## 8-BIT MICROCONTROLLER

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## 1 GENERAL DESCRIPTION

The W79E825 series are an 8-bit Turbo 51 microcontroller which has an in-system programmable Flash EPROM which Flash EPROM can program by ICP (In Circuit Program) or by hardware writer. The instruction set of the W79E825 series are fully compatible with the standard 8052. The W79E825 series contain a $16 \mathrm{~K} / 8 \mathrm{~K} / 4 \mathrm{~K} / 2 \mathrm{~K}$ bytes of main Flash EPROM; a $256 / 128$ bytes of RAM; 256/128 bytes NVM Data Flash EPROM; two 8-bit bi-directional, one 2-bit bi-directional and bit-addressable I/O ports; two 16-bit timer/counters; 4-channel multiplexed 10-bit A/D convert; 4-channel 10-bit PWM; two serial ports that include a I2C and an enhanced full duplex serial port. These peripherals are supported by 13 sources four-level interrupt capability. To facilitate programming and verification, the Flash EPROM inside the W79E825 series allow the program memory to be programmed and read electronically. Once the code is confirmed, the user can protect the code for security.

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## 2 FEATURES

- Fully static design 8-bit Turbo 51 CMOS microcontroller up to 20 MHz when $\mathrm{VDD}=4.5 \mathrm{~V}$ to 5.5 V , 12 MHz when VDD=2.7V to 5.5 V
- $\mathbf{1 6 K} / 8 K / 4 K / 2 K$ bytes of AP Flash EPROM, with ICP and external writer programmable mode.
- 256/128 bytes of on-chip RAM.
- 256/128 bytes NVM Data Flash EPROM for customer data storage used and 10K writer cycles.
- Instruction-set compatible with MSC-51.
- Two 8-bit bi-directional and one 2-bit bi-directional ports.
- Two 16-bit timer/counters.
- 13 interrupts source with four levels of priority.
- One enhanced full duplex serial port with framing error detection and automatic address recognition.
- The 4 outputs mode and TTL/Schmitt trigger selectable Port.
- Programmable Watchdog Timer.
- Four-channel 10-bit PWM (Pulse Width Modulator).
- Four-channel multiplexed with 10-bits A/D convert.
- One I2C communication port (Master / Slave).
- Eight keypad interrupt inputs.
- Two analog comparators.
- Configurable on-chip oscillator.
- LED drive capability ( 20 mA ) on all port pins.
- Low Voltage Detect interrupt and reset.
- Development Tools:
- JTAG ICE(In Circuit Emulation) tool
- ICP(In Circuit Programming) writer
- Packages:
- Lead Free (RoHS) DIP 20: W79E825ADG
- Lead Free (RoHS) DIP 20: W79E825ADG
- Lead Free (RoHS) SOP 20: W79E825ASG
- Lead Free (RoHS) DIP 20: W79E824ADG
- Lead Free (RoHS) SOP 20: W79E824ASG
- Lead Free (RoHS) DIP 20: W79E823BDG
- Lead Free (RoHS) SOP 20: W79E823BSG
- Lead Free (RoHS) DIP 20: W79E822BDG
- Lead Free (RoHS) SOP 20: W79E822BSG

3 PARTS INFORMATION LIST
3.1 Lead Free (RoHS) Parts information list

| PART NO. | EPROM FLASH <br> SIZE | RAM | NVM FLASH <br> EPROM | PACKAGE | REMARK |
| :---: | :---: | :---: | :---: | :---: | :---: |
| W79E825ADG | $16 K B$ | $256 B$ | $256 B$ | DIP-20 Pin |  |
| W79E825ASG | $16 K B$ | $256 B$ | $256 B$ | SOP-20 Pin |  |
| W79E824ADG | $8 K B$ | $256 B$ | $256 B$ | DIP-20 Pin |  |
| W79E824ASG | $8 K B$ | $256 B$ | $256 B$ | SOP-20 Pin |  |
| W79E823BDG | $4 K B$ | $128 B$ | $128 B$ | DIP-20 Pin |  |
| W79E823BSG | $4 K B$ | $128 B$ | $128 B$ | SOP-20 Pin |  |
| W79E822BDG | $2 K B$ | $128 B$ | $128 B$ | DIP-20 Pin |  |
| W79E822BSG | $2 K B$ | $128 B$ | $128 B$ | SOP-20 Pin |  |

Table 3-1: Lead Free (RoHS) Parts information list

## 4 PIN CONFIGURATION



Figure 4-1: Pin Configuration

## 5 PIN DESCRIPTION

| SYMBOL | TYPE | DESCRIPTIONS |
| :---: | :---: | :--- |
| $\overline{\text { RST (P1.5) }}$ | I | RESET: A low on this pin for two machine cycles while the oscillator is <br> running resets the device. |
| XTAL1(P2.1) | I/O | CRYSTAL1: This is the crystal oscillator input. This pin may be driven by <br> an external clock or configurable I/O pin. |
| XTAL2(P2.0) | I/O | CRYSTAL2: This is the crystal oscillator output. It is the inversion of <br> XTAL1 or configurable I/O pin. |
| VSS | P | GROUND: Ground potential |
| VDD | P | POWER: SUPPLY: Supply voltage for operation. |
| P0.0-P0.7 | I/O | PORT 0: Port 0 is four mode output pin and two mode input. The <br> P0.3~P0.6 are 4-channel input ports (ADC0-ADC3) for ADC used. |
| P1.0-P1.7 | I/O | PORT 1: Port 1 is four mode output pin and two mode input. The P1.2 <br> (SCL) and P1.3 (SDA) is only open drain circuit, and P1.5 only input pin. |

* TYPE: P: power, I: input, O: output, I/O: bi-directional, H: pull-high, L: pull-low, D: open-drain.

Table 5-1: Pin Description

## 6 FUNCTIONAL DESCRIPTION

The W79E825 series architecture consist of a 4T 8051 core controller surrounded by various registers, 16K/8K/4K/2K bytes Flash EPROM, 256/128 bytes of RAM, 256/128 bytes NVM Data Flash EPROM, three general purpose I/O ports, two timer/counters, one serial port, one I2C serial I/O, 4 channel PWM with 10-bit counter, 4-channel multiplexed with 10-bit ADC analog input, Flash EPROM program by Writer and ICP.

### 6.1 On-Chip Flash EPROM

The W79E825 series include one $\mathbf{1 6 K / 8 K / 4 K / 2 K}$ bytes of main Flash EPROM for application program. A Writer or ICP programming board is required to program the Flash EPROM or NVM Data Flash EPROM.
This ICP (In-Circuit Programming) feature makes the job easy and efficient when the application's firmware needs to be updated frequently. In some applications, the in-circuit programming feature makes it possible for the end-user to easily update the system firmware without opening the chassis.

### 6.2 I/O Ports

The W79E825 series have two 8-bit and one 2-bit port, up to $18 \mathrm{I} / \mathrm{O}$ pins using on-chip oscillator \& /RST is input only by reset options. All ports can be used as four outputs mode when it may set by PxM1.y and PxM2.y SFR's registers, it has strong pull-ups and pull-downs, and does not need any external pull-ups. Otherwise it can be used as general I/O port as open drain circuit. All ports can be used bi-directional and these are as I/O ports. These ports are not true I/O, but rather are pseudo-I/O ports. This is because these ports have strong pull-downs and weak pull-ups.

### 6.3 Serial I/O

The W79E825 series have one serial port that is functionally similar to the serial port of the original 8032 family. However the serial port on the W79E825 series can operate in different modes in order to obtain timing similarity as well. The Serial port has the enhanced features of Automatic Address recognition and Frame Error detection.

### 6.4 Timers

The W79E825 series have two 16-bit timers that are functionally and similar to the timers of the 8052 family. When used as timers, the user has a choice of 12 or 4 clocks per count that emulates the timing of the original 8052.

### 6.5 Interrupts

The Interrupt structure in the W79E825 series is slightly different from that of the standard 8052. Due to the presence of additional features and peripherals, the number of interrupt sources and vectors has been increased.

### 6.6 Data Pointers

The data pointers of W79E825 series are same as 8052 that has dual 16-bit Data Pointers (DPTR) by setting DPS bit at AUXR1.0. The figure of dual DPTR is as below diagram.


Figure 6-1: Dual DPTR

### 6.7 Architecture

The W79E825 series are based on the standard 8052 device. It is built around an 8-bit ALU that uses internal registers for temporary storage and control of the peripheral devices. It can execute the standard 8052 instruction set.

### 6.7.1 ALU

The ALU is the heart of the W79E825 series. It is responsible for the arithmetic and logical functions. It is also used in decision making, in case of jump instructions, and is also used in calculating jump addresses. The user cannot directly use the ALU, but the Instruction Decoder reads the op-code, decodes it, and sequences the data through the ALU and its associated registers to generate the required result. The ALU mainly uses the ACC which is a special function register (SFR) on the chip. Another SFR, namely B register is also used in Multiply and Divide instructions. The ALU generates several status signals which are stored in the Program Status Word register (PSW).

### 6.7.2 Accumulator

The Accumulator (ACC) is the primary register used in arithmetic, logical and data transfer operations in the W79E825 series. Since the Accumulator is directly accessible by the CPU, most of the high speed instructions make use of the ACC as one argument.

### 6.7.3 B Register

This is an 8-bit register that is used as the second argument in the MUL and DIV instructions. For all other instructions it can be used simply as a general purpose register.

### 6.7.4 Program Status Word:

This is an 8-bit SFR that is used to store the status bits of the ALU. It holds the Carry flag, the Auxiliary Carry flag, General purpose flags, the Register Bank Select, the Overflow flag, and the Parity flag.

### 6.7.5 Scratch-pad RAM

The W79E825 series have a 256/128 bytes on-chip scratch-pad RAM. These can be used by the user for temporary storage during program execution. A certain section of this RAM is bit addressable, and can be directly addressed for this purpose.

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### 6.7.6 Stack Pointer

The W79E825 series have an 8-bit Stack Pointer which points to the top of the Stack. This stack resides in the Scratch Pad RAM in the W79E825 series. Hence the size of the stack is limited by the size of this RAM.

### 6.8 Power Management

Power Management like the standard 8052, the W79E825 series also have the IDLE and POWER DOWN modes of operation. In the IDLE mode, the clock to the CPU is stopped while the timers, serial ports and interrupt lock continue to operate. In the POWER DOWN mode, all clocks are stopped and the chip operation is completely stopped. This is the lowest power consumption state.

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## 7 MEMORY ORGANIZATION

The W79E825 series separate the memory into two separate sections, the Program Memory and the Data Memory. The Program Memory is used to store the instruction op-codes, while the Data Memory is used to store data or for memory mapped devices.

### 7.1 Program Memory (on-chip Flash)

The Program Memory on the W79E825 series can be up to $\mathbf{1 6 K / 8 K / 4 K / 2 K}$ bytes long. All instructions are fetched for execution from this memory area. The MOVC instruction can also access this memory region.

### 7.2 Data Memory

The NVM Data Memory of Flash EPROM on the W79E825 series can be up to 256/128 bytes long. The W79E825 series read the content of data memory by using "MOVC A, @A+DPTR". To write data is by NVMADDR, NVMDAT and NVMCON SFR's registers.


Figure 7-1: W79E825/824 Memory Map

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Figure 7-2: W79E823/822 Memory Map

### 7.3 Register Map

As mentioned before the W79E825 series have separate Program and Data Memory areas. The onchip 256/128 bytes scratch pad RAM is in addition to the internal memory. There are also several Special Function Registers (SFRs) which can be accessed by software. The SFRs can be accessed only by direct addressing, while the on-chip RAM can be accessed by either direct or indirect addressing.

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Figure 7-3: W79E825 and W79E824 RAM and SFR memory map


Figure 7-4: W79E823 and W79E822 RAM and SFR memory map

Since the scratch-pad RAM is only $\mathbf{2 5 6 / 1 2 8}$ bytes it can be used only when data contents are small. There are several other special purpose areas within the scratch-pad RAM. These are described as follows.

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Figure 7-5: Scratch pad RAM

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### 7.4 Working Registers

There are four sets of working registers, each consisting of eight 8 -bit registers. These are termed as Banks 0, 1, 2, and 3 . Individual registers within these banks can be directly accessed by separate instructions. These individual registers are named as R0, R1, R2, R3, R4, R5, R6 and R7. However, at one time the W79E825 series can work with only one particular bank. The bank selection is done by setting RS1-RS0 bits in the PSW. The R0 and R1 registers are used to store the address for indirect accessing.

### 7.5 Bit addressable Locations

The Scratch-pad RAM area from location 20h to 2 Fh is byte as well as bit addressable. This means that a bit in this area can be individually addressed. In addition some of the SFRs are also bit addressable. The instruction decoder is able to distinguish a bit access from a byte access by the type of the instruction itself. In the SFR area, any existing SFR whose address ends in a 0 or 8 is bit addressable.

### 7.6 Stack

The scratch-pad RAM can be used for the stack. This area is selected by the Stack Pointer (SP), which stores the address of the top of the stack. Whenever a jump, call or interrupt is invoked the return address is placed on the stack. There is no restriction as to where the stack can begin in the RAM. By default however, the Stack Pointer contains 07h at reset. The user can then change this to any value desired. The SP will point to the last used value. Therefore, the SP will be incremented and then address saved onto the stack. Conversely, while popping from the stack the contents will be read first, and then the SP is decreased.

## 8 SPECIAL FUNCTION REGISTERS

The W79E825 series uses Special Function Registers (SFRs) to control and monitor peripherals and their Modes. The SFRs reside in the register locations $80-\mathrm{FFh}$ and are accessed by direct addressing only. Some of the SFRs are bit addressable. This is very useful in cases where users wish to modify a particular bit without changing the others. The SFRs that are bit addressable are those whose addresses end in 0 or 8 . The W79E825 series contain all the SFRs present in the standard 8052. However some additional SFRs are added. In some cases the unused bits in the original 8052, have been given new functions. The list of the SFRs is as follows.

| F8 | IP1 |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| F0 | B |  |  |  |  |  | POID | IP1H |
| E8 | EIE |  |  |  |  |  |  |  |
| E0 | ACC | ADCCON | ADCH |  |  |  |  |  |
| D8 | WDCON | PWMPL | PWM0L | PWM1L | PWMCON1 | PWM2L | PWM3L | PWMCON2 |
| D0 | PSW | PWMPH | PWM0H | PWM1H |  | PWM2H | PWM3H | PWMCON3 |
| C8 |  |  |  |  |  |  | NVMCON | NVMDAT |
| C0 | I2CON | I2ADDR |  |  |  | NVMADDR | TA |  |
| B8 | IP0 | SADEN |  |  | I2DAT | I2STATUS | I2CLK | I2TIMER |
| B0 |  | P0M1 | P0M2 | P1M1 | P1M2 | P2M1 | P2M2 | IP0H |
| A8 | IE | SADDR |  |  | CMP1 | CMP2 |  |  |
| A0 | P2 | KBI | AUXR1 |  |  |  |  |  |
| 98 | SCON | SBUF |  |  |  |  |  |  |
| 90 | P1 |  |  |  |  | DIVM |  |  |
| 88 | TCON | TMOD | TL0 | TL1 | TH0 | TH1 | CKCON |  |
| $80 ~$ | P0 | SP | DPL | DPH |  |  |  | PCON |

Table 8-1: Special Function Register Location Table

Note: 1. The SFRs in the column with dark borders are bit-addressable
2. The table is condensed with eight locations per row. Empty locations indicate that these are no registers at these addresses. When a bit or register is not implemented, it will read high.

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| SYMBOL | DEFINITION | $\begin{aligned} & \text { ADDR } \\ & \text { ESS } \end{aligned}$ | $\begin{aligned} & \text { MSB } \\ & \text { LSB } \end{aligned}$ | BIT_ADDRESS, SYMBOL |  |  |  |  |  |  | RESET |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IP1 | Interrupt priority 1 | F8H | (FF) | (FE) | (FD) PPWM | $\begin{aligned} & \text { (FC) } \\ & \text { PWDI } \end{aligned}$ | $\begin{aligned} & \text { (FB) } \\ & \text { PC2 } \end{aligned}$ | $\begin{aligned} & \text { (FA) } \\ & \mathrm{PC} 1 \end{aligned}$ | $\begin{aligned} & \text { (F9) } \\ & \text { PKB } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (F8) } \\ \text { PI2 } \end{array}$ | xx000000B |
| IP1H | Interrupt high priority 1 | F7H | - | - | PPWMH | PWDIH | PC 2 H | PC 1 H | PKBH | PI 2 H | xx000000B |
| POIDS | Port 0 Digital Input Disable | F6H |  |  |  |  |  |  |  |  | 00000000B |
| B | B register | FOH | (F7) | (F6) | (F5) | (F4) | (F3) | (F2) | (F1) | (F0) | 00000000B |
| EIE | Interrupt enable 1 | E8H | (EF) | (EE) | $\begin{aligned} & \text { (ED) } \\ & \text { EPWM } \end{aligned}$ | $\begin{aligned} & \text { (EC) } \\ & \text { EWDI } \end{aligned}$ | $\begin{aligned} & \hline(E B) \\ & E C 2 \end{aligned}$ | $\begin{aligned} & \hline(E A) \\ & E C 1 \end{aligned}$ | $\begin{aligned} & \text { (E9) } \\ & \text { EKB } \end{aligned}$ | $\begin{aligned} & \text { (E8) } \\ & \mathrm{El2} \end{aligned}$ | xx000000B |
| ADCH | ADC converter result | E2H | ADC. 9 | ADC. 8 | ADC. 7 | ADC. 6 | ADC. 5 | ADC. 4 | ADC. 3 | ADC. 2 | xxxxxxxxB |
| ADCCON | ADC control register | E1H | ADC. 1 | ADC. 0 | ADCEX | ADCI | ADCS | RCCLK | AADR1 | AADR0 | xx000x00B |
| ACC | Accumulator | EOH | (E7) | (E6) | (E5) | (E4) | (E3) | (E2) | (E1) | (E0) | 00000000B |
| PWMCON2 | PWM control register 2 | DFH | BKCH | BKPS | BPEN | BKEN | PWM3B | PWM2B | PWM1B | PWM0B | 00000000B |
| PWM3L | PWM 3 low bits register | DEH | PWM3.7 | PWM3.6 | PWM3.5 | PWM3.4 | PWM3.3 | PWM3.2 | PWM3.1 | PWM3.0 | 00000000B |
| PWM2L | PWM 2 low bits register | DDH | PWM2.7 | PWM2.6 | PWM2.5 | PWM2.4 | PWM2.3 | PWM2.2 | PWM2.1 | PWM2.0 | 00000000B |
| PWMCON1 | PWM control register 1 | DCH | PWMRUN | load | CF | CLRPWM | PWM3I | PWM2I | PWM1I | PWMOI | 00000000B |
| PWM1L | PWM 1 low bits register | DBH | PWM1.7 | PWM1.6 | PWM1.5 | PWM1.4 | PWM1.3 | PWM1.2 | PWM1.1 | PWM1.0 | 00000000B |
| PWMOL | PWM 0 low bits register | DAH | PWM0.7 | PWM0.6 | PWM0.5 | PWM0.4 | PWM0.3 | PWM0.2 | PWM0.1 | PWM0.0 | 00000000B |
| PWMPL | PWM counter low register | D9H | PWMP0.7 | PWMP0.6 | PWMP0.5 | PWMP0.4 | PWMP0.3 | PWMP0.2 | PWMP0.1 | PWMP0.0 | 00000000B |
| WDCON | Watch-Dog control | D8H | (DF) WDRUN | (DE) | $\begin{aligned} & \hline \text { (DD) } \\ & \text { WD1 } \end{aligned}$ | $\begin{aligned} & \hline \text { (DC) } \\ & \text { WDO } \end{aligned}$ | (DB) WDIF | (DA) WTRF | (D9) EWRST | (D8) WDCLR | 0x000000B |
| PWMCON3 | PWM control register 3 | D7H | - | - |  | - |  | - |  | BKF | xxxxxxx0B |
| PWM3H | PWM 3 high bits register | D6H | - | - |  | - |  | - | PWM3.9 | PWM3.8 | xxxxxx00B |
| PWM2H | PWM 2 high bits register | D5H | - | - | - | - |  | - | PWM2.9 | PWM2.8 | xxxxxx00B |
| PWM1H | PWM 1 high bits register | D3H | - | - |  | - |  | - | PWM1.9 | PWM1.8 | xxxxxx00B |
| PWMOH | PWM 0 high bits register | D2H | - | - | - | - | - | - | PWM0.9 | PWM0.8 | xxxxxx00B |
| PWMPH | PWM counter high register | D1H | - | - | - | - | - | - | $\begin{aligned} & \hline \text { PWMPO. } \\ & 9 \end{aligned}$ | $\left.\right\|_{8} ^{\mathrm{PWWMPO}} \mathrm{l}$ | xxxxxx00B |
| PSW | Program status word | DOH | $\begin{aligned} & \text { (D7) } \\ & \mathrm{CY} \end{aligned}$ | $\begin{aligned} & \text { (D6) } \\ & \text { AC } \end{aligned}$ | $\begin{aligned} & \hline \text { (D5) } \\ & \text { F0 } \end{aligned}$ | $\begin{aligned} & \text { (D4) } \\ & \text { RS1 } \end{aligned}$ | $\begin{aligned} & \text { (D3) } \\ & \text { RSO } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (D2) } \\ \text { OV } \\ \hline \end{array}$ | $\begin{array}{\|l\|} \hline \text { (D1) } \\ \hline \text { F1 } \end{array}$ | $\begin{aligned} & \text { (D0) } \\ & \hline P \end{aligned}$ | 00000000B |
| NVMDATA | NVM Data | CFH |  |  |  |  |  |  |  |  | 00000000B |
| NVMCON | NVM Control | CEH | EER | EWR |  | - |  | - |  | - | 00000000B |
| TA | Timed Access Protection | C7H | TA. 7 | TA. 6 | TA. 5 | TA. 4 | TA. 3 | TA. 2 | TA. 1 | TA. 0 | 11111111B |
| NVMADDR | NVM address | C6H |  |  |  |  |  |  |  |  | 00000000B |
| I2ADDR | I2C address1 | C1H | ADDR. 7 | ADDR. 6 | ADDR. 5 | ADDR. 4 | ADDR. 3 | ADDR. 2 | ADDR. 1 | GC | xxxxxxx0B |
| I2CON | 12C Control register | COH | (CF) | $\begin{aligned} & \text { (CE) } \\ & \text { ENS1 } \end{aligned}$ | $\begin{aligned} & \hline \text { (CD) } \\ & \text { STA } \end{aligned}$ | $\begin{aligned} & \hline \text { (CC) } \\ & \text { STO } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (CB) } \\ \hline \mathrm{SI} \\ \hline \end{array}$ | $\begin{aligned} & \hline \text { (CA) } \\ & \text { AA } \end{aligned}$ | (C9) | (C8) | x00000xxB |
| I2TIMER | I2C Timer Counter register | BFH | - | - | - | - | - | ENTI | DIV4 | TIF | 00000000B |
| 12CLK | I2C Clock Rate | BEH | I2CLK. 7 | I2CLK. 6 | I2CLK. 5 | I2CLK. 4 | I2CLK. 3 | I2CLK. 2 | I2CLK. 1 | I2CLK. 0 | 00000000B |
| I2STATUS |  | BDH |  |  |  |  |  |  |  |  | 00000000B |
| I2DAT |  | BCH | I2DAT. 7 | I2DAT. 6 | I2DAT. 5 | I2DAT. 4 | I2DAT. 3 | I2DAT. 2 | I2DAT. 1 | I2DAT. 0 | xxxxxxxxB |
| SADEN | Slave address mask | B9H |  |  |  |  |  |  |  |  | 00000000B |
| IP0 | Interrupt priority | B8H | (BF) | $\begin{aligned} & \hline \text { (BE) } \\ & \text { PADC } \end{aligned}$ | $\begin{array}{\|l\|} \hline(\mathrm{BD}) \\ \mathrm{PBO} \\ \hline \end{array}$ | $\begin{array}{\|l} \hline(\mathrm{BC}) \\ \mathrm{PS} \\ \hline \end{array}$ | $\begin{aligned} & \hline(\mathrm{BB}) \\ & \mathrm{PT1} \end{aligned}$ | $\begin{aligned} & \hline(\mathrm{BA}) \\ & \mathrm{PX1} \end{aligned}$ | $\begin{aligned} & \hline \text { (B9) } \\ & \text { PT0 } \end{aligned}$ | $\begin{aligned} & \hline \text { (B8) } \\ & \text { PX0 } \end{aligned}$ | x0000000B |
| IPOH | Interrupt high priority | B7H | - | PADCH | PBOH | PSH | PT1H | PX1H | PTOH | PXOH | x0000000B |
| P2M2 | Port 2 output mode 2 | B6H | - | - | - | - | - | - | P2M2.1 | P2M2.0 | xxxxxx00B |
| P2M1 | Port 2 output mode 1 | B5H | P2S | P1S | POS | ENCLK | T1OE | TOOE | P2M1.1 | P2M1.0 | 00000000B |
| P1M2 | Port 1 output mode 2 | B4H | P1M2.7 | P1M2.6 | - | P1M2.4 | P1M2.3 | P1M2.2 | P1M2.1 | P1M2.0 | 00000000B |
| P1M1 | Port 1 output mode 1 | B3H | P1M1.7 | P1M1.6 | - | P1M1.4 | P1M1.3 | P1M1.2 | P1M1.1 | P1M1.0 | 00000000B |
| POM2 | Port 0 output mode 2 | B2H | P0M2.7 | POM2.6 | P0M2.5 | POM2.4 | POM2.3 | POM2.2 | P0M2.1 | POM2.0 | 00000000B |
| P0M1 | Port 0 output mode 1 | B1H | P0M1.7 | P0M1.6 | P0M1.5 | P0M1.4 | P0M1.3 | P0M1.2 | P0M1.1 | P0M1.0 | 00000000B |

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Continued

| SYMBOL | DEFINITION | $\begin{aligned} & \text { ADDR } \\ & \text { ESS } \end{aligned}$ | $\begin{aligned} & \text { MSB } \\ & \text { LSB } \end{aligned}$ | BIT_ADDRESS, SYMBOL |  |  |  |  |  |  | RESET |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CMP2 | Comparator 2 control register | ADH | - | - | CE2 | CP2 | CN2 | OE2 | CO2 | CMF2 | 00000000B |
| CMP1 | Comparator 1 control register | ACH | - | - | CE1 | CP1 | CN1 | OE1 | CO1 | CMF1 | 00000000B |
| SADDR | Slave address | A9H |  |  |  |  |  |  |  |  | 00000000B |
| IE | Interrupt enable | A8H | $\begin{array}{\|l\|} \hline \text { (AF) } \\ \hline \end{array}$ | $\begin{aligned} & \text { (AE) } \\ & \text { EADC } \end{aligned}$ | $\begin{aligned} & \text { (AD) } \\ & \text { EBO } \end{aligned}$ | $\begin{aligned} & \text { (AC) } \\ & \text { ES } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { (AB) } \\ & \mathrm{ET} 1 \end{aligned}$ | $\begin{aligned} & \text { (AA) } \\ & \mathrm{EX1} \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (A9) } \\ \text { ETO } \end{array}$ | $\begin{aligned} & \text { (A8) } \\ & \text { EXO } \end{aligned}$ | 00000000B |
| AUXR1 | AUX function register | A2H | KBF | BOD | BOI | LPBOV | SRST | ADCEN | 0 | DPS | 000X0000B |
| KBI | Keyboard Interrupt | A1H |  |  |  |  |  |  |  |  | 00000000B |
| P2 | Port 2 | AOH | (A7) | (A6) | (A5) | (A4) | (A3) | (A2) | $\left\lvert\, \begin{aligned} & \text { (A1) } \\ & \text { XTAL1 } \end{aligned}\right.$ | $\begin{aligned} & \hline \text { (AO) } \\ & \text { XTAL2 } \\ & \text { CLKOUT } \end{aligned}$ | xxxxxx11B |
| SBUF | Serial buffer | 99H |  |  |  |  |  |  |  |  | xxxxxxxxB |
| SCON | Serial control | 98H | (9F) SM0/FE | $\begin{aligned} & \text { (9E) } \\ & \text { SM1 } \end{aligned}$ | $\begin{aligned} & \hline \text { (9D) } \\ & \text { SM2 } \end{aligned}$ | $\begin{aligned} & \text { (9C) } \\ & \text { REN } \end{aligned}$ | $\begin{aligned} & \text { (9B) } \\ & \text { TB8 } \end{aligned}$ | $\begin{aligned} & \text { (9A) } \\ & \text { RB8 } \end{aligned}$ | $\begin{aligned} & \hline(99) \\ & \mathrm{TI} \end{aligned}$ | $\begin{aligned} & \hline \text { (98) } \\ & \text { RI } \end{aligned}$ | 00000000B |
| DIVM | uC clock divide register | 95H |  |  |  |  |  |  |  |  | 00000000B |
| P1 | Port 1 | 90 H | (97) <br> PWM2 | (96) PWM1 | $\begin{aligned} & (95) \\ & \text { /RST } \end{aligned}$ | $\begin{aligned} & \text { (94) } \\ & \text { /INT1 } \end{aligned}$ | $\begin{aligned} & \hline \text { (93) } \\ & \text { /INT0 } \\ & \text { SDL } \end{aligned}$ | $\begin{aligned} & \hline \text { (92) } \\ & \text { TO } \\ & \text { SCL } \end{aligned}$ | $\begin{aligned} & (91) \\ & \text { RXD } \end{aligned}$ | $\begin{aligned} & (90) \\ & \text { TXD } \end{aligned}$ | 11111111B |
| CKCON | Clock control | 8EH | - | - | - | T1M | TOM | - | - | - | xxx00xxxB |
| TH1 | Timer high 1 | 8DH |  |  |  |  |  |  |  |  | 00000000B |
| TH0 | Timer high 0 | 8CH |  |  |  |  |  |  |  |  | 00000000B |
| TL1 | Timer low 1 | 8BH |  |  |  |  |  |  |  |  | 00000000B |
| TLO | Timer low 0 | 8AH |  |  |  |  |  |  |  |  | 00000000B |
| TMOD | Timer mode | 89H | GATE | C/T | M1 | M0 | GATE | C/T | M1 | M0 | 00000000B |
| TCON | Timer control | 88H | $\begin{aligned} & \text { (8F) } \\ & \text { TF1 } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (8E) } \\ \text { TR1 } \end{array}$ | $\begin{aligned} & \hline \text { (8D) } \\ & \text { TF0 } \end{aligned}$ | $\begin{aligned} & \text { (8C) } \\ & \text { TR0 } \end{aligned}$ | $\begin{array}{\|l\|} \hline \text { (8B) } \\ \text { IE1 } \end{array}$ | $\begin{array}{\|l\|} \hline \text { (8A) } \\ \text { IT1 } \end{array}$ | $\begin{array}{\|l\|} \hline(89) \\ \text { IE0 } \end{array}$ | $\begin{array}{\|l\|} \hline \text { (88) } \\ \text { ITO } \end{array}$ | 00000000B |
| PCON | Power control | 87H | SMOD | SMOD0 | BOF | POR | GF1 | GF0 | PD | IDL | 00xx0000B |
| DPH | Data pointer high | 83H |  |  |  |  |  |  |  |  | 00000000B |
| DPL | Data pointer low | 82H |  |  |  |  |  |  |  |  | 00000000B |
| SP | Stack pointer | 81H |  |  |  |  |  |  |  |  | 00000111B |
| P0 | Port 0 | 80H | $\begin{array}{\|l\|} \hline(87) \\ \mathrm{T} 1 \end{array}$ | $\begin{aligned} & \hline(86) \\ & \text { AD3 } \\ & \text { CMP1 } \end{aligned}$ | (85) <br> AD2 <br> CMPREF | (84) AD1 CIN1A | (83) ADO CIN1B | (82) <br> BRAKE CIN2A | $\begin{aligned} & \text { (81) } \\ & \text { PWM0 } \\ & \text { CIN2B } \end{aligned}$ | (80) PWM3 CMP2 | 11111111B |

Table 8-2: Special Function Registers

PORT 0

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P0.7 | P0.6 | P0.5 | P0.4 | P0.3 | P0.2 | P0.1 | P0.0 |

Mnemonic: PO
Address: 80h
P0.7-0: General purpose Input/Output port. Most instructions will read the port pins in case of a port read access, however in case of read-modify-write instructions, the port latch is read. These alternate functions are described below:

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 7 | P0.7 | Timer 1 pin or KBI.7 pin of keypad input. |
| 6 | P0.6 | CMP1 pin of analog comparator or KBI.6 pin of keypad input. |
| 5 | P0.5 | CMPREF pin of analog comparator or KBI.5 pin of keypad input. |
| 4 | P0.4 | CIN1A pin of analog comparator or KBI.4 pin of keypad input. |
| 3 | P0.3 | CIN1B pin of analog comparator or KBI.3 pin of keypad input. |
| 2 | P0.2 | BRAKE pin of PWM or CIN2A pin of analog comparator or KBI.2 pin of keypad input. |
| 1 | P0.1 | PWM0 pin or CIN2B pin of analog comparator or KBI.1 pin of keypad input. |
| 0 | P0.0 | PWM3 pin or CMP2 pin of analog comparator or KBI.0 pin of keypad input. |

Note: The initial value of the port is set by CONFIG1.PRHI bit. The default setting for CONFIG1.PRHI $=1$ which the alternative function output is turned on upon reset. If CONFIG1.PRHI is set to 0 , the user has to write a 1 to port SFR to turn on the alternative function output.

## STACK POINTER



## DATA POINTER LOW



Mnemonic: DPL
Address: 82h

| BIT | NAME | FUNCTION |
| :---: | :---: | :---: |
| $7-0$ | DPL.[7:0] | This is the low byte of the standard 8052 16-bit data pointer. |

## DATA POINTER HIGH



| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| $7-0$ | DPH.[7:0] | This is the high byte of the standard 8052 16-bit data pointer. <br> This is the high byte of the DPTR 16-bit data pointer. |

POWER CONTROL

| Bit: | 7 | 6 | 5 | 4 |  | 3 | 1 | 0 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | SMOD | SMODO | BOF | POR | GF1 | GF0 | PD | IDL |

Mnemonic: PCON
Address: 87h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | SMOD | 1: This bit doubles the serial port baud rate in mode 1, 2, and 3. |
| 6 | SMOD0 | 0: Framing Error Detection Disable. SCON.7 (SMO/FE) bit is used as SM0 <br> (standard 8052 function). <br> 1: Framing Error Detection Enable. SCON.7 (SMO/FE) bit is used to reflect as <br> Frame Error (FE) status flag. |
| 5 | BOF | 0: Cleared by software. <br> 1: Set automatically when a brownout reset or interrupt has occurred. Also set at <br> power on. |
| 4 | POR | 0: Cleared by software. <br> 1: Set automatically when a power-on reset has occurred. |
| 3 | GF1 | General purpose user flags. |
| 2 | GF0 | General purpose user flags. |
| 1 | PD | 1: The CPU goes into the POWER DOWN mode. In this mode, all the clocks are <br> stopped and program execution is frozen. |
| 0 | IDL | 1: The CPU goes into the IDLE mode. In this mode, the clocks CPU clock <br> stopped, so program execution is frozen. But the clock to the serial, timer and <br> interrupt blocks is not stopped, and these blocks continue operating. |

## TIMER CONTROL

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TF1 | TR1 | TFO | TR0 | IE1 | IT1 | IE0 | ITO |
| Mnemonic: TCON |  |  |  |  |  |  |  | Address: 88h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | TF1 | Timer 1 Overflow Flag. This bit is set when Timer 1 overflows. It is cleared automatically when the program does a timer 1 interrupt service routine. Software can also set or clear this bit. |  |  |  |  |  |  |
| 6 | TR1 | Timer 1 Run Control. This bit is set or cleared by software to turn timer/counter on or off. |  |  |  |  |  |  |

Continued

| BIT | NAME | FUNCTION |
| :--- | :--- | :--- |
| 5 | TF0 | Timer 0 Overflow Flag. This bit is set when Timer 0 overflows. It is cleared <br> automatically when the program does a timer 0 interrupt service routine. Software <br> can also set or clear this bit. |
| 4 | TR0 | Timer 0 Run Control. This bit is set or cleared by software to turn timer/counter on <br> or off. |
| 3 | IE1 | Interrupt 1 Edge Detect Flag: Set by hardware when an edge/level is detected on <br> INT1. This bit is cleared by hardware when the service routine is vectored to only if <br> the interrupt was edge triggered. Otherwise it follows the inverse of the pin. |
| 2 | IT1 | Interrupt 1 Type Control. Set/cleared by software to specify falling edge/ low level <br> triggered external inputs. |
| 1 | IE0 | Interrupt 0 Edge Detect Flag. Set by hardware when an edge/level is detected on <br> INT0 . This bit is cleared by hardware when the service routine is vectored to only if <br> the interrupt was edge triggered. Otherwise it follows the inverse of the pin. |
| 0 | IT0 | Interrupt 0 Type Control: Set/cleared by software to specify falling edge/ low level <br> triggered external inputs. |

TIMER MODE CONTROL

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | GATE | $\mathrm{C} / \overline{\mathrm{T}}$ | M1 | M0 | GATE | C/T | M1 | M0 |
|  | TIMER1 |  |  |  | TIMER0 |  |  |  |

Mnemonic: TMOD
Address: 89h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | GATE | Gating control: When this bit is set, Timer/counter 1 is enabled only while the INT1 <br> pin is high and the TR1 control bit is set. When cleared, the INT1 pin has no effect, <br> and Timer 1 is enabled whenever TR1 control bit is set. |
| 6 | C/T | Timer or Counter Select: When clear, Timer 1 is incremented by the internal clock. <br> When set, the timer counts falling edges on the T1 pin. |
| 5 | M1 | Timer 1 mode select bit 1. See table below. |
| 4 | M0 | Timer 1 mode select bit 0. See table below. |
| 3 | GATE | Gating control: When this bit is set, Timer/counter 0 is enabled only while the $\overline{\text { INT0 }}$ <br> pin is high and the TR0 control bit is set. When cleared, the INT0 pin has no effect, <br> and Timer 0 is enabled whenever TR0 control bit is set. |
| 2 | C/T | Timer or Counter Select: When clear, Timer 0 is incremented by the internal clock. <br> When set, the timer counts falling edges on the TO pin. |
| 1 | M1 | Timer 0 mode select bit 1. See table below. |
| 0 | M0 | Timer 0 mode select bit 0. See table below. |

M1, M0: Mode Select bits:

| M1 | M0 | MODE |
| :---: | :---: | :--- |
| 0 | 0 | Mode 0: 8-bit timer/counter TLx serves as 5-bit pre-scale. |
| 0 | 1 | Mode 1: 16-bit timer/counter, no pre-scale. |
| 1 | 0 | Mode 2: 8-bit timer/counter with auto-reload from THx. |
| 1 | 1 | Mode 3: (Timer 0) TL0 is an 8-bit timer/counter controlled by the standard Timer0 <br> control bits. TH0 is an 8-bit timer only controlled by Timer1 control bits. <br> (Timer 1) Timer/Counter 1 is stopped. |

TIMER 0 LSB


| BIT | NAME |  |
| :---: | :---: | :--- |
| $7-0$ | TL0.[7:0] | Timer 0 LSB. |

TIMER 1 LSB

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TL1.7 | TL1.6 | TL1.5 | TL1.4 | TL1.3 | TL1.2 | TL1.1 | TL1.0 |
| Mnemonic: TL1 |  |  |  |  |  |  |  | Address: 8Bh |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | TL1.[7:0] | Timer 1 LSB. |  |  |  |  |  |  |

## TIMER 0 MSB

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TH0.7 | TH0.6 | TH0.5 | TH0. 4 | TH0.3 | TH0. 2 | TH0.1 | TH0.0 |
| Mnemonic: TH0 |  |  |  |  |  |  |  | Address: 8Ch |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | TH0.[7:0] | Timer 0 MSB. |  |  |  |  |  |  |

## TIMER 1 MSB

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TH1.7 | TH1.6 | TH1.5 | TH1.4 | TH1.3 | TH1.2 | TH1.1 | TH1.0 |
| Mnemonic: TH1 |  |  |  |  |  |  |  | Address: 8Dh |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | TH1.[7:0] | Timer 1 MSB. |  |  |  |  |  |  |

## CLOCK CONTROL

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | T1M | TOM | - | - | - |

Mnemonic: CKCON
Address: 8Eh

| BIT | NAME | FUNCTION |
| :--- | :---: | :--- |
| $7-5$ | - | Reserved. |
| 4 | T1M | Timer 1 clock select: <br> 0: Timer 1 uses a divide by 12 clocks. <br> 1: Timer 1 uses a divide by 4 clocks. |
| 3 | T0M | Timer 0 clock select: <br> 0: Timer 0 uses a divide by 12 clocks. <br> 1: Timer 0 uses a divide by 4 clocks. |
| $2-0$ | - | Reserved. |

## PORT 1

|  | Bit: | 7 | 6 | 5 | 4 | 2 | 1 | 0 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | P1.7 | P 1.6 | P 1.5 | P 1.4 | P 1.3 | P 1.2 | P 1.1 | P 1.0 |

Mnemonic: P1
Address: 90h
P1.7-0: General purpose Input/Output port. Most instructions will read the port pins in case of a port read access, however in case of read-modify-write instructions, the port latch is read. These alternate functions are described below:

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 7 | P1.7 | PWM 2 Pin. |
| 6 | P1.6 | PWM 1 Pin. |
| 5 | P1.5 | /RST Pin or Input Pin by alternative. |
| 4 | P1.4 | /INT1 interrupt. |
| 3 | P1.3 | /INT0 interrupt or SDA of I2C. |
| 2 | P1.2 | Timer 0 or SCL of I2C. |
| 1 | P1.1 | RXD of Serial port. |
| 0 | P1.0 | TXD of Serial port. |

Note: The initial value of the port is set by CONFIG1.PRHI bit. The default setting for CONFIG1.PRHI =1 which the alternative function output is turned on upon reset. If CONFIG1.PRHI is set to 0 , the user has to write a 1 to port SFR to turn on the alternative function output.

DIVIDER CLOCK

| Bit: | 7 | 6 | 5 | 4 | 3 |  |  |  |  |  |  | 2 | 1 | 0 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  | DIVM. 7 | DIVM.6 | DIVM.5 | DIVM.4 | DIVM.3 | DIVM.2 | DIVM. 1 | DIVM.0 |  |  |  |  |  |  |


| BIT | NAME | FUNCTION |
| :---: | :---: | :---: |
| $7-0$ | DIVM.[7:0] | The DIVM register is clock divider of uC. Refer OSCILLATOR chapter. |

## SERIAL PORT CONTROL

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SM0/FE | SM1 | SM2 | REN | TB8 | RB8 | TI | RI |


| BIT | NAME | FUNCTION |
| :---: | :---: | :---: |
| 7 | SMO/FE | Serial port mode select bit 0 or Framing Error Flag: The SMODO bit in PCON SFR determines whether this bit acts as SMO or as FE. The operation of SMO is described below. When used as FE, this bit will be set to indicate an invalid stop bit. This bit must be manually cleared in software to clear the FE condition. |
| 6 | SM1 | Serial Port mode select bit 1. See table below. |
| 5 | SM2 | Multiple processors communication. Setting this bit to 1 enables the multiprocessor communication feature in mode 2 and 3 . In mode 2 or 3 , if SM2 is set to 1 , then RI will not be activated if the received 9th data bit (RB8) is 0 . In mode 1 , if $S M 2=1$, then RI will not be activated if a valid stop bit was not received. In mode 0, the SM2 bit controls the serial port clock. If set to 0 , then the serial port runs at a divide by 12 clock of the oscillator. This gives compatibility with the standard 8052. When set to 1 , the serial clock become divide by 4 of the oscillator clock. This results in faster synchronous serial communication. |
| 4 | REN | Receive enable: <br> 0 : Disable serial reception. <br> 1: Enable serial reception. |
| 3 | TB8 | This is the 9th bit to be transmitted in modes 2 and 3 . This bit is set and cleared by software as desired. |
| 2 | RB8 | In modes 2 and 3 this is the received 9th data bit. In mode 1, if SM2 $=0$, RB8 is the stop bit that was received. In mode 0 it has no function. |
| 1 | TI | Transmit interrupt flag: This flag is set by hardware at the end of the 8th bit time in mode 0 , or at the beginning of the stop bit in all other modes during serial transmission. This bit must be cleared by software. |
| 0 | RI | Receive interrupt flag: This flag is set by hardware at the end of the 8th bit time in mode 0, or halfway through the stop bits time in the other modes during serial reception. However the restrictions of SM2 apply to this bit. This bit can be cleared only by software. |

# W79E825A/824A/823B/822B Data Sheet 

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SM0, SM1: Mode Select bits

| MODE | SM0 | SM1 | DESCRIPTION | LENGTH | BAUD RATE |
| :---: | :---: | :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | Synchronous | 8 | Tclk divided by 4 or 12 |
| 1 | 0 | 1 | Asynchronous | 10 | Variable |
| 2 | 1 | 0 | Asynchronous | 11 | Tclk divided by 32 or 64 |
| 3 | 1 | 1 | Asynchronous | 11 | Variable |

## SERIAL DATA BUFFER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SBUF. 7 | SBUF. 6 | SBUF. 5 | SBUF. 4 | SBUF. 3 | SBUF. 2 | SBUF. 1 | SBUF. 0 |


| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 7-0 | SBUF.[7:0] | Serial data on the serial port is read from or written to this location. It actually <br> consists of two separate internal 8-bit registers. One is the receive resister, and <br> the other is the transmit buffer. Any read access gets data from the receive data <br> buffer, while write access is to the transmit data buffer. |

## PORT 2



Mnemonic: P2
Address: AOh

| BIT | NAME |  |
| :---: | :---: | :--- |
| $7-2$ | - | Reserved |
| 1 | P2.1 | XTAL1 clock input pin. |
| 0 | P2.0 | XTAL2 or CLKOUT pin by alternative. |

Note: The initial value of the port is set by CONFIG1.PRHI bit. The default setting for CONFIG1.PRHI $=1$ which the alternative function output is turned on upon reset. If CONFIG1.PRHI is set to 0 , the user has to write a 1 to port SFR to turn on the alternative function output.
KEYBOARD INTERRUPT

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | KBI. 7 | KBI. 6 | KBI. 5 | KBI. 4 | KBI. 3 | KBI. 2 | KBI. 1 | KBI. 0 |
| Mnemonic: KBI |  |  |  |  |  |  |  | Address: A1h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | KBI. 7 | 1: Enable P0.7 as a cause of a Keyboard interrupt. |  |  |  |  |  |  |
| 6 | KBI. 6 | 1: Enable P0.6 as a cause of a Keyboard interrupt. |  |  |  |  |  |  |
| 5 | KBI. 5 | 1: Enable P0.5 as a cause of a Keyboard interrupt. |  |  |  |  |  |  |
| 4 | KBI. 4 | 1: Enable P0.4 as a cause of a Keyboard interrupt. |  |  |  |  |  |  |

Continued.

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 3 | KBI.3 | 1: Enable P0.3 as a cause of a Keyboard interrupt. |
| 2 | KBI.2 | 1: Enable P0.2 as a cause of a Keyboard interrupt. |
| 1 | KBI.1 | 1: Enable P0.1 as a cause of a Keyboard interrupt. |
| 0 | KBI.0 | 1: Enable P0.0 as a cause of a Keyboard interrupt. |

## AUX FUNCTION REGISTER 1

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | KBF | BOD | BOI | LPBOV | SRST | ADCEN | 0 | DPS |
| Mnemonic: AUXR1 |  |  |  |  |  |  |  | dress: A2h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | KBF | Keyboard Interrupt Flag: <br> 1: When any pin of port 0 that is enabled for the Keyboard Interrupt function goes low. Must be cleared by software. |  |  |  |  |  |  |
| 6 | BOD | Brown Out Disable: <br> 0: Enable Brownout Detect function. <br> 1: Disable Brownout Detect function and save power. |  |  |  |  |  |  |
| 5 | BOI | Brown Out Interrupt: <br> 0: Disable Brownout Detect Interrupt function and it will cause chip reset when BOF is set. <br> 1: This prevents Brownout Detection from causing a chip reset and allows the Brownout Detect function to be used as an interrupt. |  |  |  |  |  |  |
| 4 | LPBOV | Low Power Brown Out Detect control: <br> 0 : When BOD is enable, the Brown Out detect is always turned on by normal run or Power Down mode. <br> 1: When BOD is enable, the Brown Out detect circuit is turned on by Power Down mode. This control can help save 15/16 of the Brownout circuit power. When uC is in Power Down mode, the BOD will enable internal RC OSC ( $2 \mathrm{MHz} \sim 0.5 \mathrm{MHZ}$ ) |  |  |  |  |  |  |
| 3 | SRST | Software reset: <br> 1: reset the chip as if a hardware reset occurred. |  |  |  |  |  |  |
| 2 | ADCEN | 0: Disable ADC circuit. <br> 1: Enable ADC circuit. |  |  |  |  |  |  |
| 1 | 0 | Reserved. |  |  |  |  |  |  |
| 0 | DPS | Dual Data Pointer Select <br> 0: To select DPTR of standard 8051. <br> 1: To select DPTR1 |  |  |  |  |  |  |

## INTERRUPT ENABLE



Mnemonic: IE
Address: A8h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | EA | Global enable. Enable/Disable all interrupts. |
| 6 | EADC | Enable ADC interrupt. |
| 5 | EBO | Enable Brown Out interrupt. |
| 4 | ES | Enable Serial Port interrupt. |
| 3 | ET1 | Enable Timer 1 interrupt. |
| 2 | EX1 | Enable external interrupt 1. |
| 1 | ET0 | Enable Timer 0 interrupt. |
| 0 | EX0 | Enable external interrupt 0. |

SLAVE ADDRESS

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SADDR. 7 | SADDR. 6 | SADDR. 5 | SADDR. 4 | SADDR. 3 | SADDR. 2 | SADDR. 1 | SADDR. 0 |

Mnemonic: SADDR
Address: A9h

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| $7-0$ | SADDR.[7:0] | The SADDR should be programmed to the given or broadcast address for <br> serial port to which the slave processor is designated. |

## COMPARATOR 1 CONTROL REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | CE1 | CP1 | CN1 | OE1 | CO1 | CMF1 |

Mnemonic: CMP1
Address: ACh

| BIT | NAME | FUNCTION |
| ---: | :--- | :--- |
| 7 | - | Reserved. |
| 6 | - | Reserved. |
| 5 | CE1 | Comparator enable: <br> 0: Disable Comparator. <br> 1: Enabled Comparator. Comparator output need wait stable 10 us after CE1 is <br> first set. |
| 4 | CP1 | Comparator positive input select: <br> 0: CIN1A is selected as the positive comparator input. <br> 1: CIN1B is selected as the positive comparator input. |

Continued.

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 3 | CN1 | Comparator negative input select: <br> 0: The comparator reference pin CMPREF is selected as the negative <br> comparator input. <br> 1: The internal comparator reference Vref is selected as the negative comparator <br> input. |
| 2 | OE1 | Output enable: <br> 1: The comparator output is connected to the CMP1 pin if the comparator is <br> enabled (CE1 = 1). This output is asynchronous to the CPU clock. |
| 1 | CO1 | Comparator output: <br> Synchronized to the CPU clock to allow reading by software. Cleared when the <br> comparator is disabled (CE1 = 0). |
| 0 | CMF1 | Comparator interrupt flag: <br> This bit is set by hardware whenever the comparator output CO1 changes state. <br> This bit will cause a hardware interrupt if enabled and of sufficient priority. <br> Cleared by software and when the comparator is disabled (CE1 = 0). |

## COMPARATOR 2 CONTROL REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | CE2 | CP2 | CN2 | OE2 | CO 2 | CMF2 |
| Mnemonic: CMP2 |  |  |  |  |  |  |  | dress: |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | - | Reserved. |  |  |  |  |  |  |
| 6 | - | Reserved. |  |  |  |  |  |  |
| 5 | CE2 | Comparator enable: <br> 0: Disable Comparator. <br> 1: Enabled Comparator. Comparator output need wait stable 10 us after CE2 is first set. |  |  |  |  |  |  |
| 4 | CP2 | Comparator positive input select: <br> 0: CIN2A is selected as the positive comparator input. <br> 1: CIN2B is selected as the positive comparator input. |  |  |  |  |  |  |
| 3 | CN2 | Comparator negative input select: <br> 0 : The comparator reference pin CMPREF is selected as the negative comparator input. <br> 1: The internal comparator reference Vref is selected as the negative comparator input. |  |  |  |  |  |  |

Continued .

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 2 | OE2 | Output enable: <br> 1: The comparator output is connected to the CMP2 pin if the comparator is <br> enabled (CE2 = 1). This output is asynchronous to the CPU clock. |
| 1 | CO2 | Comparator output: <br> Synchronized to the CPU clock to allow reading by software. Cleared when the <br> comparator is disabled (CE2 = 0). |
| 0 | CMF2 | Comparator interrupt flag: <br> This bit is set by hardware whenever the comparator output CO2 changes state. <br> This bit will cause a hardware interrupt if enabled and of sufficient priority. <br> Cleared by software and when the comparator is disabled (CE2 = 0). |

## PORT 0 OUTPUT MODE 1

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P0M1.7 | P0M1.6 | P0M1.5 | P0M1.4 | P0M1.3 | P0M1.2 | P0M1.1 | P0M1.0 |
| Mnemonic: P0M1 |  |  |  |  |  |  |  | Address: B1h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | P0M1.[7:0] | To control the output configuration of P0 bits [7:0] |  |  |  |  |  |  |

## PORT 0 OUTPUT MODE 2

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | POM2.7 | P0M2.6 | P0M2.5 | P0M2.4 | P0M2.3 | P0M2.2 | P0M2.1 | P0M2.0 |
| Mnemonic: POM2 |  |  |  |  |  |  |  | Address: B2h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | POM2.[7:0] | To control the output configuration of P0 bits [7:0] |  |  |  |  |  |  |

## PORT 1 OUTPUT MODE 1



## PORT 1 OUTPUT MODE 2

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P1M2.7 | P1M2.6 | - | P1M2.4 | P1M2.3 | P1M2.2 | P1M2.1 | P1M2.0 |

Mnemonic: P1M2
Address: B4h

| BIT | NAME | FUNCTION |
| :---: | :---: | :---: |
| $7-0$ | P1M2.[7:0] | To control the output configuration of P1 bits [7:0] |

PORT 2 OUTPUT MODE 1

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P2S | P1S | POS | ENCLK | T1OE | T0OE | P2M1.1 | P2M1.0 |
| Mnemonic: P2M1 |  |  |  |  |  |  |  | Address: B5h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | P2S | 0: Disable Schmitt trigger inputs on port 2 and enable TTL inputs on port 2. <br> 1: Enables Schmitt trigger inputs on Port 2. |  |  |  |  |  |  |
| 6 | P1S | 0: Disable Schmitt trigger inputs on port 1 and enable TTL inputs on port 1. 1: Enables Schmitt trigger inputs on Port 1. |  |  |  |  |  |  |
| 5 | POS | 0: Disable Schmitt trigger inputs on port 0 and enable TTL inputs on port 0 <br> 1: Enables Schmitt trigger inputs on Port 0. |  |  |  |  |  |  |
| 4 | ENCLK | 1: Enabled clock output to XTAL2 pin (P2.0) |  |  |  |  |  |  |
| 3 | T10E | 1: The P0.7 pin is toggled whenever Timer 1 overflows. The output frequency is therefore one half of the Timer 1 overflow rate. |  |  |  |  |  |  |
| 2 | T0OE | 1: The P1.2 pin is toggled whenever Timer 0 overflows. The output frequency is therefore one half of the Timer 0 overflow rate. |  |  |  |  |  |  |
| 1 | P2M1.1 | To control the output configuration of P2.1. |  |  |  |  |  |  |
| 0 | P2M1.0 | To control the output configuration of P2.0. |  |  |  |  |  |  |

## PORT 2 OUTPUT MODE 2

| Bit: | 7 | 6 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - - | - | - | - | P2M2.1 | P2M2.0 |
| Mnemonic: P2M2 |  |  |  |  |  | Address: B6h |  |
| BIT | NAME | FUNCTION |  |  |  |  |  |
| 7-2 | - | Reserved. |  |  |  |  |  |
| 1-0 | P2M2.[1:0] | To control the output configuration of P2 bits [1:0] |  |  |  |  |  |

## W79E825A/824A/823B/822B Data Sheet

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Port Output Configuration Settings:

| PXM1.Y | PXM2.Y | PORT INPUT/OUTPUT MODE |
| :---: | :---: | :--- |
| 0 | 0 | Quasi-bidirectional |
| 0 | 1 | Push-Pull |
| 1 | 0 | Input Only (High Impedance) <br> P2M1.PxS=0, TTL input <br> P2M1.PxS=1, Schmitt input |
| 1 | 1 | Open Drain |

## INTERRUPT HIGH PRIORITY

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | PADCH | PBOH | PSH | PT1H | PX1H | PTOH | PXOH |

Mnemonic: IPOH
Address: B7h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | - | This bit is un-implemented and will read high. |
| 6 | PADCH | 1: To set interrupt high priority of ADC is highest priority level. |
| 5 | PBOH | 1: To set interrupt high priority of Brown Out Detector is highest priority level. |
| 4 | PSH | 1: To set interrupt high priority of Serial port 0 is highest priority level. |
| 3 | PT1H | 1: To set interrupt high priority of Timer 1 is highest priority level. |
| 2 | PX1H | 1: To set interrupt high priority of External interrupt 1 is highest priority level. |
| 1 | PTOH | 1: To set interrupt high priority of Timer 0 is highest priority level. |
| 0 | PXOH | 1: To set interrupt high priority of External interrupt 0 is highest priority level. |

## INTERRUPT PRIORITY 0



SLAVE ADDRESS MASK ENABLE

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SADEN. 7 | SADEN. 6 | SADEN. 5 | SADEN. 4 | SADEN. 3 | SADEN. 2 | SADEN. 1 | SADEN. 0 |

Mnemonic: SADEN
Address: B9h

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| $7-0$ | SADEN [7:0] | This register enables the Automatic Address Recognition feature of the Serial <br> port 0. When a bit in the SADEN is set to 1, the same bit location in SADDR <br> will be compared with the incoming serial data. When SADEN is 0, then the <br> bit becomes a "don't care" in the comparison. This register enables the <br> Automatic Address Recognition feature of the Serial port 0. When all the bits <br> of SADEN are 0, interrupt will occur for any incoming address. |

## SLAVE ADDRESS MASK ENABLE 1



Mnemonic: I2DAT
Address: BCh

| BIT | NAME |  |
| :---: | :---: | :--- |
| $7-0$ | I2DAT.[7:0] | The data register of I2C. |

## I2C STATUS REGISTER

| Bit: | 76 | $\begin{array}{lll}5 & 4\end{array}$ | 2 |  | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | I2STATUS. 7 I2S |  |  |  |  |
| Mnemonic: I2STATUS Address: BDh |  | Address: BDh |  |  |  |
| BIT | NAME | FUNCTION |  |  |  |
| 7-0 | I2STATUS.[7:0] | The status register of I2C: <br> The three least significant bits are always 0 . The five most significant bits contain the status code. There are 23 possible status codes. When I2STATUS contains F 8 H , no serial interrupt is requested. All other I2STATUS values correspond to defined I2C states. When each of these states is entered, a status interrupt is requested $(S I=1)$. A valid status code is present in I2STATUS one machine cycle after SI is set by hardware and is still present one machine cycle after SI has been reset by software. In addition, states 00 H stands for a Bus Error. A Bus Error occurs when a START or STOP condition is present at an illegal position in the formation frame. Example of illegal position are during the serial transfer of an address byte, a data byte or an acknowledge bit. |  |  |  |

## I2C BAUD RATE CONTROL REGISTER



Mnemonic: I2CLK
Address: BEh

| BIT | NAME |  | FUNCTION |
| :---: | :---: | :--- | :--- |
| $7-0$ | I2CLK.[7:0] | The I2C clock rate bits. |  |

## I2C TIMER COUNTER REGISTER



Mnemonic: I2TIMER
Address: BFh

| BIT | NAME | FUNCTION |
| :--- | :--- | :--- |
| $7 \sim 3$ | - | Reserved. |
| 2 | ENTI | Enable I2C 14-bits Timer Counter: <br> 0: Disable 14-bits Timer Counter count. <br> 1: Enable 14-bits Timer Counter count. After enable ENTI and ENSI, the 14-bit <br> counter will be counted. When SI flag of I2C is set, the counter will stop to <br> count and 14-bits Timer Counter will be cleared. |
| 1 | DIV4 | I2C Timer Counter clock source divide function: <br> 0: The 14-bits Timer Counter source clock is Fosc clock. <br> 1: The 14-bits Timer Counter source clock is divided by 4. |
| 0 | TIF | The I2C Timer Counter count flag: <br> 0: The 14-bits Timer Counter is not overflow. <br> 1: The 14-bits Timer Counter is overflow. Before enable I2C Timer (both ENTI, <br> ENSI = [1,1]) the SI must be cleared. If I2C interrupt is enabled. The I2C <br> interrupt service routine will be executed. This bit is cleared by software. |

## I2C CONTROL REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | ENSI | STA | STO | SI | AA |  | - |
| Mnemonic: 12 CON |  |  |  |  |  |  | Address: COh |  |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7 | - | Reserved. |  |  |  |  |  |  |
| 6 | ENS1 | 0: Disable I2C Serial Function. The SDA and SCL output are in a high impedance state. SDA and SCL input signals are ignored, I2C is not in the addressed slave mode or it is not addressable, and STO bit in I2CON is forced to " 0 ". No other bits are affected. P1.2 (SCL) and P1.3 (SDA) may be used as open drain I/O ports. <br> 1: Enable I2C Serial Function. The P1.2 and P1.3 port latches must be to logic 1. |  |  |  |  |  |  |

Continued

| BIT | NAME | FUNCTION |
| :--- | :--- | :--- |
| 5 | $\begin{array}{l}\text { The START flag. } \\ \text { O: The STA bit is reset, no START condition or repeated START condition will be } \\ \text { generated. }\end{array}$ |  |
| 1: The STA bit is set to enter a master mode. The I2C hardware checks the status of |  |  |
| I2C bus and generates a START condition if the bus is free. If bus is not free, then |  |  |
| I2C waits for a STOP condition and generates a START condition after a delay. If |  |  |
| STA is set while I2C is already in a master mode and one or more bytes are |  |  |
| transmitted or received, I2C transmits a repeated START condition. STA may be |  |  |
| set any time. STA may also be set when I2C interface is an addressed slave |  |  |
| mode. |  |  |$\}$

## I2C ADDRESS REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 |  |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | I2ADDR. 7 | I2ADDR. 6 | I2ADDR. 5 | I2ADDR.4 | I2ADDR. 3 | I2ADDR. 2 | I2ADDR. 1 | GC |

Mnemonic: I2ADDR
Address: C1h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| $7 \sim 1$ | I2ADDR.[7:1] | I2C Address register: <br> The 8051 uC can read from and write to this 8-bit, directly addressable <br> SFR. The content of this register is irrelevant when I2C is in master mode. <br> In the slave mode, the seven most significant bits must be loaded with the <br> MCU's own address. The I2C hardware will react if either of the address is <br> matched. |
| 0 | GC | General Call Function. <br> 0: Disable General Call Function. <br> 1: Enable General Call Function. |

## NVM ADDRESS



## TIMED ACCESS

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | TA. 7 | TA. 6 | TA. 5 | TA. 4 | TA. 3 | TA. 2 | TA. 1 | TA. 0 |
| Mnemonic: TA |  |  |  |  |  |  |  | Address: C7h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | TA.[7:0] | The Timed Access register: <br> The Timed Access register controls the access to protected bits. To access protected bits, the user must first write AAH to the TA. This must be immediately followed by a write of 55 H to TA. Now a window is opened in the protected bits for three machine cycles, during which the user can write to these bits. |  |  |  |  |  |  |

## NVM CONTROL



| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | EER | NVM page(n) erase bit: <br> 0: Without erase NVM page(n). <br> 1: Set this bit to erase page(n) of NVM. The NVM has 4 pages and each page <br> have 64 bytes data memory. Before select page by NVMADDR register that <br> will automatic enable page area, after set this bit, the page will be erased and <br> program counter will halt at this instruction. After finished, program counter will <br> kept next instruction then executed. The NVM page's address is defined in <br> table below. |
| 6 | EWR | NVM data write bit: <br> 0: Without write NVM data. <br> 1: Set this bit to write NVM bytes and program counter will halt at this instruction. <br> After write is finished, program counter will kept next instruction then executed. |
| $5-0$ | - | Reserved |

NVM Page(n) Area Definition Table:

| PAGE | START ADDRESS | END ADDRESS |
| :---: | :---: | :---: |
| 0 | 00 H | $3 F H$ |
| 1 | 40 H | 7 FH |
| 2 | $\mathbf{8 0 H}$ | BFH |
| 3 | $\mathbf{C O H}$ | FFH |

Note: The W79E823 and W79E822 without page 2 and page 3.

## NVM DATA

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | NVMDAT.7 | NVMDAT.6 | NVMDAT.5 | NVMDAT.4 | NVMDAT3 | NVMDAT.2 | NVMDAT.1 | NVMDAT.0 |

Mnemonic: NVMDATA
Address: CFh

| BIT | NAME | FUNCTION |
| :--- | :---: | :---: |
| $7 \sim 0$ | NVMDAT.[7:0] | The NVM data write register. The read NVM data is by MOVC instruction. |

PROGRAM STATUS WORD


Continued

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 6 | AC | Auxiliary carry: <br> Set when the previous operation resulted in a carry from the high order nibble. |
| 5 | F0 | User flag 0: <br> The General purpose flag that can be set or cleared by the user. |
| $4 \sim 3$ | RS1~RS0 | Register bank select bits. |
| 2 | OV | Overflow flag: <br> Set when a carry was generated from the seventh bit but not from the 8th bit as <br> a result of the previous operation, or vice-versa. |
| 1 | F1 | User Flag 1: <br> The General purpose flag that can be set or cleared by the user software. |
| 0 | P | Parity flag: <br> Set/cleared by hardware to indicate odd/even number of 1's in the accumulator. |

RS.1-0: Register Bank Selection Bits:

| RS1 | RS0 | REGISTER BANK | ADDRESS |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | $00-07 \mathrm{~h}$ |
| 0 | 1 | 1 | $08-0 \mathrm{Fh}$ |
| 1 | 0 | 2 | $10-17 \mathrm{~h}$ |
| 1 | 1 | 3 | $18-1 \mathrm{Fh}$ |

PWMP COUNTER HIGH BITS REGISTER

| Bit: | 76 |  | 4 | 3 | 2 | 10 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - - | - | - | - | - | PWMP. 9 | PWMP. 8 |
| Mnemonic: PWMPH |  |  |  |  |  | Address: D1h |  |
| BIT | NAME | FUNCTION |  |  |  |  |  |
| 7-2 | - | Reserved. |  |  |  |  |  |
| 1-0 | PWMP.[9:8] | The PWM Counter Register bits 9~8. |  |  |  |  |  |

## PWM 0 HIGH BITS REGISTER

| Bit: | $7 \quad 6$ |  | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | PWM0.9 | PWM0.8 |
| Mnemonic: PWMOH |  |  |  |  |  | Address: D2h |  |
| BIT | NAME | FUNCTION |  |  |  |  |  |
| 7~2 | - | Reserved. |  |  |  |  |  |
| 1~0 | PWMO.[9:8] | The PWM 0 High Bits Register bit 9~8. |  |  |  |  |  |

## PWM 1 HIGH BITS REGISTER



## PWM 2 HIGH BITS REGISTER



## PWM 3 HIGH BITS REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | - | - | - | - | PWM3.9 | PWM3.8 |


| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| $7 \sim 2$ | - | Reserved. |
| $1 \sim 0$ | PWM3.[9:8] | The PWM 3 High Bits Register bit 9~8. |

## PWM CONTROL REGISTER 3



Mnemonic: PWMCON3
Address: D7h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| $7-6$ | - | Reserved. |
| 0 | BKF | The external brake pin Flag. <br> 0: The PWM is not brake. <br> 1: The PWM is brake by external brake pin. It will be cleared by software. |

## WATCHDOG CONTROL




The WDCON SFR is set to 0x000000B on a reset. WTRF (WDCON.2) is set to a 1 on a Watchdog timer reset, but to a 0 on power on/down resets. WTRF (WDCON.2) is not altered by an external reset. EWRST (WDCON.1) is set to 0 on a Power-on reset, reset pin reset, and Watch Dog Timer reset.
All the bits in this SFR have unrestricted read access. WDRUN, WDO, WD1, EWRST, WDIF and WDCLR require Timed Access procedure to write. The remaining bits have unrestricted write accesses. Please refer TA register description.

| TA | REG | C7H |
| :--- | :--- | :--- |
| WDCON | REG |  |
| MOV | TA, \#AAH |  |
| MOV | TA, \#55H | ; To access protected bits |
| SETB | WDCON.0 |  |
| ORL | WDCON, \#00110000B | ; Seset watchdog timer 26 bits watchdog timer |
| MOV | TA, \#AAH |  |
| MOV | TA, \#55H |  |
| ORL | WDCON, \#00000010B | ; Enable watchdog |

## PWMP COUNTER LOW BITS REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PWMP. 7 | PWMP. 6 | PWMP. 5 | PWP. 4 | PWMP. 3 | PWMP. 2 | PWMP. 1 | PWMP. 1 |

Mnemonic: PWMPL
Address: D9h

| BIT | NAME | FUNCTION |
| :--- | :---: | :---: |
| $7 \sim 0$ | PWMP.[7:0] | PWM Counter Low Bits Register. |

## PWMO LOW BITS REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | 7 | PWM0.7 | PWM0.6 | PWM0.5 | PWM0.4 | PWM0.3 | PWM0.2 | PWM0.1 |
|  | PWM0.1 |  |  |  |  |  |  |  |

Mnemonic: PWMOL
Address: DAh

| BIT | NAME |  | FUNCTION |
| :--- | :---: | :--- | :--- |
| $7 \sim 0$ | PWM0.[7:0] | PWM 0 Low Bits Register. |  |

## PWM1 LOW BITS REGISTER

| Bit: | 76 |  | 5 4 |  | 32 |  | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PWM1.7 | PWM1.6 | PWM1.5 | PWM1.4 | PWM1.3 | PWM1.2 | PWM1.1 | PWM1.0 |
| Mnemonic: PWM1L |  |  |  |  |  |  |  | Address: DBh |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7~0 | PWM1.[7:0] | PWM 1 Low Bits Register. |  |  |  |  |  |  |

PWM CONTROL REGISTER 1

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PWMRUN | Load | CF | CLRPWM | PWM3I | PWM2I | PWM1I | PWMOI |
| Mnemonic: PWMCON1 Address: D |  |  |  |  |  |  |  |  |


| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | PWMRUN | 0: The PWM is not running. <br> 1: The PWM counter is running. |
| 6 | Load | 0: The registers value of PWMP and PWMn are never loaded to counter and <br> Comparator registers. <br> 1: The PWMP and PWMn registers load value to counter and compare registers <br> at the counter underflow. This bit is auto cleared by hardware at next clock <br> cycle. |
| 5 | CF | 0: The 10-bit counter down count is not underflow. <br> 1: The 10-bit counter down count is underflow. This bit is Software clear. |
| 4 | CLRPWM | 1: Clear 10-bit PWM counter to 000H. This bit is auto cleared by hardware. |$|$| 0: PWM3 out is non-inverted. |
| :--- |
| 1: PWM3 output is inverted. |

PWM2 LOW BITS REGISTER

| Bit: | 76 |  | 5 4 |  | 32 |  | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PWM2.7 | PWM2.6 | PWM2.5 | PWM2.4 | PWM2.3 | PWM2.2 | PWM2.1 | PWM2.0 |
| Mnemonic: PWM2L |  |  |  |  |  |  |  | Address: DDh |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7~0 | PWM2.[7:0] | PWM 2 Low Bits Register. |  |  |  |  |  |  |

## PWM3 LOW BITS REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | PWM3.7 | PWM3.6 | PWM3.5 | PWM3.4 | PWM3.3 | PWM3.2 | PWM3.1 | PWM3.0 |

Mnemonic: PWM3L
Address: DEh

| BIT | NAME |  |
| :--- | :---: | :--- |
| $7 \sim 0$ | PWM3.[7:0] | PWM 3 Low Bits Register. |

## PWM CONTROL REGISTER 2



| BIT | NAME |  |
| :---: | :--- | :--- |
| 7 | BKCH | See the below table, when BKEN is set. |
| 6 | BKPS | 0: Brake is asserted if P0.2 is low. <br> 1: Brake is asserted if P0.2 is high |
| 5 | BPEN | See the below table, when BKEN is set. |
| 4 | BKEN | 0: The Brake is never asserted. <br> 1: The Brake is enabled, and see the below table. |
| 3 | PWM3B | 0: The PWM3 output is low, when Brake is asserted. <br> 1: The PWM3 output is high, when Brake is asserted. |
| 2 | PWM2B | 0: The PWM2 output is low, when Brake is asserted. <br> 1: The PWM2 output is high, when Brake is asserted. |
| 1 | PWM1B | 0: The PWM1 output is low, when Brake is asserted. <br> 1: The PWM1 output is high, when Brake is asserted. |
| 0 | PWM0B | 0: The PWM0 output is low, when Brake is asserted. <br> 1: The PWM0 output is high, when Brake is asserted. |

## Brake Condition Table:

| BPEN | BKCH | BRAKE CONDITION |
| :---: | :---: | :--- |
| 0 | 0 | Brake On (software brake and keeping brake). <br> Software brake condition. When active (BPEN=BKCH=0, and BKEN=1), PWM <br> output follows PWMnB setting. This brake has no effect on PWMRUN bit, <br> therefore, internal PWM generator continues to run. When the brake is released, <br> the state of PWM output depends on the current state of PWM generator output <br> during the release. |
| 0 | 1 | Brake On, when PWM is not running (PWMRUN=0), the PWM output condition <br> follows PWMnB setting. <br> When the brake is released (by disabling BKEN = 0), the PWM output resumes <br> to the state when PWM generator stop running prior to enabling the brake. <br> Brake Off, when PWM is running (PWMRUN=1). |
| 1 | 0 | Brake On, when Brake Pin asserted. <br> External pin brake condition. When active (by external pin), PWM output follows <br> PWMnB setting, PWMRUN will be cleared by hardware, and BKF flag will be set. <br> When the brake is released (by de-asserting the external pin and disabling <br> BKEN = 0), the PWM output resumes to the state of the PWM generator output <br> prior to the brake. |
| 1 | 1 | No any active. |

## ACCUMULATOR

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ACC. 7 | ACC. 6 | ACC. 5 | ACC. 4 | ACC. 3 | ACC. 2 | ACC. 1 | ACC. 0 |


| BIT | NAME | FUNCTION |
| :---: | :---: | :---: |
| $7-0$ | ACC.[7:0] | The A or ACC register is the standard 8052 accumulator |

## ADC CONTROL REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ADC. 1 | ADC. 0 | ADCEX | ADCI | ADCS | RCCLK | AADR1 | AADR0 |

Mnemonic: ADCCON
Address: E1h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | ADC.1 | The ADC conversion result. |
| 6 | ADC. 0 | The ADC conversion result. |
| 5 | ADCEX | $\begin{array}{l}\text { Enable STADC-triggered conversion } \\ \text { 0: Conversion can only be started by software (i.e., by setting ADCS). } \\ \text { 1: Conversion can be started by software or by a rising edge on STADC (pin } \\ \text { P1.4). }\end{array}$ |
| 4 | ADCI | $\begin{array}{l}\text { ADC Interrupt flag: } \\ \text { This flag is set when the result of an A/D conversion is ready. This generates an } \\ \text { ADC interrupt, if it is enabled. The flag may be cleared by the ISR. While this flag } \\ \text { is 1, the ADC cannot start a new conversion. ADCI can not be set by software. }\end{array}$ |
| 3 | $\begin{array}{l}\text { ADC Start and Status: Set this bit to start an A/D conversion. It may also be set } \\ \text { by STADC if ADCEX is 1. This signal remains high while the ADC is busy and is } \\ \text { reset right after ADCI is set. } \\ \text { Notes: } \\ \text { 1. It is recommended to clear ADCI before ADCS is set. However, if ADCI is } \\ \text { cleared and ADCS is set at the same time, a new A/D conversion may start }\end{array}$ |  |
| 2 | RCCLK | $\begin{array}{l}\text { on the same channel. }\end{array}$ |
| 1 | 2. Software clearing of ADCS will abort conversion in progress. |  |
| 3. ADC cannot start a new conversion while ADCS or ADCI is high. |  |  |$]$

The ADCI and ADCS control the ADC conversion as below:

| ADCI | ADCS | ADC STATUS |
| :---: | :---: | :--- |
| 0 | 0 | ADC not busy; A conversion can be started. |
| 0 | 1 | ADC busy; Start of a new conversion is blocked. |
| 1 | 0 | Conversion completed; Start of a new conversion requires ADCI $=0$. |
| 1 | 1 | This is an internal temporary state that user can ignore it. |

AADR1, AADR0: ADC Analog Input Channel select bits:
These bits can only be changed when ADCI and ADCS are both zero.

| AADR1 | AADR0 | SELECTED ANALOG INPUT CHANNEL |
| :---: | :---: | :---: |
| 0 | 0 | AD0 (P0.3) |
| 0 | 1 | AD1 (P0.4) |
| 1 | 0 | AD2 (P0.5) |
| 1 | 1 | AD3 (P0.6) |

## ADC CONVERTER RESULT HIGH REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ADC. 9 | ADC. 8 | ADC. 7 | ADC. 6 | ADC. 5 | ADC. 4 | ADC. 3 | ADC. 2 |
| Mnemonic: ADCH |  |  |  |  |  |  |  | Address: E2h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | ADC.[9:2] | The ADC conversion result. |  |  |  |  |  |  |

## INTERRUPT ENABLE REGISTER 1

| Bit: | 7 | 6 | 5 | 4 |  |  |  |  |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | 7 | - | EPWM | EWDI | EC2 | EC1 | EKB | EI2 |
|  | - | - |  |  |  |  |  |  |

Mnemonic: EIE
Address: E8h

| BIT | NAME |  |
| :---: | :--- | :--- |
| 7 | - | Reserved. |
| 6 | - | Reserved. |
| 5 | EPWM | 0: Disable PWM Interrupt when external brake pin was brake. <br> 1: Enable PWM Interrupt when external brake pin was brake. |
| 4 | EWDI | 0: Disable Watchdog Timer Interrupt. <br> 1: Enable Watchdog Timer Interrupt. |
| 3 | EC2 | 0: Disable Comparator 2 Interrupt. <br> 1: Enable Comparator 2 Interrupt. |
| 2 | EC1 | 0: Disable Comparator 1 Interrupt. <br> 1: Enable Comparator 1 Interrupt. |
| 1 | EKB | 0: Disable Keypad Interrupt. <br> 1: Enable Keypad Interrupt. |
| 0 | El2 | 0: Disable I2C Interrupt. <br> 1: Enable I2C Interrupt. |

## B REGISTER

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | B. 7 | B. 6 | B. 5 | B. 4 | B. 3 | B. 2 | B. 1 | B. 0 |
| Mnemonic: B |  |  |  |  |  |  |  | dress: FOh |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7-0 | B. [7:0] | The B register is the standard 8052 register that serves as a second accumulator. |  |  |  |  |  |  |

## PORT 0 DIGITAL INPUT DISABLE

| Bit: | 76 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | P0ID. 7 | POID. 6 | POID. 5 | POID. 4 | POID. 3 | POID. 2 | POID. 1 | POID. 0 |
| Mnemonic: POID |  |  |  |  |  |  |  | dress: F6h |
| BIT | NAME | FUNCTION |  |  |  |  |  |  |
| 7~0 | POID.[7:0] | Enable/Disable Port 0 digital inputs. <br> 0: Enable Port 0 digital inputs. <br> 1: Disable Port 0 digital inputs. |  |  |  |  |  |  |

## INTERRUPT HIGH PRIORITY 1

| Bit: | 7 | 6 | 5 | 4 | 3 |  | 4 | 2 |
| :---: | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | - | - | PPWMH | PWDIH | PC2H | PC1H | PKBH | PI2H |

Mnemonic: IP1H
Address: F7h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | - | Reserved. |
| 6 | - | Reserved. |
| 5 | PPWMH | 1: To set interrupt high priority of PWM's brake is highest priority level. |
| 4 | PWDIH | 1: To set interrupt high priority of Watchdog is highest priority level. |
| 3 | PC2H | 1: To set interrupt high priority of Comparator 2 is highest priority level. |
| 2 | PC1H | 1: To set interrupt high priority of Comparator 1 is highest priority level. |
| 1 | PKBH | 1: To set interrupt high priority of Keypad is highest priority level. |
| 0 | PI2H | 1: To set interrupt high priority of I2C is highest priority level. |

## EXTENDED INTERRUPT PRIORITY

| Bit: | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | - | - | PPWM | PWDI | PC2 | PC1 | PKB | PI2 |

Mnemonic: IP1
Address: F8h

| BIT | NAME | FUNCTION |
| :---: | :--- | :--- |
| 7 | - | Reserved. |
| 6 | - | Reserved. |
| 5 | PPWM | 1: To set interrupt priority of PWM's external brake is higher priority level. |
| 4 | PWDI | 1: To set interrupt priority of Watchdog is higher priority level. |
| 3 | PC2 | 1: To set interrupt priority of Comparator 2 is higher priority level. |
| 2 | PC1 | 1: To set interrupt priority of Comparator 1 is higher priority level. |
| 1 | PKB | 1: To set interrupt priority of Keypad is higher priority level. |
| 0 | PI2 | 1: To set interrupt priority of I2C is higher priority level. |

## 9 INSTRUCTION SET

The W79E825 series execute all the instructions of the standard 8052 family. The operations of these instructions, as well as their effects on flag and status bits, are exactly the same. However, the timing of these instructions is different in two ways. Firstly, the machine cycle is four clock periods, while the standard-8051/52 machine cycle is twelve clock periods. Secondly, it can fetch only once per machine cycle (i.e., four clocks per fetch), while the standard 8051/52 can fetch twice per machine cycle (i.e., six clocks per fetch).
The timing differences create an advantage for the W79E825 series. There is only one fetch per machine cycle, so the number of machine cycles is usually equal to the number of operands in the instruction. (Jumps and calls do require an additional cycle to calculate the new address.) As a result, the W79E825 series reduces the number of dummy fetches and wasted cycles, and therefore improves overall efficiency, compared to the standard 8051/52.

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 SERIES CLOCK CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| NOP | 00 | 1 | 1 | 4 | 12 | 3 |
| ADD A, R0 | 28 | 1 | 1 | 4 | 12 | 3 |
| ADD A, R1 | 29 | 1 | 1 | 4 | 12 | 3 |
| ADD A, R2 | 2A | 1 | 1 | 4 | 12 | 3 |
| ADD A, R3 | 2B | 1 | 1 | 4 | 12 | 3 |
| ADD A, R4 | 2C | 1 | 1 | 4 | 12 | 3 |
| ADD A, R5 | 2D | 1 | 1 | 4 | 12 | 3 |
| ADD A, R6 | 2E | 1 | 1 | 4 | 12 | 3 |
| ADD A, R7 | 2F | 1 | 1 | 4 | 12 | 3 |
| ADD A, @R0 | 26 | 1 | 1 | 4 | 12 | 3 |
| ADD A, @R1 | 27 | 1 | 1 | 4 | 12 | 3 |
| ADD A, direct | 25 | 2 | 2 | 8 | 12 | 1.5 |
| ADD A, \#data | 24 | 2 | 2 | 8 | 12 | 1.5 |
| ADDC A, R0 | 38 | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R1 | 39 | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R2 | 3A | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R3 | 3B | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R4 | 3C | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R5 | 3D | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R6 | 3E | 1 | 1 | 4 | 12 | 3 |
| ADDC A, R7 | 3F | 1 | 1 | 4 | 12 | 3 |
| ADDC A, @R0 | 36 | 1 | 1 | 4 | 12 | 3 |
| ADDC A, @R1 | 37 | 1 | 1 | 4 | 12 | 3 |

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 <br> SERIES <br> CLOCK <br> CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADDC A, direct | 35 | 2 | 2 | 8 | 12 | 1.5 |
| ADDC A, \#data | 34 | 2 | 2 | 8 | 12 | 1.5 |
| SUBB A, R0 | 98 | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R1 | 99 | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R2 | 9A | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R3 | 9B | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R4 | 9C | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R5 | 9D | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R6 | 9E | 1 | 1 | 4 | 12 | 3 |
| SUBB A, R7 | 9F | 1 | 1 | 4 | 12 | 3 |
| SUBB A, @R0 | 96 | 1 | 1 | 4 | 12 | 3 |
| SUBB A, @R1 | 97 | 1 | 1 | 4 | 12 | 3 |
| SUBB A, direct | 95 | 2 | 2 | 8 | 12 | 1.5 |
| SUBB A, \#data | 94 | 2 | 2 | 8 | 12 | 1.5 |
| INC A | 04 | 1 | 1 | 4 | 12 | 3 |
| INC RO | 08 | 1 | 1 | 4 | 12 | 3 |
| INC R1 | 09 | 1 | 1 | 4 | 12 | 3 |
| INC R2 | OA | 1 | 1 | 4 | 12 | 3 |
| INC R3 | OB | 1 | 1 | 4 | 12 | 3 |
| INC R4 | OC | 1 | 1 | 4 | 12 | 3 |
| INC R5 | OD | 1 | 1 | 4 | 12 | 3 |
| INC R6 | OE | 1 | 1 | 4 | 12 | 3 |
| INC R7 | OF | 1 | 1 | 4 | 12 | 3 |
| INC @R0 | 06 | 1 | 1 | 4 | 12 | 3 |
| INC @R1 | 07 | 1 | 1 | 4 | 12 | 3 |
| INC direct | 05 | 2 | 2 | 8 | 12 | 1.5 |
| INC DPTR | A3 | 1 | 2 | 8 | 24 | 3 |
| DEC A | 14 | 1 | 1 | 4 | 12 | 3 |
| DEC R0 | 18 | 1 | 1 | 4 | 12 | 3 |
| DEC R1 | 19 | 1 | 1 | 4 | 12 | 3 |
| DEC R2 | 1A | 1 | 1 | 4 | 12 | 3 |
| DEC R3 | 1B | 1 | 1 | 4 | 12 | 3 |

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กUVOTO!

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 SERIES CLOCK CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DEC R4 | 1C | 1 | 1 | 4 | 12 | 3 |
| DEC R5 | 1D | 1 | 1 | 4 | 12 | 3 |
| DEC R6 | 1E | 1 | 1 | 4 | 12 | 3 |
| DEC R7 | 1F | 1 | 1 | 4 | 12 | 3 |
| DEC @R0 | 16 | 1 | 1 | 4 | 12 | 3 |
| DEC @R1 | 17 | 1 | 1 | 4 | 12 | 3 |
| DEC direct | 15 | 2 | 2 | 8 | 12 | 1.5 |
| MUL AB | A4 | 1 | 5 | 20 | 48 | 2.4 |
| DIV AB | 84 | 1 | 5 | 20 | 48 | 2.4 |
| DA A | D4 | 1 | 1 | 4 | 12 | 3 |
| ANL A, R0 | 58 | 1 | 1 | 4 | 12 | 3 |
| ANL A, R1 | 59 | 1 | 1 | 4 | 12 | 3 |
| ANL A, R2 | 5A | 1 | 1 | 4 | 12 | 3 |
| ANL A, R3 | 5B | 1 | 1 | 4 | 12 | 3 |
| ANL A, R4 | 5C | 1 | 1 | 4 | 12 | 3 |
| ANL A, R5 | 5D | 1 | 1 | 4 | 12 | 3 |
| ANL A, R6 | 5E | 1 | 1 | 4 | 12 | 3 |
| ANL A, R7 | 5F | 1 | 1 | 4 | 12 | 3 |
| ANL A, @R0 | 56 | 1 | 1 | 4 | 12 | 3 |
| ANL A, @R1 | 57 | 1 | 1 | 4 | 12 | 3 |
| ANL A, direct | 55 | 2 | 2 | 8 | 12 | 1.5 |
| ANL A, \#data | 54 | 2 | 2 | 8 | 12 | 1.5 |
| ANL direct, A | 52 | 2 | 2 | 8 | 12 | 1.5 |
| ANL direct, \#data | 53 | 3 | 3 | 12 | 24 | 2 |
| ORL A, R0 | 48 | 1 | 1 | 4 | 12 | 3 |
| ORL A, R1 | 49 | 1 | 1 | 4 | 12 | 3 |
| ORL A, R2 | 4A | 1 | 1 | 4 | 12 | 3 |
| ORL A, R3 | 4B | 1 | 1 | 4 | 12 | 3 |
| ORL A, R4 | 4C | 1 | 1 | 4 | 12 | 3 |
| ORL A, R5 | 4D | 1 | 1 | 4 | 12 | 3 |
| ORL A, R6 | 4E | 1 | 1 | 4 | 12 | 3 |
| ORL A, R7 | 4F | 1 | 1 | 4 | 12 | 3 |

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 <br> SERIES <br> CLOCK <br> CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ORL A, @R0 | 46 | 1 | 1 | 4 | 12 | 3 |
| ORL A, @R1 | 47 | 1 | 1 | 4 | 12 | 3 |
| ORL A, direct | 45 | 2 | 2 | 8 | 12 | 1.5 |
| ORL A, \#data | 44 | 2 | 2 | 8 | 12 | 1.5 |
| ORL direct, A | 42 | 2 | 2 | 8 | 12 | 1.5 |
| ORL direct, \#data | 43 | 3 | 3 | 12 | 24 | 2 |
| XRL A, R0 | 68 | 1 | 1 | 4 | 12 | 3 |
| XRL A, R1 | 69 | 1 | 1 | 4 | 12 | 3 |
| XRL A, R2 | 6A | 1 | 1 | 4 | 12 | 3 |
| XRL A, R3 | 6B | 1 | 1 | 4 | 12 | 3 |
| XRL A, R4 | 6C | 1 | 1 | 4 | 12 | 3 |
| XRL A, R5 | 6D | 1 | 1 | 4 | 12 | 3 |
| XRL A, R6 | 6E | 1 | 1 | 4 | 12 | 3 |
| XRL A, R7 | 6F | 1 | 1 | 4 | 12 | 3 |
| XRL A, @R0 | 66 | 1 | 1 | 4 | 12 | 3 |
| XRL A, @R1 | 67 | 1 | 1 | 4 | 12 | 3 |
| XRL A, direct | 65 | 2 | 2 | 8 | 12 | 1.5 |
| XRL A, \#data | 64 | 2 | 2 | 8 | 12 | 1.5 |
| XRL direct, A | 62 | 2 | 2 | 8 | 12 | 1.5 |
| XRL direct, \#data | 63 | 3 | 3 | 12 | 24 | 2 |
| CLR A | E4 | 1 | 1 | 4 | 12 | 3 |
| CPL A | F4 | 1 | 1 | 4 | 12 | 3 |
| RL A | 23 | 1 | 1 | 4 | 12 | 3 |
| RLC A | 33 | 1 | 1 | 4 | 12 | 3 |
| RR A | 03 | 1 | 1 | 4 | 12 | 3 |
| RRC A | 13 | 1 | 1 | 4 | 12 | 3 |
| SWAP A | C4 | 1 | 1 | 4 | 12 | 3 |
| MOV A, R0 | E8 | 1 | 1 | 4 | 12 | 3 |
| MOV A, R1 | E9 | 1 | 1 | 4 | 12 | 3 |
| MOV A, R2 | EA | 1 | 1 | 4 | 12 | 3 |
| MOV A, R3 | EB | 1 | 1 | 4 | 12 | 3 |
| MOV A, R4 | EC | 1 | 1 | 4 | 12 | 3 |

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 SERIES CLOCK CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MOV A, R5 | ED | 1 | 1 | 4 | 12 | 3 |
| MOV A, R6 | EE | 1 | 1 | 4 | 12 | 3 |
| MOV A, R7 | EF | 1 | 1 | 4 | 12 | 3 |
| MOV A, @R0 | E6 | 1 | 1 | 4 | 12 | 3 |
| MOV A, @R1 | E7 | 1 | 1 | 4 | 12 | 3 |
| MOV A, direct | E5 | 2 | 2 | 8 | 12 | 1.5 |
| MOV A, \#data | 74 | 2 | 2 | 8 | 12 | 1.5 |
| MOV R0, A | F8 | 1 | 1 | 4 | 12 | 3 |
| MOV R1, A | F9 | 1 | 1 | 4 | 12 | 3 |
| MOV R2, A | FA | 1 | 1 | 4 | 12 | 3 |
| MOV R3, A | FB | 1 | 1 | 4 | 12 | 3 |
| MOV R4, A | FC | 1 | 1 | 4 | 12 | 3 |
| MOV R5, A | FD | 1 | 1 | 4 | 12 | 3 |
| MOV R6, A | FE | 1 | 1 | 4 | 12 | 3 |
| MOV R7, A | FF | 1 | 1 | 4 | 12 | 3 |
| MOV R0, direct | A8 | 2 | 2 | 8 | 12 | 1.5 |
| MOV R1, direct | A9 | 2 | 2 | 8 | 12 | 1.5 |
| MOV R2, direct | AA | 2 | 2 | 8 | 12 | 1.5 |
| MOV R3, direct | $A B$ | 2 | 2 | 8 | 12 | 1.5 |
| MOV R4, direct | AC | 2 | 2 | 8 | 12 | 1.5 |
| MOV R5, direct | AD | 2 | 2 | 8 | 12 | 1.5 |
| MOV R6, direct | AE | 2 | 2 | 8 | 12 | 1.5 |
| MOV R7, direct | AF | 2 | 2 | 8 | 12 | 1.5 |
| MOV R0, \#data | 78 | 2 | 2 | 8 | 12 | 1.5 |
| MOV R1, \#data | 79 | 2 | 2 | 8 | 12 | 1.5 |
| MOV R2, \#data | 7A | 2 | 2 | 8 | 12 | 1.5 |
| MOV R3, \#data | 7B | 2 | 2 | 8 | 12 | 1.5 |
| MOV R4, \#data | 7C | 2 | 2 | 8 | 12 | 1.5 |
| MOV R5, \#data | 7D | 2 | 2 | 8 | 12 | 1.5 |
| MOV R6, \#data | 7E | 2 | 2 | 8 | 12 | 1.5 |
| MOV R7, \#data | 7F | 2 | 2 | 8 | 12 | 1.5 |
| MOV @R0, A | F6 | 1 | 1 | 4 | 12 | 3 |

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 SERIES CLOCK CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MOV @R1, A | F7 | 1 | 1 | 4 | 12 | 3 |
| MOV @R0, direct | A6 | 2 | 2 | 8 | 12 | 1.5 |
| MOV @R1, direct | A7 | 2 | 2 | 8 | 12 | 1.5 |
| MOV @R0, \#data | 76 | 2 | 2 | 8 | 12 | 1.5 |
| MOV @R1, \#data | 77 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, A | F5 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R0 | 88 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R1 | 89 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R2 | 8A | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R3 | 8B | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R4 | 8C | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R5 | 8D | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R6 | 8E | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, R7 | 8F | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, @R0 | 86 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, @R1 | 87 | 2 | 2 | 8 | 12 | 1.5 |
| MOV direct, direct | 85 | 3 | 3 | 12 | 24 | 2 |
| MOV direct, \#data | 75 | 3 | 3 | 12 | 24 | 2 |
| MOV DPTR, \#data 16 | 90 | 3 | 3 | 12 | 24 | 2 |
| MOVC A, @A+DPTR | 93 | 1 | 2 | 8 | 24 | 3 |
| MOVC A, @A+PC | 83 | 1 | 2 | 8 | 24 | 3 |
| MOVX A, @R0 | E2 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| MOVX A, @R1 | E3 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| MOVX A, @DPTR | E0 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| MOVX @R0, A | F2 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| MOVX @R1, A | F3 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| MOVX @DPTR, A | F0 | 1 | 2-9 | 8-36 | 24 | 3-0.66 |
| PUSH direct | C0 | 2 | 2 | 8 | 24 | 3 |
| POP direct | D0 | 2 | 2 | 8 | 24 | 3 |
| XCH A, R0 | C8 | 1 | 1 | 4 | 12 | 3 |
| XCH A, R1 | C9 | 1 | 1 | 4 | 12 | 3 |

Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 SERIES CLOCK CYCLES | $\begin{gathered} 8032 \\ \text { CLOCK } \\ \text { CYCLES } \end{gathered}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| XCH A, R2 | CA | 1 | 1 | 4 | 12 | 3 |
| XCH A, R3 | CB | 1 | 1 | 4 | 12 | 3 |
| XCH A, R4 | CC | 1 | 1 | 4 | 12 | 3 |
| XCH A, R5 | CD | 1 | 1 | 4 | 12 | 3 |
| XCH A, R6 | CE | 1 | 1 | 4 | 12 | 3 |
| XCH A, R7 | CF | 1 | 1 | 4 | 12 | 3 |
| XCH A, @R0 | C6 | 1 | 1 | 4 | 12 | 3 |
| XCH A, @R1 | C7 | 1 | 1 | 4 | 12 | 3 |
| XCHD A, @R0 | D6 | 1 | 1 | 4 | 12 | 3 |
| XCHD A, @R1 | D7 | 1 | 1 | 4 | 12 | 3 |
| XCH A, direct | C5 | 2 | 2 | 8 | 12 | 1.5 |
| CLR C | C3 | 1 | 1 | 4 | 12 | 3 |
| CLR bit | C2 | 2 | 2 | 8 | 12 | 1.5 |
| SETB C | D3 | 1 | 1 | 4 | 12 | 3 |
| SETB bit | D2 | 2 | 2 | 8 | 12 | 1.5 |
| CPL C | B3 | 1 | 1 | 4 | 12 | 3 |
| CPL bit | B2 | 2 | 2 | 8 | 12 | 1.5 |
| ANL C, bit | 82 | 2 | 2 | 8 | 24 | 3 |
| ANL C, /bit | B0 | 2 | 2 | 6 | 24 | 3 |
| ORL C, bit | 72 | 2 | 2 | 8 | 24 | 3 |
| ORL C, /bit | A0 | 2 | 2 | 6 | 24 | 3 |
| MOV C, bit | A2 | 2 | 2 | 8 | 12 | 1.5 |
| MOV bit, C | 92 | 2 | 2 | 8 | 24 | 3 |
| ACALL addr11 | $\begin{aligned} & \text { 71, 91, B1, } \\ & \text { 11, 31, 51, } \\ & \text { D1, F1 } \end{aligned}$ | 2 | 3 | 12 | 24 | 2 |
| LCALL addr16 | 12 | 3 | 4 | 16 | 24 | 1.5 |
| RET | 22 | 1 | 2 | 8 | 24 | 3 |
| RETI | 32 | 1 | 2 | 8 | 24 | 3 |
| AJMP ADDR11 | $\begin{aligned} & 01,21,41, \\ & 61,81, \mathrm{~A} 1, \\ & \mathrm{C} 1, \mathrm{E} 1 \end{aligned}$ | 2 | 3 | 12 | 24 | 2 |
| LJMP addr16 | 02 | 3 | 4 | 16 | 24 | 1.5 |
| JMP @A+DPTR | 73 | 1 | 2 | 6 | 24 | 3 |

W79E825A/824A/823B/822B Data Sheet
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Continued

| OP-CODE | HEX CODE | BYTES | W79E825 SERIES MACHINE CYCLE | W79E825 <br> SERIES <br> CLOCK <br> CYCLES | $\begin{aligned} & 8032 \\ & \text { CLOCK } \\ & \text { CYCLES } \end{aligned}$ | W79E825 SERIES VS. 8032 SPEED RATIO |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SJMP rel | 80 | 2 | 3 | 12 | 24 | 2 |
| JZ rel | 60 | 2 | 3 | 12 | 24 | 2 |
| JNZ rel | 70 | 2 | 3 | 12 | 24 | 2 |
| JC rel | 40 | 2 | 3 | 12 | 24 | 2 |
| JNC rel | 50 | 2 | 3 | 12 | 24 | 2 |
| JB bit, rel | 20 | 3 | 4 | 16 | 24 | 1.5 |
| JNB bit, rel | 30 | 3 | 4 | 16 | 24 | 1.5 |
| JBC bit, rel | 10 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE A, direct, rel | B5 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE A, \#data, rel | B4 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE @R0, \#data, rel | B6 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE @R1, \#data, rel | B7 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R0, \#data, rel | B8 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R1, \#data, rel | B9 | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R2, \#data, rel | BA | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R3, \#data, rel | BB | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R4, \#data, rel | BC | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R5, \#data, rel | BD | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R6, \#data, rel | BE | 3 | 4 | 16 | 24 | 1.5 |
| CJNE R7, \#data, rel | BF | 3 | 4 | 16 | 24 | 1.5 |
| DJNZ R0, rel | D8 | 2 | 3 | 12 | 24 | 2 |
| DJNZ R1, rel | D9 | 2 | 3 | 12 | 24 | 2 |
| DJNZ R5, rel | DD | 2 | 3 | 12 | 24 | 2 |
| DJNZ R2, rel | DA | 2 | 3 | 12 | 24 | 2 |
| DJNZ R3, rel | DB | 2 | 3 | 12 | 24 | 2 |
| DJNZ R4, rel | DC | 2 | 3 | 12 | 24 | 2 |
| DJNZ R6, rel | DE | 2 | 3 | 12 | 24 | 2 |
| DJNZ R7, rel | DF | 2 | 3 | 12 | 24 | 2 |
| DJNZ direct, rel | D5 | 3 | 4 | 16 | 24 | 1.5 |

Table 9-1: Instruction Set for W79E825/824/823/822

### 9.1 Instruction Timing

This section is important because some applications use software instructions to generate timing delays. It also provides more information about timing differences between the W79E825 series and the standard 8051/52.
In W79E825 series, each machine cycle is four clock periods long. Each clock period is called a state, and each machine cycle consists of four states: C1, C2 C3 and C4, in order. Both clock edges are used for internal timing, so the duty cycle of the clock should be as close to $50 \%$ as possible to avoid timing conflicts.
The W79E825 series does one op-code fetch per machine cycle, so, in most instructions, the number of machine cycles required is equal to the number of bytes in the instruction. There are 256 available op-codes. 128 of them are single-cycle instructions, so many op-codes are executed in just four clocks period. Some of the other op-codes are two-cycle instructions, and most of these have two-byte opcodes. However, there are some instructions that have one-byte instructions yet take two cycles to execute. One important example is the MOVX instruction.
In the standard 8052, the MOVX instruction is always two machine cycles long. However, in the W79E825 series each machine cycle is made of only 4 clock periods compared to the 12 clock periods for the standard 8052. Therefore, even though the number of categories has increased, each instruction is at least 1.5 to 3 times faster than the standard 8052 in terms of clock periods.


Figure 9-1: Single Cycle Instruction Timing


Figure 9-2: Two Cycles Instruction Timing


Figure 9-3: Three Cycles Instruction Timing


Figure 9-4: Four Cycles Instruction Timing


Figure 9-5: Five Cycles Instruction Timing

## W79E825A/824A/823B/822B Data Sheet

## 10 POWER MANAGEMENT

The W79E825 series has several features that help the user to control the power consumption of the device. These modes are discussed in the next two sections.

### 10.1 Idle Mode

The user can put the device into idle mode by writing 1 to the bit PCON.O. The instruction that sets the idle bit is the last instruction that will be executed before the device goes into Idle Mode. In the Idle mode, the clock to the CPU is halted, but not to the Interrupt, Timer, Watchdog timer, I2C, PWM and Serial port blocks. This forces the CPU state to be frozen; the Program counter, the Stack Pointer, the Program Status Word, the Accumulator and the other registers hold their contents. The port pins hold the logical states they had at the time Idle was activated. The Idle mode can be terminated in two ways. Since the interrupt controller is still active, the activation of any enabled interrupt can wake up the processor. This will automatically clear the Idle bit, terminate the Idle mode, and the Interrupt Service Routine (ISR) will be executed. After the ISR, execution of the program will continue from the instruction which put the device into Idle Mode.
The Idle mode can also be exited by activating the reset. The device can put into reset either by applying a low on the external /RST pin, a Power on reset condition or a Watchdog timer reset. The external reset pin has to be held low for at least two machine cycles i.e. 8 clock periods to be recognized as a valid reset. In the reset condition the program counter is reset to 0000h and all the SFRs are set to the reset condition. Since the clock is already running there is no delay and execution starts immediately. In the Idle mode, the Watchdog timer continues to run, and if enabled, a time-out will cause a watchdog timer interrupt which will wake up the device. The software must reset the Watchdog timer in order to preempt the reset which will occur after 512 clock periods of the time-out. When the W79E825 series are exiting from an Idle Mode with a reset, the instruction following the one which put the device into Idle Mode is not executed. So there is no danger of unexpected writes.

### 10.2 Power Down Mode

The device can be put into Power Down mode by writing 1 to bit PCON.1. The instruction that does this will be the last instruction to be executed before the device goes into Power Down mode. In the Power Down mode, all the clocks are stopped and the device comes to a halt. All activity is completely stopped and the power consumption is reduced to the lowest possible value. The port pins output the values held by their respective SFRs.
The W79E825 series will exit the Power Down mode with a reset or by an external interrupt pin enabled as level detected. An external reset can be used to exit the Power down state. The low on /RST pin terminates the Power Down mode, and restarts the clock. The program execution will restart from 0000h. In the Power down mode, the clock is stopped, so the Watchdog timer cannot be used to provide the reset to exit Power down mode when its clock source is external OSC or crystal.
The sources that can wake up from the power down mode are external interrupts, keyboard interrupt (KBI), brownout reset (BOR), and comparator interrupt (CMF1, CMF2).
The W79E825 series can be waken up from the Power Down mode by forcing an external interrupt pin activation, provided the corresponding interrupt is enabled, while the global enable (EA) bit is set. If these conditions are met, then either a low-level or a falling-edge at external interrupt pin will re-start the oscillator. The device will then execute the interrupt service routine for the corresponding external interrupt. After the interrupt service routine is completed, the program execution returns to the instruction after one which put the device into Power Down mode and continues from there. During Power down mode, if AUXR1.LPBOV $=1$ and AUXR1.BOD $=0$, the internal RC clock will be enabled and hence save power.

## W79E825A/824A/823B/822B Data Sheet

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## 11 RESET CONDITIONS

The user has several hardware related options for placing the W79E825 series into reset condition. In general, most register bits go to their reset value irrespective of the reset condition, but there are a few flags whose state depends on the source of reset. The user can use these flags to determine the cause of reset using software.

### 11.1 Sources of reset

### 11.1.1 External Reset

The device samples the /RST pin every machine cycle during state C4. The /RST pin must be held low for at least two machine cycles before the reset circuitry applies an internal reset signal. Thus, this reset is a synchronous operation and requires the clock to be running.
The device remains in the reset state as long as /RST is low and remains low up to two machine cycles after /RST is deactivated. Then, the device begins program execution at 0000h. There are no flags associated with the external reset, but, since the other two reset sources do have flags, the external reset is the cause if those flags are clear.

### 11.1.2 Power-On Reset (POR)

The software must clear the POR flag after reading it. Otherwise it will not be possible to correctly determine future reset sources. If the power fails, then the device will once again go into reset state. When the power returns to the proper operating levels, the device will again perform a power on reset delay and set the POR flag.

### 11.1.3 Watchdog Timer Reset

The Watchdog Timer is a free-running timer with programmable time-out intervals. The program must clear the Watchdog Timer before the time-out interval is reached to restart the count. If the time-out interval is reached, an interrupt flag is set. 512 clocks later, if the Watchdog Reset is enabled and the Watchdog Timer has not been cleared, the Watchdog Timer generates a reset. The reset condition is maintained by the hardware for two machine cycles, and the WTRF bit in WDCON is set. Afterwards, the device begins program execution at 0000h.

### 11.2 Reset State

When the device is reset, most registers return to their initial state. The Watchdog Timer is disabled if the reset source was a power-on reset. The port registers are set to FFh, which puts most of the port pins in a high state. The Program Counter is set to 0000h, and the stack pointer is reset to 07h. After this, the device remains in the reset state as long as the reset conditions are satisfied.
Reset does not affect the on-chip RAM, however, so RAM is preserved as long as VDD remains above approximately 2 V , the minimum operating voltage for the RAM. If VDD falls below 2 V , the RAM contents are also lost. In either case, the stack pointer is always reset, so the stack contents are lost.

## W79E825A/824A/823B/822B Data Sheet

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SFR RESET VALUE

| SFR NAME | RESET VALUE | SFR NAME | RESET VALUE |
| :---: | :---: | :---: | :---: |
| P0 | 1111 1111B | I2DAT | xxxx xxxxB |
| SP | 0000 0111B | I2STATUS | 0000 0xxxB |
| DPL | 0000 0000B | I2TIMER | 0000 0000B |
| DPH | 0000 0000B | I2CLK | 0000 0000B |
| PCON | 00xx 0000B | I2CON | 0000 0000B |
| TCON | 0000 0000B | I2ADDR | xxxx xxxxB |
| TMOD | 0000 0000B | TA | 1111 1111B |
| TLO | 0000 0000B | PSW | 0000 0000B |
| TL1 | 0000 0000B | PWMP1 | xxxx xx00B |
| TH0 | 0000 0000B | PWMOH | xxxx xx00B |
| TH1 | 0000 0000B | PWM1H | xxxx xx00B |
| CKCON | 0000 0000B | PWM2H | xxxx xx00B |
| P1 | 1111 xx11B | PWM3H | xxxx xx00B |
| DIVM | 0000 0000B | WDCON | 0x00 0000B |
| SCON | 0000 0000B | PWMPO | 0000 0000B |
| SBUF | xxxx xxxxB | PWMOL | 0000 0000B |
| P2 | xxx xx11B | PWM1L | 0000 0000B |
| KBI | 0000 0000B | PWMCON1 | 0000 0000B |
| AUXR1 | 0000 0000B | PWM2L | 0000 0000B |
| IE | 0000 0000B | PWM3L | 0000 0000B |
| SADDR | 0000 0000B | PWMCON2 | 0000 0000B |
| CMP1 | 0000 0000B | PWMCON3 | xxxxxxx0B |
| CMP2 | 0000 0000B | ACC | 0000 0000B |
| P0M1 | 0000 0000B | ADCCON | xx00 0x00B |
| POM2 | 0000 0000B | ADCH | xxxx xxxxB |
| P1M1 | 0000 0000B | EIE | xx000 000B |
| P1M2 | 0000 0000B | B | 0000 0000B |
| P2M1 | 0000 0000B | POIDS | 0000 0000B |
| P2M2 | xxxx xx00B | IPH | xx00 0000B |
| IPOH | x000 0000B | IP1 | xx00 0000B |
| IP0 | x000 0000B | NVMADDR | 0000 0000B |
| SADEN | 0000 0000B | NVMDAT | 0000 0000B |
|  |  | NVMCON | 00xx xxxxB |

Table 11-1: SFR Reset Value

## W79E825A/824A/823B/822B Data Sheet

The WDCON SFR bits are set/cleared in reset condition depending on the source of the reset.

| External reset | Watchdog reset | Power on reset |
| :--- | :--- | :--- |
| $0 \times 0 \times 0 \times 0 \mathrm{~B}$ | $0 \times 0 \times 01 \times 0 \mathrm{~B}$ | 01000000 B |

The WDCON SFR is set to a $0 \times 00$ 0000B on the reset. WTRF (WDCON.2) is set to a 1 on a Watchdog timer reset, but to a 0 on power on/down resets. WTRF (WDCON.2) is not altered by external reset. EWRST (WDCON.1) is cleared by any reset. Software or any reset will clear WDIF(WDCON.3) bit.
Some of the bits in the WDCON SFR (WDRUN, WDCLR, EWRST, WDIF, WD0 and WD1) have unrestricted read access which required Timed Access procedure to write. The remaining bits have unrestricted write accesses. Please refer TA register description.


Figure 11-1: Internal reset and VDD monitor timing diagram


Figure 11-2: External reset timing diagram

## 12 INTERRUPTS

The W79E825 series have four priority level interrupts structure with 13 interrupt sources. Each of the interrupt sources has an individual priority bit, flag, interrupt vector and enable bit. In addition, the interrupts can be globally enabled or disabled.

### 12.1 Interrupt Sources

The External Interrupts INTO and INT1 can be either edge triggered or level triggered, depending on bits ITO and IT1. The bits IEO and IE1 in the TCON register are the flags which are checked to generate the interrupt. In the edge triggered mode, the INTx inputs are sampled in every machine cycle. If the sample is high in one cycle and low in the next, then a high to low transition is detected and the interrupts request flag IEx in TCON is set. The flag bit requests the interrupt. Since the external interrupts are sampled every machine cycle, they have to be held high or low for at least one complete machine cycle. The IEx flag is automatically cleared when the service routine is called. If the level triggered mode is selected, then the requesting source has to hold the pin low till the interrupt is serviced. The IEx flag will not be cleared by the hardware on entering the service routine. If the interrupt continues to be held low even after the service routine is completed, then the processor may acknowledge another interrupt request from the same source.
The Timer 0 and 1 Interrupts are generated by the TFO and TF1 flags. These flags are set by the overflow in the Timer 0 and Timer 1. The TF0 and TF1 flags are automatically cleared by the hardware when the timer interrupt is serviced. The Watchdog timer can be used as a system monitor or a simple timer. In either case, when the time-out count is reached, the Watchdog Timer interrupt flag WDIF (WDCON.3) is set. If the interrupt is enabled by the enable bit EIE.4, then an interrupt will occur.
The Serial block can generate interrupt on reception or transmission. There are two interrupt sources from the Serial block, which are obtained by the RI and TI bits in the SCON SFR. These bits are not automatically cleared by the hardware, and the user will have to clear these bits by software.
All the bits that generate interrupts can be set or reset by software, and thereby software initiated interrupts can be generated. Each of the individual interrupts can be enabled or disabled by setting or clearing a bit in the IE SFR. IE also has a global enable/disable bit EA, which can be cleared to disable all interrupts.
The ADC can generate interrupt after finished ADC converter. There is one interrupt source, which is obtained by the ADCI bit in the ADCCON SFR. This bit is not automatically cleared by the hardware, and the user will have to clear this bit using software.
The two comparators can generate interrupt after comparator output has toggle occurs by CMF1 and CMF2. These bits are not automatically cleared by the hardware, and the user will have to clear these bits using software.
The I2C function can generate interrupt, if EI2C and EA bits are enabled, when SI Flag is set due to a new I2C status code is generated, SI flag is generated by hardware and must be cleared by software.
The PWM function can generate interrupt by BKF flag, after external brake pin has brake occurred. This bit will be cleared by software.
The interrupt flags are sampled every machine cycle. In the same machine cycle, the sampled interrupts are polled and their priority is resolved. If certain conditions are met then the hardware will execute an internally generated LCALL instruction which will vector the process to the appropriate interrupt vector address. The conditions for generating the LCALL are;

1. An interrupt of equal or higher priority is not currently being serviced.
2. The current polling cycle is the last machine cycle of the instruction currently being execute.
3. The current instruction does not involve a write to IE, EIE, IP0, IPOH, IP1 or IPH1 registers and is not a RETI.

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If any of these conditions are not met, then the LCALL will not be generated. The polling cycle is repeated every machine cycle, with the interrupts sampled in the same machine cycle. If an interrupt flag is active in one cycle but not responded to, and is not active when the above conditions are met, the denied interrupt will not be serviced. This means that active interrupts are not remembered; every polling cycle is new.

The processor responds to a valid interrupt by executing an LCALL instruction to the appropriate service routine. This may or may not clear the flag which caused the interrupt. In case of Timer interrupts, the TF0 or TF1 flags are cleared by hardware whenever the processor vectors to the appropriate timer service routine. In case of external interrupts, $\overline{\text { INTO }}$ and INT1, the flags are cleared only if they are edge triggered. In case of Serial interrupts, the flags are not cleared by hardware. The Watchdog timer interrupt flag WDIF has to be cleared by software. The hardware LCALL behaves exactly like the software LCALL instruction. This instruction saves the Program Counter contents onto the Stack, but does not save the Program Status Word PSW. The PC is reloaded with the vector address of that interrupt which caused the LCALL. These address of vector for the different sources are as follows:

VECTOR LOCATIONS FOR INTERRUPT SOURCES

| SOURCE | VECTOR ADDRESS | SOURCE | VECTOR ADDRESS |
| :--- | :--- | :--- | :--- |
| External Interrupt 0 | 0003 h | Timer 0 Overflow | 000 Bh |
| External Interrupt 1 | 0013 h | Timer 1 Overflow | 001 Bh |
| Serial Port | 0023 h | Brownout Interrupt | 002 Bh |
| I2C Interrupt | 0033 h | KBI Interrupt | 003 Bh |
| Comparator 2 Interrupt | 0043 h | - | 004 Bh |
| Watchdog Timer | 0053 h | ADC Interrupt | 005 Bh |
| Comparator 1 Interrupt | 0063 h | - | 006 Bh |
| PWM Brake Interrupt | 0073 h | - | 007 Bh |

Table 12-1: Vector locations for interrupt sources

Execution continues from the vectored address till an RETI instruction is executed. On execution of the RETI instruction the processor pops the Stack and loads the PC with the contents at the top of the stack. The user must take care that the status of the stack is restored to what it was after the hardware LCALL, if the execution is return to the interrupted program. The processor does not notice anything if the stack contents are modified and will proceed with execution from the address put back into PC. Note that a RET instruction would perform exactly the same process as a RETI instruction, but it would not inform the Interrupt Controller that the interrupt service routine is completed, and would leave the controller still thinking that the service routine is underway.

### 12.2 Priority Level Structure

The W79E825 series uses a four priority level interrupt structure (highest, high, low and lowest) and supports up to 13 interrupt sources. The interrupt sources can be individually set to either high or low levels. Naturally, a higher priority interrupt cannot be interrupted by a lower priority interrupt. However there exists a pre-defined hierarchy amongst the interrupts themselves. This hierarchy comes into play when the interrupt controller has to resolve simultaneous requests having the same priority level. This hierarchy is defined as table below. This allows great flexibility in controlling and handling many interrupt sources.

| PRIORITY BITS |  | INTERRUPT PRIORITY LEVEL |  |
| :---: | :---: | :--- | :---: |
| IPXH | IPX |  |  |
| 0 | 0 | Level 0 (lowest priority) |  |
| 0 | 1 | Level 1 |  |
| 1 | 0 | Level 2 |  |
| 1 | 1 | Level 3 (highest priority) |  |

Table 12-2: Four-level interrupt priority
Each interrupt source can be individually programmed to one of four priority levels by setting or clearing bits in the IPO, IPOH, IP1, and IP1H registers. An interrupt service routine in progress can be interrupted by a higher priority interrupt, but not by another interrupt of the same or lower priority. The highest priority interrupt service cannot be interrupted by any other interrupt source. So, if two requests of different priority levels are received simultaneously, the request of higher priority level is serviced.

If requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. This is called the arbitration ranking. Note that the arbitration ranking is only used to resolve simultaneous requests of the same priority level.

As below Table summarizes the interrupt sources, flag bits, vector addresses, enable bits, priority bits, arbitration ranking, and whether each interrupt may wake up the CPU from Power Down mode.

| SOURCE | FLAG | VECTOR <br> ADDRESS | INTERRUPT <br> ENABLE <br> BITS | INTERRUPT <br> PRIORITY | FLAG <br> CLEARED BY | ARBITRATI <br> ON <br> RANKING | POWER <br> DOWN <br> WAKEUP |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: |
| External <br> Interrupt 0 | IE0 | 0003 H | EX0 (IE0.0) | IP0H.0, IP0.0 | Hardware, <br> Follow the <br> inverse of pin | 1(highest) | Yes |
| Brownout <br> Detect | BOF | 002 BH | EBO (IE.5) | IP0H.5, IP0.5 | Software | 2 | Yes |
| Watchdog <br> Timer | WDIF | $0053 H$ | EWDI (EIE.4) | IP1H.4, IP1.4 | Software | 3 | No |
| Timer 0 <br> Interrupt | TF0 | $000 B H$ | ET0 (IE.1) | IP0H.1, IP0.1 | Hardware, <br> software | 4 | No |
| I2C Interrupt | SI | $0033 H$ | EI2 (EIE.0) | IP1H.0, IP1.0 | Software | 5 | No |
| ADC Converter | ADCI | $005 B H$ | EAD (IE.6) | IP0H.6, IP0.6 | Hardware | 6 | Yes ${ }^{(1)}$ |

Continued .

| Source | Flag | Vector <br> address | Interrupt <br> Enable Bits | Interrupt <br> Priority | Flag cleared <br> by | Arbitration <br> Ranking | Power <br> Down <br> Wakeup |
| :--- | :--- | :--- | :--- | :--- | :--- | :---: | :---: |
| External <br> Interrupt 1 | IE1 | 0013 H | EX1 (IE.2) | IP0H.2, IP0.2 | Hardware, <br> Follow the <br> inverse of pin | 7 | Yes |
| KBI Interrupt | KBF | 003 BH | EKB (EIE.1) | IP1H.1, IP1.1 | Software | 8 | Yes |
| Comparator 1 <br> Interrupt | CMF1 | $0063 H$ | EC1 (EIE.2) | IP1H.2, IP1.2 | Software | 9 | Yes |
| Timer 1 Interrupt | TF1 | 001 BH | ET1 (IE.3) | IP0H.3, IP0.3 | Hardware, <br> software | 10 | No |
| Comparator 2 <br> Interrupt | CMF2 | $0043 H$ | EC2 (EIE.3) | IP1H.3, IP1.3 | Software | 11 | Yes |
| Serial Port Tx <br> and Rx | TI \& RI | $0023 H$ | ES (IE.4) | IP0H.4, IP0.4 | Software | 12 | No |
| PWM Interrupt | BKF | $0073 H$ | EPWM (EIE.5) | IP1H.5, IP1.5 | Software | 13 (lowest) | No |

Note: 1. The ADC Converter can wake up Power Down Mode when its clock source is from internal RC.
Table 12-3: Vector location for Interrupt sources and power down wakeup

### 12.3 Response Time

The response time for each interrupt source depends on several factors, such as the nature of the interrupt and the instruction underway. In the case of external interrupts $\overline{\operatorname{INTO}}$ and $\overline{\operatorname{INT} 1}$, they are sampled at C3 of every machine cycle and then their corresponding interrupt flags IEx will be set or reset. The Timer 0 and 1 overflow flags are set at C3 of the machine cycle in which overflow has occurred. These flag values are polled only in the next machine cycle. If a request is active and all three conditions are met, then the hardware generated LCALL is executed. This LCALL itself takes four machine cycles to be completed. Thus there is a minimum time of five machine cycles between the interrupt flag being set and the interrupt service routine being executed.

A longer response time should be anticipated if any of the three conditions are not met. If a higher or equal priority is being serviced, then the interrupt latency time obviously depends on the nature of the service routine currently being executed. If the polling cycle is not the last machine cycle of the instruction being executed, then an additional delay is introduced. The maximum response time (if no other interrupt is in service) occurs if the W79E825 series are performing a write to IE, EIE, IPO, IPOH, IP1 or IP1H and then executes a MUL or DIV instruction. From the time an interrupt source is activated, the longest reaction time is 12 machine cycles. This includes 1 machine cycle to detect the interrupt, 2 machine cycles to complete the IE, EIE, IP0, IPOH, IP1 or IP1H access, 5 machine cycles to complete the MUL or DIV instruction and 4 machine cycles to complete the hardware LCALL to the interrupt vector location.

Thus in a single-interrupt system the interrupt response time will always be more than 5 machine cycles and not more than 12 machine cycles. The maximum latency of 12 machine cycles is 48 clock cycles. Note that in the standard 8051 the maximum latency is 8 machine cycles which equals 96 machine cycles. This is a $50 \%$ reduction in terms of clock periods.

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### 12.4 Interrupt Inputs

The W79E825 series have 13 interrupts source, and two individual interrupt inputs sources, one is for IE0, IE1, BOF, KBF, WDT, ADC, CMF1 and CMF2, and other is IF0, IF1, RI+TI ,SI and BKF. Two interrupt inputs are identical to those present on the standard 80C51 microcontroller as show in below figures.

If an external interrupt is enabled when the W79E825 series are put into Power Down or Idle mode, the interrupt will cause the processor to wake up and resume operation.


Figure 12-1: Interrupt Sources that can wake up from Power Down Mode

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Figure 12-2: Interrupt Sources that cannot wake up from Power Down Mode

## 13 PROGRAMMABLE TIMERS/COUNTERS

The W79E825 series have two 16-bit programmable timer/counters and one programmable Watchdog Timer. The Watchdog Timer is operationally quite different from the other two timers. It's timer/counters have additional timer 0 or timer 1 overflow toggle output enable feature as compare to conventional timer/counters. This timer overflow toggle output can be configured to automatically toggle T0 or T1 pin output whenever a timer overflow occurs.

### 13.1 Timer/Counters 0 \& 1

The W79E825 series have two 16-bit Timer/Counters. Each of these Timer/Counters has two 8 bit registers which form the 16 bit counting register. For Timer/Counter 0 they are THO, the upper 8 bits register, and TLO, the lower 8 bit register. Similarly Timer/Counter 1 has two 8 bit registers, TH1 and TL1. The two can be configured to operate either as timers, counting machine cycles or as counters counting external inputs.
When configured as a "Timer", the timer counts clock cycles. The timer clock can be programmed to be thought of as $1 / 12$ of the system clock or $1 / 4$ of the system clock. In the "Counter" mode, the register is incremented on the falling edge of the external input pin, T0 for Timer 0, and T1 for Timer 1. The T0 and T1 inputs are sampled in every machine cycle at C4. If the sampled value is high in one machine cycle and low in the next, then a valid high to low transition on the pin is recognized and the count register is incremented. Since it takes two machine cycles to recognize a negative transition on the pin, the maximum rate at which counting will take place is $1 / 8$ of the master clock frequency. In either the "Timer" or "Counter" mode, the count register will be updated at C3. Therefore, in the "Timer" mode, the recognized negative transition on pin T0 and T1 can cause the count register value to be updated only in the machine cycle following the one in which the negative edge was detected.
The "Timer" or "Counter" function is selected by the " $C / \bar{T}$ " bit in the TMOD Special Function Register. Each Timer/Counter has one selection bit for its own; bit 2 of TMOD selects the function for Timer/Counter 0 and bit 6 of TMOD selects the function for Timer/Counter 1. In addition each Timer/Counter can be set to operate in any one of four possible modes. The mode selection is done by bits M 0 and M 1 in the TMOD SFR.

### 13.1.1 Time-Base Selection

The W79E825 series can operate like the standard 8051/52 family, counting at the rate of $1 / 12$ of the clock speed, or in turbo mode, counting at the rate of $1 / 4$ clock speed. The speed is controlled by the T0M and T1M bits in CKCON, and the default value is zero, which uses the standard 8051/52 speed.

### 13.1.2 Mode 0

In Mode 0, the timer/counter is a 13-bit counter. The 13-bit counter consists of THx (8 MSB) and the five lower bits of TLx (5 LSB). The upper three bits of TLx are ignored. The timer/counter is enabled when TRx is set and either GATE is 0 or $\overline{I N T x}$ is 1 . When $C / \bar{T}$ is 0 , the timer/counter counts clock cycles; when $\mathrm{C} / \overline{\mathrm{T}}$ is 1, it counts falling edges on T0 (P1.2 for Timer 0 ) or T 1 ( P 0.7 for Timer 1). For clock cycles, the time base may be $1 / 12$ or $1 / 4$ clock speed, and the falling edge of the clock increments the counter. When the 13-bit value moves from 1FFFh to 0000h, the timer overflow flag TFx is set, and an interrupt occurs if enabled. This is illustrated in next figure below.
In "Timer" mode, if output toggle enable bit of P2M1.T0OE or P2M1.T1OE is enabled, T0 or T1 output pin will toggle whenever a timer overflow occurs.

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Figure 13-1: Timer/Counters 0 \& 1 in Mode 0

### 13.1.3 Mode 1

Mode 1 is similar to Mode 0 except that the counting register forms a 16-bit counter, rather than a 13bit counter. This means that all the bits of THx and TLx are used. Roll-over occurs when the timer moves from a count of FFFFh to 0000h. The timer overflow flag TFx of the relevant timer is set and if enabled an interrupt will occur. The selection of the time-base in the timer mode is similar to that in Mode 0 . The gate function operates similarly to that in Mode 0.


Figure 13-2: Timer/Counters 0 \& 1 in Mode 1

### 13.1.4 Mode 2

In Mode 2, the timer/counter is in the Auto Reload Mode. In this mode, TLx acts as 8-bit count register, while THx holds the reload value. When the TLx register overflows from FFh to 00h, the TFx bit in TCON is set and TLx is reloaded with the contents of THx , and the counting process continues from here. The reload operation leaves the contents of the THx register unchanged. Counting is enabled by the TRx bit and proper setting of GATE and INTx pins. As in the other two modes 0 and 1 mode 2 allows counting of either clock cycles (clock/12 or clock/4) or pulses on pin Tn .
In "Timer" mode, if output toggle enable bit of P2M1.T0OE or P2M1.T1OE is enabled, T0 or T1 output pin will toggle whenever a timer overflow occurs.


Figure 13-3: Timer/Counter 0 \& 1 in Mode 2

### 13.1.5 Mode 3

Mode 3 has different operating methods for the two timer/counters. For timer/counter 1, mode 3 simply freezes the counter. Timer/Counter 0, however, configures TLO and TH0 as two separate 8 bit count registers in this mode. The logic for this mode is shown in the figure. TLO uses the Timer/Counter 0 control bits $\mathrm{C} / \overline{\mathrm{T}}$, GATE, TRO, INTO and TFO. The TLO can be used to count clock cycles (clock/12 or clock/4) or 1-to-0 transitions on pin T0 as determined by C/T (TMOD.2). TH0 is forced as a clock cycle counter (clock/12 or clock/4) and takes over the use of TR1 and TF1 from Timer/Counter 1. Mode 3 is used in cases where an extra 8 bit timer is needed. With Timer 0 in Mode 3, Timer 1 can still be used in Modes 0,1 and 2, but its flexibility is somewhat limited. While its basic functionality is maintained, it no longer has control over its overflow flag TF1 and the enable bit TR1. Timer 1 can still be used as a timer/counter and retains the use of GATE and INT1 pin. In this condition it can be turned on and off by switching it out of and into its own Mode 3. It can also be used as a baud rate generator for the serial port.
In "Timer" mode, if output toggle enable bit of P2M1.T0OE or P2M1.T1OE is enabled, T0 or T1 output pin will toggle whenever a timer overflow occurs.

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Figure 13-4: Timer/Counter Mode 3

## 14 NVM MEMORY

The W79E825 series have NVM data memory of 256/128 bytes for customer's data store used. The NVM data memory has four/two pages area and each page has 64 bytes as below figure. The Page 0 address is from FC00h ~ FC3Fh, Page 1 address is from FC40h ~ FC7Fh, Page 2 address is from FC80h ~ FCBFh, and Page 3 address is from FCCOh ~ FCFFh.

The NVM memory can be read/write by customer program to access. Read NVM data is by MOVC A,@A+DPTR instruction, and write data is by SFR of NVMADDR, NVMDAT and NVMCON. Before write data to NVM memory, the page must be erased by providing page address on NVMADDR, which low byte address of On-Chip Code Memory space will decode, then set EER of NVMCON.7. This will automatically hold fetch program code and PC Counter, and execute page erase. After finished, this bit will be cleared by hardware. The erase time is $\sim 5 \mathrm{~ms}$.

For writing data to NVM memory, user must set address and data to NVMADDR and NVMDAT, then set EWR of NVMCON. 6 to initiate nvm data write. The uC will hold program code and PC Counter, and then write data to mapping address. Upon write completion, the EWR bit will be cleared by hardware, the uC will continue execute next instruction. The program time is $\sim 50$ us.


Figure 14-1: W79E825/824 Memory Map

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Figure 14-2: W79E823/822 Memory Map

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## 15 WATCHDOG TIMER

The Watchdog Timer is a free-running Timer which can be programmed by the user to serve as a system monitor, a time-base generator or an event timer. It is basically a set of dividers that divide the system clock. The divider output is selectable and determines the time-out interval. When the time-out occurs a flag is set, which can cause an interrupt if enabled, and a system reset can also be caused if it is enabled. The interrupt will occur if the individual interrupt enable and the global enable are set. The interrupt and reset functions are independent of each other and may be used separately or together depending on the user's software.


Figure 15-1: Watchdog Timer
The Watchdog Timer should first be restarted by using WDCLR. This ensures that the timer starts from a known state. The WDCLR bit is used to restart the Watchdog Timer. This bit is self clearing, i.e. after writing a 1 to this bit the software will automatically clear it. The Watchdog Timer will now count clock cycles. The time-out interval is selected by the two bits WD1 and WDO (WDCON. 5 and WDCON.4). When the selected time-out occurs, the Watchdog interrupt flag WDIF (WDCON.3) is set. After the time-out has occurred, the Watchdog Timer waits for an additional 512 clock cycles. If the Watchdog Reset EWRST (WDCON.1) is enabled, then 512 clocks after the time-out, if there is no WDCLR, a system reset due to Watchdog Timer will occur. This will last for two machine cycles, and the Watchdog Timer reset flag WTRF (WDCON.2) will be set. This indicates to the software that the Watchdog was the cause of the reset.

When used as a simple timer, the reset and interrupt functions are disabled. The timer will set the WDIF flag each time the timer completes the selected time interval. The WDIF flag is polled to detect a time-out and the WDCLR allows software to restart the timer. The Watchdog Timer can also be used as a very long timer. The interrupt feature is enabled in this case. Every time the time-out occurs an interrupt will occur if the global interrupt enable EA is set.

The main use of the Watchdog Timer is as a system monitor. This is important in real-time control applications. In case of some power glitches or electro-magnetic interference, the processor may begin to execute errant code. If this is left unchecked the entire system may crash. Using the watchdog timer interrupt during software development will allow the user to select ideal watchdog reset locations. The code is first written without the watchdog interrupt or reset. Then the Watchdog interrupt is enabled to identify code locations where interrupt occurs. The user can now insert instructions to reset the Watchdog Timer, which will allow the code to run without any Watchdog Timer interrupts. Now the Watchdog Timer reset is enabled and the Watchdog interrupt may be disabled. If

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any errant code is executed now, then the reset Watchdog Timer instructions will not be executed at the required instants and Watchdog reset will occur.
The Watchdog Timer time-out selection will result in different time-out values depending on the clock speed. The reset, when enabled, will occur when 512 clocks after time-out has occurred.

| WD1 | WD0 | INTERRUPT TIME-OUT | RESET TIME-OUT | NUMBER OF CLOCKS | TIME <br> @ 10 MHZ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | $2^{17}$ | $2^{17}+512$ | 131072 | 13.11 mS |
| 0 | 1 | $2^{20}$ | $2^{20}+512$ | 1048576 | 104.86 mS |
| 1 | 0 | $2^{23}$ | $2^{23}+512$ | 8388608 | 838.86 mS |
| 1 | 1 | $2^{26}$ | $2^{26}+512$ | 67108864 | 6710.89 mS |

Table 15-2: Time-out values for the Watchdog Timer

The Watchdog Timer will de disabled by a power-on/fail reset. The Watchdog Timer reset does not disable the Watchdog Timer, but will restart it. In general, software should restart the timer to put it into a known state. The control bits that support the Watchdog Timer are discussed below.

### 15.1 WATCHDOG CONTROL

WDIF: WDCON. 3 - Watchdog Timer Interrupt flag. This bit is set whenever the time-out occurs in the Watchdog Timer. If the Watchdog interrupt is enabled (EIE.4), then an interrupt will occur (if the global interrupt enable is set and other interrupt requirements are met). Software or any reset can clear this bit.
WTRF: WDCON. 2 - Watchdog Timer Reset flag. This bit is set whenever a watchdog reset occurs. This bit is useful for determined the cause of a reset. Software must read it, and clear it manually. A Power-fail reset will clear this bit. If EWRST $=0$, then this bit will not be affected by the Watchdog Timer.
EWRST: WDCON. 1 - Enable Watchdog Timer Reset. This bit when set to 1 will enable the Watchdog Timer reset function. Setting this bit to 0 will disable the Watchdog Timer reset function, but will leave the timer running.
WDCLR: WDCON. 0 - Reset Watchdog Timer. This bit is used to clear the Watchdog Timer and to restart it. This bit is self-clearing, so after the software writes 1 to it the hardware will automatically clear it. If the Watchdog Timer reset is enabled, then the WDCLR has to be set by the user within 512 clocks of the time-out. If this is not done then a Watchdog Timer reset will occur.

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### 15.2 CLOCK CONTROL of Watchdog

WD1, WDO: WDCON.5, WDCON. 4 - Watchdog Timer Mode select bits. These two bits select the time-out interval for the watchdog timer. The reset time is 512 clocks longer than the interrupt time-out value.
The default Watchdog time-out is $2^{17}$ clocks, which is the shortest time-out period. The WDRUN, WDO, WD1, EWRST, WDIF and WDCLR bits are protected by the Timed Access procedure. This prevents software from accidentally enabling or disabling the watchdog timer. More importantly, it makes it highly improbable that errant code can enable or disable the Watchdog Timer.

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## 16 SERIAL PORT (UART)

Serial port in the W79E825 series is a full duplex port. The W79E825 series provide the user with additional features such as the Frame Error Detection and the Automatic Address Recognition. The serial ports are capable of synchronous as well as asynchronous communication. In Synchronous mode the W79E825 series generate the clock and operates in a half duplex mode. In the asynchronous mode, full duplex operation is available. This means that it can simultaneously transmit and receive data. The transmit register and the receive buffer are both addressed as SBUF Special Function Register. However any write to SBUF will be to the transmit register, while a read from SBUF will be from the receiver buffer register. The serial port can operate in four different modes as described below.

### 16.1 MODE 0

This mode provides synchronous communication with external devices. In this mode serial data is transmitted and received on the RXD line. TXD is used to transmit the shift clock. The TxD clock is provided by the W79E825 series whether the device is transmitting or receiving. This mode is therefore a half duplex mode of serial communication. In this mode, 8 bits are transmitted or received per frame. The LSB is Transmitted/Received first. The baud rate is fixed at $1 / 12$ or $1 / 4$ of the oscillator frequency. This Baud Rate is determined by the SM2 bit (SCON.5). When this bit is set to 0, then the serial port runs at $1 / 12$ of the clock. When set to 1 , the serial port runs at $1 / 4$ of the clock. This additional facility of programmable baud rate in mode 0 is the only difference between the standard 8051 and the W79E825 series.
The functional block diagram is shown below. Data enters and leaves the Serial port on the RxD line. The TxD line is used to output the shift clock. The shift clock is used to shift data into and out of the W79E825 series and the device at the other end of the line. Any instruction that causes a write to SBUF will start the transmission. The shift clock will be activated and data will be shifted out on the $R \times D$ pin till all 8 bits are transmitted. If $S M 2=1$, then the data on $R \times D$ will appear 1 clock period before the falling edge of shift clock on TxD. The clock on TxD then remains low for 2 clock periods, and then goes high again. If $\mathrm{SM} 2=0$, the data on RxD will appear 3 clock periods before the falling edge of shift clock on TxD. The clock on TxD then remains low for 6 clock periods, and then goes high again. This ensures that at the receiving end the data on RxD line can either be clocked on the rising edge of the shift clock on TxD or latched when the TxD clock is low.


Figure 16-1: Serial Port Mode 0

The TI flag is set high in C1 following the end of transmission of the last bit. The serial port will receive data when REN is 1 and RI is zero. The shift clock (TxD) will be activated and the serial port will latch data on the rising edge of shift clock. The external device should therefore present data on the falling edge on the shift clock. This process continues till all the 8 bits have been received. The RI flag is set in C1 following the last rising edge of the shift clock on TxD. This will stop reception, till the RI is cleared by software.

### 16.2 MODE 1

In Mode 1, the full duplex asynchronous mode is used. Serial communication frames are made up of 10 bits transmitted on TXD and received on RXD. The 10 bits consist of a start bit (0), 8 data bits (LSB first), and a stop bit (1). On received, the stop bit goes into RB8 in the SFR SCON. The baud rate in this mode is variable. The serial baud can be programmed to be $1 / 16$ or $1 / 32$ of the Timer 1 overflow. Since the Timer 1 can be set to different reload values, a wide variation in baud rates is possible.

Transmission begins with a write to SBUF. The serial data is brought out on to TxD pin at C1 following the first roll-over of divide by 16 counter. The next bit is placed on TxD pin at C1 following the next rollover of the divide-by-16 counter. Thus the transmission is synchronized to the divide-by-16 counter and not directly to the write to SBUF signal. After all 8 bits of data are transmitted, the stop bit is transmitted. The TI flag is set in the C1 state after the stop bit has been put out on TxD pin. This will be at the 10th rollover of the divide-by-16 counter after a write to SBUF.

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Reception is enabled only if REN is high. The serial port actually starts the receiving of serial data, with the detection of a falling edge on the RxD pin. The 1-to-0 detector continuously monitors the RxD line, sampling it at the rate of 16 times the selected baud rate. When a falling edge is detected, the divide-by- 16 counter is immediately reset. This helps to align the bit boundaries with the rollovers of the divide-by-16 counter.
The 16 states of the counter effectively divide the bit time into 16 slices. The bit detection is done on a best of three bases. The bit detector samples the RxD pin, at the 8th, 9th and 10th counter states. By using a majority 2 of 3 voting system, the bit value is selected. This is done to improve the noise rejection feature of the serial port. If the first bit detected after the falling edge of RxD pin is not 0 , then this indicates an invalid start bit, and the reception is immediately aborted. The serial port again looks for a falling edge in the RxD line. If a valid start bit is detected, then the rest of the bits are also detected and shifted into the SBUF.

After shifting in 8 data bits, there is one more shift to do, after which the SBUF and RB8 are loaded and RI is set. However certain conditions must be met before the loading and setting of RI can be done.

1. RI must be 0 and
2. Either $\mathrm{SM} 2=0$, or the received stop bit $=1$.

If these conditions are met, then the stop bit goes to RB8, the 8 data bits go into SBUF and RI is set. Otherwise the received frame may be lost. After the middle of the stop bit, the receiver goes back to looking for a 1-to-0 transition on the RxD pin.


Figure 16-2: Serial Port Mode 1

### 16.3 MODE 2

This mode uses a total of 11 bits in asynchronous full-duplex communication. The functional description is shown in the figure below. The frame consists of one start bit ( 0 ), 8 data bits (LSB first), a programmable 9th bit (TB8) and a stop bit (0). The 9th bit received is put into RB8. The baud rate is programmable to $1 / 32$ or $1 / 64$ of the oscillator frequency, which is determined by the SMOD bit in PCON SFR. Transmission begins with a write to SBUF. The serial data is brought out on to TxD pin at C1 following the first roll-over of the divide-by-16 counter. The next bit is placed on TxD pin at C1 following the next rollover of the divide-by-16 counter. Thus the transmission is synchronized to the divide-by-16 counter, and not directly to the write to SBUF signal. After all 9 bits of data are transmitted, the stop bit is transmitted. The TI flag is set in the C 1 state after the stop bit has been put out on TXD pin. This will be at the 11th rollover of the divide-by-16 counter after a write to SBUF. Reception is enabled only if REN is high. The serial port actually starts the receiving of serial data, with the detection of a falling edge on the RxD pin. The 1-to-0 detector continuously monitors the RxD line, sampling it at the rate of 16 times the selected baud rate. When a falling edge is detected, the divide- by-16 counter is immediately reset. This helps to align the bit boundaries with the rollovers of the divide-by-16 counter. The 16 states of the counter effectively divide the bit time into 16 slices. The bit detection is done on a best of three bases. The bit detector samples the RxD pin, at the 8th, 9th and 10th counter states. By using a majority 2 of 3 voting system, the bit value is selected. This is done to improve the noise rejection feature of the serial port.


Figure 16-3: Serial Port Mode 2

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If the first bit detected after the falling edge of RxD pin, is not 0 , then this indicates an invalid start bit, and the reception is immediately aborted. The serial port again looks for a falling edge in the RxD line. If a valid start bit is detected, then the rest of the bits are also detected and shifted into the SBUF. After shifting in 9 data bits, there is one more shift to do, after which the SBUF and RB8 are loaded and RI is set. However certain conditions must be met before the loading and setting of RI can be done.

1. RI must be 0 and
2. Either $\mathrm{SM} 2=0$, or the received stop bit $=1$.

If these conditions are met, then the stop bit goes to RB8, the 8 data bits go into SBUF and RI is set. Otherwise the received frame may be lost. After the middle of the stop bit, the receiver goes back to looking for a 1-to-0 transition on the RxD pin.

### 16.4 MODE 3

This mode is similar to Mode 2 in all aspects, except that the baud rate is programmable. The user must first initialize the Serial related SFR SCON before any communication can take place. This involves selection of the Mode and baud rate. The Timer 1 should also be initialized if modes 1 and 3 are used. In all four modes, transmission is started by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI $=0$ and REN $=1$. This will generate a clock on the TxD pin and shift in 8 bits on the RxD pin. Reception is initiated in the other modes by the incoming start bit if REN $=1$. The external device will start the communication by transmitting the start bit.


Figure 16-4: Serial Port Mode 3

| SM0 | SM1 | MODE | TYPE | BAUD CLOCK | FRAME <br> SIZE | START <br> BIT | STOP <br> BIT | 9TH BIT <br> FUNCTION |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | Synch. | 4 or 12 TCLKS | 8 bits | No | No | None |
| 0 | 1 | 1 | Asynch. | Timer 1 | 10 bits | 1 | 1 | None |
| 1 | 0 | 2 | Asynch. | 32 or 64 TCLKS | 11 bits | 1 | 1 | 0,1 |
| 1 | 1 | 3 | Asynch. | Timer 1 | 11 bits | 1 | 1 | 0,1 |

Table 16-5: Serial Port Mode Summary Table

### 16.5 Framing Error Detection

A Frame Error occurs when a valid stop bit is not detected. This could indicate incorrect serial data communication. Typically the frame error is due to noise and contention on the serial communication line. The W79E825 series have the facility to detect such framing errors and set a flag which can be checked by software.

The Frame Error FE bit is located in SCON.7. This bit is normally used as SM0 in the standard 8051 family. However, in the W79E825 series it serves a dual function and is called SM0/FE. There are actually two separate flags, one for SMO and the other for FE. The flag that is actually accessed as SCON. 7 is determined by SMODO (PCON.6) bit. When SMODO is set to 1, then the FE flag is indicated in SMO/FE. When SMODO is set to 0 , then the SMO flag is indicated in SMO/FE.

The FE bit is set to 1 by hardware but must be cleared by software. Note that SMODO must be 1 while reading or writing to $F E$. If $F E$ is set, then any following frames received without any error will not clear the FE flag. The clearing has to be done by software.

### 16.6 Multiprocessor Communications

Multiprocessor communications makes use of the 9th data bit in modes 2 and 3. In the W79E825 series, the RI flag is set only if the received byte corresponds to the Given or Broadcast address. This hardware feature eliminates the software overhead required in checking every received address, and greatly simplifies the software programmer task.

In the multiprocessor communication mode, the address bytes are distinguished from the data bytes by transmitting the address with the 9th bit set high. When the master processor wants to transmit a block of data to one of the slaves, it first sends out the address of the targeted slave (or slaves). All the slave processors should have their SM2 bit set high when waiting for an address byte. This ensures that they will be interrupted only by the reception of an address byte. The Automatic address recognition feature ensures that only the addressed slave will be interrupted. The address comparison is done in hardware not software.

The addressed slave clears the SM2 bit, thereby clearing the way to receive data bytes. With $\mathrm{SM} 2=0$, the slave will be interrupted on the reception of every single complete frame of data. The unaddressed slaves will be unaffected, as they will be still waiting for their address. In Mode 1, the 9th bit is the stop bit, which is 1 in case of a valid frame. If SM2 is 1 , then RI is set only if a valid frame is received and the received byte matches the Given or Broadcast address.

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The Master processor can selectively communicate with groups of slaves by using the Given Address. All the slaves can be addressed together using the Broadcast Address. The addresses for each slave are defined by the SADDR and SADEN SFRs. The slave address is an 8 -bit value specified in the SADDR SFR. The SADEN SFR is actually a mask for the byte value in SADDR. If a bit position in SADEN is 0 , then the corresponding bit position in SADDR is don't care. Only those bit positions in SADDR whose corresponding bits in SADEN are 1 are used to obtain the Given Address. This gives the user flexibility to address multiple slaves without changing the slave address in SADDR.

The following example shows how the user can define the Given Address to address different slaves. Slave 1:

SADDR1010 0100
SADEN 11111010
Given 1010 0x0x
Slave 2:
SADDR1010 0111
SADEN 11111001
Given $10100 x \times 1$

The Given address for slave 1 and 2 differ in the LSB. For slave 1, it is a don't care, while for slave 2 it is 1 . Thus to communicate only with slave 1, the master must send an address with LSB $=0$ (1010 0000 ). Similarly the bit 1 position is 0 for slave 1 and don't care for slave 2 . Hence to communicate only with slave 2 the master has to transmit an address with bit $1=1$ (1010 0011). If the master wishes to communicate with both slaves simultaneously, then the address must have bit $0=1$ and bit $1=0$. The bit 3 position is don't care for both the slaves. This allows two different addresses to select both slaves (1010 0001 and 1010 0101).

The master can communicate with all the slaves simultaneously with the Broadcast Address. This address is formed from the logical OR of the SADDR and SADEN SFRs. The zeros in the result are defined as don't cares. In most cases the Broadcast Address is FFh. In the previous case, the Broadcast Address is (1111111x) for slave 1 and (11111111) for slave 2.

The SADDR and SADEN SFRs are located at address A9h and B9h respectively. On reset, these two SFRs are initialized to 00h. This results in Given Address and Broadcast Address being set as xxxx xxxx (i.e. all bits don't care). This effectively removes the multiprocessor communications feature, since any selectivity is disabled.

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## 17 TIME ACCESS PROCTECTION

The W79E825 series have a new feature, like the Watchdog Timer which is a crucial to proper operation of the system. If left unprotected, errant code may write to the Watchdog control bits resulting in incorrect operation and loss of control. In order to prevent this, the W79E825 series have a protection scheme which controls the write access to critical bits. This protection scheme is done using a timed access.

In this method, the bits which are to be protected have a timed write enable window. A write is successful only if this window is active, otherwise the write will be discarded. This write enable window is open for 3 machine cycles if certain conditions are met. After 3 machine cycles, this window automatically closes. The window is opened by writing AAh and immediately 55h to the Timed Access (TA) SFR. This SFR is located at address C7h. The suggested code for opening the timed access window is
TA REG 0C7h
MOV TA, \#OAAh
MOV TA, \#055h
;Define new register TA, located at 0C7h

When the software writes AAh to the TA SFR, a counter is started. This counter waits for 3 machine cycles looking for a write of 55h to TA. If the second write (55h) occurs within 3 machine cycles of the first write (AAh), then the timed access window is opened. It remains open for 3 machine cycles, during which the user may write to the protected bits. Once the window closes the procedure must be repeated to access the other protected bits.

Examples of Timed Assessing are shown below.
Example 1: Valid access

| MOV | TA, \#0AAh |
| :--- | :--- |
| MOV | TA, \#055h |
| MOV | WDCON, \#00h |

$; 3 \mathrm{M} / \mathrm{C}$ Note: $\mathrm{M} / \mathrm{C}=$ Machine Cycles
$; 3 \mathrm{M} / \mathrm{C}$
$; 3 \mathrm{M} / \mathrm{C}$

Example 2: Valid access
MOV
TA, \#OAAh
;3 M/C
MOV TA,\#055h ;3 M/C
NOP
;1 M/C
SETB EWRST
;2 M/C
Example 3: Valid access

| MOV | TA, \#OAAh | $; 3 \mathrm{M} / \mathrm{C}$ |
| :--- | :--- | :--- |
| MOV | TA, \#055h | $; 3 \mathrm{M} / \mathrm{C}$ |
| ORL | WDCON, \#00000010B | $; 3 \mathrm{M} / \mathrm{C}$ |

Example 4: Invalid access

| MOV | TA, \#OAAh | $; 3 \mathrm{M} / \mathrm{C}$ |
| :--- | :--- | :---: |
| MOV | TA, \#055h | $; 3 \mathrm{M} / \mathrm{C}$ |
| NOP |  | $; 1 \mathrm{M} / \mathrm{C}$ |
| NOP |  | $; 1 \mathrm{M} / \mathrm{C}$ |
| CLR | EWT | $; 2 \mathrm{M} / \mathrm{C}$ |

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Example 5: Invalid Access

| MOV | TA, \#OAAh | $; 3 \mathrm{M} / \mathrm{C}$ |
| :--- | :--- | :--- |
| NOP |  | $; 1 \mathrm{M} / \mathrm{C}$ |
| MOV | TA, \#055h | $; 3 \mathrm{M} / \mathrm{C}$ |
| SETB | EWT | $; 2 \mathrm{M} / \mathrm{C}$ |

In the first three examples, the writing to the protected bits is done before the 3 machine cycles window closes. In Example 4, however, the writing to the protected bit occurs after the window has closed, and so there is effectively no change in the status of the protected bit. In Example 5, the second write to TA occurs 4 machine cycles after the first write, therefore the timed access window is not opened at all, and the write to the protected bit fails.

## 18 KEYBOARD INTERRUPT (KBI)

The W79E825 series are provided 8 keyboard interrupt function to detect keypad status which key is acted, and allow a single interrupt to be generated when any key is pressed on a keyboard or keypad connected to specific pins of the W79E825 series, as shown below Figure. This interrupt may be used to wake up the CPU from Idle or Power Down modes, after chip is in Power Down or Idle Mode.

Keyboard function is supported through by Port 0. It can allow any or all pins of Port 0 to be enabled to cause this interrupt. Port pins are enabled by the setting of bits of KBIO ~KBI7 in the KBI register, as shown below Figure. The Keyboard Interrupt Flag (KBF) in the AUXR1 register is set when any enabled pin is pulled low while the KBI interrupt function is active, and the low pulse must be more than 1 machine cycle, an interrupt will be generated if it has been enabled. The KBF bit set by hardware and must be cleared by software. In order to determine which key was pressed, the KBI will allow the interrupt service routine to poll port 0.


Figure 18-1: Keyboard Interrupt

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## 19 ANALOG COMPARATORS

The W79E825 series are provided two Comparators. Input and output options allow use of the comparators in a number of different Configurations. The Comparator output is a logical one when its positive input is greater than its negative input, otherwise the output is a zero. Each Comparator can be configured to cause to an interrupt when the output value change. The block diagram is as below.

Each Comparator has a control register (CMP1 and CMP2), Both Inputs are CINnA, CINnB, CMPREF and internal reference voltage, and outputs are CMP1 and CMP2 by setting OEn bit. After enable Comparators the Comparator need waited stable time to guarantee Comparator output. If programmer used internal reference voltage, it will be set OEn bit to "1". The value of internal reference voltage (Vref) is $1.19 \mathrm{~V}+/-10 \%$.


Figure 19-1: Analog Comparators

## 20 I/O PORT CONFIGURATION

The W79E825 series have three I/O ports, port 0, port 1 and port 2. All pins of I/O ports can be configured to one of four types by software except P1.5 is only input pin. When P1.5 is configured reset pin by RPD=0 in the CONFIG 1 register, the W79E825 series can support 15 pins by use Crystal. If used on-chip RC oscillator the P1.5 is configured input pin, the W79E825 series can be supported up to 18 pins. The I/O ports configuration setting as below table.

| PXM1.Y | PXM2.Y | PORT INPUT/OUTPUT MODE |
| :---: | :---: | :---: |
| 0 | 0 | Quasi-bidirectional |
| 0 | 1 | Push-Pull |
| 1 | 0 | Input Only (High Impedance) |
| P2M1.PxS=0, TTL input |  |  |
| P2M1.PxS=1, Schmitt input |  |  |

Table 20-1: I/O port Configuration Table

All port pins can be determined to high or low after reset by configure PRHI bit in the CONFIG1 register. After reset, these pins are in quasi-bidirectional mode. The port pin of P1.5 only is a Schmitt trigger input.

Enabled toggle outputs from Timer 0 and Timer 1 by T0OE and T1OE on P2M1 register, the output frequency of Timer 0 or Timer 1 is by Timer overflow.

Each I/O port of the W79E825 series may be selected to use TTL level inputs or Schmitt inputs by $\mathrm{P}(\mathrm{n}) \mathrm{S}$ bit on P2M1 register, where n is 0,1 or 2 . When $\mathrm{P}(\mathrm{n}) \mathrm{S}$ is set to 1 , Ports are selected Schmitt trigger inputs on Port(n). The P2.0 (XTAL2) can be configured clock output when used on-chip RC or external Oscillator is clock source, and the frequency of clock output is divided by 4 on on-chip RC clock or external Oscillator.

### 20.1 Quasi-Bidirectional Output Configuration

After chip was power on or reset, the all ports output are this mode, and output is common with the 8051. This mode can be used as both an input and output without the need to reconfigure the port.

When the pin is pulled low, it is driven strongly and able to sink a fairly large current. These features are somewhat similar to an open drain output except that there are three pull-up transistors in the quasi-bidirectional output that serve different purposes.

This mode has three pull-up resisters that are "strong" pull-up, "weak" pull-up and "very weak" pull-up. The "strong" pull-up is used fast transition from logic "0" change to logic " 1 ", and it is fast latch and transition. When port pins is occur from logic " 0 " to logic " 1 ", the strong pull-up will quickly turn on two CPU clocks to pull high then turn off.

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The "weak" pull-up is turned on when the input port pin is logic " 1 " level or itself is logic " 1 ", and it provides the most source current for a quasi-bidirectional pin that output is " 1 " or port latch is logic " 0 ".

The "very weak" pull-up is turned on when the port latch is logic " 1 ". If port latch is logic " 0 ", it will be turned off. The very weak pull-up is support a very small current that will pull the pin high if it is left floating. And the quasi-bidirectional port configuration is shown as below figure.

If port pin is low, it can drives large sink current for output, and it is similar with push-pull and open drain on sink current output.


Figure 20-2: Quasi-Bidirectional Output

### 20.2 Open Drain Output Configuration

To configure this mode is turned off all pull-ups. If used similar as a logic output, the port must has an external pull-up resister. The open drain port configuration is shown as below.


Figure 20-3: Open Drain Output

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### 20.3 Push-Pull Output Configuration

The push-pull output mode has two strong pull-up and pull-down structure that support large source and sink current output. It removes "weak" pull-up and "very weak" pull-up resister and remain "strong pull-up resister on quasi-bidirectional output mode. The "strong" pull-up is always turns on when port latch is logic " 1 " to support source current. The push-pull port configuration is shown in below Figure.

The W79E825 series have three port pins that can't be configured. They are P1.2, P1.3, and P1.5. The port pins P1.2 and P1.3 are configured to open drain outputs. They may be used as inputs by writing ones to their respective port latches.


Figure 20-4: Push-Pull Output

### 20.4 Input Only Configuration

By configure this mode, the ports are only digital input and disable digital output. The W79E825 series can select input pin to Schmitt trigger or TTL level input by PxM1.y and PxM2.y registers.

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## 21 OSCILLATOR

The W79E825 series provides three oscillator input option. These are configured at CONFIG register (CONFIG1) that include On-Chip RC Oscillator Option, External Clock Input Option and Crystal Oscillator Input Option. The Crystal Oscillator Input frequency may be supported from 4 MHz to 20 MHz , and without capacitor or resistor.


Figure 21-1: Oscillator

### 21.1 On-Chip RC Oscillator Option

The On-Chip RC Oscillator is fixed at $6 \mathrm{MHz}+/-25 \%$ frequency to support clock source. When FOSC1, FOSC0 $=01 \mathrm{~b}$, the On-Chip RC Oscillator is enabled. A clock output on P2.0 (XTAL2) may be enabled when On-Chip RC oscillator is used.

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### 21.2 External Clock Input Option

The clock source pin (XTAL1) is from External Clock Input by FOSC1, FOSC0 $=11 \mathrm{H}$, and frequency range is form 0 Hz up to 20 MHz . A clock output on P2.0 (XTAL2) may be enabled when External Clock Input is used.
The W79E825 series supports a clock output function when either the on-chip RC oscillator or the external clock input options is selected. This allows external devices to synchronize to the W79E825 serial. When enabled, via the ENCLK bit in the P2M1 register, the clock output appears on the XTAL2/CLKOUT pin whenever the on-chip oscillator is running, including in Idle Mode. The frequency of the clock output is $1 / 4$ of the CPU clock rate. If the clock output is not needed in Idle Mode, it may be turned off prior to entering Idle mode, saving additional power. The clock output may also be enabled when the external clock input option is selected.

### 21.3 CPU Clock Rate select

The CPU clock of W79E825 series may be selected by the DIVM register. If DIVM $=00 \mathrm{H}$, the CPU clock is running at 4 CPU clock per machine cycle, and without any division from source clock (Fosc). When the DIVM register is set to $N$ value, the CPU clock is divided by 2(DVIM+1), so CPU clock frequency division is from 4 to 512. The user may use this feature to set CPU at a lower speed rate for reducing power consumption. This is very similar to the situation when CPU has entered Idle mode. In addition this frequency division function affect all peripheral timings as they are all sourcing from the CPU clock(Fcpu).

## 22 POWER MONITORING FUNCTION

## Power-On Detect and Brownout are two additional power monitoring functions implemented in W79E825 series to prevent incorrect operation during power up and power drop or loss.

### 22.1 Power On Detect

The Power-On Detect function is a designed to detect power up after power voltage reaches to a level where Brownout Detect can work. After power on detect, the POR (PCON.4) will be set to " 1 " to indicate an initial power up condition. The POR flag will be cleared by software.

### 22.2 Brownout Detect

The Brownout Detect function is detect power voltage is drops to brownout voltage level, and allows preventing some process work or indicate power warming. The W79E825 series have two brownout voltage levels to select by BOV (CONFIG1.4). If $\mathrm{BOV}=0$ that brownout voltage level is 3.8 V , If $\mathrm{BOV}=$ 1 that brownout voltage level is 2.5 V . When the Brownout voltage is drop to select level, the brownout detector will detect and keeps this active until VDD is returns to above brownout Detect voltage. The Brownout Detect block is as follow.


Figure 22-1: Brownout Detect Block

When Brownout Detect is enabled by BOD (AUXR1.6), the BOF (PCON.5) flag will be set and brownout reset will occur. If BOI (AUXR1.5) is set to " 1 ", the brownout detect will cause interrupt via the EA (IE.7) and EBO (IE.5) bits is set. BOF is cleared by software.
In order to guarantee a correct detection of Brownout, The VDD fall time must be slower than $50 \mathrm{mV} / \mathrm{us}$, and rise time is slower than $2 \mathrm{mV} / \mathrm{us}$ to ensure a proper reset.

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## 23 PULSE-WIDTH-MODULATED (PWM) OUTPUTS

The W79E825 series have four Pulse Width Modulated (PWM) channels, and the PWM outputs are PWM0 (P0.1), PWM1 (P1.6), PWM2 (P1.7) and PWM3 (P0.0). The initial PWM outputs level correspondingly depend on the PRHI level set prior to the chip reset. When PRHI set to high, PWM output will initialize to high after chip reset; if PRHI set to low, PWM output will be initialize to low after chip reset.

The W79E825 series support 10-bits down counter with cpu clock as its input. The PWM counter clock, has the same frequency as the clock source $\mathrm{F}_{\mathrm{CPU}}=\mathrm{F}_{\mathrm{Osc}}$. When the counter reaches underflow it will automatic reloaded from counter register. The PWM frequency is given by: $f_{P W M}=F_{C P U} /$ (PWMP+1), where PWMP is 10-bits register of PWMPH.1, PWMPH. 0 and PWMPL.7~PWMPL. 0.

The counter register will be loaded with the PWMP register value when PWMRUN, load and CF are equal to 1 ; the load bit will be automatically cleared to zero on the next clock cycle, and at the same time the counter register value will be loaded to the 10 bits down counter. CF flag is 10-bits down counter reaches underflow, the CF flag will be cleared by software.

The pulse width of each PWM output is determined by the Compare registers of PWMOL through PWM3L and PWMOH through PWM3H. When PWM compare register is greater than 10-bits counter register, the PWM output is low. Load bit has to be set to 1 for alteration of PWMn width. After the new values are written to the PWMn registers, and if load bit is set to 1 , the new PWMn values will be loaded to the PWMn registers upon the next underflow. The PWM output high pulses width is given by:
$\mathrm{t}_{\mathrm{HI}}=(\mathrm{PWMP}-\mathrm{PWMn}+1)$. Notice, if compare register is set to 000 H , the PWMn output will stay at high, and if compare register is set to 3FFH, the PWMn output will stuck at low until there is a change in the compare register.

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Figure 23-1: W79E825/824/823/822 PWM Block Diagram

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The W79E825 series devices support brake function which can be activated by software or external pin (P0.2). The Brake function is controlled by the PWMCON2 register. The setting and details description of software brake and external pin brake can be found at the brake condition table at the SFR section.
As for external brake, the user program can poll the brake flag (BKF) or enable PWM's brake interrupt to determine when the external Brake Pin is asserted and causes a brake to occur. The brake pin (P0.2) can be set to trigger the brake function by either low or high level, by clearing or setting the PWMCON2.6 (BKPS) bit respectively. The details description of varies brake functions can be found in the brake condition table.

Since the Brake Pin being asserted will automatically clear the Run bit of PWMCON1.7 and BKF (PWMCON3.0) flag will be set, the user program can poll this bit or enable PWM's brake interrupt to determine when the Brake Pin causes a brake to occur. The other method for detecting a brake caused by the Brake Pin would be to tie the Brake Pin to one of the external interrupt pins. This latter approach is needed if the Brake signal is of insufficient length to ensure that it can be captured by a polling routine. When, after being asserted, the condition causing the brake is removed, the PWM outputs go to whatever state that had immediately prior to the brake. This means that in order to go from brake being asserted to having the PWM run without going through an indeterminate state, care must be taken. If the Brake Pin causes brake to be asserted, the following prototype code will allow the PWM to go from brake and then run smoothly after brake is released.


Figure 23-2: PWM Brake Function

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## 24 ANALOG-TO-DIGITAL CONVERTER

The ADC contains a DAC which converts the contents of a successive approximation register to a voltage (VDAC) which is compared to the analog input voltage (Vin). The output of the comparator is fed to the successive approximation control logic which controls the successive approximation register. A conversion is initiated by setting ADCS in the ADCCON register. There are two triggering methods by ADC to start conversion, either by purely software start or external pin STADC triggering.
The software start mode is used to trigger ADC conversion regardless of ADCCON. 5 (ADCEX) bit is set or cleared. A conversion will start simply by setting the ADCCON. 3 (ADCS) bit. As for the external STADC pin triggering mode, ADCCON. 5 (ADCEX) bit has to be set and a rise edge pulse has to apply to STADC pin to trigger the ADC conversion. For the rising edge triggering method, a minimum of at least 2 machine cycles symmetrical pulse is required.

The low-to-high transition of STADC is recognized at the end of a machine cycle, and the conversion commences at the beginning of the next cycle. When a conversion is initiated by software, the conversion starts at the beginning of the machine cycle which follows the instruction that sets ADCS. ADCS is actually implemented with tpw flip-flops: a command flip-flop which is affected by set operations, and a status flag which is accessed during read operations.
The next two machine cycles are used to initiate the converter. At the end of the first cycle, the ADCS status flag is set end a value of " 1 " will be returned if the ADCS flag is read while the conversion is in progress. Sampling of the analog input commences at the end of the second cycle.

During the next eight machine cycles, the voltage at the previously selected pin of one of analog input pin is sampled, and this input voltage should be stable in order to obtain a useful sample. In any event, the input voltage slew rate must be less than $10 \mathrm{~V} / \mathrm{ms}$ in order to prevent an undefined result.

The successive approximation control logic first sets the most significant bit and clears all other bits in the successive approximation register (10 0000 0000b). The output of the DAC ( $50 \%$ full scale) is compared to the input voltage Vin. If the input voltage is greater than VDAC, then the bit remains set; otherwise if is cleared.

The successive approximation control logic now sets the next most significant bit (11 00000000 b or 010000 0000b, depending on the previous result), and the VDAC is compared to Vin again. If the input voltage is greater then VDAC, then the bit remains set; otherwise it is cleared. This process is repeated until all ten bits have been tested, at which stage the result of the conversion is held in the successive approximation register. The conversion takes four machine cycles per bit.

The end of the 10-bit conversion is flagged by control bit ADCCON. 4 (ADCI). The upper 8 bits of the result are held in special function register ADCH , and the two remaining bits are held in ADCCON. 7 (ADC.1) and ADCCON. 6 (ADC.0). The user may ignore the two least significant bits in ADCCON and use the ADC as an 8-bit converter ( 8 upper bits in ADCH). In any event, the total actual conversion time is 52 machine cycles. ADCI will be set and the ADCS status flag will be reset 52 cycles after the ADCS is set. Control bits ADCCON. 0 and ADCCON. 1 are used to control an analog multiplexer which selects one of 4 analog channels. An ADC conversion in progress is unaffected by an external or software ADC start. The result of a completed conversion remains unaffected provided ADCI = logic 1; a new ADC conversion already in progress is aborted when the idle or power down mode is entered. The result of a completed conversion ( $\mathrm{ADCI}=\operatorname{logic} 1$ ) remains unaffected when entering the idle mode.


Figure 24-1: Successive Approximation ADC

### 24.1 ADC Resolution and Analog Supply:

The ADC circuit has its own supply pins (AVDD and AVSS) and one pins (Vref+) connected to each end of the DAC's resistance-ladder that the AVDD and Vref+ are connected to VDD and AVSS is connected to VSS. The ladder has 1023 equally spaced taps, separated by a resistance of " R ". The first tap is located $0.5 \times \mathrm{R}$ above AVSS, and the last tap is located $0.5 \times \mathrm{R}$ below Vref+. This gives a total ladder resistance of $1024 \times$ R. This structure ensures that the DAC is monotonic and results in a symmetrical quantization error.

For input voltages between VSS and [(Vref+) + $1 / 2$ LSB], the 10 -bit result of an A/D conversion will be $0000000000 \mathrm{~B}=000 \mathrm{H}$. For input voltages between [(Vref+) - 3/2 LSB] and Vref+, the result of a conversion will be 1111111111B $=3 F F H$. Vref+ and AVSS may be between AVDD +0.2 V and VSS 0.2 V . Vref+ should be positive with respect to VSS, and the input voltage (Vin) should be between Vref+ and VSS.

The result can always be calculated from the following formula:
Result $=1024 \times \frac{\text { Vin }}{\text { Vref }+}$ or Result $=1024 \times \frac{\text { Vin }}{\text { VDD }}$

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Figure 24-2: The ADC Block Diagram

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## 25 I2C SERIAL CONTROL

The I2C bus uses two wires (SDA and SCL) to transfer information between devices connected to the bus. The main features of the bus are:

- Bidirectional data transfer between masters and slaves
- Multimaster bus (no central master)
- Arbitration between simultaneously transmitting masters without corruption of serial data on the bus
- Serial clock synchronization allows devices with different bit rates to communicate via one serial bus
- Serial clock synchronization can be used as a handshake mechanism to suspend and resume serial transfer
- The I2C bus may be used for test and diagnostic purposes


Figure 25-1: I2C Bus Timing

The device's on-chip I2C logic provides the serial interface that meets the I2C bus standard mode specification. The I2C logic handles bytes transfer autonomously. It also keeps track of serial transfers, and a status register (I2STATUS) reflects the status of the I2C bus.
The I2C port, SCL and SDA are at P1.2 and P1.3. When the I/O pins are used as I2C port, user must set the pins to logic high in advance. When I2C port is enabled by setting ENS to high, the internal states will be controlled by I2CON and I2C logic hardware. Once a new status code is generated and stored in I2STATUS, the I2C interrupt flag (SI) will be set automatically. If both EA and EI2C are also in logic high, the I2C interrupt is requested. The 5 most significant bits of I2STATUS stores the internal state code, the lowest 3 bits are always zero and the content keeps stable until SI is cleared by software.

### 25.1 SIO Port

The SIO port is a serial I/O port, which supports all transfer modes from and to the I2C bus. The SIO port handles byte transfers autonomously. To enable this port, the bit ENS1 in I2CON should be set to '1'. The CPU interfaces to the SIO port through the following six special function registers: I2CON (control register, COH), I2STATUS (status register, BDH), I2DAT (data register, $B C H$ ), I2ADDR (address registers, C1H), I2CLK (clock rate register BEH) and I2TIMER (Timer counter register, BFH). The SIO H/W interfaces to the I2C bus via two pins: SDA (P1.3, serial data line) and SCL (P1.2, serial clock line). Pull up resistor is needed for Pin P1.2 and P1.3 for I2C operation as these are 2 open drain pins.

### 25.2 The I2C Control Registers:

The I2C has 1 control register (I2CON) to control the transmit/receive flow, 1 data register (I2DAT) to buffer the Tx/Rx data, 1 status register (I2STATUS) to catch the state of Tx/Rx, recognizable slave address register for slave mode use and 1 clock rate control block for master mode to generate the variable baud rate.

### 25.2.1 The Address Registers, I2ADDR

I2C port is equipped with one slave address register. The contents of the register are irrelevant when I2C is in master mode. In the slave mode, the seven most significant bits must be loaded with the MCU's own slave address. The I2C hardware will react if the contents of I2ADDR are matched with the received slave address.
The I2C ports support the "General Call" function. If the GC bit is set the I2C port1 hardware will respond to General Call address $(00 \mathrm{H})$. Clear GC bit to disable general call function.
When GC bit is set, the I2C is in Slave mode, it can be received the general call address by 00 H after Master send general call address to I2C bus, then it will follow status of GC mode. If it is in Master mode, the AA bit must be cleared when it will send general call address of 00 H to I2C bus.

### 25.2.2 The Data Register, I2DAT

This register contains a byte of serial data to be transmitted or a byte which has just been received. The CPU can read from or write to this 8 -bit directly addressable SFR while it is not in the process of shifting a byte. This occurs when SIO is in a defined state and the serial interrupt flag (SI) is set. Data in I2DAT remains stable as long as SI bit is set. While data is being shifted out, data on the bus is simultaneously being shifted in; I2DAT always contains the last data byte present on the bus. Thus, in the event of arbitration lost, the transition from master transmitter to slave receiver is made with the correct data in I2DAT.

I2DAT and the acknowledge bit form a 9-bit shift register, the acknowledge bit is controlled by the SIO hardware and cannot be accessed by the CPU. Serial data is shifted through the acknowledge bit into I2DAT on the rising edges of serial clock pulses on the SCL line. When a byte has been shifted into I2DAT, the serial data is available in I2DAT, and the acknowledge bit (ACK or NACK) is returned by the control logic during the ninth clock pulse. Serial data is shifted out from I2DAT on the falling edges of SCL clock pulses, and is shifted into I2DAT on the rising edges of SCL clock pulses.

I2C Data Register:


### 25.2.3 The Control Register, I2CON

The CPU can read from and write to this 8 -bit, directly addressable SFR. Two bits are affected by hardware: the SI bit is set when the I2C hardware requests a serial interrupt, and the STO bit is cleared when a STOP condition is present on the bus. The STO bit is also cleared when ENS = " 0 ".
ENSI Set to enable I2C serial function block. When ENS=1 the I2C serial function enables. The port latches of SDA1 and SCL1 must be set to logic high.
STA I2C START Flag. Setting STA to logic 1 to enter master mode, the I2C hardware sends a START or repeat START condition to bus when the bus is free.
STO I2C STOP Flag. In master mode, setting STO to transmit a STOP condition to bus then I2C hardware will check the bus condition if a STOP condition is detected this flag will be cleared by hardware automatically. In a slave mode, setting STO resets I2C hardware to the defined "not addressed" slave mode. This means it is NO LONGER in the slave receiver mode to receive data from the master transmit device.
SI I2C Port 1 Interrupt Flag. When a new SIO state is present in the S1STA register, the SI flag is set by hardware, and if the EA and EI2C1 bits are both set, the I2C1 interrupt is requested. SI must be cleared by software.
AA Assert Acknowledge control bit. When AA=1 prior to address or data received, an acknowledged (low level to SDA) will be returned during the acknowledge clock pulse on the SCL line when 1.) A slave is acknowledging the address sent from master, 2.) The receiver devices are acknowledging the data sent by transmitter. When $A A=0$ prior to address or data received, a Not acknowledged (high level to SDA) will be returned during the acknowledge clock pulse on the SCL line.

### 25.2.4 The Status Register, I2STATUS

I2STATUS is an 8 -bit read-only register. The three least significant bits are always 0 . The five most significant bits contain the status code. There are 23 possible status codes. When I2STATUS contains F8H, no serial interrupt is requested. All other I2STATUS values correspond to defined SIO states. When each of these states is entered, a status interrupt is requested $(\mathrm{SI}=1)$. A valid status code is present in I2STATUS one machine cycle after SI is set by hardware and is still present one machine cycle after SI has been reset by software.

### 25.2.5 The I2C Clock Baud Rate Bits, I2CLK

The data baud rate of I2C is determines by I2CLK register when SIO is in a master mode. It is not important when SIO is in a slave mode. In the slave modes, SIO will automatically synchronize with any clock frequency up to 400 KHz from master I2C device.

The data baud rate of I2C setting is Data Baud Rate of I2C = Fcpu / (I2CLK+1). The Fcpu=Fosc/4. If Fosc $=16 \mathrm{MHz}$, the $\operatorname{I2CLK}=40(28 \mathrm{H})$, so data baud rate of $\mathrm{I} 2 \mathrm{C}=16 \mathrm{MHz} /(4 \mathrm{X}(40+1))=$ $97.56 \mathrm{Kbits} / \mathrm{sec}$. The block diagram is as below figure.


Figure 25-2: I2C Timer Count Block Diagram

### 25.3 Modes of Operation

The on-chip I2C ports support five operation modes, Master transmitter, Master receiver, Slave transmitter, Slave receiver, and GC call.
In a given application, I2C port may operate as a master or as a slave. In the slave mode, the I2C port hardware looks for its own slave address and the general call address. If one of these addresses is detected, and if the slave is willing to receive or transmit data from/to master(by setting the AA bit), acknowledge pulse will be transmitted out on the 9th clock, hence an interrupt is requested on both master and slave devices if interrupt is enabled. When the microcontroller wishes to become the bus master, the hardware waits until the bus is free before the master mode is entered so that a possible slave action is not interrupted. If bus arbitration is lost in the master mode, I2C port switches to the slave mode immediately and can detect its own slave address in the same serial transfer.

### 25.3.1 Master Transmitter Mode

Serial data output through SDA while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device ( 7 bits) and the data direction bit. In this case the data direction bit (R/W) will be logic 0 , and it is represented by " $W$ " in the flow diagrams. Thus the first byte transmitted is SLA+W. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an acknowledge bit is received. START and STOP conditions are output to indicate the beginning and the end of a serial transfer.

### 25.3.2 Master Receiver Mode

In this case the data direction bit (R/W) will be logic 1, and it is represented by " $R$ " in the flow diagrams. Thus the first byte transmitted is SLA+R. Serial data is received via SDA while SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are output to indicate the beginning and end of a serial transfer.

### 25.3.3 Slave Receiver Mode

Serial data and the serial clock are received through SDA and SCL. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are recognized as the beginning and end of a serial transfer. Address recognition is performed by hardware after reception of the slave address and direction bit.

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### 25.3.4 Slave Transmitter Mode

The first byte is received and handled as in the slave receiver mode. However, in this mode, the direction bit will indicate that the transfer direction is reversed. Serial data is transmitted via SDA while the serial clock is input through SCL. START and STOP conditions are recognized as the beginning and end of a serial transfer.

### 25.4 Data Transfer Flow in Five Operating Modes

The five operating modes are: Master/Transmitter, Master/Receiver, Slave/Transmitter, Slave/Receiver and GC Call. Bits STA, STO and AA in I2CON register will determine the next state of the SIO hardware after SI flag is cleared. Upon complexion of the new action, a new status code will be updated and the SI flag will be set. If the I2C interrupt control bits (EA and EI2) are enable, appropriate action or software branch of the new status code can be performed in the Interrupt service routine.
Data transfers in each mode are shown in the following figures.
*** Legend for the following five figures:


Figure 25-3: Legen for the following four figures

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Figure 25-4: Master Transmitter Mode

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Figure 25-5: Master Receiver Mode

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Figure 25-6: Slave Transmitter Mode

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Figure 25-7:Slave Receiver Mode


Figure 25-8:GC Mode

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## 26 ICP(IN-CIRCUIT PROGRAM) FLASH PROGRAM

The contexts of flash in W79E825 series are empty by default. User must program the flash EPROM by external Writer device or by ICP (In-Circuit Program) tool.

In the ICP tool, the user must take note of ICP's program pins used in system board. In some application circuits, the pins are located at P1.5, P0.4 and P0.5, as below figure. During ICP programming, P1.5 must be set to high voltage ( $\sim 10.5 \mathrm{~V}$ ), and keeping this voltage to update code, data and/or configure CONFIG bits. After programming completion, the high voltage of P1.5 should be released. So, it is highly recommended user power off then power on after ICP programming has completed on the system board.

Upon entry into ICP program mode, all pin will be set to quasi-bidirectional mode, and output to level "1".

The W79E825 series support programming of Flash EPROM ( $\mathbf{1 6 K / 8 K / 4 K / 2 K}$ bytes AP Flash EPROM) and NVM data memory ( $\mathbf{2 5 6 / 1 2 8}$ bytes). User has the option to program the AP flash and NVM either individually or both.


Figure 26-1: Application Circuit of ICP

Note: 1. When using ICP to upgrade code, the P1.5, P0.4 and P0.5 must be taken within design system board.
2. After program finished by ICP, to suggest system power must power off and remove ICP connector then power on.
3. It is recommended that user performs erase function and programming configure bits continuously without any interruption.

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## 27 CONFIG BITS

The W79E825 series have two CONFIG bits (CONFIG1, CONFIG2) that must be define at power up and can not be set after the program start of execution. Those features are configured through the use of two flash EPROM bytes, and the flash EPROM can be programmed and verified repeatedly. Until the code inside the Flash EPROM is confirmed OK, the code can be protected. The protection of flash EPROM (CONFIG2) and those operations on it are described below. The data of these bytes may be read by the MOVC instruction at the addresses.

### 27.1 CONFIG1

## CONFIG 1 :



RPD: Reset Pin Disable Bit.
PRHI: Port Reset High or Low Bit.
BOV: Brownout voltage select Bit.
Fosc1: CPU oscillator type select Bit 1.
Fosc0: CPU oscillator type select Bit 0 .

CONFIG Register 1

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 7 | - | Reserved. |
| 6 | RPD | Reset Pin Disable bit: <br> 0: Enable Reset function of Pin 1.5. <br> 1: Disable Reset function of Pin 1.5, and it to be used as an input port pin. |
| 5 | PRHI | Port Reset High or Low bit: <br> 0: Port reset to low state. <br> 1: Port reset to high state. |
| 4 | BOV | Brownout Voltage Select bit: <br> 0: Brownout detect voltage is 3.8V. <br> 1: Brownout detect voltage is 2.5 V. |

Continued

| BIT | NAME | FUNCTION |
| :---: | :---: | :--- |
| 3 | - | Reserved. |
| 2 | - | Reserved. |
| 1 | Fosc1 | CPU Oscillator Type Select bit 1 |
| 0 | Fosc0 | CPU Oscillator Type Select bit 0 |

Oscillator Configuration bits:

| FOSC1 | FOSC0 | OSC SOURCE |
| :---: | :---: | :---: |
| 0 | 0 | $4 \mathrm{MHz} \sim 20 \mathrm{MHz}$ crystal |
| 0 | 1 | Internal RC Oscillator |
| 1 | 0 | Reserved |
| 1 | 1 | External Oscillator in XTAL1 |

### 27.2 CONFIG2

CONFIG 2 :


C7: 16K/8K/4K/2K Flash EPROM Code Lock Bit.
C6: 256/128 byte Data Lock Bit.

## CONFIG Register 2

## C7: 16K/8K/4K/2K Flash EPROM Lock bit

This bit is used to protect the customer's program code. It may be set after the programmer finishes the programming and verifies sequence. Once this bit is set to logic 0, both the Flash EPROM data and CONFIG Registers can not be accessed again.
C6: 256/128 byte Data Flash EPROM Lock bit
This bit is used to protect the customer's data code. It may be set after the programmer finishes the programming and verifies sequence. Once this bit is set to logic 0 , both the data Flash EPROM and CONFIG Registers can not be accessed again.

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| BIT 7 | BIT 6 | FUNCTION DESCRIPTION |
| :---: | :---: | :--- |
| 1 | 1 | Both security of 16KB/8KB/4KB/2KB program code and 256/128 Bytes data <br> area are unlocked. They can be erased, programmed or read by Writer or ICP. |
| 0 | 1 | The 16KB/8KB/4KB/2KB program code area is locked. It can't be read by Writer <br> or ICP. |
| 1 | 0 | Don't support (Invalid). |
| 0 | 0 | Both security of 16KB/8KB/4KB/2KB program code and 256/128 Bytes data <br> area are locked. They can't be read by Writer or ICP. |

## 28 ELECTRICAL CHARACTERISTICS

### 28.1 Absolute Maximum Ratings

| SYMBOL | PARAMETER | MIN | MAX | UNIT |
| :--- | :---: | :---: | :---: | :---: |
| DC Power Supply | VDD-VSS | -0.3 | +7.0 | V |
| Input Voltage | VIN | VSS-0.3 | VDD+0.3 | V |
| Operating Temperature | TA | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | Tst | -55 | +150 | ${ }^{\circ} \mathrm{C}$ |
| Maximum Current into VDD |  | - | 120 | mA |
| Maximum Current out of $\mathrm{V}_{\mathrm{Ss}}$ |  |  | 120 | mA |
| Maximum Current suck by I <br> I/O pin |  |  | 25 | mA |
| Maximum Current sourced by <br> a I/O pin |  |  | 75 | mA |
| Maximum Current suck by <br> total I/O pins |  |  | mA |  |
| Maximum Current sourced by <br> total I/O pins |  |  | mA |  |

Note: Exposure to conditions beyond those listed under absolute maximum ratings may adversely affects the lift and reliability

### 28.2 DC ELECTRICAL CHARACTERISTICS

( $\mathrm{V}_{\mathrm{ss}}=0 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40 \sim 85^{\circ} \mathrm{C}$, unless otherwise specified; Typical value is test at $\mathrm{TA}=25^{\circ} \mathrm{C}$ )

| PARAMETER | SYMBOL | SPECIFICATION |  |  |  | TEST CONDITIONS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | TYP. | MAX. | UNIT |  |
| Operating Voltage | $V_{\text {DD }}$ | 2.7 |  | 5.5 | V | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V} \sim 5.5 \mathrm{~V} @ \\ & 20 \mathrm{MHz} \\ & \mathrm{~V}_{\mathrm{DD}}=2.7 \mathrm{~V} \sim 5.5 \mathrm{~V} @ \\ & 12 \mathrm{MHz} \end{aligned}$ |
| Operating Current | $\mathrm{I}_{\mathrm{DD} 1}$ |  | 18 | 25 | mA | No load, $/ \mathrm{RST}=\mathrm{V}_{\mathrm{SS}}, \mathrm{V}_{\mathrm{DD}}=$ 5.0V @ 20MHz |
|  |  |  | 6 | 8 | mA | No load, /RST = V ${ }_{\text {Ss }}$, $V_{D D}=3.0 \mathrm{~V} @ 12 \mathrm{MHz}$ |
|  | $\mathrm{I}_{\mathrm{DD} 2}$ |  | 23 | 29 | mA | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} @ 20 \mathrm{MHz}$, No load, /RST = V ${ }_{\text {DD }}$, Run NOP |
|  |  |  | 6.5 | 9 | mA | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V} @ 12 \mathrm{MHz}$, No load, /RST = VDD, Run NOP |
| Idle Current | $l_{\text {IDLE }}$ |  | 11.5 | 15 | mA | No load, $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}$ <br> @ 20MHz |
|  |  |  | 5 | 6.5 | mA | No load, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ <br> @ 12MHz |
| Power Down Current | IpwdN |  | 1 | 10 | $\mu \mathrm{A}$ | No load, $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}$ <br> @ Disable BOV function |
|  |  |  | 1 | 10 | uA | No load, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ <br> @ Disable BOV function |
| Input Current P0, P1, P2 | $\mathrm{I}_{\text {IN } 1}$ | -50 | - | +15 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}, 0<\mathrm{V}_{\text {IN }}<\mathrm{V}_{\text {DD }}$ |
| Input Current /RST ${ }^{\left[{ }^{*}\right]}$ | IIN2 | -55 | -45 | -30 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=0.45 \mathrm{~V}$ |
| Input Leakage Current P0, P1, P2 (Open Drain) | ILK | -10 | - | +10 | $\mu \mathrm{A}$ | $V_{\text {DD }}=5.5 \mathrm{~V}, 0<\mathrm{V}_{\text {IN }}<\mathrm{V}_{\mathrm{DD}}$ |
| Logic 1 to 0 Transition Current P0, P1, P2 | $\mathrm{I}_{\text {TL }}{ }^{[* 3]}$ | -500 | - | -200 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}, \mathrm{VIN}<2.0 \mathrm{~V}$ |
| Input Low Voltage P0, P1, P2 <br> (TTL input) | $\mathrm{V}_{\text {ILI }}$ | 0 | - | 0.8 | V | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}$ |
|  |  | 0 | - | 0.6 | V | $V_{D D}=3.0 \mathrm{~V}$ |

DC ELECTRICAL CHARACTERISTICS, continued

| PARAMETER | SYMBOL | SPECIFICATION |  |  |  | TEST CONDITIONS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | TYP. | MAX. | UNIT |  |
| Input Low Voltage XTAL1 ${ }^{\left[{ }^{* 2}\right]}$ | $V_{\text {IL3 }}$ | 0 | - | 0.8 | V | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}$ |
|  |  | 0 | - | 0.4 | V | $V_{D D}=3.0 \mathrm{~V}$ |
| Input High Voltage XTAL1 ${ }^{\left[{ }^{* 2}\right]}$ | $\mathrm{V}_{\text {IH3 }}$ | 3.5 | - | $\mathrm{V}_{\mathrm{DD}}+0.2$ | V | $V_{D D}=5.5 \mathrm{~V}$ |
|  |  | 2.4 | - | $\mathrm{V}_{\mathrm{DD}}+0.2$ | V | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$ |
| Input High Voltage PO, <br> P1, P2 (TTL input) | $\mathrm{V}_{\mathrm{HH} 1}$ | 2.4 | - | $\mathrm{V}_{\mathrm{DD}}+0.2$ | V | $\mathrm{V}_{\mathrm{DD}}=5.5 \mathrm{~V}$ |
|  |  | 2.0 | - | $\mathrm{V}_{\mathrm{DD}}+0.2$ | V | $V_{D D}=3.0 \mathrm{~V}$ |
| Negative going threshold (Schmitt input) | VILS | -0.5 | - | $0.3 \mathrm{~V}_{\text {DD }}$ | V |  |
| Positive going threshold (Schmitt input) | $\mathrm{V}_{\text {IHS }}$ | $0.7 \mathrm{~V}_{\mathrm{DD}}$ | - | $\mathrm{V}_{\mathrm{DD}}+0.5$ | V |  |
| Hysteresis voltage | $\mathrm{V}_{\mathrm{HY}}$ |  | $0.2 \mathrm{~V}_{\mathrm{DD}}$ |  | V |  |
| Source Current P0, P1, P2 <br> (Quasi-bidirectional Mode) | $\mathrm{I}_{\text {sr1 }}$ | -180 | -210 | -360 | uA | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}}=2.4 \mathrm{~V}$ |
| Sink Current P0, P1, P2 (Quasi-bidirectional Mode) | ISK2 | 13 | 18.5 | 24 | mA | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{S}}=0.45 \mathrm{~V}$ |
| Output Low Voltage P0, | $\mathrm{V}_{\text {OL1 }}$ | - | 0.5 | 0.9 | V | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}, \mathrm{l}_{\mathrm{OL}}=20 \mathrm{~mA}$ |
| (PUSH-PULL Mode) |  | - | 0.1 | 0.4 | V | $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}, \mathrm{l}_{\mathrm{OL}}=3.2 \mathrm{~mA}$ |
| Output High Voltage P0, P1, P2 <br> (PUSH-PULL Mode) | VOH | 2.4 | 3.4 | - | V | $\mathrm{V}_{\mathrm{DD}}=4.5 \mathrm{~V}, \mathrm{I}_{\mathrm{OH}}=-16 \mathrm{~mA}$ |
|  |  | 1.9 | 2.4 | - | V | $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}, \mathrm{I}_{\mathrm{OH}}=-3.2 \mathrm{~mA}$ |
| Brownout voltage with $\mathrm{BOV}=1$ | $V_{\text {BO2.5 }}$ | 2.4 | - | 2.7 | V | $\mathrm{TA}=-0$ to $70^{\circ} \mathrm{C}$ |
| Brownout voltage with $\mathrm{BOV}=0$ | $\mathrm{V}_{\text {воз.8 }}$ | 3.5 | - | 4.0 | V | $\mathrm{TA}=-0$ to $70^{\circ} \mathrm{C}$ |
| Comparator Reference Voltage | Vref | 1.02 | 1.20 | 1.31 | V |  |

Notes: *1. /RST pin is a Schmitt trigger input.
*2. XTAL1 is a CMOS input.
*3. Pins of P0, P1 and P2 can source a transition current when they are being externally driven from 1 to 0 . The transition current reaches its maximum value when Vin approximates to 2 V .

### 28.3 The ADC Converter DC ELECTRICAL CHARACTERISTICS

$\left(V_{D D}-V_{S S}=3.0 \sim 5 \mathrm{~V}, \mathrm{TA}=-40 \sim 85^{\circ} \mathrm{C}\right.$, Fosc $=20 \mathrm{MHz}$, unless otherwise specified. $)$

| PARAMETER | SYMBOL | SPECIFICATION |  |  | TEST CONDITIONS |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | TYP. | MAX. |  |  |
| Analog input | AVIN | $\mathrm{V}_{\mathrm{SS}}-0.2$ |  | $\mathrm{~V}_{\mathrm{DD}}+0.2$ | V |  |
| ADC clock | ADCCLK | 200 KH <br> $z$ |  | 5 MHz | Hz | ADC circuit input <br> clock |
| Conversion time | $\mathrm{t}_{\mathrm{C}}$ |  | $52 \mathrm{t}_{\mathrm{ADC}}{ }^{[1]}$ |  | us |  |
| Differential non-linearity | DNL | -1 | - | +1 | LSB |  |
| Integral non-linearity | INL | -2 | - | +2 | LSB |  |
| Offset error | Ofe | -1 | - | +1 | LSB |  |
| Gain error | Ge | -1 | - | +1 | $\%$ |  |
| Absolute voltage error | Ae | -3 | - | +3 | LSB |  |

Notes: 1. tADC: The period time of ADC input clock.

### 28.4 The COMPARATOR ELECTRICAL CHARACTERISTICS

$\left(V_{D D}-V_{S S}=3.0 \sim 5 \mathrm{~V}, \mathrm{TA}=-40 \sim 85^{\circ} \mathrm{C}, \mathrm{FoSC}=20 \mathrm{MHz}\right.$, unless otherwise specified.)

| PARAMETER | SYMBOL | SPECIFICATION |  |  |  | TEST CONDITIONS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | TYP. | MAX. | UNIT |  |
| Common mode range comparator inputs | $\mathrm{V}_{\mathrm{CR}}$ | 0 |  | $\mathrm{V}_{\mathrm{DD}}-0.3$ | V |  |
| Common mode rejection ratio | CMRR |  |  | -50 | dB |  |
| Response time | $\mathrm{t}_{\mathrm{RS}}$ | - | 30 | 100 | ns |  |
| Comparator enable to output valid time | $t_{E N}$ | - | 1 | 5 | us |  |
| Input leakage current, comparator | $I_{\text {IL }}$ | -10 | 0 | 10 | uA | $0<\mathrm{V}_{\text {IN }}<\mathrm{V}_{\mathrm{DD}}$ |

### 28.5 AC ELECTRICAL CHARACTERISTICS



Note: Duty cycle is 50\%.

### 28.6 EXTERNAL CLOCK CHARACTERISTICS

| PARAMETER | SYMBOL | MIN. | TYP. | MAX. | UNITS | NOTES |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Clock High Time | $\mathrm{t}_{\mathrm{CHCX}}$ | 12.5 | - | - | nS |  |
| Clock Low Time | $\mathrm{t}_{\mathrm{CLCX}}$ | 12.5 | - | - | nS |  |
| Clock Rise Time | $\mathrm{t}_{\mathrm{CLCH}}$ | - | - | 10 | nS |  |
| Clock Fall Time | $\mathrm{t}_{\mathrm{CHCL}}$ | - | - | 10 | nS |  |

### 28.7 AC SPECIFICATION

| PARAMETER | SYMBOL | VARIABLE CLOCK MIN. | VARIABLE CLOCK MAX. | UNITS |
| :---: | :---: | :---: | :---: | :---: |
| Oscillator Frequency | $1 / \mathrm{t}_{\mathrm{CLCL}}$ | 0 | 20 | MHz |

### 28.8 TYPICAL APPLICATION CIRCUITS

| CRYSTAL | C1 | C2 | R |
| :---: | :---: | :---: | :---: |
| $4 \mathrm{MHz} \sim 20 \mathrm{MHz}$ | without | without | without |

The above table shows the reference values for crystal applications.


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## 29 PACKAGE DIMENSIONS

29.1 20-pin SOP


Figure 29-1: 20L SOP-300mil

### 29.2 20-pin DIP



| Symbol | Dimension in inch |  | Dimension in mm |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min | Nom | Max | Min | Nom | Max |
| $\mathbf{A}$ | - | - | 0.175 | - | - | 4.45 |
| $\mathbf{A}_{\mathbf{1}}$ | 0.010 | - | - | 0.25 | - | - |
| $\mathbf{A}_{\mathbf{2}}$ | 0.125 | 0.130 | 0.135 | 3.18 | 3.30 | 3.43 |
| $\mathbf{B}$ | 0.016 | 0.018 | 0.022 | 0.41 | 0.46 | 0.56 |
| $\mathbf{B}_{\mathbf{1}}$ | 0.058 | 0.060 | 0.064 | 1.47 | 1.52 | 1.63 |
| $\mathbf{C}$ | 0.008 | 0.010 | 0.014 | 0.20 | 0.25 | 0.36 |
| $\mathbf{D}$ | - | 1.026 | 1.040 | - | 20.06 | 26.42 |
| $\mathbf{E}$ | 0.290 | 0.300 | 0.310 | 7.37 | 7.62 | 7.87 |
| $\mathbf{E}_{\mathbf{1}}$ | 0.245 | 0.250 | 0.255 | 6.22 | 6.35 | 6.48 |
| $\mathbf{e}_{\mathbf{1}}$ | 0.090 | 0.100 | 0.110 | 2.29 | 2.54 | 2.79 |
| $\mathbf{L}$ | 0.120 | 0.130 | 0.140 | 3.05 | 3.30 | 3.56 |
| $\boldsymbol{\alpha}$ | 0 | - | 15 | 0 | - | 15 |
| $\mathbf{e}_{\mathbf{A}}$ | 0.335 | 0.355 | 0.375 | 8.51 | 9.02 | 9.53 |
| $\mathbf{S}^{2}$ | - | - | 0.075 | - | - | 1.91 |

Figure 29-2: 20L PDIP 300mil

## 30 REVISION HISTORY

| VERSION | DATE | PAGE | DESCRIPTION |
| :---: | :---: | :---: | :--- |
| A1 | Mar. 30, 2006 | - | Initial Issued |
| A2 | June 04, 2006 |  | To add W79E823/822 parts and function description |
| A3 | June 19, 2006 | $4 \sim 5$ | To remove normal package of part number list |
| A4 | June 23, 2006 | $86 \sim 92$ | To revise PWM function of W79E823 and W79E822 to 4 <br> PWM output channel |
| A5 | Oct. 04, 2006 | 8 <br> 98 <br> $112 \sim 114$ <br> 96 <br> 89 | 1. To remove block diagram. <br> 2. To add GC call function description. <br> 3. Revise DC characteristics spec. <br> 4. Revise Timer Counter of I2C block diagram. <br> 5. Revise "Release brake of PWM" flow chart. |
| A6 | Mar. 12, 2007 | 112 | 1. Added note for icp mode. <br> 2. Change SFR name IE1 to EIE due to conflict with IE1 <br> bit name in TCON.3. |
| 3. |  | 3FR Reset Value Table: Added PWMCON3, <br> NVMADDR, NVMDAT \& NVMCON. Rename PWMP1 <br> to PWMPH, PWMP0 to PWMPL. |  |
| 4. Port 2 SFR bit P2.0 and P2.1 descriptions reverse |  |  |  |
| order. |  |  |  |

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| A9 | Feb. 22, 2008 |  | 1. Updated keyboard interrupt diagram. IE1 changed to EIE. <br> 2. Update part name in header with version appended <br> 3. Update Maximum rating table <br> 4. Modify power down wakeup by external interrupt pins <br> 5. Revise max of $\mathrm{I}_{\mathrm{IN} 1}$ to 15 uA <br> 6. Change to header 6.3 for "Serial I/O" <br> 7. Update SFR PWMCON1 bit names PWMF, PWMOL, PWM1L, PWM2L, PWM3L to CF, PWM0I, PWM1I, PWM2I, PWM3I <br> 8. Update EWRST $=0, T A=F F H$ initial value after reset <br> 9. Rename INTO and INT1 to /INTO and /INT1 <br> 10.Type error: <br> a. 13 interrupt sources. <br> b. Corrected grammar. <br> c. Changed EWDRST to EWRST. <br> d. Changed ENT0 and ENT1 bits to T0OE and T1OE. <br> e. "pre machine cycle" should be "per machine cycle" <br> f. Change "fail time" to "fall time" <br> g. P2M1 change to P2M2 <br> h. Mnemonic PWM2L change to PWM3L <br> 11. Added Idd2 in dc spec. <br> 12. Removed 24 pin package dimension <br> 13. Update dual data pointer diagram <br> 14. Update WDT block diagram: IE1.4 to EIE.4, WDRF to WTRF <br> 15. Internal oscillator update $($ Fosc1,Fosc0 $)=01 b$ <br> 16. Update ADCS bit descriptions <br> 17. All WDCON bits except WTRF are TA protected <br> 18. Revise table 12-3 Vector location for Interrupt sources and power down wakeup, for watchdog timerdoes not support power dowm mode wakeup <br> 19. Add Figure11-1, 11-2 for reset and VDD monitor timing diagram <br> 20. Revise Chapter 11.1.2 Power-ON reset description <br> 21. Revise Figure 15-1 Watchdog Timer diagram <br> 22. Remove Config1 bit7 WDTE function <br> 23. Revise Figure 24-2 ADC Block diagram |
| :---: | :---: | :---: | :---: |
| A10 | August 4, 2008 |  | 1. Modify the edition format. |

## W79E825A/824A/823B/822B Data Sheet

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