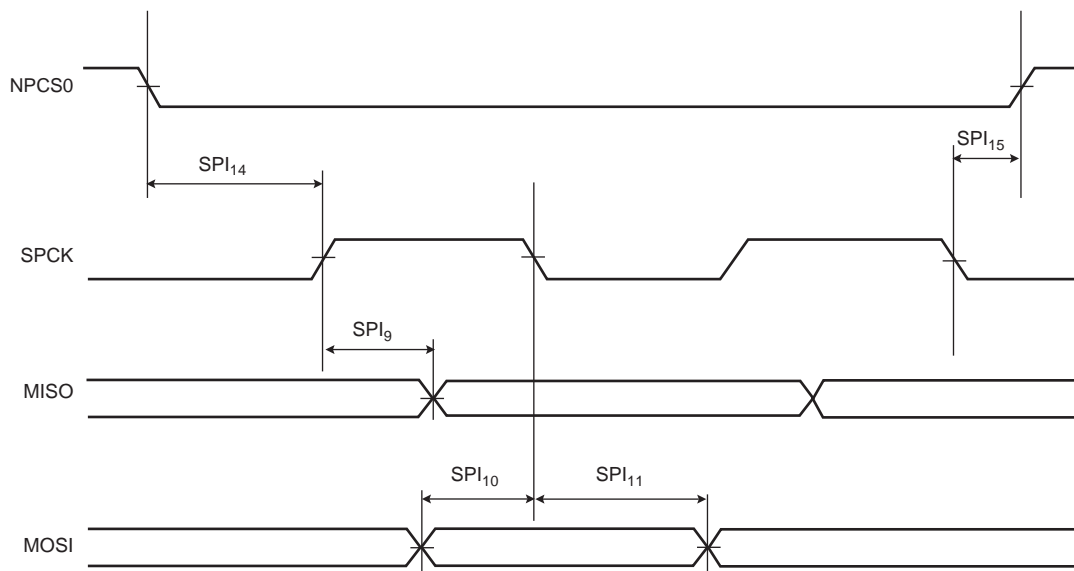


Figure 45-3. SPI Slave Mode with (CPOL=0 and NCPHA=1) or (CPOL=1 and NCPHA=0)





**Figure 45-4. SPI Slave Mode with (CPOL = NCPHA = 0) or (CPOL= NCPHA= 1)**



#### 45.4.3.1 Maximum SPI Frequency

The formulas that follow give maximum SPI frequency in Master write and read modes and in Slave read and write modes.

##### Master Write Mode

The SPI is only sending data to a slave device such as an LCD, for example. The limit is given by SPI<sub>2</sub> (or SPI<sub>5</sub>) timing. Since it gives a maximum frequency above the maximum pad speed (see [Section 45.4.2 "I/O AC Characteristics"](#)), the max SPI frequency is the one from the pad.

##### Master Read Mode

$$f_{SPCK}^{Max} = \frac{1}{SPI_0(orSPI_3) + t_{valid}}$$

t<sub>valid</sub> is the slave time response to output data after detecting an SPCK edge. For Atmel SPI DataFlash (AT45DB642D), t<sub>valid</sub> (or t<sub>v</sub>) is 12 ns Max.

In the formula above, f<sub>SPCK</sub>Max = 40 MHz @ VDDIO = 3.3V.

##### Slave Read Mode

In slave mode, SPCK is the input clock for the SPI. The max SPCK frequency is given by setup and hold timings SPI<sub>7</sub>/SPI<sub>8</sub>(or SPI<sub>10</sub>/SPI<sub>11</sub>). Since this gives a frequency well above the pad limit, the limit in slave read mode is given by SPCK pad.

##### Slave Write Mode

$$f_{SPCK}^{Max} = \frac{1}{2x(SPI_{6max}(orSPI_{9max}) + t_{setup})}$$

t<sub>setup</sub> is the setup time from the master before sampling data.

### 45.4.3.2 SPI Timings

**Table 45-8. SPI Timings**

Symbol	Parameter	Conditions <sup>(1)(2)</sup>	Min	Max	Unit
SPI <sub>0</sub>	MISO setup time before SPCK rises (master)	3.3V domain	12.7	–	ns
		1.8V domain	15.0	–	ns
SPI <sub>1</sub>	MISO hold time after SPCK rises (master)	3.3V domain	-3.9	–	ns
		1.8V domain	-4.6	–	ns
SPI <sub>2</sub>	SPCK rising to MOSI delay (master)	3.3V domain	-3.1	1.5	ns
		1.8V domain	-3.5	1.5	ns
SPI <sub>3</sub>	MISO setup time before SPCK falls (master)	3.3V domain	17.3	–	ns
		1.8V domain	20.0	–	ns
SPI <sub>4</sub>	MISO hold time after SPCK falls (master)	3.3V domain	-8.1	–	ns
		1.8V domain	-8.9	–	ns
SPI <sub>5</sub>	SPCK falling to MOSI delay (master)	3.3V domain	-8.7	-4.2	ns
		1.8V domain	-8.5	-4.2	ns
SPI <sub>6</sub>	SPCK falling to MISO delay (slave)	3.3V domain	3.9	13.4	ns
		1.8V domain	4.5	15.1	ns
SPI <sub>7</sub>	MOSI setup time before SPCK rises (slave)	3.3V domain	-0.7	–	ns
		1.8V domain	-0.5	–	ns
SPI <sub>8</sub>	MOSI hold time after SPCK rises (slave)	3.3V domain	3.9	–	ns
		1.8V domain	3.9	–	ns
SPI <sub>9</sub>	SPCK rising to MISO delay (slave)	3.3V domain	4.1	14.4	ns
		1.8V domain	4.6	15.2	ns
SPI <sub>10</sub>	MOSI setup time before SPCK falls (slave)	3.3V domain	0.1	–	ns
		1.8V domain	-0.3	–	ns
SPI <sub>11</sub>	MOSI hold time after SPCK falls (slave)	3.3V domain	2.8	–	ns
		1.8V domain	2.8	–	ns
SPI <sub>12</sub>	NPCS setup to SPCK rising (slave)	3.3V domain	2.1	–	ns
		1.8V domain	2.2	–	ns
SPI <sub>13</sub>	NPCS hold after SPCK falling (slave)	3.3V domain	-29.9	–	ns
		1.8V domain	-29.9	–	ns
SPI <sub>14</sub>	NPCS setup to SPCK falling (slave)	3.3V domain	3.4	–	ns
		1.8V domain	3.1	–	ns
SPI <sub>15</sub>	NPCS hold after SPCK falling (slave)	3.3V domain	-28.7	–	ns
		1.8V domain	-28.7	–	ns

Notes: 1. 3.3V domain: V<sub>VDDIO</sub> from 2.85 V to 3.6V, maximum external capacitor = 10 pF.

2. 1.8V domain: V<sub>VDDIO</sub> from 1.65V to 1.95V, maximum external capacitor = 10 pF.

Note that in SPI master mode, the MCU does not sample the data (MISO) on the opposite edge where data clocks out (MOSI) but the same edge is used as shown in [Figure 45-1](#) and [Figure 45-2](#).

#### 45.4.4 SMC Timings

Timings are given in the following domains:

- 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF
- 3.3V domain: VDDIO from 2.85V to 3.6V, maximum external capacitor = 10 pF

Timings are given assuming a capacitance load on data, control and address pads.

In tables that follow,  $t_{CPMCK}$  is MCK period.

##### 45.4.4.1 Read Timings

**Table 45-9. SMC Read Signals - NRD Controlled (READ\_MODE = 1)**

Symbol	Parameter	Min		Max		Unit
	VDDIO Supply	1.8V <sup>(1)</sup>	3.3V <sup>(2)</sup>			
<b>NO HOLD SETTINGS (nrd hold = 0)</b>						
SMC <sub>1</sub>	Data setup before NRD high	16.6	16.6	–	–	ns
SMC <sub>2</sub>	Data hold after NRD high	-6.7	-6.7	–	–	ns
<b>HOLD SETTINGS (nrd hold ≠ 0)</b>						
SMC <sub>3</sub>	Data setup before NRD high	10.5	10.5	–	–	ns
SMC <sub>4</sub>	Data hold after NRD high	-6.4	-6.4	–	–	ns
<b>HOLD or NO HOLD SETTINGS (nrd hold ≠ 0, nrd hold = 0)</b>						
SMC <sub>5</sub>	NBS0/A0, NBS1, A1 - A23 Valid before NRD high	(nrd setup + nrd pulse) * $t_{CPMCK} - 2$	(nrd setup + nrd pulse) * $t_{CPMCK} - 2$	–	–	ns
SMC <sub>6</sub>	NCS low before NRD high	(nrd setup + nrd pulse - ncs rd setup) * $t_{CPMCK} - 0.4$	(nrd setup + nrd pulse - ncs rd setup) * $t_{CPMCK} - 0.4$	–	–	ns
SMC <sub>7</sub>	NRD pulse width	nrd pulse * $t_{CPMCK} - 2.9$	nrd pulse * $t_{CPMCK} - 2.9$	–	–	ns

Notes: 1. 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF.

2. 3.3V domain: VDDIO from 3.0V to 3.6V, maximum external capacitor = 10 pF.

**Table 45-10. SMC Read Signals - NCS Controlled (READ\_MODE= 0)**

Symbol	Parameter	Min		Max		Unit
	VDDIO supply	1.8V <sup>(1)</sup>	3.3V <sup>(2)</sup>			
<b>NO HOLD SETTINGS (ncs rd hold = 0)</b>						
SMC <sub>8</sub>	Data setup before NCS high	28.3	28.3	–	–	ns
SMC <sub>9</sub>	Data hold after NCS high	-9.9	-9.9	–	–	ns
<b>HOLD SETTINGS (ncs rd hold ≠ 0)</b>						
SMC <sub>10</sub>	Data setup before NCS high	23	23	–	–	ns
SMC <sub>11</sub>	Data hold after NCS high	-9.5	-9.5	–	–	ns
<b>HOLD or NO HOLD SETTINGS (ncs rd hold ≠ 0, ncs rd hold = 0)</b>						
SMC <sub>12</sub>	NBS0/A0, NBS1, A1 - A23 valid before NCS high	(ncs rd setup + ncs rd pulse) * $t_{CPMCK} - 4$	(ncs rd setup + ncs rd pulse) * $t_{CPMCK} - 4$	–	–	ns
SMC <sub>13</sub>	NRD low before NCS high	(ncs rd setup + ncs rd pulse - nrd setup) * $t_{CPMCK}$ - 1.5	(ncs rd setup + ncs rd pulse - nrd setup) * $t_{CPMCK} - 1.5$	–	–	ns
SMC <sub>14</sub>	NCS pulse width	ncs rd pulse length * $t_{CPMCK} - 3.1$	ncs rd pulse length * $t_{CPMCK} - 3.1$	–	–	ns

Notes: 1. 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF.

2. 3.3V domain: VDDIO from 3.0V to 3.6V, maximum external capacitor = 10 pF.

#### 45.4.4.2 Write Timings

**Table 45-11. SMC Write Signals - NWE Controlled (WRITE\_MODE = 1)**

Symbol	Parameter	Min		Max		Unit
		1.8V <sup>(1)</sup>	3.3V <sup>(2)</sup>			
HOLD or NO HOLD SETTINGS (nwe hold ≠ 0, nwe hold = 0)						
SMC <sub>15</sub>	Data out valid before NWE high	nwe pulse * t <sub>CPMCK</sub> - 2.7	nwe pulse * t <sub>CPMCK</sub> - 2.7	—	—	ns
SMC <sub>16</sub>	NWE pulse width	nwe pulse * t <sub>CPMCK</sub> +9.5	nwe pulse * t <sub>CPMCK</sub> +9.5	—	—	ns
SMC <sub>17</sub>	NBS0/A0 NBS1, A1 - A23 valid before NWE low	nwe setup * t <sub>CPMCK</sub> -3.3	nwe setup * t <sub>CPMCK</sub> -3.3	—	—	ns
SMC <sub>18</sub>	NCS low before NWE high	(nwe setup - ncs rd setup + nwe pulse) * t <sub>CPMCK</sub> -0.5	(nwe setup - ncs rd setup + nwe pulse) * t <sub>CPMCK</sub> - 0.5	—	—	ns
HOLD SETTINGS (nwe hold ≠ 0)						
SMC <sub>19</sub>	NWE high to data OUT, NBS0/A0 NBS1, A1 - A23 change	nwe hold * t <sub>CPMCK</sub> - 3.9	nwe hold * t <sub>CPMCK</sub> - 3.9	—	—	ns
SMC <sub>20</sub>	NWE high to NCS Inactive <sup>(1)</sup>	(nwe hold - ncs wr hold)* t <sub>CPMCK</sub> - 2.3	(nwe hold - ncs wr hold)* t <sub>CPMCK</sub> - 2.3	—	—	ns
NO HOLD SETTINGS (nwe hold = 0)						
SMC <sub>21</sub>	NWE high to data OUT, NBS0/A0 NBS1, A1 - A23, NCS change <sup>(3)</sup>	5.5	5.5	—	—	ns

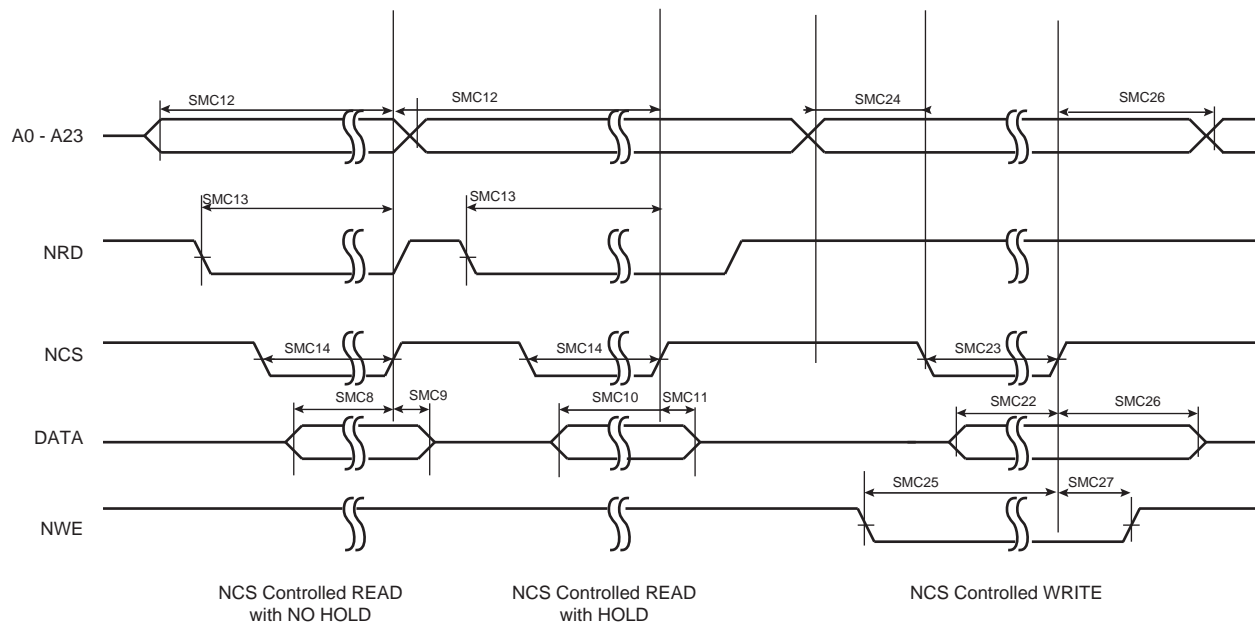
- Notes: 1. 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF.  
2. 3.3V domain: VDDIO from 3.0V to 3.6V, maximum external capacitor = 10 pF.  
3. hold length = total cycle duration - setup duration - pulse duration. "hold length" is for "ncs wr hold length" or "NWE hold length".

**Table 45-12. SMC Write NCS Controlled (WRITE\_MODE = 0)**

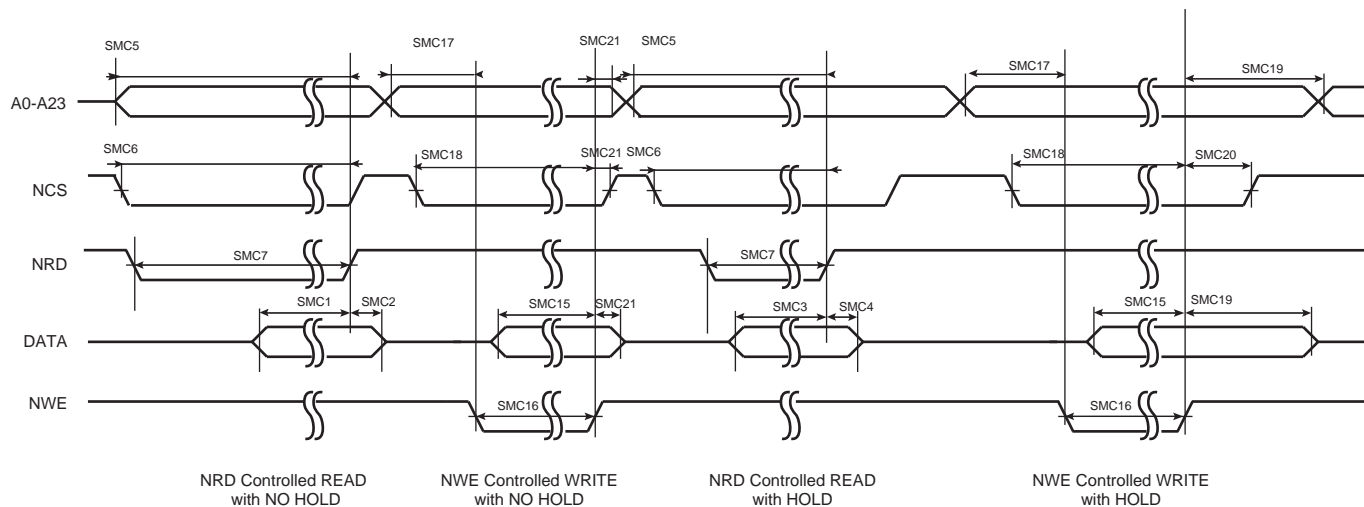
Symbol	Parameter	Min		Max		Units
		1.8V <sup>(1)</sup>	3.3V <sup>(2)</sup>			
SMC <sub>22</sub>	Data out valid before NCS High	ncs wr pulse * t <sub>CPMCK</sub> - 1.6	ncs wr pulse * t <sub>CPMCK</sub> - 1.6	—	—	ns
SMC <sub>23</sub>	NCS pulse width	ncs wr pulse * t <sub>CPMCK</sub> - 3.1	ncs wr pulse * t <sub>CPMCK</sub> - 3.1	—	—	ns
SMC <sub>24</sub>	NBS0/A0 NBS1, A1 - A23 valid before NCS low	ncs wr setup * t <sub>CPMCK</sub> - 4	ncs wr setup * t <sub>CPMCK</sub> - 4	—	—	ns
SMC <sub>25</sub>	NWE low before NCS high	(ncs wr setup - nwe setup + ncs pulse)* t <sub>CPMCK</sub> - 27.4	(ncs wr setup - nwe setup + ncs pulse)* t <sub>CPMCK</sub> - 17.8	—	—	ns
SMC <sub>26</sub>	NCS high to data out, NBS0/A0, NBS1, A1 - A23, change	ncs wr hold * t <sub>CPMCK</sub> - 7.6	ncs wr hold * t <sub>CPMCK</sub> - 7.6	—	—	ns
SMC <sub>27</sub>	NCS high to NWE inactive	(ncs wr hold - nwe hold)* t <sub>CPMCK</sub> - 7.8	(ncs wr hold - nwe hold)* t <sub>CPMCK</sub> - 7.8	—	—	ns

- Notes: 1. 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF.  
2. 3.3V domain: VDDIO from 3.0V to 3.6V, maximum external capacitor = 10 pF.

**Figure 45-5. SMC Timings - NCS Controlled Read and Write**



**Figure 45-6. SMC Timings - NRD Controlled Read and NWE Controlled Write**

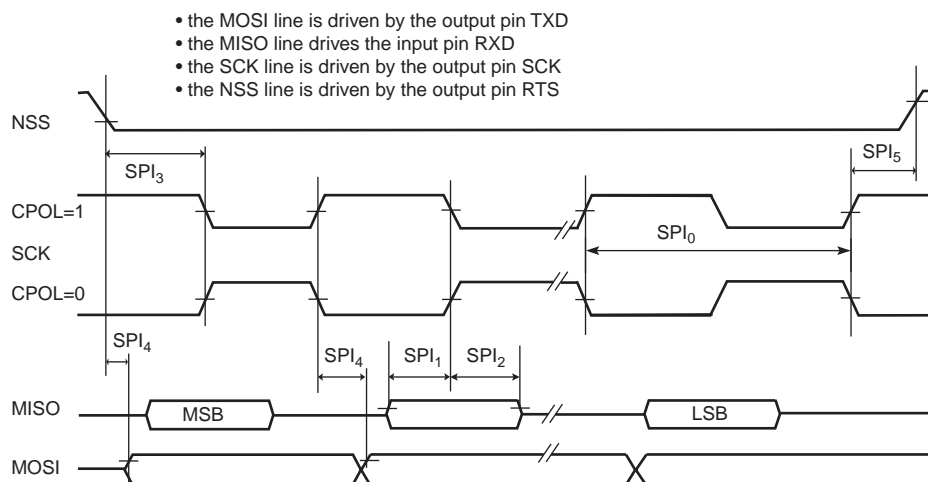


### 45.4.5 USART in SPI Mode Timings

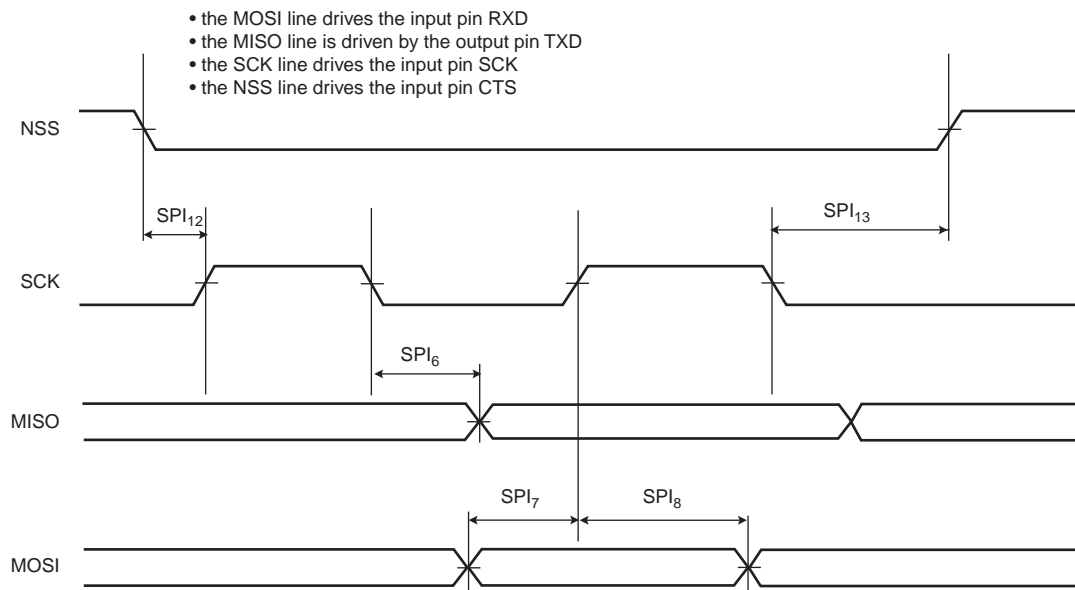
Timings are given in the following domains:

- 1.8V domain: VDDIO from 1.65V to 1.95V, maximum external capacitor = 10 pF
- 3.3V domain: VDDIO from 2.85V to 3.6V, maximum external capacitor = 10 pF

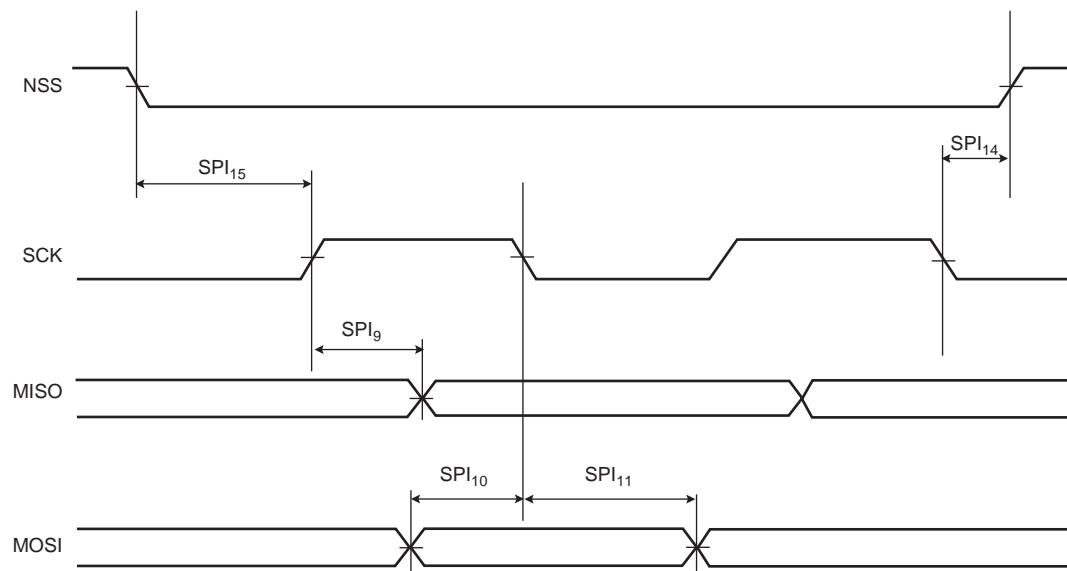
**Figure 45-7. USART SPI Master Mode**



**Figure 45-8. USART SPI Slave Mode: (Mode 1 or 2)**



**Figure 45-9. USART SPI Slave mode: (Mode 0 or 3)**





#### 45.4.5.1 USART SPI Timings

Table 45-13. USART SPI Timings

Symbol	Parameter	Conditions	Min	Max	Units
<b>Master Mode</b>					
SPI <sub>0</sub>	SCK period	1.8V domain 3.3V domain	MCK/6	–	–
SPI <sub>1</sub>	Input data setup time	1.8V domain 3.3V domain	0.5 * MCK + 1.1 0.5 * MCK + 1.1	–	–
SPI <sub>2</sub>	Input data hold time	1.8V domain 3.3V domain	1.5 * MCK + 0.7 1.5 * MCK + 0.7	–	–
SPI <sub>3</sub>	Chip select active to serial clock	1.8V domain 3.3V domain	1.5 * SPCK - 0.1 1.5 * SPCK - 0.1	–	–
SPI <sub>4</sub>	Output data setup time	1.8V domain 3.3V domain	- 4.7 - 4.7	7.1 7.1	ns
SPI <sub>5</sub>	Serial clock to chip select inactive	1.8V domain 3.3V domain	1 * SPCK - 5.1 1 * SPCK - 5.1		ns
<b>Slave Mode</b>					
SPI <sub>6</sub>	SCK falling to MISO	1.8V domain 3.3V domain	6.8 6.8	13.3 13.3	ns
SPI <sub>7</sub>	MOSI setup time before SCK rises	1.8V domain 3.3V domain	2 * MCK + 0.2 2 * MCK + 0.2	–	ns
SPI <sub>8</sub>	MOSI hold time after SCK rises	1.8V domain 3.3V domain	1.2 1.2	–	ns
SPI <sub>9</sub>	SCK rising to MISO	1.8V domain 3.3V domain	7.5 7.5	13.8 13.8	ns
SPI <sub>10</sub>	MOSI setup time before SCK falls	1.8V domain 3.3V domain	2 * MCK + 1 2 * MCK + 1	–	ns
SPI <sub>11</sub>	MOSI hold time after SCK falls	1.8V domain 3.3V domain	0.7 0.7	–	ns
SPI <sub>12</sub>	NPCS0 setup to SCK rising	1.8V domain 3.3V domain	2.5 * MCK - 0.5 2.5 * MCK - 0.5	–	ns
SPI <sub>13</sub>	NPCS0 hold after SCK falling	1.8V domain 3.3V domain	1.5 * MCK + 1 1.5 * MCK + 1	–	ns
SPI <sub>14</sub>	NPCS0 setup to SCK falling	1.8V domain 3.3V domain	2.5 * MCK + 0.2 2.5 * MCK + 0.2	–	ns
SPI <sub>15</sub>	NPCS0 hold after SCK rising	1.8V domain 3.3V domain	1.5 * MCK + 1.4 1.5 * MCK + 1.4	–	ns

## 45.5 Embedded Analog Peripherals Characteristics

### 45.5.1 Core Voltage Regulator

Table 45-14. Core Voltage Regulator Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDIN}$	Supply voltage range (VDDIN)		1.62	3.3	3.6	V
$V_{VDDOUT}$	DC output voltage	Normal Mode Standby Mode	–	1.2 0	–	V
$I_{LOAD}$	Maximum DC output current	$V_{VDDIN} > 2.0V$ , $T_J = 100^\circ C$ $V_{VDDIN} \leq 2.0V$ , $T_J = 100^\circ C$	–	–	120 40	mA
ACC	Output voltage total accuracy	$I_{LOAD} = 0.8\text{ mA to }120\text{ mA}$ $V_{VDDIN} = 2.0V\text{ to }3.6V$ $T_J = [-40^\circ C\text{ to }100^\circ C]$	-5	–	5	%
$I_{INRUSH}$	Inrush current	$I_{LOAD} = 0$ . See Note <sup>(3)</sup>	–	–	400	mA
$I_{DDIN}$	Current consumption (VDDIN)	Normal mode; $I_{Load} = 0\text{ mA}$ Normal mode; $I_{Load} = 120\text{ mA}$ Standby mode	–	5 500 0.02	1	$\mu A$
$C_{IN}$	Input decoupling capacitor <sup>(1)</sup>		1	–	–	$\mu F$
$C_{OUT}$	Output capacitor <sup>(2)</sup>	Capacitance	0.7	2.2	10	$\mu F$
		ESR	0.01		10	$\Omega$
$t_{ON}$	Turn-on time	$C_{OUT} = 2.2\mu F$ , $V_{VDDOUT}$ reaches 1.2V (+/- 3%)	–	500		$\mu s$
$t_{OFF}$	Turn-off time	$C_{OUT} = 2.2\mu F$	–		40	ms

- Notes:
1. A ceramic capacitor must be connected between VDDIN and the closest GND pin of the device. This decoupling capacitor is mandatory to reduce inrush current and to improve transient response and noise rejection.
  2. To ensure stability, an external output capacitor,  $C_{OUT}$  must be connected between the VDDOUT and the closest GND pin of the device. The ESR (Equivalent Series Resistance) of the capacitor must be in the range of 0.01 to 10 $\Omega$ . Solid tantalum, and multilayer ceramic capacitors are all suitable as output capacitor. An additional 100 nF bypass capacitor between VDDOUT and the closest GND pin of the device helps decrease output noise and improves the load transient response.
  3. Current needed to charge external bypass/decoupling capacitor network.

### 45.5.2 Automatic Power Switch

Table 45-15. Automatic Power Switch Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IT+}$	Positive-going input threshold voltage (VDDIO)		1.9	–	2.2	V
$V_{IT-}$	Negative-going input threshold voltage (VDDIO)		1.8	–	2.1	V
$V_{IT\_HYS}$	Threshold hysteresis		–	100	–	mV

### 45.5.3 LCD Voltage Regulator and LCD Output Buffers

The LCD Voltage Regulator is a complete solution to drive a LCD display. It integrates a low-power LDO regulator with programmable output voltage and some buffers to drive the LCD lines. A 1uF capacitor is required at the LDO regulator output (VDDLCD). This regulator can be either set in active (normal) mode, or in bypass mode (HiZ mode), or in OFF mode.

- In normal mode, the VDDLCD LDO regulator output can be selected from 2.4V to 3.5V using LCDVROUT bits in the Supply Controller Mode Register (SUPC\_MR), with the conditions:
  - $VDDLCD \leq VDDIO$  and,
  - $VDDLCD \leq VDDIN - 150 \text{ mV}$ .
- In bypass mode (HiZ mode), the VDDLCD is set in high impedance (through the LCDMODE bits in SUPC\_MR register), and can be forced externally. This mode can be used to save the LDO operating current (4 uA).
- In OFF mode, the VDDLCD Output is pulled down.

**IMPORTANT:** When using an external or the internal voltage regulator, VDDIO and VDDIN must be still supplied with the conditions:  $2.4V \leq VDDLCD \leq VDDIO/VDDIN$  and  $VDDIO/VDDIN \geq 2.5V$ .

**Table 45-16. LCD Voltage Regulator Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDIN}$	Supply voltage range (VDDIN)		2.5	–	3.6	V
$V_{VDDLCD}$	Programmable output range	See Table 45-17.	2.4		3.5	V
	Output voltage accuracy		-10		+10	%
$I_{DDIN}$	Current consumption (VDDIN)	LDO enabled	–	–	4	μA
$d_{VOUT}/d_{VDDIN}$	VDDLCD variation with VDDIN		–	-50	-70	mV/V
$I_{LOAD}$	Output current	DC or transient load averaged by the external decoupling capacitor	–	–	2	mA
$C_{OUT}$	Output capacitor on VDDLCD		1	–	10	μF
$t_{ON}$	Start-up time	$C_{OUT} = 1\mu F$	–	–	1	ms

**Table 45-17. VDDLCD Voltage Selection at VDDIN = 3.6V**

LCD VROUT	VDDLCD (V)	LCD VROUT	VDDLCD (V)	LCD VROUT	VDDLCD (V)	LCD VROUT	VDDLCD (V)
0	2.86	4	2.57	8	3.45	12	3.16
1	2.79	5	2.50	9	3.38	13	3.09
2	2.72	6	2.43	10	3.31	14	3.02
3	2.64	7	2.36	11	3.23	15	2.95

**Table 45-18. LCD Buffers Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DDIN}$	Current consumption (VDDIN)	LDO enabled	–	25	35	μA
$Z_{OUT}$	Buffer output impedance	GPIO in LCD mode (SEG or COM)	200	500	1000	Ω
$C_{LOAD}$	Capacitive output load		10p	–	50n	F
$t_r / t_f$	Rising or falling time 95% convergence	$C_{LOAD} = 10 \text{ pF}$ $C_{LOAD} = 50 \text{ nF}$	–	–	3 150	μs

## 45.5.4 VDDCORE Brownout Detector

Table 45-19. Core Power Supply Brownout Detector Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IT-}$	Negative-going input threshold voltage (VDDCORE) <sup>(1)</sup>		0.98	1	1.04	V
$V_{IT+}$	Positive-going input threshold voltage (VDDCORE)	—	0.8	1.0	1.08	V
$V_{HYS}$	Hysteresis voltage	$V_{IT+} - V_{IT-}$	—	25	50	mV
$t_{d-}$	$V_{IT-}$ detection propagation time	$V_{DDCORE} = V_{IT+}$ to ( $V_{IT-} - 100\text{mV}$ )	—	200	300	ns
$t_{START}$	Start-up time	From disabled state to enabled state	—	—	300	$\mu\text{s}$
$I_{DDCORE}$	Current consumption (VDDCORE)	Brownout Detector enabled	—	—	15	$\mu\text{A}$
$I_{DDIO}$	Current consumption (VDDIO)	Brownout Detector enabled	—	—	18	$\mu\text{A}$

Note: 1. The product is guaranteed to be functional at  $V_{IT-}$ .

Figure 45-10. Core Brownout Output Waveform

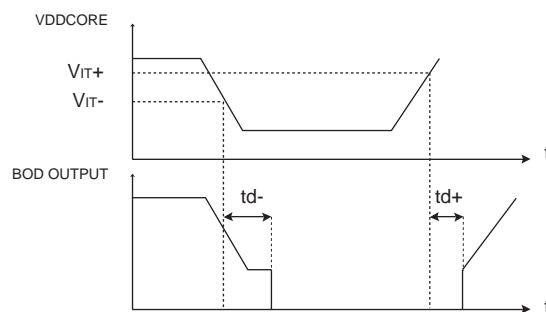
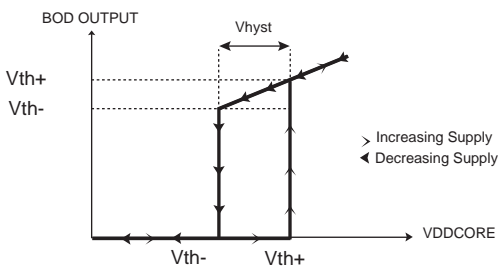


Figure 45-11. Core Brownout Transfer Characteristics



### 45.5.5 VDDCORE Power-On-Reset

Table 45-20. Core Power Supply Power-On-Reset Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IT-}$	Negative-going input threshold voltage (VDDCORE)		0.71	0.9	1.02	V
$V_{IT+}$	Positive-going input threshold voltage (VDDCORE)		0.8	1.0	1.08	V
$V_{HYS}$	Hysteresis voltage	$V_{IT+} - V_{IT-}$	–	60	110	mV
$t_{d-}$	$V_{IT-}$ detection propagation time	VDDCORE = $V_{IT+}$ to ( $V_{IT-} - 100\text{mV}$ )	–	–	15	$\mu\text{s}$
$t_{START}$	Start-up time	VDDCORE rising from 0 to final value. Time to release reset signal.	–	–	300	$\mu\text{s}$
$I_{DDCORE}$	Current consumption (VDDCORE)		–	–	6	$\mu\text{A}$
$I_{DDIO}$	Current consumption (VDDIO)		–	–	9	$\mu\text{A}$

## 45.5.6 VDDIO Supply Monitor

**Table 45-21. VDDIO Supply Monitor**

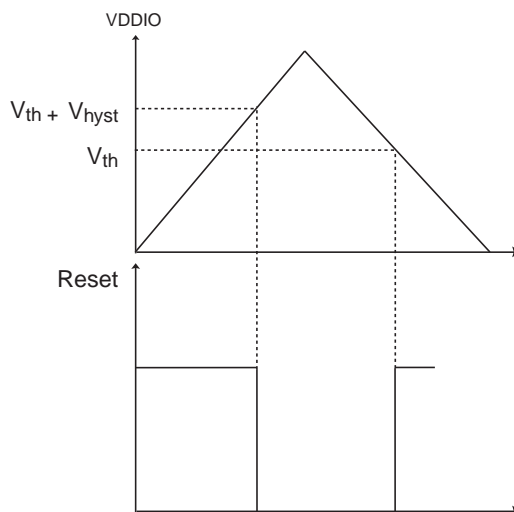
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{TH-}$	Programmable range of negative-going input threshold voltage (VDDIO)	16 selectable steps	1.6	–	3.4	V
ACC	$V_{TH-}$ accuracy	With respect to programmed value	-2	–	+2	%
$V_{HYST}$	Hysteresis		–	30	40	mV
$I_{DDON}$	Current consumption (VDDIO) <sup>(1)</sup>	On with a 100% duty cycle.	–	20	40	$\mu$ A
$t_{ON}$	Start-up time	From OFF to ON	–	–	300	$\mu$ s

Notes: 1. The average current consumption can be reduced by using the Supply monitor in sampling mode. See Supply Controller Section.

**Table 45-22. VDDIO Supply Monitor Threshold Selection**

Digital Code	Threshold Typ (V)	Digital Code	Threshold Typ (V)	Digital Code	Threshold Typ (V)	Digital Code	Threshold Typ (V)
0000	1.6	0100	2.08	1000	2.56	1100	3.04
0001	1.72	0101	2.2	1001	2.68	1101	3.16
0010	1.84	0110	2.32	1010	2.8	1110	3.28
0011	1.96	0111	2.44	1011	2.92	1111	3.4

**Figure 45-12. VDDIO Supply Monitor**

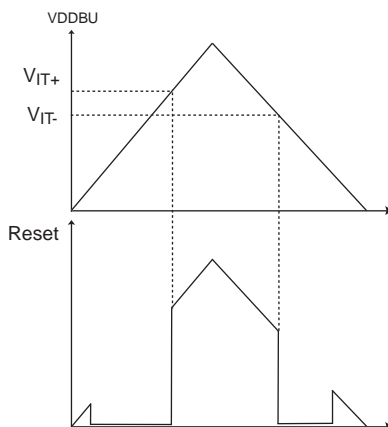


### 45.5.7 VDDBU Power-On-Reset

Table 45-23. Zero-Power-On POR (Backup POR) Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{TH+}$	Positive-going input threshold voltage (VDDBU)	At startup	1.45	1.53	1.59	V
$V_{TH-}$	Negative-going input threshold voltage (VDDBU)		1.35	1.45	1.55	V
$I_{DDBU}$	Current consumption	Enabled	–	300	700	nA
$t_{RES}$	Reset time-out period		100	240	500	$\mu$ s

Figure 45-13. Zero-Power-On Reset Characteristics



### 45.5.8 VDDIO Power-On-Reset

Table 45-24. Zero-Power-On POR (VDDIO POR) Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{TH+}$	Positive-going input threshold voltage (VDDIO)	At Startup	1.45	1.53	1.59	V
$V_{TH-}$	Negative-going input threshold voltage (VDDIO)		1.35	1.45	1.55	V
$I_{DDIO}$	Current consumption		–	300	700	nA
$t_{RES}$	Reset time-out period		100	240	500	$\mu$ s

### 45.5.9 32 kHz RC Oscillator

**Table 45-25. 32 kHz RC Oscillator Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDBU}$	Supply voltage range (VDDBU)		1.62	3.3	3.6	V
$f_0$	Frequency initial accuracy	$V_{DDBU} = 3.3V$ , $T_A = 27^\circ C$	26	32	39	kHz
dF/dV	Frequency drift with VDDBU	VDDBU from 1.6V to 3.6V	0.5	1.5	2.5	%/V
dF/dT	Frequency drift with temperature	$T_A = [-40^\circ C \text{ to } +27^\circ C]$		+8	+17	%
		$T_A = [27^\circ C \text{ to } 85^\circ C]$		+6	+13	
Duty	Duty cycle		48	50	52	%
$t_{ON}$	Startup time		–	–	100	$\mu s$
$I_{DDON}$	Current consumption (VDDBU)		–	150	300	nA

### 45.5.10 4/8/12 MHz RC Oscillator

**Table 45-26. 4/8/12 MHz RC Oscillators Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDCORE}$	Supply voltage range (VDDCORE)		1.08	1.2	1.32	V
$f_{Range}$	Output frequency range <sup>(1)</sup>		4		12	MHz
$ACC_4$	4 MHz total accuracy	$-40^\circ C < T_A < +85^\circ C$	–	–	$\pm 30$	%
$ACC_8$	8 MHz total accuracy	$T_A = 25^\circ C$	–	–	$\pm 0.5$	%
		$0^\circ C < T_A < +70^\circ C$			$\pm 2.5$	
		$-40^\circ C < T_A < +85^\circ C$			$\pm 5$	
$ACC_{12}$	12 MHz total accuracy	$T_A = 25^\circ C$	–	–	$\pm 0.5$	%
		$0^\circ C < T_A < +70^\circ C$			$\pm 2.5$	
		$-40^\circ C < T_A < +85^\circ C$			$\pm 5$	
$f_{STEP}$	Frequency trimming step size	8 MHz	–	47	–	kHz
		12 MHz		64		
Duty	Duty cycle		45	50	55	%
$t_{ON}$	Startup time, MOSCRCEEN from 0 to 1		–	–	10	$\mu s$
$t_{Stab}$	Stabilization time on RC frequency change (MOSCRCEEN)		–	–	5	$\mu s$
$I_{DDON}$	Active current consumption (VDDCORE)	4 MHz	–	50	68	$\mu A$
		8 MHz		65	86	
		12 MHz		82	102	

Note: 1. Frequency range can be configured in the PMC Clock Generator.



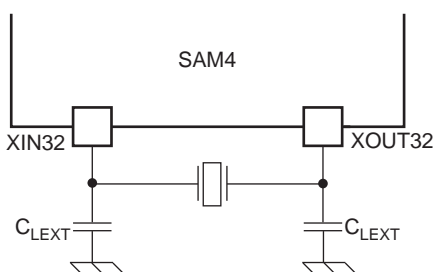
## 45.5.11 32.768 kHz Crystal Oscillator

**Table 45-27. 32.768 kHz Crystal Oscillator Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>VDDBU</sub>	Supply voltage range (VDDBU)		1.62	3.3	3.6	V
f <sub>req</sub>	Operating frequency	Normal mode with crystal	–	–	32.768	kHz
Duty	Duty cycle		40	50	60	%
t <sub>ON</sub>	Startup time	R <sub>s</sub> <sup>(1)</sup> < 50 kΩ C <sub>crystal</sub> = 12.5 pF C <sub>crystal</sub> = 6 pF  R <sub>s</sub> <sup>(1)</sup> < 100 kΩ C <sub>crystal</sub> = 12.5 pF C <sub>crystal</sub> = 6 pF	–	–	900 300 1200 500	ms
I <sub>DDON</sub>	Current consumption (VDDBU)	R <sub>s</sub> <sup>(1)</sup> < 65 kΩ C <sub>crystal</sub> = 12.5 pF C <sub>crystal</sub> = 6 pF  R <sub>s</sub> <sup>(1)</sup> < 100 kΩ C <sub>crystal</sub> = 6 pF  R <sub>s</sub> <sup>(1)</sup> < 20 kΩ C <sub>crystal</sub> = 6 pF	–	450 280 350 220	950 850 1050	nA
P <sub>ON</sub>	Drive level		–	–	0.1	μW
R <sub>f</sub>	Internal resistor	Between XIN32 and XOUT32	–	10	–	MΩ
C <sub>LEXT</sub>	External capacitor on XIN32 and XOUT32		6	–	12.5	pF
C <sub>para</sub>	Internal parasitic capacitance		0.6	0.7	0.8	pF

Note: 1. R<sub>S</sub> is the series resistor.

**Figure 45-14. 32.768 kHz Crystal Oscillator Schematic**



$$C_{LEXT} = 2 \times (C_{CRYSTAL} - C_{para} - C_{PCB})$$

where C<sub>PCB</sub> is the capacitance of the printed circuit board (PCB) track layout from the crystal to the SAM4 pin.

Table 45-28 below summarizes recommendations for 32.768 kHz crystal selection.

**Table 45-28. Recommended Crystal Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
ESR	Equivalent Series Resistor (R <sub>S</sub> )	Crystal @ 32.768 kHz	–	50	100	kΩ
C <sub>M</sub>	Motional capacitance	Crystal @ 32.768 kHz	0.6	–	3	fF
C <sub>SHUNT</sub>	Shunt capacitance	Crystal @ 32.768 kHz	0.6	–	2	pF

### 45.5.12 3 to 20 MHz Crystal Oscillator

Table 45-29. 3 to 20 MHz Crystal Oscillator Characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>VDDIO</sub>	Supply voltage range (VDDIO)		1.62	3.3	3.6	V
V <sub>VDDPLL</sub>	Supply voltage range (VDDPLL)		1.08	1.2	1.32	V
f <sub>OSC</sub>	Operating frequency range	Normal mode with crystal	3	16	20	MHz
Duty	Duty cycle		40	50	60	%
t <sub>ON</sub>	Start-up time	3 MHz, C <sub>SHUNT</sub> = 3pF 8 MHz, C <sub>SHUNT</sub> = 7pF 16 MHz, C <sub>SHUNT</sub> = 7pF with Cm = 8fF 16 MHz, C <sub>SHUNT</sub> = 7pF with Cm = 1.6fF 20 MHz, C <sub>SHUNT</sub> = 7pF	–	–	14.5 4 1.4 2.5 1	ms
I <sub>DD_ON</sub>	Current consumption On VDDIO  On VDDPLL	3 MHz <sup>(1)</sup> 8 MHz <sup>(2)</sup> 16 MHz <sup>(3)</sup> 20 MHz <sup>(4)</sup>  3 MHz <sup>(1)</sup> 8 MHz <sup>(2)</sup> 16 MHz <sup>(3)</sup> 20 MHz <sup>(4)</sup>	–	230 300 390 450  6 12 20 24	350 400 470 560  7 14 23 30	μA
P <sub>ON</sub>	Drive level	3 MHz 8 MHz 16 MHz, 20 MHz	–	–	15 30 50	μW
R <sub>f</sub>	Internal resistor	Between XIN and XOUT		0.5		MΩ
C <sub>LEXT</sub>	External capacitor on XIN and XOUT		12.5	–	17.5	pF
C <sub>L</sub>	Internal equivalent load capacitance	Integrated load capacitance (XIN and XOUT in series)	7.5	9.5	10.5	pF

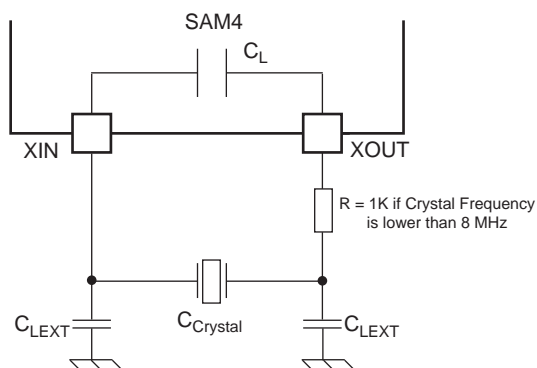
Notes: 1. Rs = 100-200 Ohms; Cs = 2.0 - 2.5 pF; Cm = 2 – 1.5 fF(typ, worst case) using 1 kΩ serial resistor on XOUT.

2. Rs = 50-100 Ohms; Cs = 2.0 - 2.5 pF; Cm = 4 - 3 fF(typ, worst case).

3. Rs = 25-50 Ohms; Cs = 2.5 - 3.0 pF; Cm = 7 - 5 fF (typ, worst case).

4. Rs = 20-50 Ohms; Cs = 3.2 - 4.0 pF; Cm = 10 - 8 fF(typ, worst case).

**Figure 45-15. 3 to 20 MHz Crystal Oscillator Schematic**



$$C_{LEXT} = 2 \times (C_{CRYSTAL} - C_L - C_{PCB})$$

where  $C_{PCB}$  is the capacitance of the printed circuit board (PCB) track layout from the crystal to the SAM4 pin.

Table 45-30 summarizes recommendations to be followed when choosing a crystal..

**Table 45-30. Recommended Crystal Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
ESR	Equivalent Series Resistor (Rs)	Fundamental @ 3 MHz	—	—	200	$\Omega$
		Fundamental @ 8 MHz			100	
		Fundamental @ 12 MHz			80	
		Fundamental @ 16 MHz			80	
		Fundamental @ 20 MHz			50	
$C_M$	Motional capacitance		—	—	8	fF
$C_{SHUNT}$	Shunt capacitance		—	—	7	pF

#### 45.5.13 Crystal Oscillator Design Considerations

When choosing a crystal for the 32768 Hz Slow Clock Oscillator or for the 3-20 MHz Oscillator, several parameters must be taken into account. Important parameters are as follows:

- Crystal Load Capacitance.**  
 The total capacitance loading the crystal, including the oscillator's internal parasitics and the PCB parasitics, must match the load capacitance for which the crystal's frequency is specified. Any mismatch in the load capacitance with respect to the crystal's specification will lead to inaccurate oscillation frequency.
- Crystal Drive Level.**  
 Use only crystals with the specified drive levels greater than the specified MCU oscillator drive level. Applications that do not respect this criterion may damage the crystal.
- Crystal Equivalent Series Resistor (ESR).**  
 Use only crystals with the specified ESR lower than the specified MCU oscillator ESR. In applications where this criterion is not respected, the crystal oscillator may not start.
- Crystal Shunt Capacitance.**  
 Use only crystal with the specified shunt capacitance lower than the specified MCU oscillator shunt capacitance. In applications where this criterion is not respected, the crystal oscillator may not start.
- PCB Layout Considerations.**  
 To minimize inductive and capacitive parasitics associated with XIN, XOUT, XIN32, XOUT32 nets, it is recommended to route them as short as possible. It is also of prime importance to keep those nets away from noisy switching signals (clock, data, PWM, etc...). A good practice is to shield them with a quiet ground net to avoid coupling to neighbour signals.

## 45.5.14 PLLA, PLLB Characteristics

**Table 45-31. PLLA Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDPLL}$	Supply voltage range (VDDPLL)		1.08	1.2	1.32	V
$f_{IN}$	Input frequency range		30	32.768	34	kHz
$f_{OUT}$	Output frequency range		7.5	8.192	8.5	MHz
$N_{RATIO}$	Frequency multiplying ratio (MULA +1)		–	250	–	–
$J_p$	Period jitter	Peak value	–	4	–	ns
$t_{ON}$	Start-up time	From OFF to output oscillations (Output frequency within 10% of target frequency)	–	–	250	$\mu$ s
$t_{LOCK}$	Lock time	From OFF to PLL locked	–	–	2.5	ms
$I_{PLLON}$	Active mode current consumption (VDDPLL)	$f_{OUT} = 8.192$ MHz	–	50	–	$\mu$ A
$I_{PLLOFF}$	OFF mode current consumption (VDDPLL)	@25°C Over the temperature range	–	0.05 0.05	0.30 5	$\mu$ A

## 45.5.15

**Table 45-32. PLLB Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDPLL}$	Supply voltage range (VDDPLL)		1.08	1.2	1.32	V
$f_{IN}$	Input frequency range		3	–	32	MHz
$f_{OUT}$	Output frequency range		80	–	240	MHz
$N_{RATIO}$	Frequency multiplying ratio (MULB +1)		3	–	62	–
$Q_{RATIO}$	Frequency dividing ratio (DIVB)		2	–	24	–
$T_{ON}$	Start-up time		–	60	150	$\mu$ s
$IDD_{PLL}$	Current consumption on VDDPLL	Active mode @ 80 MHz @1.2V Active mode @ 96 MHz @1.2V Active mode @ 160 MHz @1.2V Active mode @ 240 MHz @1.2V	–	0.94 1.2 2.1 3.34	1.2 1.5 2.5 4	mA

## 45.5.16 Temperature Sensor Characteristics

The temperature sensor provides an output voltage ( $V_T$ ) that is proportional to absolute temperature (PTAT). This voltage can be measured through the channel number 7 of the 10-bit ADC. Improvement of the raw performance of the temperature sensor acquisition can be achieved by performing a single temperature point calibration to remove the initial inaccuracies ( $V_T$  and ADC offsets).

**Table 45-33. Temperature Sensor Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDIN}$	Supply voltage range (VDDIN)		2.4		3.6	V
$V_T$	Output voltage	$T_J = 27^\circ\text{C}$	1.34	1.44	1.54	V
$dV_T/dT$	Output voltage sensitivity to Temperature		4.2	4.7	5.2	mV/°C
$dV_T/dV$	$V_T$ variation with VDDIN	VDDIN from 2.4V to 3.6V			1	mV/V
$t_S$	$V_T$ settling time	When $V_T$ is sampled by the 10-bit ADC, the required track time to ensure $1^\circ\text{C}$ accurate settling			1	$\mu\text{s}$
$T_{ACC}$	Temperature accuracy <sup>(1)</sup>	After offset calibration Over $T_J$ range $[-40^\circ\text{C} / +85^\circ\text{C}]$		$\pm 5$		$^\circ\text{C}$
		After offset calibration Over $T_J$ range $[0^\circ\text{C} / +80^\circ\text{C}]$		$\pm 4$		$^\circ\text{C}$
$t_{ON}$	Start-up time		–	5	10	$\mu\text{s}$
$I_{VDDIN}$	Current consumption		50	70	80	$\mu\text{A}$

Note: 1. Does not include errors due to A/D conversion process.

## 45.5.17 Optical UART RX Transceiver Characteristics

Table 45-34 gives the description of the optical link transceiver for electrically isolated serial communication with hand-held equipment, such as calibrators compliant with ANSI-C12.18 or IEC62056-21 norms (only available on UART1).

**Table 45-34. Transceiver Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDIO}$	Supply voltage range (VDDIO)		3	3.3	3.6	V
$I_{DD}$	Current consumption	ON OFF		25	35 0.1	$\mu\text{A}$
$V_{TH}$	Comparator threshold	According to the programmed threshold. See the OPT_CMPH bit in the UART Mode Register (UART1)	-20		+20	mV
$V_{HYST}$	Hysteresis		10	20	40	mV
$t_{PROP}$	Propagation time	With 100 mVpp square wave input around threshold			5	$\mu\text{s}$
$t_{ON}$	Start-up time				100	$\mu\text{s}$

## 45.5.18 10-bit ADC Characteristics

**Table 45-35. ADC Power Supply Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{VDDIN}$	Supply voltage range (VDDIN)		2.4	3.0	3.6	V
$I_{VDDIN}$	Current consumption on VDDIN	ADC ON <sup>(1)</sup> , internal voltage reference ON generating ADVREF = 3.0V	–	450	700	$\mu$ A
		ADC ON <sup>(1)</sup> , internal voltage reference OFF, ADVREF externally supplied		220	350	

Note: 1. Average current consumption performing conversion in free run mode @ 16 MHz ADC Clock.  $F_S = 510$  kS/s.

**Table 45-36. ADC Voltage Reference Input Characteristics (ADVREF pin)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{ADVREF}$	ADVREF input voltage range <sup>(1)</sup>	Internal voltage reference OFF	1.62		$V_{VDDIN}$	V
$R_{ADVREF}$	ADVREF input resistance	ADC ON, internal voltage reference OFF	9	14	19	k $\Omega$
$I_{ADVREF}$	Current consumption on ADVREF	ADVREF = 2.4V	-35%	170	+35%	$\mu$ A
		ADVREF = 3.3V		235		
		ADVREF = 3.6V		260		
$C_{ADVREF}$	Decoupling capacitor on ADVREF		100			nF

Note: 1. ADVREF input range limited to VDDIO if VDDIO < VDDIN.

**Table 45-37. ADC Timing Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{CK\_ADC}$	ADC clock frequency	$3.0V \leq V_{DDIN} \leq 3.6$	–	–	16	MHz
		$2.4V \leq V_{DDIN} \leq 3.0$			14	
$t_{CONV}$	ADC conversion time <sup>(1)</sup>	$f_{CK\_ADC} = 16$ MHz, $T_{TRACK} = 500$ ns	1.95	–	–	$\mu$ s
$F_S$	Sampling rate <sup>(2)</sup>	$V_{VDDIN} > 3V$ , $f_{CK\_ADC} = 16$ MHz	–	–	510	kS/s
		$V_{VDDIN} > 2.4V$ , $f_{CK\_ADC} = 14$ MHz			380	
$t_{ON}$	Start-up time	ADC only	–	–	40	$\mu$ s
$t_{TRACK}$	Track and hold time <sup>(3)</sup>	$2.4V \leq V_{DDIN} \leq 3.0$	1000	–	–	ns
		$3.0V < V_{DDIN} < 3.6V$	500			

Notes: 1.  $t_{CONV} = (TRACKTIM + 24) / f_{CK\_ADC}$ .

2.  $F_S = 1 / t_{CONV}$ .

3. Refer to [Section 45.5.18.1 “Track and Hold Time versus Source Output Impedance, Effective Sampling Rate”](#).

**Table 45-38. ADC Analog Input Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
FSR	Analog Input Full Scale Range <sup>(1)</sup>		0	–	$V_{ADVREF}$	V
$C_{IN}$	Input capacitance <sup>(2)</sup>	Accounts for I/O input capacitance + ADC Sampling capacitor	–	–	10	pF

Notes: 1. If  $V_{DDIO} < V_{ADVREF}$ , full scale range is limited to  $V_{DDIO}$ .

2. See [Figure 45-16 “Simplified Acquisition Path”](#) .

**Table 45-39. Static Performance Characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_{ADC}$	Native ADC resolution		–	10	–	Bits
$R_{ADC\_AV}$	Resolution with digital averaging	See “ADC Controller” section	10	–	12	Bits
INL	Integral non linearity	$f_{CK\_ADC} = 16 \text{ MHz}$	-2	–	+2	LSB
DNL	Differential non linearity		-1	–	+1	LSB
OE	Offset error	Errors with respect to the best fit line method	-5	–	5	LSB
GE	Gain Error		-3	–	+3	LSB

Note: 1. In this table, values expressed in LSB refer to the Native ADC resolution (i.e. a 10-bit LSB).

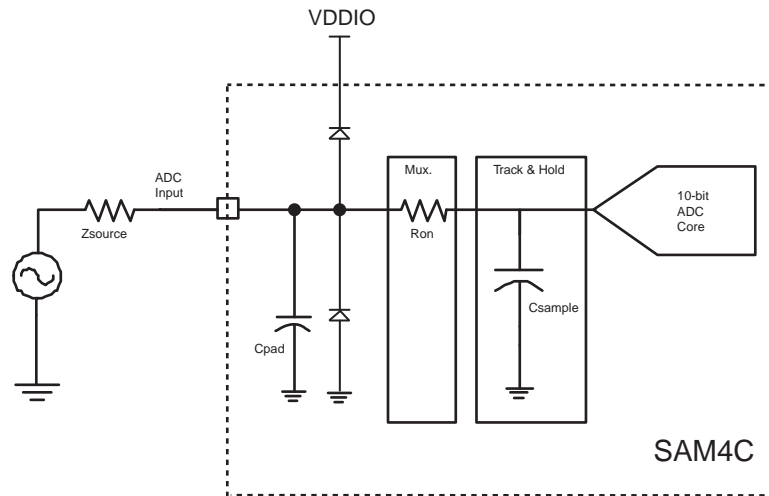
**Table 45-40. Dynamic Performance Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
SNR	Signal to Noise Ratio	$f_{CK\_ADC} = 16 \text{ MHz},$ $V_{ADVREF} = V_{DDIN},$ $f_{IN} = 50 \text{ kHz},$ $V_{INPP} = 0.95 \times V_{ADVREF}$	57	60	–	dB
THD	Total Harmonic Distortion		–	-68	-55	dB
SINAD	Signal to Noise and Distortion		52	59	–	dB
ENOB	Effective Number of Bits		8.3	9.6	–	Bits

#### 45.5.18.1 Track and Hold Time versus Source Output Impedance, Effective Sampling Rate

The following figure gives a simplified view of the acquisition path.

**Figure 45-16. Simplified Acquisition Path**



During its tracking phase, the 10-bit ADC charges its sampling capacitor through various serial resistors: source output resistor, multiplexer series resistor and the sampling switch series resistor. In case of high output source resistance (low power resistive divider, for example), the track time must be increased to ensure full settling of the sampling capacitor voltage. The following formulas give the minimum track time that guarantees a 10-bit accurate settling:

- $V_{DDIN} > 3.0V$ :  $t_{TRACK} (ns) = 0.12 \times R_{SOURCE}(\Omega) + 500$
- $V_{DDIN} \leq 3.0V$ :  $t_{TRACK} (ns) = 0.12 \times R_{SOURCE}(\Omega) + 1000$

According to the calculated track time ( $t_{TRACK}$ ), the actual track time of the ADC must be adjusted through the TRACKTIM field in the ADC\_MR register. TRACKTIM is obtained by the following formula:

$$TRACKTIM = \text{floor} \left( \frac{T_{TRACK}}{T_{CK\_ADC}} \right)$$

with  $t_{CK\_ADC} = 1 / f_{CK\_ADC}$  and floor(x) the mathematical function that rounds x to the greatest previous integer.

The actual conversion time of the converter is obtained by the following formula:

$$T_{CONV} = (TRACKTIM + 24) \times T_{CK\_ADC}$$

When converting in free run mode, the actual sampling rate of the converter is  $(1 / t_{CONV})$  or as defined by the following formula:

$$F_S = \frac{F_{CK\_ADC}}{(TRACKTIM + 24)}$$

The maximum source resistance with the actual TRACKTIM setting is:

- $R_{SOURCE\_MAX}(\Omega) = ((TRACKTIM + 1) \times t_{CK\_ADC}(ns) - 500) / 0.12$  for  $V_{DDIN} > 3.0V$ ; or
- $R_{SOURCE\_MAX}(\Omega) = ((TRACKTIM + 1) \times t_{CK\_ADC}(ns) - 1000) / 0.12$  for  $V_{DDIN} \leq 3.0V$



**Example:** Calculated track time is lower than actual ADC clock period

- Assuming:  $f_{CK\_ADC} = 1 \text{ MHz}$  ( $t_{CK\_ADC} = 1 \mu\text{s}$ ),  $R_{SOURCE} = 100\Omega$  and  $V_{VDDIN} = 3.3\text{V}$
- The minimum required track time is:  $t_{TRACK} = 0.12 \times 100 + 500 = 512 \text{ ns}$
- $t_{TRACK}$  begin less than  $t_{CK\_ADC}$ , TRACKTIM is set to 0. Actual track time is  $t_{CK\_ADC} = 1 \mu\text{s}$
- The calculated sampling rate is:  $f_s = 1 \text{ MHz} / 24 = 41.7 \text{ kHz}$
- The maximum allowable source resistance is:  $R_{SOURCE\_MAX} = (1000 - 500) / 0.12 = 4.1 \text{ k}\Omega$

**Example:** Calculated track time is greater than actual ADC clock period

- Assuming:  $f_{CK\_ADC} = 16 \text{ MHz}$  ( $t_{CK\_ADC} = 62.5\text{ns}$ ),  $R_{SOURCE} = 600\Omega$  and  $V_{VDDIN} = 2.8\text{V}$
- The minimum required track time is:  $t_{TRACK} = 0.12 \times 600 + 1000 = 1072 \text{ ns}$
- $\text{TRACKTIM} = \text{floor}(1072 / 62.5) = 17$ . Actual track time is:  $(17 + 1) \times t_{CK\_ADC} = 1.125 \mu\text{s}$
- The calculated sampling rate is:  $f_s = 16 \text{ MHz} / (24 + 17) = 390.2 \text{ kHz}$
- The maximum allowable source resistance is :  $R_{SOURCE\_MAX} = (1125 - 1000) / 0.12 = 1.04 \text{ k}\Omega$

#### 45.5.18.2 10-bit ADC with Averager

Table 45-41 indicates the typical ADC performance when used with the embedded digital averager. This table is built with the following conditions:

- $T_A = 25^\circ\text{C}$
- $f_{adc} = 10 \text{ MHz}$
- $V_{DDIN} = 2.4\text{V}$

**Table 45-41. ADC Resolution with Digital Averager**

Parameter Averaging Resolution RES (ADC_EMR)	OSR (Oversampling Ratio)	Mode (bits)	INL (LSB)	DNL (LSB)	SNR (dB)	THD (dB)	ENOB (bits)	FS (kS/s)
RES = 0	1	10	TBD	TBD	TBD	TBD	TBD	TBD
RES = 1	4	11	TBD	TBD	63	-76	10.2	TBD
RES = 2	16	12	TBD	TBD	TBD	TBD	TBD	TBD

### 45.5.19 Programmable Voltage Reference Characteristics

SAM4C embeds a programmable voltage reference designed to drive the 10-bit ADC ADVREF input. Table 45-42 shows the electrical characteristics of this internal voltage reference. In case of need, this voltage reference can be bypassed with some level of configurability: the user can either choose to feed the ADVREF input with an external voltage source or with the VDDIO internal power rail. See programming details in the ADC Analog Control Register (ADC\_ACR) in the ADC controller section.

**Table 45-42. Programmable Voltage Reference Characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>VDDIN</sub>	Voltage reference supply range		2	–	3.6	V
V <sub>ADVREF</sub>	Programmable output range	See Table 45-43.	1.6	–	3.4	V
ACC	Reference voltage accuracy	With respect to the programmed value. V <sub>VDDIN</sub> = 3.3V; T <sub>J</sub> = 25°C	-3		3	%
T <sub>C</sub>	Temperature coefficient	Box method <sup>(1)</sup>	–	–	250	ppm/°C
t <sub>ON</sub>	Start-up time	V <sub>VDDIN</sub> = 2.4V V <sub>VDDIN</sub> = 3V V <sub>VDDIN</sub> = 3.6V	–	–	100 70 40	μs
Z <sub>LOAD</sub>	Load impedance	Resistive	4			kΩ
		Capacitive	0.1		1	μF
I <sub>VDDIN</sub>	Current consumption on VDDIN <sup>(2)</sup>	ADC is OFF	–	20	30	μA

Notes: 1.  $TC = ( \max(V_{ADVREF}) - \min(V_{ADVREF}) ) / ( (T_{MAX} - T_{MIN}) * V_{ADVREF}(25^{\circ}C) )$ .

2. Does not include the current consumed by the ADC's ADVREF input if ADC is ON.

**Table 45-43. Programmable Voltage Reference Selection Values**

Sel. Value <sup>(1)</sup>	ADVREF	Notes
0	2.40	Default value
1	2.28	–
2	2.16	–
3	2.04	–
4	1.92	–
5	1.80	–
6	1.68	–
7	1.55	Min value
8	3.38	Max value
9	3.25	–
A	3.13	–
B	3.01	–
C	2.89	–
D	2.77	–
E	2.65	–
F	2.53	–

Note: 1. Voltage reference values are configurable in the ADC\_ACR register (the IRVS field).

## 45.6 Embedded Flash Characteristics

### 45.6.1 Embedded Flash DC characteristics

Table 45-44. DC Flash Characteristics

Symbol	Parameter	Conditions	Typ	Max	Unit
$I_{CC}$	Active current	Random 128-bit Read:			
		Maximum Read Frequency onto VDDCORE = 1.2V @ 25°C	16	25	mA
		Maximum Read Frequency onto VDDIO = 3.3V @ 25°C	3	5	
		Random 64-bit Read:			
		Maximum Read Frequency onto VDDCORE = 1.2V @ 25°C	10	18	mA
		Maximum Read Frequency onto VDDIO = 3.3V @ 25°C	3	5	
		Program:			
		- Onto VDDCORE = 1.2V @ 25°C	3	5	mA
		- Onto VDDIO = 3.3V @ 25°C	10	15	
		Erase:			
		- Onto VDDCORE = 1.2V @ 25°C	3	5	mA
		- Onto VDDIO = 3.3V @ 25°C	10	15	

### 45.6.2 Embedded Flash AC Characteristics

Table 45-45. AC Flash Characteristics

Parameter	Conditions	Min	Typ	Max	Unit
Program/ Erase Operation Cycle Time	Write page (512 Bytes)	–	1.5	3	ms
	Erase page	–	10	50	ms
	Erase block (4 Kbytes)	–	50	200	ms
	Erase sector	–	400	950	ms
	Full chip erase				
	- 1 Mbytes	–	9	18	s
	- 512 Kbytes		5.5	11	
	Lock/unlock time per region	–	1.5	3	ms
Data Retention	Not powered or powered	–	20	–	Years
Endurance	Write/erase cycles per page, block or sector @ 85°C	10K	–	–	Cycles

#### 45.6.2.1 Flash Wait States and Operating Frequency

The maximum operating frequency given in [Table 45-46](#) below is limited by the Embedded Flash access time when the processor is fetching code out of it. The table give the device maximum operating frequency depending on the FWS field of the EFC\_FMR register. This field defines the number of wait states required to access the Embedded Flash Memory.

**Table 45-46. Flash Wait State Versus Operating Frequency**

FWS (Flash Wait State)	Maximum Operating Frequency (MHz) @ T° = 85°C			
	VDDCORE = 1.08V VDDIO = 1.62V to 3.6V	VDDCORE = 1.2V VDDIO = 1.62V to 3.6V	VDDCORE = 1.08V VDDIO = 2.7V to 3.6V	VDDCORE = 1.2V VDDIO = 2.7V to 3.6V
0	16	17	20	21
1	33	35	40	42
2	51	52	61	63
3	67	70	81	85
4	85	87	98	106
5	100	105	–	120
6	–	121	–	–

## 45.7 Power Supply Current Consumption

This section provides information about the current consumption on different power supply rails of the device. It gives current consumption in low-power modes (Backup mode, Wait mode, Sleep mode) and in active mode (the application running from memory, by peripheral).

### 45.7.1 Backup Mode Current Consumption

Backup mode configuration and measurements are defined as follows: the Configuration A is used to achieve the lowest possible current consumption in this mode, the Configurations B, C and D are typical use cases with crystal oscillator, LCD and anti-tampering enabled.

Reminder: In Backup mode, the core voltage regulator is off and thus all the digital logics powered by VDDCORE is off. The backup mode is a low-power mode with the lowest possible expenses of the start-up time.

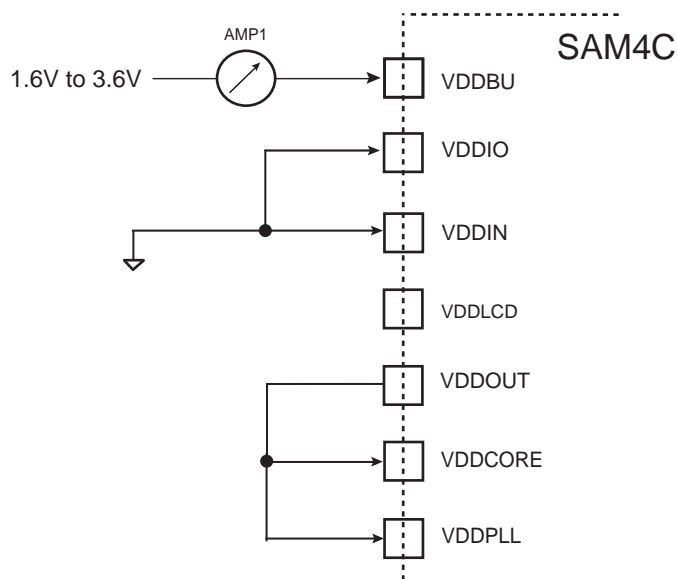
#### 45.7.1.1 Backup Mode Configuration A: Embedded Slow Clock RC Oscillator Enabled

- POR backup on VDDBU is disabled
- RTC running
- RTT enabled on 1 Hz mode
- Force wake-up (FWUP) enabled
- Current measurement as per [Figure 45-17](#)

#### 45.7.1.2 Backup Mode Configuration B: 32.768 kHz Crystal Oscillator Enabled

- POR backup on VDDBU is disabled
- RTC running
- RTT enabled on 1 Hz mode
- Force wake-up (FWUP) enabled
- Anti-tamper Input TMP0 enabled
- Current measurement as per [Figure 45-17](#)

Figure 45-17. Measurement Setup for Configurations A and B



**Table 45-47. Typical Current Consumption Values for Backup Mode Configurations A and B**

Conditions	Configuration A	Configuration B	Unit
VDDBU = 3.6V @25°C	580	760	nA
VDDBU = 3.3V @25°C	520	700	
VDDBU = 3.0V @25°C	480	680	
VDDBU = 2.5V @25°C	440	640	
VDDBU = 1.6V @25°C	400	600	
VDDBU = 3.6V @85°C	1.57	1.8	μA
VDDBU = 3.3V @85°C	1.5	1.7	
VDDBU = 3.0V @85°C	1.44	1.65	
VDDBU = 2.5V @85°C	1.3	1.56	
VDDBU = 1.6V @85°C	1.16	1.43	

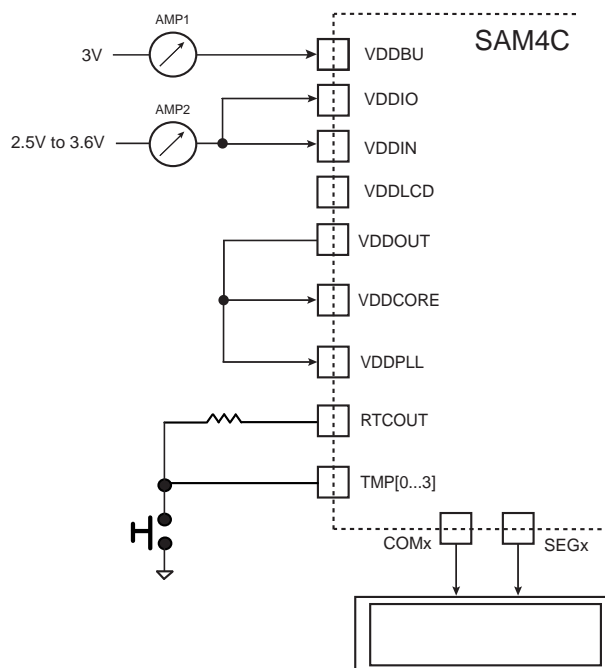
#### 45.7.1.3 Backup Mode Configuration C: 32.768 kHz Crystal Oscillator Enabled

- POR backup on VDDBU is disabled
- RTC running
- RTT enabled on 1 Hz mode
- Force wake-up (FWUP) enabled
- Anti-tamper input TMP0, TMP1, TMP2, TMP3 and RTCOUT0 enabled
- Main crystal oscillator disabled
- System IO lines PA30, PA31, PB[0...3] in GPIO Input Pull-up mode
- All other GPIO lines in default state (See PIO Multiplexing Table)
- Current measurement as per [Figure 45-18](#)

#### 45.7.1.4 Backup Mode Configuration D: 32.768 kHz Crystal Oscillator and LCD Enabled

- POR backup on VDDBU is disabled
- RTC running
- RTT enabled on 1 Hz mode
- LCD controller in low power mode, static bias and x64 slow clock buffer on-time drive time
- LCD voltage regulator used
- Force wake-up (FWUP) enabled
- Anti-tamper input TMP0, TMP1, TMP2, TMP3 and RTCOUT0 enabled
- Main crystal oscillator disabled
- System IO lines PA30, PA31, PB[0...3] in GPIO Input Pull-up mode
- All other GPIO lines in default state (See PIO Multiplexing Table)
- Current measurement as per [Figure 45-18](#)

**Figure 45-18. Measurement Setup for Configuration C and D**



**Table 45-48. Typical Current Consumption Values for Backup Mode Configurations C and D**

Conditions	Configuration C		Configuration D		Unit
	IDD_BU - AMP1	IDD_IN/IO - AMP2	IDD_BU - AMP1	IDD_IN/IO - AMP2	
VDDBU = 3.6V @25°C	0.05	2.6	0.05	9.5	μA
VDDBU = 3.3V @25°C		2.4		9	
VDDBU = 3.0V @25°C		2.1		8.5	
VDDBU = 2.5V @25°C		1.8		7.7	
VDDBU = 3.6V @85°C	0.09	6.7	0.1	14.8	
VDDBU = 3.3V @85°C		6.2		14	
VDDBU = 3.0V @85°C		5.8		13.5	
VDDBU = 2.5V @85°C		5		12.5	

## 45.7.2 Wait Mode Current Consumption

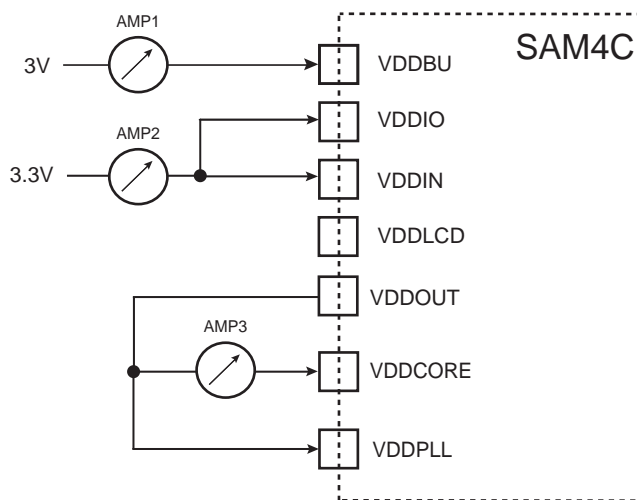
Wait mode configuration and measurements are defined in [Section 45.7.2.1 “Wait Mode Configuration”](#).

Reminder: In wait mode, the core voltage regulator is on, but the device is not clocked. Flash power mode can be either in stand-by mode or deep power-down mode. Wait mode provides a much faster wake-up compared to backup mode.

### 45.7.2.1 Wait Mode Configuration

- 32.768 kHz crystal oscillator running
- Fast RC oscillator, main crystal and PLLs stopped
- RTC running
- RTT enabled on 1 Hz mode.
- One wake-up pin (WKUPx) used in Fast Wake-up mode
- Anti-tamper inputs TMP0, TMP1, TMP2, TMP3 and RTCOUT0 enabled
- System IO lines PA30, PA31, PB[0...3] in GPIO Input Pull-up mode
- All other GPIO lines in default state
- Current measurement as per [Figure 45-19](#)

**Figure 45-19. Measurement Setup for Wait Mode Configuration**



**Table 45-49. Typical Current Consumption in Wait Mode**

Conditions	IDD_BU - AMP1		IDD_IN/IO - AMP2		IDD_CORE - AMP3		Unit
	@25°C	@85°C	@25°C	@85°C	@25°C	@85°C	
Flash in Read-Idle mode	0.003	0.09	68	476	45	446	μA
Flash in Standby mode	0.003	0.09	66	474	45	446	
Flash in Deep Power-down mode	0.003	0.09	62	469	45	446	

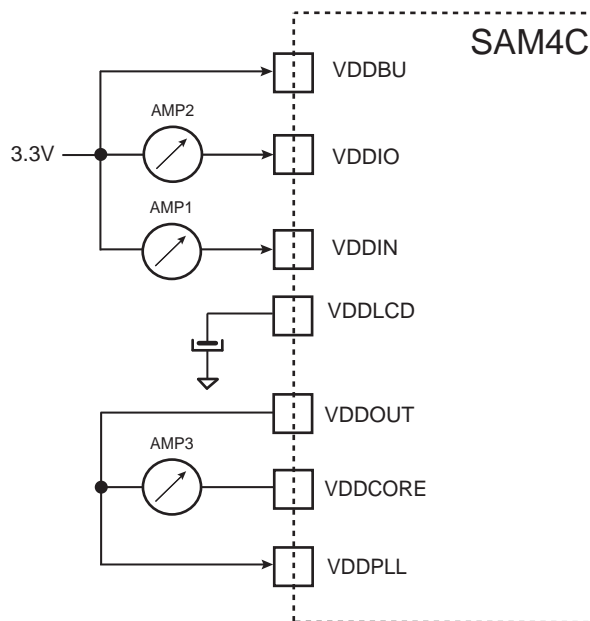


### 45.7.3 Sleep Mode Current Consumption

Sleep mode configuration and measurements are defined further on in this section.

Reminder: The purpose of sleep mode is to optimize power consumption of the device versus response time. In this mode, only the core clocks of CM4P0 and/or CM4P1 are stopped.

**Figure 45-20. Measurement Setup for Sleep Mode**



- VDDIO = VDDIN = 3.3V
- VDDCORE = 1.2V (Internal voltage regulator used)
- $T_A = 25^\circ\text{C}$
- Core 0 clock (HCLK) and Core 1 (CPHCLK) clock stopped
- Sub-system 0 Master Clock (MCK), Sub-system 1 Master Clock (CPBMCK) running at various frequencies (PLLB used for frequencies above 8 MHz and fast RC osc at 8 MHz divided by 1/2/4/8/16/32 for lower frequencies)
- All peripheral clocks deactivated
- No activity on I/O lines
- VDDPLL not taken into account. See PLL characteristics for further details
- Current measurement as per [Figure 45-20](#)

**Table 45-50. Typical Sleep Mode Current Consumption Versus Frequency**

Master Clock (MHz)	IDD_IN - AMP1	IDD_IO - AMP2	IDD_CORE - AMP3	Unit
120	14.26	0.03	10.83	mA
100	11.96	0.03	9.09	
84	10.1	0.03	7.68	
64	7.78	0.03	5.92	
48	5.93	0.03	4.48	
32	5.02	0.03	3.16	
24	3.85	0.03	2.4	
12	1.26	0.03	1.21	

**Table 45-50. Typical Sleep Mode Current Consumption Versus Frequency**

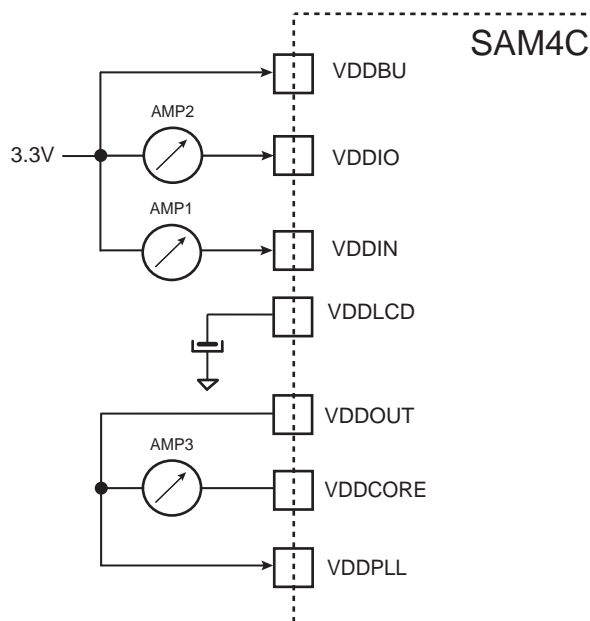
Master Clock (MHz)	IDD_IN - AMP1	IDD_IO - AMP2	IDD_CORE - AMP3	Unit
8	0.88	0.03	0.83	mA
4	0.5	0.03	0.45	
2	0.32	0.03	0.27	
1	0.26	0.03	0.22	
0.5	0.22	0.03	0.2	
0.25	0.19	0.03	0.18	

#### 45.7.4 Active Mode Power Consumption

The current consumption configuration for active mode, i.e. Core executing codes, are as follows:

- VDDIO = VDDIN = 3.3V
- VDDCORE = 1.2V (Internal Voltage regulator used)
- T<sub>A</sub> = 25°C
- Sub-system 0 Master Clock (MCK) and Core Clock (HCLK), Sub-system 1 Master Clock (CPBMCK) and Core Clock (CPHCLK) running at various frequencies (PLL used for frequencies above 8 MHz and fast RC oscillator at 8 MHz divided by 1/2/4/8/16/32 for lower frequencies)
- All Peripheral clocks deactivated
- No activity on IO lines
- Flash Wait State (FWS) in EEFC\_FMR adjusted versus core frequency
- Current measurement as per [Figure 45-21](#)

**Figure 45-21. Measurement Setup for Active Mode**



#### 45.7.4.1 Test Setup 1: CoreMark™

- CoreMark on Core 0 (CM4P0) running out of flash in 128-bit or 64-bit access mode with and without Cache Enabled. Cache is enabled above 0 WS.
- Sub-system 1 Master Clock (CPBMCK) and Core Clock (CPHCLK) stopped and in reset state

**Table 45-51. Test Setup 1 Current Consumption**

Clock (MHz)	128-bit Flash Access						64-bit Flash Access						Unit
	Cache Enabled			Cache Disabled			Cache Enabled			Cache Disabled			
	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	
120	26	0.3	23	29	2	26	26	0.27	23	26	1.9	22	mA
100	22	0.3	19	25	1.8	22.5	22	0.27	19	21	1.75	18.4	
84	19	0.3	17	22	1.7	19	19	0.27	17	19.5	1.65	16.2	
64	15	0.3	12	16	1.5	14	15	0.27	12	14	1.44	13.4	
48	11.5	0.3	9.85	13.5	1.4	12	11.5	0.27	9.85	12	1.32	12	
32	9.5	0.3	7.5	10	1.2	8.5	9.5	0.27	7.5	9	1.2	8	
24	5.6	0.3	4.35	7.7	1.1	6.7	5.6	0.27	4.35	7	1.17	6	
12	2.4	0.05	2.3	3.2	0.9	3	2.4	0.09	2.3	3.1	1	3.1	
8	1.65	0.05	1.61	2.1	0.7	2	1.65	0.09	1.61	2.1	0.9	2.1	
4	1	0.05	1	1.4	0.5	1.36	1	0.09	1	1.36	0.8	1.4	
2	0.7	0.05	0.68	0.9	0.4	0.9	0.7	0.09	0.68	0.7	0.7	0.7	
1	0.55	0.05	0.53	0.65	0.3	0.65	0.55	0.09	0.53	0.65	0.4	0.65	
0.5	0.47	0.05	0.45	0.5	0.2	0.5	0.47	0.09	0.45	0.6	0.2	0.6	
0.25	0.25	0.05	0.24	0.26	0.1	0.25	0.25	0.09	0.24	0.36	0.1	0.25	

#### 45.7.4.2 Test Setup 2: CoreMark

- CoreMark on Core 1 (CM4P1) running out of SRAM1 (Code) / SRAM2 (Data)
- Core 0 (CM4P0) in Sleep Mode.

**Table 45-52. Test Setup 2 Current Consumption**

Clock (MHz)	SRAM1, SRAM2			Unit
	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	
120	22.3	0.05	19	mA
100	19	0.05	16	
84	16	0.05	13.83	
64	12.2	0.05	10.66	
48	9.2	0.05	7.9	
32	7.15	0.05	5.5	
24	5.4	0.05	4.15	
12	2.1	0.05	2	
8	1.5	0.05	1.4	
4	0.8	0.05	0.75	
2	0.5	0.05	0.45	
1	0.32	0.05	0.38	
0.5	0.21	0.05	0.2	
0.25	0.13	0.05	0.12	

#### 45.7.4.3 Test Setup 3: CoreMark

- CoreMark on Core 0 (CM4P0) running out of Flash in 128-bit or 64-bit access mode with and without Cache Enabled. Cache is enabled above 0 WS.
- CoreMark on Core 1 (CM4P1) running out of SRAM1 (Code) / SRAM2 (Data)

**Table 45-53. Test Setup 3 Current Consumption**

Clock (MHz)	128-bit Flash Access						64-bit Flash Access						Unit
	Cache Enabled			Cache Disabled			Cache Enabled			Cache Disabled			
	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	
120	32.5	0.28	29	35	2	30	31.2	0.28	28	30.8	1.8	27.4	mA
100	27.8	0.28	32.3	29.8	1.75	27	26.4	0.28	23.6	27	1.7	24.3	
84	23.8	0.28	28.5	26.3	1.68	24	22.3	0.28	20	24	1.7	21.8	
64	18.6	0.28	23.5	20.9	1.5	19.3	17.2	0.28	15.6	19.7	1.6	18	
48	14.4	0.28	18.3	16.5	1.4	15.3	13	0.28	11.8	16	1.5	14.7	
32	10.8	0.28	14.2	12.6	1.2	10.9	9.8	0.28	8.15	12.5	1.4	10.6	
24	8.3	0.28	12	9.5	1.1	8.25	7.4	0.28	6.2	9.4	1.2	8.2	
12	3.6	0.1	6.1	4.2	0.9	4	3.1	0.1	3.1	4.2	1.1	4.2	
8	2.1	0.1	2	2.85	0.8	2.8	2.1	0.1	2.1	2.8	1	2.7	
4	1.1	0.1	1	1.5	0.6	1.45	1.2	0.1	1.2	1.5	0.9	1.4	
2	0.65	0.1	0.61	0.82	0.4	0.8	0.65	0.1	0.6	0.8	0.66	0.75	
1	0.38	0.1	0.35	0.47	0.25	0.45	0.38	0.1	0.37	0.47	0.38	0.4	
0.5	0.25	0.1	0.24	0.3	0.18	0.28	0.25	0.1	0.23	0.3	0.25	0.28	
0.25	0.14	0.1	0.13	0.16	0.1	0.14	0.15	0.1	0.12	0.16	0.15	0.13	

#### 45.7.4.4 Test Setup 4: DSP Application (Five Cascaded 4th Order Biquad Filters) from ARM CMSIS DSP Library

- Application running on Core 1 (CM4P1) out of SRAM1 (Code) / SRAM2 (Data)
- Core 0 (CM4P0) in Sleep Mode.

**Table 45-54. Test Setup 4 Current Consumption**

Clock (MHz)	DSP Application			Unit
	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	
120	22.8	0.05	19.4	mA
100	19.2	0.05	17	
84	16.25	0.05	13.83	
64	12.53	0.05	10.66	
48	9.53	0.05	8.08	
32	7.44	0.05	5.59	
24	5.68	0.05	4.23	
12	2.17	0.05	2.13	
8	1.48	0.05	1.45	
4	0.8	0.05	0.76	
2	0.47	0.05	0.43	
1	0.3	0.05	0.27	
0.5	0.25	0.05	0.22	
0.25	0.18	0.05	0.16	

#### 45.7.4.5 Test Setup 5: DSP Application (Five Cascaded 4th Order Biquad Filters) from ARM CMSIS DSP Library

- Application running on Core 0 (CM4P0) out of Flash in 128-bit or 64-bit access mode with and without Cache Enabled. Cache is enabled above 0 WS.
- Sub-system 1 Master Clock (CPBMCK) and Core Clock (CPHCLK) stopped and in reset.
- VDDIO = VDDIN = 3V

**Table 45-55. Test Setup 5 Current Consumption**

Clock (MHz)	128-bit Flash Access						Unit
	Cache Enabled			Cache Disabled			
	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	IDD_IN (AMP1)	IDD_I0 (AMP2)	IDD_CORE (AMP3)	
120	24	0.31	22	26.5	2.1	23	mA
100	21	0.31	19	23.7	2	21	
84	18	0.31	16	21.2	1.9	19	
64	16	0.31	14	17.2	1.8	15.5	
48	13	0.31	9	14	1.6	12.7	
32	9	0.31	6	10.6	1.4	9	
24	6	0.31	4.5	8.7	1.2	7.45	
12	3	0.05	2.6	4	0.82	3.8	
8	1.8	0.05	1.77	2.6	1.25	2.5	
4	1	0.05	0.9	1.5	0.9	1.4	
2	0.56	0.05	0.54	0.85	0.5	0.75	
1	0.38	0.05	0.37	0.475	0.27	0.47	
0.5	0.27	0.05	0.25	0.375	0.15	0.32	
0.25	0.22	0.05	0.2	0.275	0.1	0.24	

## 45.7.5 Peripheral Power Consumption in Active Mode

**Table 45-56. Power Consumption on  $V_{DDCORE}$** <sup>(1)</sup>

Peripheral	Consumption (Typical)	Unit
PIO Controller: PIO Read	TBD	$\mu\text{A}/\text{MHz}$
UART0: Tx/Rx Transfer @ 115200 with PDC	TBD	
UART1: Optilink Enable	TBD	
USART[0-4]: Tx/Rx Transfer @ 115200 with PDC, Loop back Mode	TBD	
PWM: Generating 10 kHz signal on all 4 channels	TBD	
TWI: Master Mode / Data write with TWCK 400 kHz	TBD	
SPI: SPCK @ 8 MHz with writing data @ 16 kHz frame rate	TBD	
Timer Counter (TCx): Generating 10 ms Tick	TBD	
ADC: Free run mode / 10-bit Resolution with 2-bit enhanced resolution by interpolation @ 10 kS/s	TBD	
SMC: Read/Write accesses with PSRAM @ Maximum Wait States	TBD	
SLCD: Display Rate at the maximum possible, Maximum Segments and COM used.	TBD	
AES: Performing AES256 Encryption	164	
TRNG: Performing Random Number Generation	TBD	
ICM: Performing Memory Integrity Check on SHA256 on SRAM Block	TBD	

Note: 1. Note:  $V_{DDIO} = 3.3\text{V}$ ,  $V_{DDCORE} = 1.2\text{V}$ ,  $T_A = 25^\circ\text{C}$ .

## 45.7.6 Low-power Mode Wake-up Time

Table 45-57 gives time required for the device to wake up from any low-power mode to active mode. The time is measured from a wake-up event detection to the first instruction fetch from flash memory.

**Table 45-57. Wake-up Time versus Low-power Mode**

Flash Mode	RC Osc @ 4 MHz	RC Osc @ 8 MHz	RC Osc @ 12 MHz	Unit
Read-Idle	TBD	TBD	TBD	$\mu\text{s}$
Stand-By	TBD	TBD	TBD	
Deep Power-Down	TBD	TBD	TBD	





## 46.2 Soldering Profile

Table 46-4 gives the recommended soldering profile from J-STD-020C.

**Table 46-4. Soldering Profile**

Profile Feature	Green Package
Average Ramp-up Rate (217°C to Peak)	3°C/sec. max.
Preheat Temperature 175°C ± 25°C	180 sec. max.
Temperature Maintained Above 217°C	60 sec. to 150 sec.
Time within 5°C of Actual Peak Temperature	20 sec. to 40 sec.
Peak Temperature Range	260°C
Ramp-down Rate	6°C/sec. max.
Time 25°C to Peak Temperature	8 min. max.

Note: The package is certified to be backward compatible with Pb/Sn soldering profile.  
A maximum of three reflow passes is allowed per component.

## 46.3 Packaging Resources

This section provides land pattern definition.

Refer to the following IPC standards:

- IPC-7351A and IPC-782 (*Generic Requirements for Surface Mount Design and Land Pattern Standards*)  
<http://landpatterns.ipc.org/default.asp>
- Atmel Green and RoHS Policy and Package Material Declaration Data Sheet  
<http://www.atmel.com/about/quality/package.aspx>

## 47. Ordering Information

Figure 47-1. Ordering Codes for SAM4C Devices

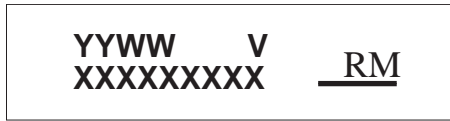
Ordering Code	MRL	Flash (Kbytes)	Package	Package Type	Temperature Operating Range
ATSAM4C16CA-AU	A	1*1024	LQFP100	Green	Industrial (-40°C to +85°C)
ATSAM4C16CA-AUR	A	1*1024	LQFP100	Green	Industrial (-40°C to +85°C)
ATSAM4C8CA-AU	A	1*512	LQFP100	Green	Industrial (-40°C to +85°C)
ATSAM4C8CA-AUR	A	1*512	LQFP100	Green	Industrial (-40°C to +85°C)

## 48. SAM4C Series Errata

### 48.1 Marking

All devices are marked with the Atmel logo and the ordering code.

Additional marking is as follows:



where

- “YY”: Manufactory year
- “WW”: Manufactory week
- “V”: Revision
- “XXXXXXXXX”: Lot number

### 48.2 Device Identification

It applies to devices marked:

- ATSAM4C16CA-AU (CHIPID: 0xA64C0CE0)
- ATSAM4C16CA-AUR (CHIPID: 0xA64C0CE0)
- ATSAM4C8CA-AU (CHIPID: 0xA64C0AE0)
- ATSAM4C8CA-AUR (CHIPID: 0xA64C0AE0)

## 48.3 Supply Controller (SUPC)

### 48.3.1 SUPC: LCD End of Frame Disable Does Not Work

LCDMODE (SUPC\_MR 0x400E1418 bit 4:5) with EOF\_LCDOFF (0x01) does not turn off SLCDC power between frames. This does not affect the other SLCD features. SLCD is fully functional.

#### Problem Fix/Workaround

None.

### 48.3.2 SUPC: Supply Monitor (SM) on VDDIO

The Supply Monitor (SM) sampling mode reducing the average current consumption on VDDIO is not functional.

#### Problem Fix/Workaround

Use the Supply Monitor in continuous mode only.

### 48.3.3 SUPC: Core Voltage Regulator Standby Mode Control

The Core Voltage Regulator standby mode controlled by the ONREG bit in SUPC\_MR is not functional. This does not prevent to power VDDCORE and VDDPL by using an external voltage regulator.

#### Problem Fix/Workaround

None. Do not use the ONREG Bit.

### 48.3.4 SUPC: Core Brownout Detector. Unpredictable Behavior if BOD is Disabled, VDDCORE is Lost and VDDIO is Powered

In active mode or in wait mode, if the Brownout Detector (BOD) is disabled (SUPC\_MR: BODDIS=1) and power is lost on VDDCORE while VDDIO is powered, the device can be reset incorrectly and its behavior becomes then unpredictable.

#### Problem Fix/Workaround

When the Brownout Detector is disabled in active or in wait mode, VDDCORE must be always powered.

## 48.4 Parallel Input Output (PIO) Controller

### 48.4.1 PIO: Schmitt Trigger

- Schmitt triggers on all PIO controllers are not enabled by default (after reset) as stated in the product datasheet
- Enable and disable values in the PIO Schmitt Trigger Register (for all PIO controllers) are swapped. The definition of PIO\_SCHMITT fields must be as follows:
  - 0: Schmitt Trigger is disabled.
  - 1: Schmitt Trigger is enabled.

#### Problem Fix/Workaround

None. It is up to the application to enable Schmitt trigger mode and to take into account the swapped values of the PIO\_SCHMITT fields.

## 48.5 Watchdog (WDT) / Reinforced Safety Watchdog (RSWDT)

### 48.5.1 WDT / RSWDT not stopped in WAIT mode

When the Watchdog (WDT) or the Reinforced Safety Watchdog (RSWDT) is enabled and the WAITMODE bit set to 1 is used to enter low-power wait mode, the WDT/RSWDT is not halted. If the time spent in wait mode is longer than the Watchdog (Reinforced Safety Watchdog) time-out, the device is reset provided that the WDT/RSWDT reset is enabled.

#### Problem Fix/Workaround

When entering wait mode, the Wait-For-Event (WFE) instruction of the Cortex-M4 processor must be used while the SLEEPDEEP bit of the Cortex-M System Control Register (SCB\_SCR) is set to 0.

## 48.6 Enhanced Embedded Flash Controller (EEFC)

### 48.6.1 Erase Sector (ES) command cannot be performed if a subsector is locked (ONLY in Flash sector 0)

If one of the subsectors

- small sector 0
- small sector 1
- larger sector

is locked within the Flash sector 0, the erase sector (ES) command cannot be processed on non-locked subsectors. Refer to the Flash overview in the “Memories” section of the datasheet.

#### Fix/Workaround

All the lock bits of the sector 0 must be cleared prior to issuing the ES command. After the ES command has been issued, the lock bits must be reverted to the state before clearing them.

## 49. Revision History

In the tables that follow, the most recent version of the document appears first.

**Table 49-1. SAM4C Datasheet Rev. 11102B 02-Dec-13 Revision History**

Doc. Rev. 11102B	Changes
02-Dec-13	Removed preliminary status.
	<a href="#">Section 45. "SAM4C Series Electrical Characteristics"</a>
	<a href="#">Table 45-16 "LCD Voltage Regulator Characteristics"</a> : Changed min and max values for $V_{VDDLCD}$ 'Output voltage accuracy' parameter. Added new characteristic $d_{VOUT}/d_{VDDIN}$ 'VDDLCD variation with VDDIN' with typ and max values.
	<a href="#">Table 45-18 "LCD Buffers Characteristics"</a> : Changed min, typ and max values for parameter $Z_{OUT}$ 'Buffer output impedance'. Changed convergence value and max value for $t_r / t_f$ 'Rising or falling time'.
	<a href="#">Table 45-26 "4/8/12 MHz RC Oscillators Characteristics"</a> : Modified conditions for ACC4, ACC8 and ACC12. Modified max values for ACC8 and ACC12.
	<a href="#">Table 45-31 "PLLA Characteristics"</a> : modified $t_{ON}$ 'Startup time'. Added new parameter $t_{LOCK}$ 'Lock time'.
	<a href="#">Table 45-42 "Programmable Voltage Reference Characteristics"</a> : Modified ACC 'reference voltage accuracy' min and max values and added conditions. Modified max value of $T_C$ 'Temperature coefficient'.
	<a href="#">Table 45-43 "Programmable Voltage Reference Selection Values"</a> : all ADVREF values modified.
	<a href="#">Section 48. "SAM4C Series Errata"</a>
	Added erratum on EFC: <a href="#">"Erase Sector (ES) command cannot be performed if a subsector is locked (ONLY in Flash sector 0)"</a>

**Table 49-2. SAM4C Datasheet Rev. 11102A 15-Oct-13 Revision History**

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15-Oct-13	First issue

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